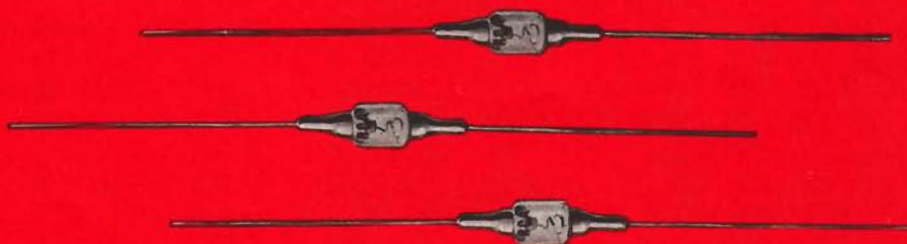


RCA Transistors and Semiconductor Diodes

- CHARACTERISTICS
- CIRCUITS
- THEORY
- INTERCHANGEABILITY
DIRECTORY



RADIO CORPORATION OF AMERICA

® SEMICONDUCTOR DIVISION

SOMERVILLE, N. J.

TRANSISTOR CONSIDERATIONS

INTRODUCTION

Transistors are a new form of electron device. They can perform many of the functions of an electron tube and, in addition, can do some things better and more efficiently than electron tubes. Unlike electron tubes which depend for their functioning on the flow of electrons through a vacuum, a gas, or a vapor, transistors make use of the flow of electric charge carriers in a solid — a semiconductor.

A semiconductor is a material having a conductivity lower than that of metals but higher than that of insulators. The conductivity of a material is determined by the number of charge carriers present. There are many varieties of semiconductors, but the two presently employed in the manufacture of transistors are germanium and silicon. Germanium in its very purest state behaves very much like an insulator because it has very few charge carriers. However, the conductivity of this material can be increased by the addition of almost infinitesimal amounts of certain impurities which cause an increase in the supply of charge carriers. Impurities such as antimony and arsenic increase the number of free electrons, or mobile negative charges while others such as gallium and indium increase the quantity of holes, or mobile positive charges.

The concept of a hole as a mobile positive charge with a definite mass and velocity dates back to 1931. This concept is really an abstraction from the equations of physics which describe the manner in which certain materials conduct electricity. These equations indicate that the material acts as if such a positive particle really existed. The term "hole" in the following discussion will denote a mobile "particle" having a positive charge. The notion of a hole is relatively new, but it is just as useful as the notion of the manner in which electricity is conducted by electrons. The conduction of electricity by electrons is also a very complex matter, though much is taken for granted about it.

A germanium crystal having more holes than free electrons is identified as p-type, because it depends on positive "particles" of electricity, holes, for conduction. On the other hand, a conducting germanium crystal having an excess of free electrons is identified as n-type, because it depends on negative particles of electricity, electrons, for conduction.

As previously implied, holes and mobile electrons are simultaneously present in both p-type and n-type materials. However, one kind of charge carrier predominates and is called the majority carrier. When a hole and a mobile electron meet, they combine with each other and cease to exist as *mobile* charges. This event is often referred to as a recombination. While this event is taking place, other electrons in the structure of the crystal are separated from the atoms they are associated with and result in new mobile electrons

and holes to restore the equilibrium condition. There are many recombinations of holes and mobile electrons like these taking place every second in both p-type and n-type materials.

PRINCIPLES OF OPERATION

To gain some facility in the use of these concepts, consider the situation shown in Fig. 1. The applied electric field causes the holes in the p-type material to drift towards its upper end, point "A". At the same time the free electrons in the wire are repelled by the negative terminal of the battery towards the same point. When electrons enter the p-type material, they readily find holes with which to recombine. This recombination depletes the region around point "A" of holes. The equilibrium condition is restored by the arrival of new holes generated by electron-hole pairs within the p-type material. The new free electrons, released by the electron-hole pairs, move towards the positive terminal of the battery, while the holes move towards the end of the p-type material nearest the negative terminal of the battery in order to maintain equilibrium with electrons arriving in this area.

Consider what happens when an n-type material and a p-type material are joined by suitable means and a voltage is applied as indicated in Fig. 2.

The holes in the p-type material are repelled by the battery voltage towards the junction between the n-type and p-type materials, and the electrons in the n-type material are also repelled by the battery voltage towards this junction. In the vicinity of the junction, the holes and electrons meet and combine with each other, and thereupon cease to exist as mobile carriers. In some cases holes will flow a short distance into the n-type material or electrons will flow a short distance into the p-type material before combining. At the opposite end of the n-type material more electrons arrive from the negative terminal of the battery to replenish the electrons that combined with holes at the junction. Similarly, at the end of the p-type material farthest from the junction, electrons leave the crystal, because new holes are created in order to replenish those cancelled at the junction.

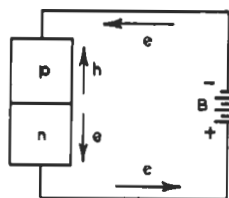


Fig. 3

cancelled at the junction.

Consider the case when the positions of the n-type and p-type materials are reversed as in Fig. 3.

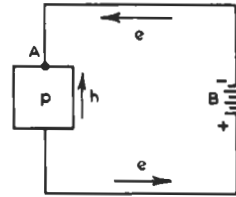


Fig. 1

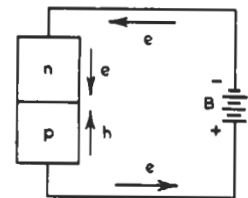


Fig. 2

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There is an initial movement of the holes in the p-type material away from the junction, due to the attraction forces established by the battery voltage. In like manner there is an initial movement of the electrons in the n-type material away from the junction. Thus, there are practically no majority carriers left at the junction and a barrier or depletion region is formed. Conduction across this region practically ceases because it has very few mobile charge carriers. The mobile charges left are minority carriers and are capable of maintaining a current of only a few microamperes.

OPERATION OF JUNCTION TRANSISTORS

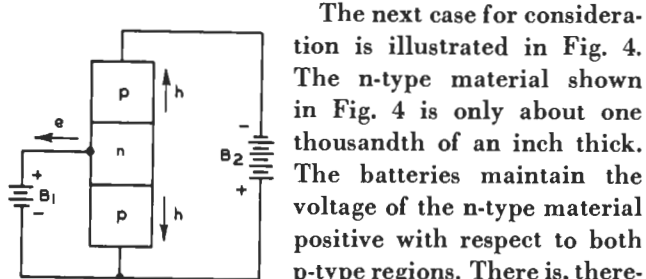


Fig. 4

The next case for consideration is illustrated in Fig. 4. The n-type material shown in Fig. 4 is only about one thousandth of an inch thick. The batteries maintain the voltage of the n-type material positive with respect to both p-type regions. There is, therefore, an initial movement of the mobile electrons in the

base away from both junctions, and an initial movement of the holes in both of the p-type materials away from the base. After these initial displacements of charge carriers have taken place there is no further current flow because in the neighborhood of both junctions there are very few mobile charge carriers.

Consider the situation illustrated by Fig. 5. Since the voltage of battery B_2 is several times greater than the voltage of B_1 , the voltage of the upper piece of p-type material is more negative than the voltage of the n-type material. Therefore, as in the preceding example, the region around the junction marked "C" is almost devoid of charge carriers. However, in the vicinity of the junction marked "E" there is a relatively high concentration of charge carriers because the lower piece of p-type material has a voltage more positive than that of the n-type material. It might seem at first that no current can flow due to the

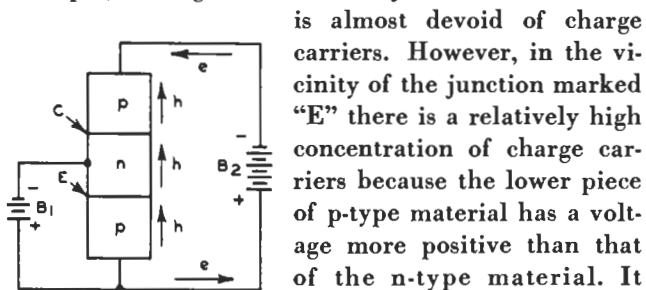


Fig. 5

relative scarcity of charge carriers in the region of "C". However, the n-type material is relatively thin and may be regarded as porous as far as holes are concerned. The thin n-type material appears to be porous because this material has very few mobile negative charges relative to the number of holes diffusing through it, and because only a few of the holes will recombine with the electrons. Holes diffuse from the lower piece of p-type material near junction "E"

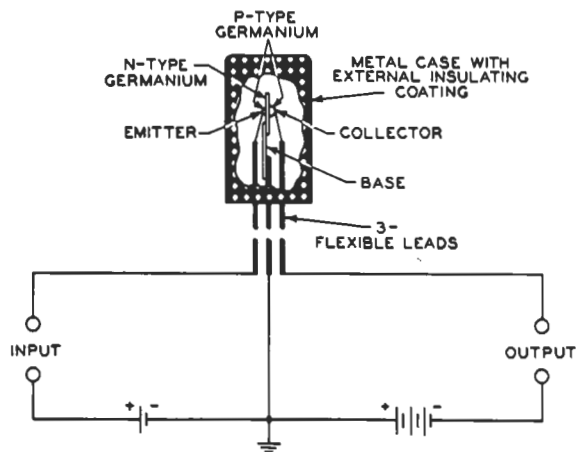


Fig. 6

through the n-type material and arrive at the upper piece of p-type material near junction "C". The lower piece of p-type material is referred to as the *Emitter*; the upper piece is referred to as the *Collector*. These names were adopted because the emitter emits charge carriers and the collector collects them. The junction marked "E" is, therefore, the emitter junction; the junction marked "C", the collector junction. The material between the two junctions is referred to as the *Base*. The complete structure constitutes the basic components of a p-n-p type of junction transistor such as the RCA-2N109, 2N175, 2N270 etc. A diagrammatic sketch showing the structural arrangement of a junction-type transistor with associated simple circuit is shown in Fig. 6.

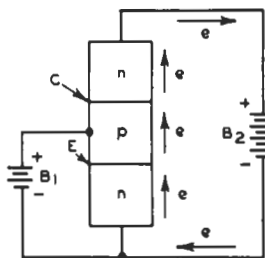


Fig. 7

The discussion of Fig. 4 and Fig. 5 illustrates the fact that the base-to-emitter voltage can, by controlling the density of holes at junction "E", also control the difference in the density of the holes between junctions "C" and "E". This control leads in turn to control of the rate of diffusion of holes through the base. The current in the collector circuit corresponds to this rate of diffusion of holes. In practice, the collector current of a p-n-p type transistor can be varied from a few microamperes up to the maximum current rating without exceeding a base-to-emitter voltage of zero volts. Moreover, the power which the base-to-emitter circuit utilizes in controlling the collector current is very small in comparison with the power available in the collector-to-emitter circuit.

A similar discussion for the n-p-n type of junction

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transistor will lead up to the circuit shown in Fig. 7, which is similar to Fig. 5.

However, observe the change in the relative positions of the n-type and p-type materials and the change in the battery polarities. In this case electrons are emitted by the n-type emitter. The electrons diffuse through the p-type base and are collected by the n-type collector. This type of transistor is referred to as an n-p-n junction type.

POTENTIAL ENERGY DIAGRAM

Much of the preceding discussion concerning the principles of operation of a transistor can be sum-

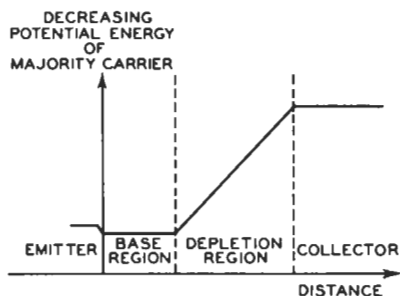


Fig. 8

marized by a potential-energy diagram. Such a diagram is shown in Fig. 8.

An electron at the negative terminal of a battery has more potential energy than an electron at the positive terminal. Similarly, if the transistor is a p-n-p type and the collector is, therefore, connected to the negative terminal of the battery, a hole has the least potential energy when it reaches the collector. This diagram also applies to an n-p-n transistor but differs in that an electron has the least potential energy when it reaches the collector. A statement covering both cases is that the majority carriers originating in the emitter region move in the direction of decreasing potential energy. It is important to note that a small amount of energy must be supplied by the base-to-emitter circuit to move a majority carrier out of the emitter into the base. Because of the addition of this small amount of energy, the charge carrier has somewhat more potential energy in the emitter-to-collector circuit than it had before. Once the charge carrier has entered the base, the carrier mingles with the other charges in this region. Those carriers which find their way to the edge of the depletion region in the base are quickly swept across the depletion region by the electric field established by the collector-to-emitter supply voltage.

"DRIFT" TRANSISTORS

As previously noted, the signal current in a conventional transistor is transmitted across the base

region by a diffusion process. The transit time of the charge carriers across this region is, therefore, relatively long. RCA has developed a technique for the manufacture of transistors which do not depend upon diffusion for the transmission of the signal across the base region. Transistors manufactured according to this new technique, such as the RCA-2N247, are known as "drift" transistors. Diffusion of charge carriers across the base region is eliminated and the charge carriers are propelled across this region by a "built-in" electric field. The resulting reduction in the transit time of the charge carriers permits "drift" transistors to be used at much higher frequencies than transistors of conventional design.

The "built-in" electric field is in the base region of the "drift" transistor. This field is achieved by utilizing an impurity density which varies from one side of the base to the other. The impurity density is high

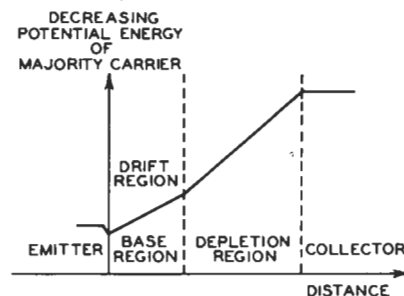


Fig. 9

next to the emitter and low next to the collector. Thus, there are more mobile electrons in the region near the emitter than in the region near the collector and they will try to diffuse evenly throughout the base. However, any displacement of the negative charge leaves a positive charge in the region from which the electrons came, because every atom of the base material was originally electrically neutral. The displacement of the charge creates an electric field that tends to prevent further electron diffusion so

that equilibrium condition is reached. The direction of this field is such as to prevent electron diffusion from the high density area near the emitter to the low density area near the collector. Therefore, holes entering the base will be accelerated from the emitter to the collector by the electric field.

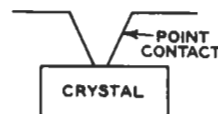


Fig. 10

Thus the diffusion of charge carriers across the base region is augmented by the built-in electric field. A potential energy diagram for a drift transistor appears in Fig. 9.

The transistors discussed so far are known as junction transistors, and they may be made as either p-n-p types or n-p-n types. The essential idea in the manu-

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facture of these transistors is to produce a sandwich-type assembly, i.e., p-type regions and n-type regions in intimate contact with each other. However, it is also possible to manufacture a transistor in which the points of certain kinds of wire touch the surface of an n-type crystal, as illustrated in Fig. 10. The points of the wires must be very close together. After the structure is completed, currents are passed through each of the two contact points in such a way that p-type material is formed in the region of the contact points. Such a transistor is referred to as a point-contact transistor, and has largely been superseded by the junction-type of transistor.

The point-contact type of semiconductor diode is also made by a similar technique, but this type of semiconductor device utilizes only a single point contact. The crystal on which the point contact is made may be either p-type or n-type.

SIGNIFICANCE OF POLARITY

In the preceding discussion, reference was made to the particular kind of charge carriers present. The following discussion will be in terms of conventional current flow, i.e. the flow of positive charge from the positive terminal of the battery to the negative terminal as shown in Fig. 11. The actual flow is a flow of negative charge (electrons) from the negative terminal of the battery to the positive terminal.

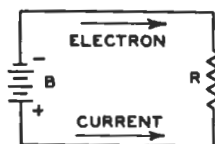


Fig. 11

Because the discussion is based on conventional current flow, the current will flow counterclockwise through the resistor and circuit. Suppose that the magnitude of the current is 10 milliamperes. If the direction of the current is important to the discussion, it can be assumed that the positive direction of the flow of the positive charges is counterclockwise around the circuit and the value of the current is + 10 milliamperes. However, if it had been assumed that the positive charges flowed in a clockwise direction, then the value of the current would be -10 milliamperes, the - sign indicating that the positive charges are actually flowing in a direction opposite to that assumed.



Fig. 12

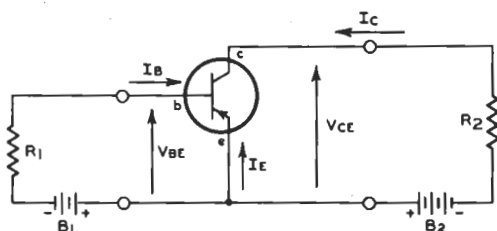


Fig. 13

It has been the custom for many years to assume positive directions of voltage and conventional current flow at the input and output of any device as indicated in Fig. 12. For example, a germanium junction transistor of the p-n-p type would be connected into a common-emitter circuit in the manner indicated in Fig. 13. The assumed positive directions of dc voltage and conventional dc currents are in conformity with the custom indicated in Fig. 12. The emitter current I_E is also assumed to flow into the transistor. Actually, the conventional base current I_B flows out of the transistor; the emitter current, into the transistor; and the collector current I_C , out of the transistor. Thus, the algebraic sign of the numerical value of the base current and of the numerical value of the collector current is negative, while the algebraic sign of the numerical value of the emitter current is positive. Also, as indicated by Fig. 13 the base voltage and collector voltage are negative with respect to the emitter. Therefore, the algebraic sign of the numerical value of V_{BE} and of

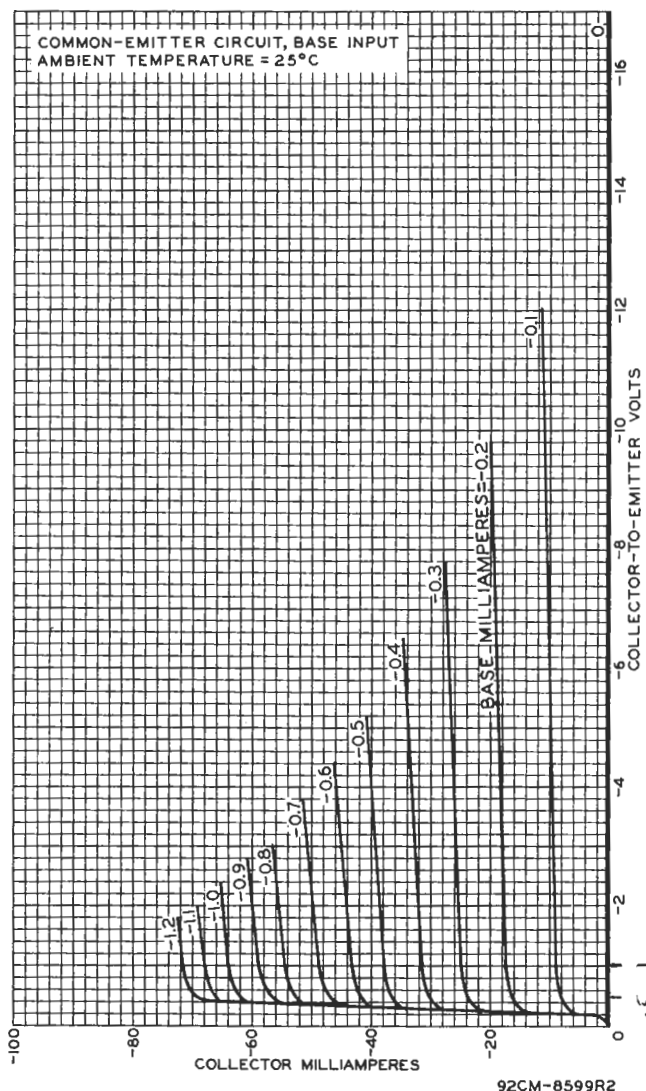


Fig. 14. Average Collector Characteristics.

TRANSISTOR CONSIDERATIONS

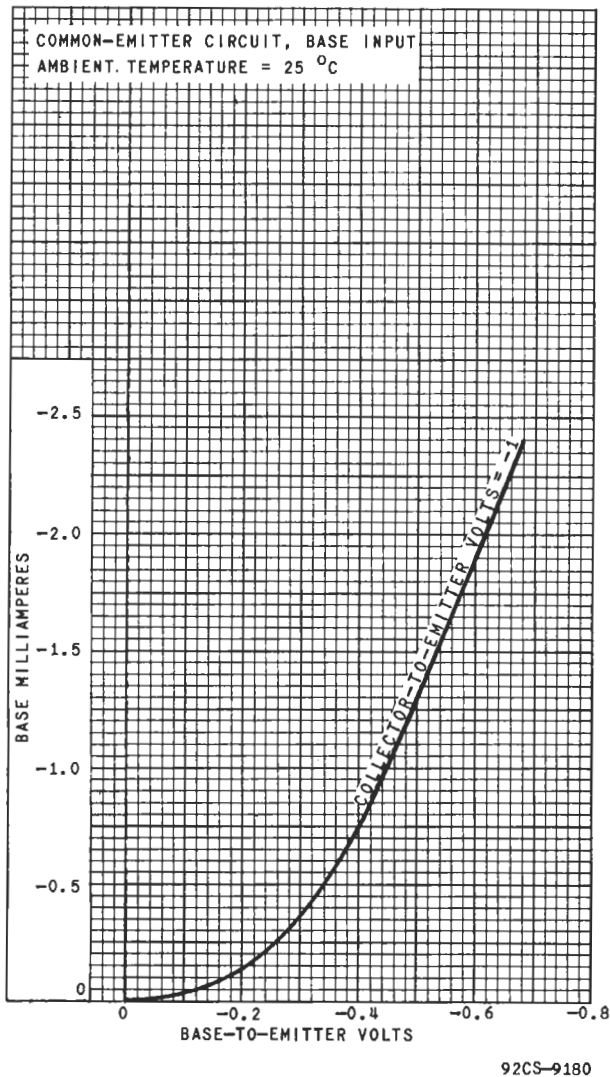


Fig. 15. Average Base Characteristics.

V_{CE} is negative.

These conventions require that the sign of the numerical values shown on the typical characteristics in Fig. 14 and Fig. 15 be negative.

The collector characteristics also show that the collector current is many times greater than the base current. For example, at a collector-to-emitter voltage of -4 volts and a base current of -0.05 milliamperes the collector current is -5 milliamperes. Since it has been assumed that all the currents flow into the transistor, the following equation can be written.

$$I_B + I_E + I_C = 0$$

$$I_E = -I_B - I_C$$

When the numerical values for the base current and collector current are substituted in this equation, the following result is obtained.

$$I_E = -(-0.05) - (-5)$$

$$I_E = 5.05 \text{ milliamperes}$$

There are two facts to be observed from this result.

First, the collector current and emitter current are nearly equal, and second, the algebraic sign of the emitter current is positive, which indicates that the emitter current flows into the transistor as assumed.

Now, suppose that the resistance in the base circuit is reduced. This would cause the base current to increase in magnitude, for example, from -0.05 milliamperes to -0.1 milliamperes. This increase in base current will cause an increase in the magnitude of the collector current by about 4.5 milliamperes.

Because in practical circuits it is desirable to amplify signals, Fig. 13 has been redrawn in Fig. 16 with slight modifications.

When no signal is applied to the input, the transistor voltages and currents are at their quiescent values, and the top of the resistor in the base circuit is positive with respect to the bottom. Suppose a signal voltage V_{BE} of -0.045 volt is applied at the input. The top of the resistor in the base circuit is now less positive by 0.045 volt than it was before the signal was applied. More of the battery voltage V_{BB} must now appear between the base and emitter. The base current must, therefore, increase in magnitude from 0.05 milliamperes to 0.1 milliamperes. The collector current increases in magnitude by about 4.5 milliamperes, and the direction of flow is from top to bottom of the resistor in the collector circuit. Thus, V_{CE} becomes more positive in the indicated direction than it was before a V_{BE} of -0.045 volt was applied at the input. The common-emitter type of circuit, with base input, produces a 180° phase reversal, and the change in the output current is much greater than the change in the input current. Moreover, the resistance in the collector circuit is normally

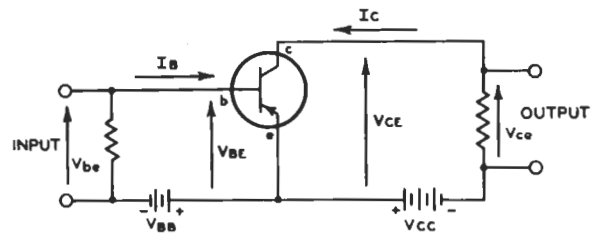


Fig. 16

much greater than the resistance in the base circuit, resulting not only in a greater change in the collector current than in the base current, but also a greater change in the collector voltage than in the base voltage. Both of these factors contribute to the increase in the signal power output as compared with the signal power input.

Fig. 17 shows some typical amplifier circuits for junction transistors of the p-n-p type. These circuits may also be used for junction transistors of the n-p-n type provided the polarities are reversed.

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DEFINITIONS (Continued)

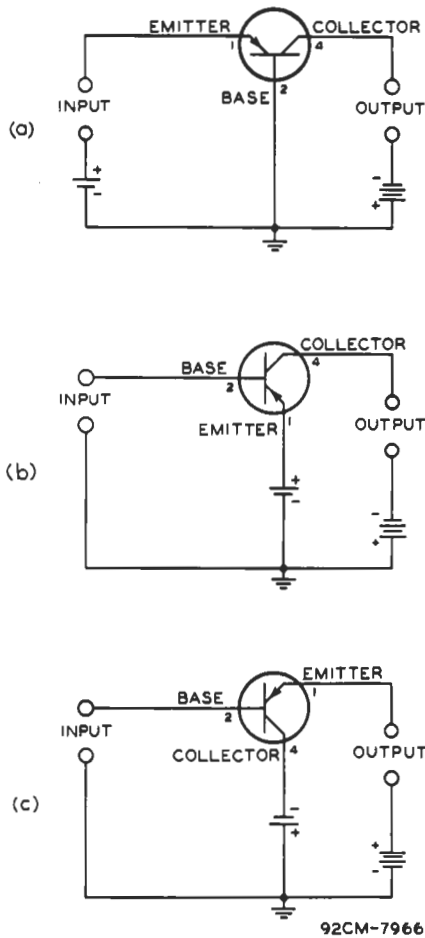


Fig. 17. Typical Amplifier Circuit Connections:
 (a) Common-Base, Emitter-Input.
 (b) Common-Emitter, Base-Input.
 (c) Common-Collector, Base-Input.

DEFINITIONS

Alpha Cutoff Frequency. Frequency at which the forward current transfer ratio drops to 0.707 times its value at 1 kc.

Class A Amplifier. An amplifier in which the bias of the input electrode and the alternating input signal are such that output current flows at all times.

Class B Amplifier. An amplifier in which the bias of the input electrode is such that the output current is approximately zero when no alternating input signal is applied, and such that when an alternating input signal is applied, the output current flows for approximately one-half cycle.

Collector Transition Capacitance. The capacitance across the collector-to-base transition region. (A transition region is a region between two homogeneous semiconductor regions, in which the impurity concentration changes).

Large-Signal DC Current Transfer Ratio. The dc output current divided by the dc input current.

Small-Signal Current Transfer Ratio. The change in output current with ac output circuit shorted divided by the change in input current, measured at 1000 cycles per second.

Unilateralization. Unilateralization is a special case of neutralization in that the feedback parameters are completely balanced out. In the case of transistors, these feedback parameters include a resistive component in addition to capacitive component. Unilateralization changes a bilateral network into a unilateral network.

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RCA TRANSISTOR

 TYPE	CLASS OF SERVICE	GENERAL DATA					MAXIMUM RATINGS (Absolute Values)							TYPICAL OPERATION Ambient Temperature = 25°C				
		Electrical			Mechanical		DC Collector-to-Base volts	DC Emitter-to-Base volts	DC Collector Current ma	DC Emitter Current ma	Collector Dissipation mw at ambient temperature of:			Ambient Storage Temperature °C	Circuit	DC Collector-to-Emitter volts	DC Collector Current ma	Current Transfer Ratio at 1 Kc
		Minimum DC Collector-to-Base volts (for stated DC Collector Current μ a)	Maximum DC Collector Current μ a (for stated DC Collector-to-Base volts)	Maximum DC Emitter Current μ a (for stated DC Emitter-to-Base volts)	Operating Position	Dimensional Outline and Basing or Lead Arrangement see pages 14-16					25°C	55°C	71°C					
Germanium, p-n-p, alloy-junction types ➔																		
2N77	Class A AF Amplifier	-30 at $I_C = -20$	-10 at $V_{CB} = -12$	—	Any	H	-25	—	-15	15	—	35	—	-65 to +85	E E C B	-4 -4 -4 -4	-0.7 -0.7 -0.7 -0.7	55 55 — 0.982
2N104	Class A AF Amplifier	-30 at $I_C = -20$	-10 at $V_{CB} = -12$	-10 at $V_{EB} = -12$	Any	B	-30	-12	-50	50	150	80	35	-65 to +85	E E C B B	-4 -6 -3 -3 -6	-0.7 -1 -0.2 -0.2 -1	— 44 — — 0.977
2N105	Class A AF Amplifier	-25 at $I_C = -10$	-7 at $V_{CB} = -12$	—	Any	C	-25	—	-15	15	—	35	—	-65 to +85	E E C B B	-1.3 -4 -1.3 -4 -1.3	-0.3 -0.7 -0.3 -0.7 -0.3	45 55 — 0.982 0.978
2N109	Large-Signal AF Amplifier	—	-14 at $V_{CB} = -25$	-14 at $V_{EB} = -12$	Any	B	-25 peak -12*	-12	-70 peak -35 avg.	70 peak 35 avg.	150	50	20	-65 to +85	E	-1	-50	75 [■]
2N139	Class A 455-Kc Amplifier	-16 at $I_C = -10$	-6 at $V_{CB} = -12$	-12 at $V_{EB} = -0.5$	Any	B	-16	-0.5	-15	15	80	35	10	-65 to +85	E E B B	-9 -9 -9 -9	-0.5 -1 -0.5 -1	45 48 0.978 0.980
2N140	540-1640 Kc Converter	-16 at $I_C = -10$	-6 at $V_{CB} = -12$	-12 at $V_{EB} = -0.5$	Any	B	-16	-0.5	-15	15	80	35	10	-65 to +85	E B	-9 -9	-0.6 -0.6	75 0.987
2N175	Class A Low-Noise AF Amplifier	—	-12 at $V_{CB} = -25$	-12 at $V_{EB} = -12$	Any	B	-10	-10	-2	2	50	20	10	-65 to +85	E B	-4 -4	-0.5 -0.5	65 0.985
2N206	Class A AF Amplifier	—	-10 at $V_{CB} = -30$	-10 at $V_{EB} = -12$	Any	H	-30	-12	-50	50	75	—	—	-65 to +85	E C B B	-5 -3 -3 -5	-1 -0.2 -0.2 -1	47 — 0.974 0.980
2N215	Like RCA-2N104 but has flexible leads.																	
2N217	Like RCA-2N109 but has flexible leads.																	
2N218	Like RCA-2N139 but has flexible leads.																	
2N219	Like RCA-2N140 but has flexible leads.																	
2N220	Like RCA-2N175 but has flexible leads.																	
2N247	Class A RF Amplifier	-40 at $I_C = -50$	-16 at $V_{CB} = -30$	-50 at $V_{EB} = -0.5$	Any	E	-35	-0.5	-10	10	80	50	35	-65 to +85	E E B	-9 -9 -9	-1 -1 -1	60 60 0.984
2N269	Low-Level Switch	-25 at $I_C = -20$	-2.5 at $V_{CB} = -2.5$	-2.5 at $V_{EB} = -2.5$	Any	D	-20	-9	-100	100	120	35	10	-65 to +85	B	-6	-1	—
2N270	Large-Signal AF Amplifier	—	-16 at $V_{CB} = -25$	-12 at $V_{EB} = -12$	Any	J	-25 peak -12*	-12	-150 peak -75 avg.	150 peak 75 avg.	250	150	60	-65 to +85	E E	-1 -6.7	-150 -19	70 [■] —
2N274	Class A RF Amplifier	-40 at $I_C = -50$	-16 at $V_{CB} = -30$	-50 at $V_{EB} = -0.5$	Any	I	-35	-0.5	-10	10	80	50	35	-65 to +85	E E B	-9 -9 -9	-1 -1 -1	60 60 0.984

♦ Cutoff Data. Collector-to-Base Voltage and Collector Current values obtained with emitter open. Emitter Current values obtained with collector open. Ambient temperature = 25°C.
 • Storage or Operating Temperatures above 85°C will affect the serviceability of the transistor.

B = Common-Base, Emitter-Input Circuit.
 C = Common-Collector, Base-Input Circuit.
 E = Common-Emitter, Base-Input Circuit.
 * For inductive load.
 ■ Large-Signal DC Current Transfer Ratio.

CHARACTERISTICS CHART I

TYPICAL OPERATION (Cont'd) Ambient Temperature = 25°C						CHARACTERISTICS For COMMON-EMITTER CIRCUIT, BASE INPUT at Ambient Temperature = 25°C Small-Signal Hybrid- π Parameters For equivalent circuit, see page 14												RCA TYPE
Power Gain		Noise Factor			Alpha Cutoff Frequency Mc (f _{αb})	DC Collector-to-Emitter volts (V _{CE})	R _S Collector Current ma (I _c)	Resistance ohms (r _{bb'})	Conductance μmhos			Capacitance μμf		Intrinsic Trans-conductance μmhos (g _m)	Frequency For Unity Power Amplification Mc			
Input Resistance ohms	Load Resistance ohms	Power Gain db	Generator Resistance ohms	Load Resistance ohms					Noise Factor db	g _{b'e}	g _{ce}	g _{b'c}	C _{b'e}			C _{b'c}		
Small-Signal H Parameters and Small-Signal T Parameters are shown for some types on page 12. Germanium, p-n-p, alloy-junction types																		
1980 2670 500000 215	100000 2670 10000 500000	44.1 34.5 17 32.5	1000 — — —	20000 — — —	6.5 — — —	— — — 0.70	— -4 — —	— -0.7 — —	240 — — —	407 — — —	6.8 — — —	0.13 — — —	5000 — — —	40 — — —	22300 — — —	1.7 — — —	2N77	
— 1400 500000 — 170	— 20000 18000 — 500000	— 41 14.3 — 32.4	518 — — — —	20000 — — — —	12 max. — — — —	— — — 0.53 0.70	— -3 -6 — —	— -0.2 -1 — —	300 — 290 — —	174 — 727 — —	1.28 — 6.62 — —	0.3 — 0.36 — —	1225 — 6900 — —	36 — 40 — —	5540 — 32000 — —	1.6 — 1.6 — —	2N104	
4700 2300 500000 180 —	4700 20000 13000 500000	32.5 42 16 33.2 —	1000 — — — —	20000 — — — —	16.5 max. — — — —	— — — 0.75 0.64	— -1.3 -4 — —	— -0.3 -0.7 — —	260 — 250 — —	220 — 380 — —	3.1 — 4.5 — —	0.20 — 0.21 — —	2500 — 4500 — —	27 — 17 — —	10000 — 21000 — —	1.9 — 2.6 — —	2N105	
For typical class B push-pull data, see page 12.																		
1000 500 — —	70000 30000 — —	38.8□ 37.8□ — —	— — — —	— — — —	4.5 spot□ 4.5 spot□ — —	— — 6.8 6.7	— -9 -9 —	— -0.5 -1 —	85 75 — —	425 800 — —	4.6 8.6 — —	0.22 0.25 — —	540 1100 — —	9.5 9.5 — —	19300 38600 — —	13 14 — —	2N139	
700 —	75000 —	32 —	(Useful conversion gain)			— 10	-9 —	-0.6 —	85 —	480 —	5.4 —	0.23 —	650 —	9.5 —	22600 —	16.5 —	2N140	
2000 —	70000 —	43 —	1000 —	20000 —	6 max. —	— 0.85	-4 —	-0.5 —	190 —	296 —	6.6 —	0.279 —	3900 —	36 —	19200 —	2.06 —	2N175	
1200 560000 — 140	20000 18000 — 500000	46 30 — 35	1000 — — —	20000 — — —	9 — — —	— — 0.62 0.78	— -3 -5 —	— -0.2 -1 —	210 200 — —	225 710 — —	2.3 7.6 — —	0.27 0.4 — —	2380 9400 — —	48 35 — —	7430 33400 — —	— — — —	2N206	
Like RCA-2N104 but has flexible leads.																		
Like RCA-2N109 but has flexible leads.																		
Like RCA-2N139 but has flexible leads.																		
Like RCA-2N140 but has flexible leads.																		
Like RCA-2N175 but has flexible leads.																		
1350 170 —	70000 4500 —	45□ 24□ —	← at 1.5 Mc ← at 10.7 Mc	8 spot — —	— — —	— — 30	-9 — —	-1 — —	40 — —	640 — —	— — —	— — —	200 — —	1.7 — —	37000 — —	132 — —	2N247	
For typical data in Low-Level Switching Service, see page 12.																		
For typical class B push-pull data, see page 12.																		
1350 170 —	70000 4500 —	45□ 24□ —	← at 1.5 Mc ← at 10.7 Mc	8 spot — —	— — —	— — 30	-9 — —	-1 — —	40 — —	640 — —	— — —	— — —	200 — —	1.7 — —	37000 — —	132 — —	2N274	

† Measured with a noise diode and a thermocouple voltmeter with an equivalent noise bandwidth of 12.3 Kc and geometric mean of 3000 cps.

□ Measured in a single-tuned unilateralized circuit matched to the generator and load impedances for maximum transfer of power (transformer insertion losses not included).

‡ This frequency (figure of merit) may be calculated from the equation:

$$f = \frac{1}{4\pi} \sqrt{\frac{g_m}{r_{bb'} C_{b'e} C_{b'c}}}$$

RCA TRANSISTOR

 TYPE	CLASS OF SERVICE	GENERAL DATA					MAXIMUM RATINGS (Absolute Values)							TYPICAL OPERATION Ambient Temperature = 25°C				
		Electrical			Mechanical		DC Collector-to-Base volts	DC Emitter-to-Base volts	DC Collector Current ma	DC Emitter Current ma	Collector Dissipation mw at ambient temperature of:			Ambient Storage Temperature °C	Circuit	DC Collector-to-Emitter volts	DC Collector Current ma	Current Transfer Ratio at 1 Kc
		Minimum DC Collector-to-Base volts (for stated DC Collector Current μa)	Maximum DC Collector Current μa (for stated DC Collector-to-Base volts)	Maximum DC Emitter Current μa (for stated DC Emitter-to-Base volts)	Operating Position	Dimensional Outline and Basing or Lead Arrangement see pages 14-15					25°C	55°C	71°C					
Germanium, p-n-p, alloy-junction types ➔																		
2N301	Large-Signal AF Power Amplifier	-40 at I _C = -2000	-220 at V _{CB} = -12	-2000 at V _{EB} = -12	Any	F	-40 peak -20*	-12	-2000 peak -1000 avg.	2000 peak 1000 avg.	—	12000	5500	-65 to +85	E E E	-1.5 -1.5 -13.6	-1000 -500 -400	70 [■] 75 —
2N301-A	Large-Signal AF Power Amplifier	-35 at I _C = -1000	-220 at V _{CB} = -12	-2000 at V _{EB} = -12	Any	F	-60 peak -30*	-12	-2000 peak -1000 avg.	2000 peak 1000 avg.	—	12000	5500	-65 to +85	E E E	-1.5 -1.5 -13.6	-1000 -500 -400	70 [■] 75 —
2N370	Class A RF Amplifier	—	-20 at V _{CB} = -12	-50 at V _{EB} = -1.5	Any	E	-20	-1.5	-10	10	80	35	10	-65 to +85	E E E	-12 -12 -12	-1 -1 -1	60 60 60
2N371	RF Oscillator	—	-20 at V _{CB} = -12	-50 at V _{EB} = -0.5	Any	E	-20	-0.5	-10	10	80	35	10	-65 to +85	B	-12	-1	0.984
2N372	RF Mixer	—	-20 at V _{CB} = -12	-50 at V _{EB} = -0.5	Any	E	-20	-0.5	-10	10	80	35	10	-65 to +85	E E E	-12 -12 -12	-1 -1 -1	60 60 60
2N384	VHF Amplifier	—	-16 at V _{CB} = -12	-50 at V _{EB} = -0.5	Any	I	-30	-0.5	-10	10	120	70	35	-65 to +85	E B	-12 -12	-1.5 -1.5	60 0.984
2N398	High-Voltage Switch	-105 at I _C = -50	-14 at V _{CB} = -2.5	-50 at V _{EB} = -50	Any	G	-105	-50	-100	100	50	10	—	-65 to +85	E	-0.35	-5	60 [■]
2N404	Low-Level Switch	-25 at I _C = -20	-5 at V _{CB} = -12	-2.5 at V _{EB} = -2.5	Any	G	-25	-12	-100	100	120	35	10	-65 to +85	B	-6	-1	—
2N405	Class A AF Driver	—	-14 at V _{CB} = -12	-14 at V _{EB} = -12	Any	B	-12	-2.5	-70	70	150	50	20	-65 to +85	E	-6	-1	35
2N406	Like RCA-2N405 but has flexible leads.																	
2N407	Large Signal AF Amplifier	—	-14 at V _{CB} = -12	-14 at V _{EB} = -2.5	Any	B	-20	-2.5	-70	70	150	50	20	-65 to +85	E	-1	-50	65 [■]
2N408	Like RCA-2N407 but has flexible leads.																	
2N409	Class A 455-Kc Amplifier	-13 at I _C = -10	-10 at V _{CB} = -12	-12 at V _{EB} = -0.5	Any	B	-12	-0.5	-15	15	80	35	10	-65 to +85	E E B B	-9 -9 -9 -9	-0.5 -1 -0.5 -1	45 48 0.978 0.980
2N410	Like RCA-2N409 but has flexible leads.																	
2N411	540-1640 Kc Converter	-13 at I _C = -10	-10 at V _{CB} = -12	-12 at V _{EB} = -0.5	Any	B	-12	-0.5	-15	15	80	35	10	-65 to +85	E B	-9 -9	-0.6 -0.6	75 0.987
2N412	Like RCA-2N411 but has flexible leads.																	

† Cutoff Data. Collector-to-Base Voltage and Collector Current values obtained with emitter open. Emitter Current values obtained with collector open. Ambient temperature = 25°C.
 • Storage or Operating Temperatures above 85°C will affect the serviceability of the transistor.

B = Common-Base, Emitter-Input Circuit.
 C = Common-Collector, Base-Input Circuit.
 E = Common-Emitter, Base-Input Circuit.
 * For inductive load.
 ■ Large-Signal DC Current Transfer Ratio.

CHARACTERISTICS CHART I (Cont'd)

TYPICAL OPERATION (Cont'd) Ambient Temperature = 25°C							CHARACTERISTICS For COMMON-EMITTER CIRCUIT, BASE INPUT at Ambient Temperature = 25°C Small-Signal Hybrid- π Parameters For equivalent circuit, see page 14										RCA	
Power Gain			Noise Factor			Alpha Cutoff Frequency Mc ($f_{\alpha b}$)	DC Collector- to-Emitter volts (V_{CE})	DC Collector Current ma (I_C)	Resistance ohms ($r_{bb'}$)	Conductance μ mhos			Capacitance μ μf		Intrinsic Trans- conductance μ mhos (g_m)	Frequency For Unity Power Amplifi- cation† Mc	TYPE	
Input Resistance ohms	Load Resistance ohms	Power Gain db	Generator Resistance ohms	Load Resistance ohms	Noise Factor‡ db					$g_{b'e}$	g_{ce}	$g_{b'c}$	$C_{b'e}$	$C_{b'c}$				
<div style="display: flex; justify-content: space-between;"> Small-Signal H Parameters and Small-Signal T Parameters are shown for some types on page 13. Germanium, p-n-p, alloy-junction types </div>																		
75	34	32.5	—	—	—	—	—	—	—	—	—	—	—	—	—	—	For typical class A and class B push-pull data, see page 13.	2N301
75	34	32.5	—	—	—	—	—	—	—	—	—	—	—	—	—	—	For typical class A and class B push-pull data, see page 13.	2N301-A
1750 200 100	180000 18000 11000	50.5□ 26.2□ 17.0□	(Maximum useful gain = 31 db) ← at 1.5 Mc (Maximum useful gain = 17.6 db) ← at 10 Mc (Maximum useful gain = 12.5 db) ← at 20 Mc				—	—	—	—	—	—	—	—	37800 21400 13700	132	—	2N370
For Oscillation applications up to 23 Mc							30	For additional data, see page 14.							132	—	2N371	
1750 200 100	180000 18000 11000	50.5□ 26.2□ 17.0□	(Maximum useful gain = 31 db) ← at 1.5 Mc (Maximum useful gain = 17.6 db) ← at 10 Mc (Maximum useful gain = 12.5 db) ← at 20 Mc				—	—	—	—	—	—	—	—	37800 21400 13700	132	—	2N372
400 30	28000 5000	34 15	← at 10.7 Mc ← at 50 Mc		—	—	100	-12	-1.5	50	960	—	—	90	1.3	56800	250	2N384
For high-voltage "on-off" control applications.							—	For design data for High-Voltage Switching Service, see page 13.							—	—	2N398	
For medium-speed "on-off" control applications.							12	For design data for Low-Level Switching Service, see page 13.							—	—	2N404	
750	85000	43	—	—	—	—	—	-6	-1	120	1070	6.5	0.295	—	40	37500	—	2N405
Like RCA-2N405 but has flexible leads.																	2N406	
This type is intended for use in class A and class B push-pull power output stages of radio receivers and audio amplifiers operating at power output levels of approximately 150 milliwatts.																	2N407	
Like RCA-2N407 but has flexible leads.																	2N408	
1000 500 —	70000 30000 —	38.8□ 37.8□ —	— — —	— — —	4.5 spot□ 4.5 spot□ —	— — —	— 6.8 6.7	-9 -9-	-0.5 -1	85 75	425 800	4.6 8.6	0.22 0.25	540 1100	9.5 9.5	19300 38600	13 14	2N409
Like RCA-2N409 but has flexible leads.																	2N410	
700	75000	32	(Useful conversion gain)			—	10	-9	-0.6	85	480	5.4	0.23	650	9.5	22600	16.5	2N411
Like RCA-2N411 but has flexible leads.																	2N412	

† Measured with a noise diode and a thermocouple voltmeter with an equivalent noise bandwidth of 12.3 Kc and geometric mean of 3000 cps.

□ Measured in a single-tuned unilateralized circuit matched to the generator and load impedances for maximum transfer of power (transformer insertion losses not included).

‡ This frequency (figure of merit) may be calculated from the equation:

$$f = \frac{1}{4\pi} \sqrt{\frac{g_m}{r_{bb'} C_{b'e} C_{b'c}}}$$

RCA TRANSISTOR CHARACTERISTICS CHART II

CHARACTERISTICS

For COMMON-EMITTER CIRCUIT, BASE INPUT at Ambient Temperature = 25°C



TYPE

Small-Signal H Parameters

Small-Signal T Parameters

DC Collector-to-Emitter volts (V _{CE})	DC Collector Current ma (I _c)	Input Resistance, Output Circuit Shorted ohms (h _i)	Reverse Voltage Transfer Ratio, Input Circuit Open (h _r)	Forward Current Transfer Ratio, Output Circuit Shorted (h _f)	Output Conductance, Input Circuit Open μmhos (h _o)	DC Collector-to-Emitter volts (V _{CE})	DC Collector Current ma (I _c)	Collector Resistance megohms (r _c)	Emitter Resistance ohms (r _e)	Base Resistance ohms (r _b)	Mutual Resistance megohms (r _m)
--	---	---	--	--	--	--	---	--	---	--	---

Germanium, p-n-p, alloy-junction types

For equivalent circuits, see page 14

2N77	-4	-0.7	2720	3.23 x 10 ⁻⁴	55	14	-4	-0.7	3.93	23	1430	3.86		
2N104	-3	-0.2	6040	17.2 x 10 ⁻⁴	32	11.1	-3	-0.2	2.95	155	960	2.86		
	-6	-1	1667	4.95 x 10 ⁻⁴	44	22.8	-6	-1	1.974	21.7	690	1.93		
2N105	-1.3	-0.3	4800	9.1 x 10 ⁻⁴	45	12.4	-1.3	-0.3	3.74	73	1400	3.66		
	-4	-0.7	2880	5.5 x 10 ⁻⁴	55	16.3	-4	-0.7	3.45	34	976	3.39		
2N109	Typical Operation in Class B Push-Pull Service for Common-Emitter, Base-Input Circuit \bullet													
	DC Supply Voltage				-4.5		-9 volts		Signal-Source Impedance (base-to-base)				1500	1500 ohms
	Zero-Signal DC Base-to-Emitter Voltage				-0.15		-0.15 volt		Load Impedance (collector-to-collector)				400	800 ohms
	Peak Collector Current (per transistor)				-35		-40 ma		Signal Frequency				1000	1000 cps
	Zero-Signal DC Collector Current (per transistor)				-2		-2 ma		Circuit Efficiency				60	69 %
	Max.-Signal DC Collector Current (per transistor)				-11.5		-13 ma		Power Gain \blacklozenge				30	33 db
								Total Harmonic Distortion				7	7 %	
								Max.-Signal Power Output				75	160 mw	
2N139	This type is intended for use in 455-Kc intermediate-frequency amplifier service.													
2N140	This type is intended for 540-1640 Kc converter service.													
2N175	-4	-0.5	3570	9.44 x 10 ⁻⁴	65	25	-4	-0.5	2.86	37.7	1085	2.82		
2N206	Derived from a Common-Base, Emitter-Input Equivalent Circuit													
	-3	-0.2	137	3.8 x 10 ⁻⁴	-0.974	0.3	-3	-0.2	3.34	104	1270	3.25		
	-5	-1	33	3.2 x 10 ⁻⁴	-0.980	0.55	-5	-1	1.82	21	580	1.78		
2N215	Like RCA-2N104 but has flexible leads.													
2N217	Like RCA-2N109 but has flexible leads.													
2N218	Like RCA-2N139 but has flexible leads.													
2N219	Like RCA-2N140 but has flexible leads.													
2N220	Like RCA-2N175 but has flexible leads.													
2N247	Interlead Capacitance between base lead and collector lead is .003 μμf with leads cut to 5/16" and interlead shield grounded.													
2N269	Maximum DC Collector Current, at 80°C (for stated DC Collector-to-Base volts), and with emitter open:						Minimum Change in Temperature of Transistor Junctions such that the collector current will double for a dc collector-to-base voltage equal to or greater than -2.5 volts (but not more than -25 volts) and with emitter open..... 10°C							
	at V _{CB} = -2.5..... -50 μa													
	at V _{CB} = -12..... -60 μa													
	Maximum temperature rise of transistor junctions (transistor in free air)..... 0.35°C/mw													
Typical Operation in Low-Level Switching Service for Common-Emitter, Base-Input Circuit														
DC Collector Current				-12		-24 ma		Max. DC Collector-to-Emitter Voltage				-0.15	-0.2 volt	
DC Base Current				-0.4		-1 ma		Max. DC Base-to-Emitter Voltage				-0.35	-0.4 volt	
2N270	Typical Operation in Class B Push-Pull Service for Common-Emitter, Base-Input Circuit \bullet													
	DC Supply Voltage				12		volts		Load Impedance (per collector)				150	ohms
	Zero-Signal DC Base-to-Emitter Voltage				-0.11		volt		Signal Frequency				1000	cps
	Peak Collector Current (per transistor)				-110		ma		Circuit Efficiency				75	%
	Zero-Signal DC Collector Current (per transistor)				-2		ma		Power Gain \blacklozenge				32	db
	Max.-Signal DC Collector Current (per transistor)				-35		ma		Total Harmonic Distortion:				10	% max.
								at 500 mw				5	% max.	
								at 10 mw				500	mw	
								Max.-Signal Power Output						
2N274	Interlead Capacitance between base lead and collector lead is .094 μμf with leads cut to 5/16" and interlead shield grounded.													

\blacklozenge Measured at the primary of the output transformer.

\bullet Unless otherwise specified, values are for 2 transistors.

RCA TRANSISTOR CHARACTERISTICS CHART II (Cont'd)

 TYPE	CHARACTERISTICS											
	For COMMON-EMITTER CIRCUIT, BASE INPUT at Ambient Temperature = 25°C											
	Small-Signal H Parameters						Small-Signal T Parameters					
	DC Collector-to-Emitter Volts (V _{CE})	DC Collector Current ma (I _c)	Input Resistance, Output Circuit Shorted ohms (h _i)	Reverse Voltage Transfer Ratio, Input Circuit Open (h _r)	Forward Current Transfer Ratio, Output Circuit Shorted (h _f)	Output Conductance, Input Circuit Open μmhos (h _o)	DC Collector-to-Emitter Volts (V _{CE})	DC Collector Current ma (I _c)	Collector Resistance megohms (r _c)	Emitter Resistance ohms (r _e)	Base Resistance ohms (r _b)	Mutual Resistance megohms (r _m)
Germanium, p-n-p, alloy-junction types												
For equivalent circuits, see page 14												
2N301 2N301-A	Typical Operation for Common-Emitter, Base-Input Circuit											
		Service Class A				Service Class B Push-Pull		Service Class A				Service Class B Push-Pull
	DC Supply Voltage				-14.4	-14.4*	volts	Signal Frequency			400	400 cps
	Zero-Signal DC Base-to-Emitter Voltage				-0.24	-0.13	volt	Circuit Efficiency			47	67* %
	Peak Collector Current				-0.8	-2	amp	Power Gain†			32.5	30* db
	Zero-Signal DC Collector Current				-0.4	-0.05	amp	Total Harmonic Distortion:				
	Max.-Signal DC Collector Current				—	-0.64	amp	at 2.7 watts			10	— % max.
	DC Collector-to-Emitter Voltage				-13.6	—	volts	at 12 watts			—	10* % max.
	Signal-Source Impedance				75	15	ohms	Max.-Signal Power Output			2.7	12* watts
	Load Impedance				34	6	ohms	Emitter Resistor (unbypassed)			1	— ohm
2N370	This type is intended for use in radio-frequency amplifier service in short-wave receivers. Interlead Capacitance between base lead and collector lead is .003 μμf with leads cut to 5/16" and interlead shield grounded.											
2N371	The 2N371, in rf oscillator service up to 23 Mc, can provide the required oscillator-injection voltage to the rf mixer stage for optimum mixing operation. In an oscillator stage utilizing the 2N371 and operating at 22 Mc, if the collector supply voltage drops from -12 to -8 volts, the frequency provided by this stage will deviate from 22 Mc by less than 7 Kc. Interlead Capacitance between base lead and collector lead is .003 μμf with leads cut to 5/16" and interlead shield grounded.											
2N372	This type is intended for use in radio-frequency mixer service in short-wave receivers. Interlead Capacitance between base lead and collector lead is .003 μμf with leads cut to 5/16" and interlead shield grounded.											
2N384	This type is intended for VHF amplifier and oscillator service. Interlead Capacitance between base lead and collector lead is .094 μμf with leads cut to 5/16" and interlead shield grounded.											
2N398	DC Collector Breakdown Voltage with dc collector current = -50 μa, dc emitter current = 0 typical..... -150 volts minimum..... -105 volts						DC Emitter Breakdown Voltage with dc emitter current = -50 μa, dc collector current = 0 typical..... -75 volts minimum..... -50 volts					
2N404	DC Collector Breakdown Voltage with dc collector current = -20 μa, dc emitter current = 0 typical..... -40 volts minimum..... -25 volts						DC Emitter Breakdown Voltage with dc emitter current = -20 μa, dc collector current = 0 typical..... -35 volts minimum..... -12 volts					
2N405 2N406	-6	-1	1115	2.93 x 10 ⁻⁴	35	17.2	-6	-1	2.1	17	500	2.04
2N407 2N408	These types are intended for use in class A and class B push-pull service.											
2N409 2N410	These types are intended for use in 455-Kc intermediate-frequency amplifier service.											
2N411 2N412	These types are intended for 540-1640 Kc converter service.											

* Value is for 2 transistors.

† Measured at the primary of the output transformer.

RCA SEMICONDUCTOR DIODES

RCA TYPE	RECTIFIER SERVICE (For Frequencies of 25 cps and above)											
	MAXIMUM RATINGS (Absolute Values)					CHARACTERISTICS (Ambient Temperature = 25°C)						
	Peak Inverse volts	Forward Current		Fault Current ^a ma	Ambient Temperature Range °C	Minimum Forward Current ma At DC volts = 1	Max. Average Inverse Current μA				Minimum Peak Inverse Voltage for Zero Dynamic Resistance volts	Capacitance Between Stud Tips (Approx.) μμf
Peak ma		Average ^b ma	At DC volts = -3				At DC volts = -10	At DC volts = -50	At DC volts = -100			
1N34-A	60	150	50	500	-50 to +75	5	—	30	500	—	75	1
1N38-A	100	150	50	500	-50 to +75	4	5	—	—	500	120	1
1N54-A	50	150	50	500	-50 to +75	5	—	7	100	—	75	1
1N58-A	100	150	50	500	-50 to +75	4	—	—	—	600	120	1

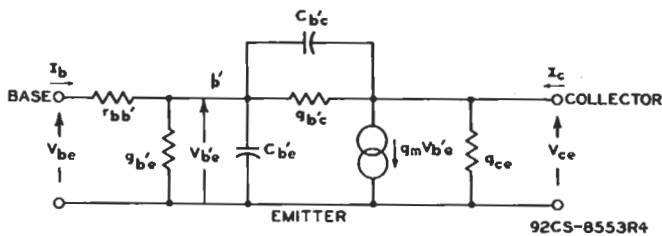
^b Averaged over one conduction cycle.

^a Maximum Fault Current is the highest value of current that should be permitted to flow through the diode under a fault condition such as load short circuit. Values are for time durations up to one second.

EQUIVALENT CIRCUITS

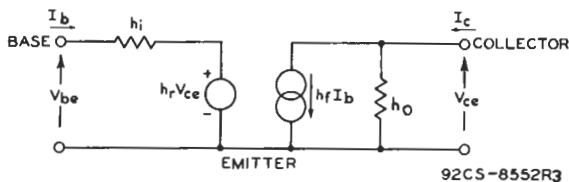
SMALL-SIGNAL HYBRID- π PARAMETERS:

Derived from the accompanying one-generator equivalent circuit and applicable over the useful frequency range of the transistor.



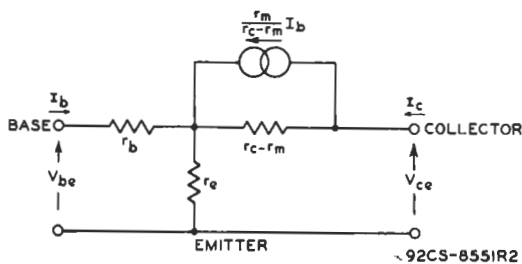
SMALL-SIGNAL H PARAMETERS:

Derived from the accompanying two-generator equivalent circuit and applicable over the audio frequency range.



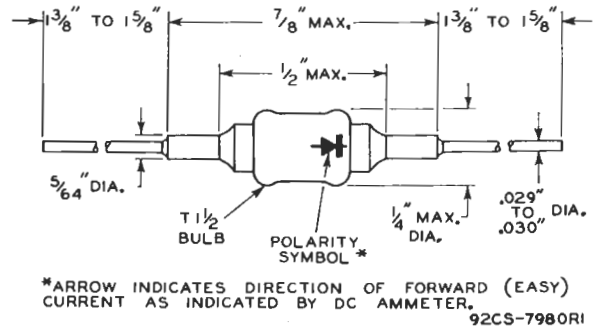
SMALL-SIGNAL T PARAMETERS:

Derived from the accompanying one-generator equivalent circuit and applicable over the audio frequency range.

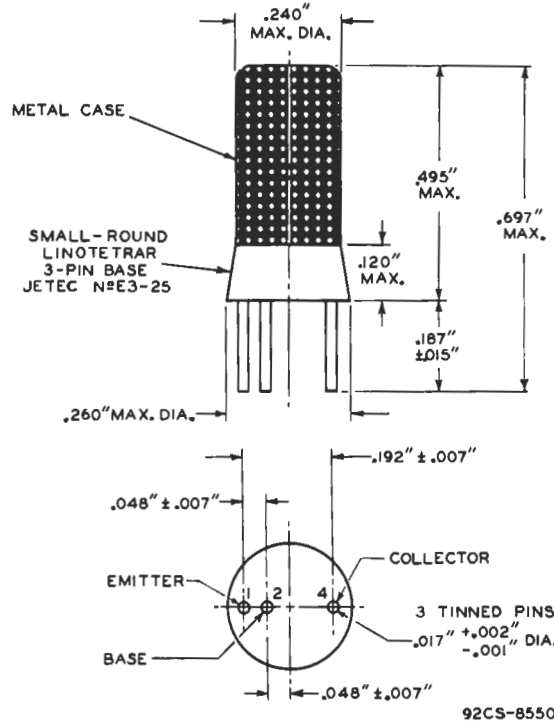


DIMENSIONAL OUTLINES

1N34-A, 1N38-A, 1N54-A, and 1N58-A

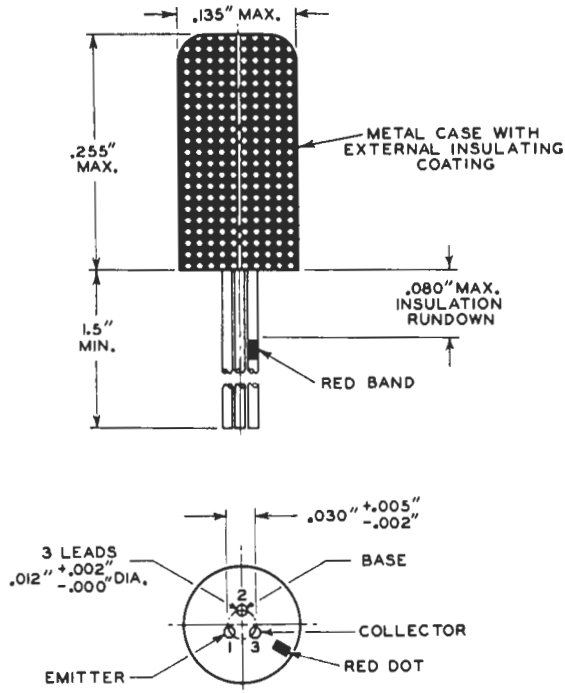


2N104, 2N109, 2N139, 2N140, 2N175, 2N405, 2N407, 2N409, and 2N411



DIMENSIONAL OUTLINES (Cont'd)

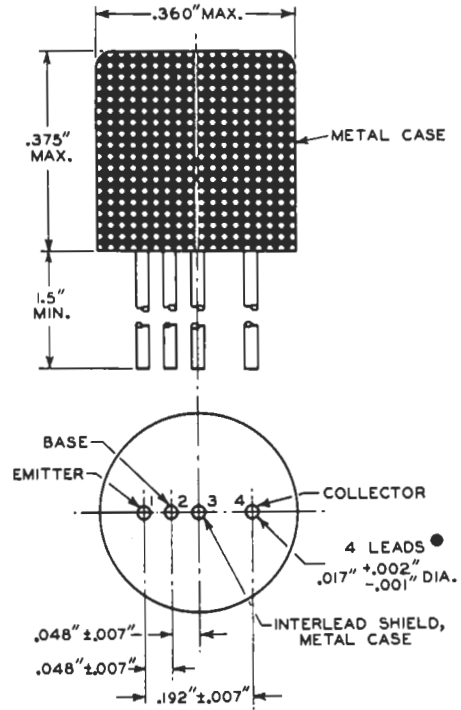
2N105



92CS-8577R2

C

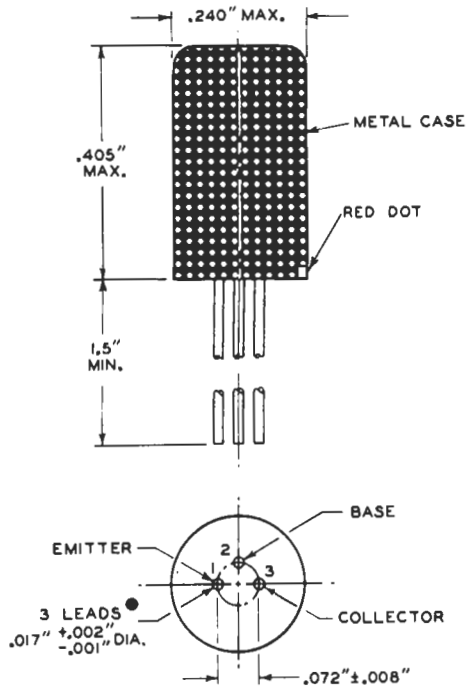
2N247, 2N370, 2N371, and 2N372



92CS-9122R2

E

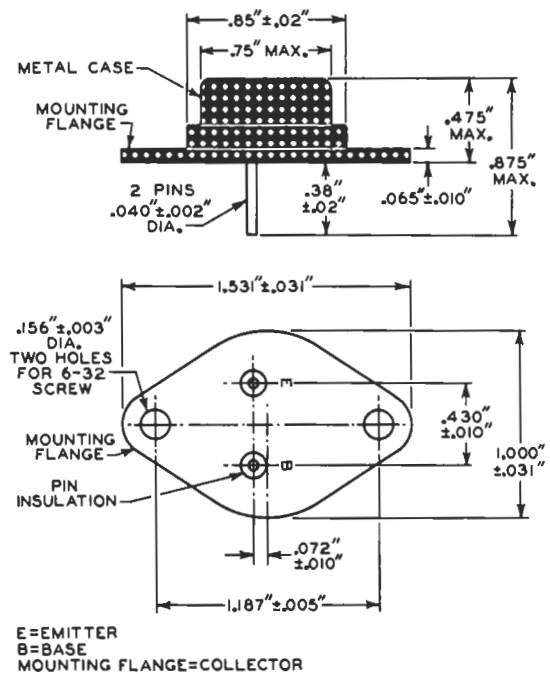
2N215, 2N217, 2N218, 2N219, 2N220, 2N269, 2N406, 2N408, 2N410, and 2N412



92CS-9148R1

D

2N301 and 2N301A



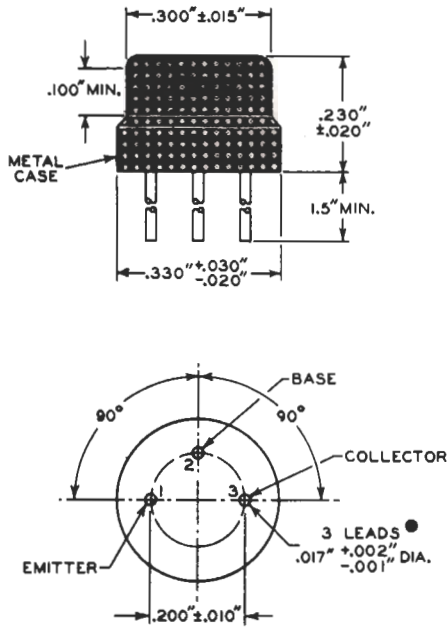
92CS-9238R2

F

• The specified lead diameter applies in the zone between 0.050" and 0.250" from the base seat. Between 0.250" and 1.50" a maximum of 0.021" diameter is held. Outside of these zones, the lead diameter is not controlled.

DIMENSIONAL OUTLINES (Cont'd)

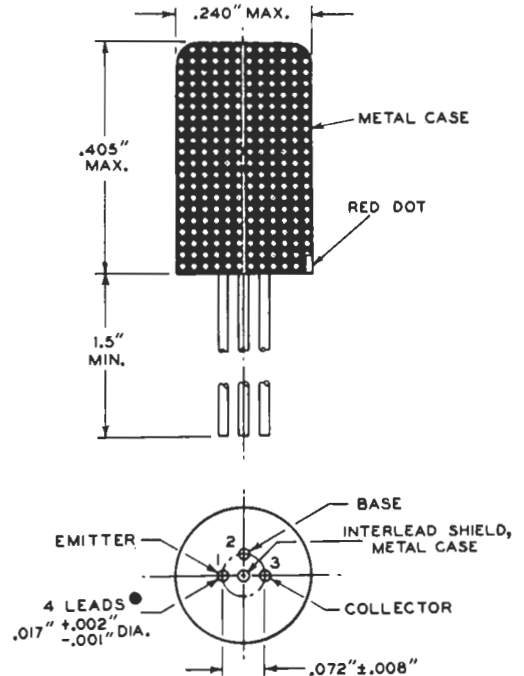
2N398 and 2N404



92CS-9371R1

G

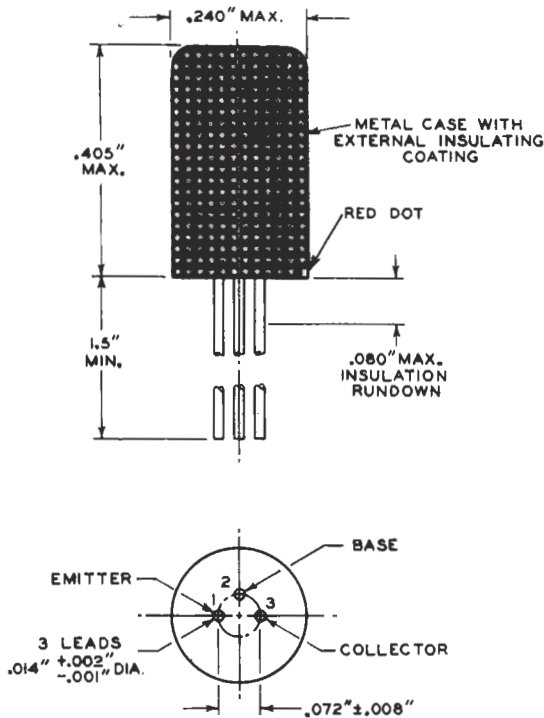
2N274 and 2N384



92CS-9266R1

I

2N77 and 2N206 ■

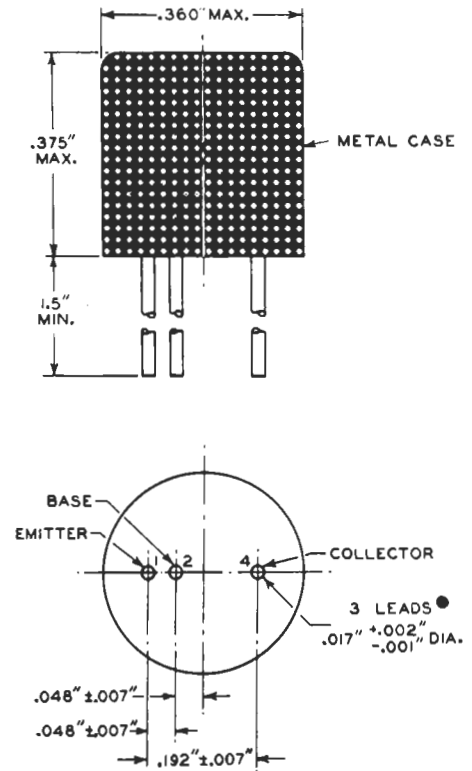


■ Lead Diameter is .017" + .002" or -.001"

92CS-8563R2

H

2N270



92CS-9176R2

J

● The specified lead diameter applies in the zone between 0.050" and 0.250" from the base seat. Between 0.250" and 1.50" a maximum of 0.021" diameter is held. Outside of these zones, the lead diameter is not controlled.

INTERCHANGEABILITY DIRECTORY

The following directory has been prepared to guide designers, experimenters, and servicemen in selecting the proper RCA transistor type as a replacement and to help identify and describe many of the transistor types which have been introduced on the market by different manufacturers. More than 500 type designations are listed including junction types, point-contact types and phototransistors.

In using the **Transistor Directory**, note that the basic type designations of different manufacturers may have been assigned according to different systems. Some basic designations consist only of a number such as 210, 355, 1032, etc.; others consist of a combination of letters and digits such as 2N77, X78, SB100, etc. In either case the basic designation may or may not have a prefix composed of one or more letters, such as CK, GT, H, OC, ZJ, etc., which indicates the particular manufacturer. For certain transistors, this prefix becomes an essential part of the type designation when as sometimes happens, two or more manufacturers utilize the same designation for different transistor types.

Identifying information about the **Type To Be Replaced** including the following: (1) manufacturer's prefix, if any, (2) the basic type designation in bold face, (3) symbol to designate the manufacturer, (4) symbol to indicate the description, for example, GPA = Germanium, p-n-p, Alloy-Junction Type (denotes the structural arrangement and kind of semiconductor materials used in the device),

and (5) class of service is charted in the first five columns. The next two columns show the **RCA Direct Replacement Type**, or the **RCA Similar Type**, respectively, when one or the other is available.

Basic designations shown in Column 2 of the tabulation are listed in numerical — alphabetical sequence. Those starting with a digit are given first; those starting with a letter appear at the end of the tabulation.

Types listed in the Similar RCA Type column (Column 6) are **not directly interchangeable** with the types listed in the Basic Designation column because of mechanical and/or electrical differences. For more information as to degree of similarity refer to respective transistor data.

How to Use

1. Look in Column 2 for basic designation of type to be replaced.

2. If type to be replaced has a prefix, look for that prefix in Column 1.

For example: If type CK-762 is to be replaced, find the basic designation 762 in Column 2 and the prefix CK in Column 1.

3. Consult Column 6 for corresponding **RCA Direct Replacement Type**.

4. If no **Direct RCA Replacement Type** is shown consult Column 7 for RCA Similar Type and obtain respective transistor data to determine degree of similarity.

KEY TO SYMBOLS IN COLUMN 3

A = Amperex
B = Bendix
BOG = Bogue (Germanium Products)
CBS = CBS-Hytron
CC = Cleveite Corporation
DEL = Delco
GE = General Electric
GT = General Transistor
HA = Hughes Aircraft

M = Motorola
MAL = Mallory
MH = Minneapolis-Honeywell
N = Nucleonics
NA = National Aircraft
NU = National Union
P = Philco
RCA = Radio Corporation of America
RK = Raytheon

RR = Radio Receptor
S = Sylvania
SPR = Sprague
SS = Scientific Specialities
TEC = Transiron
TI = Texas Instruments
TS = Tung-Sol
WE = Western Electric
WL = Westinghouse

KEY TO SYMBOLS IN COLUMN 4

GC = Germanium, Point-Contact Type
GNA = Germanium, n-p-n, Alloy-Junction Type
GNG = Germanium, n-p-n, Grown-Junction Type
GPA = Germanium, p-n-p, Alloy-Junction Type
GPD = Germanium, p-n-p, "Drift" Type
GPG = Germanium, p-n-p, Grown-Junction Type
GPS = Germanium, p-n-p, Surface-Barrier Type

SNA = Silicon, n-p-n, Alloy-Junction Type
SNG = Silicon, n-p-n, Grown-Junction Type
SPA = Silicon, p-n-p, Alloy-Junction Type
SPG = Silicon, p-n-p, Grown-Junction Type
SD = Semiconductor Diode

* RCA types shown in this column are direct replacements under all circumstances for corresponding types to be replaced.

† RCA types shown in this column are not directly interchangeable with the types to be replaced because of mechanical and/or electrical differences. For more information as to degree of interchangeability, refer to respective type data or write to Commercial Engineering, RCA, Somerville, New Jersey.

Information contained herein has been carefully checked and is believed to be reliable but no responsibility is assumed for inaccuracies. The reporting of errors to Commercial Engineering, RCA, Somerville, N. J., will be appreciated.

CIRCUITS

The circuits shown in the following pages are included in this Booklet to illustrate some of the more important applications of RCA transistors and semiconductor diodes. These circuits are not necessarily examples of commercial practice. They have been conservatively designed and are capable of excellent performance. Electrical specifications are given for circuit components to assist those interested in home construction. Layouts and mechanical details are omitted because they vary widely with the requirements of individual set builders and with the sizes and shapes of the components employed.

Performance of these circuits depends as much on

the quality of the components selected and the care employed in layout and construction as on the circuits themselves. Good signal reproduction from receivers and amplifiers requires the use of good-quality speakers, transformers, inductors, and input sources (microphones, phonograph pickups, etc.).

Information on the characteristics of each RCA transistor and semiconductor diode will be found on pages 8, 9, 10, 11 and 12. This information will prove of assistance in understanding and utilizing the circuits.

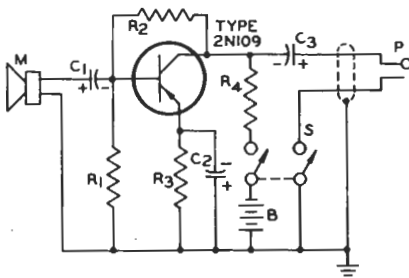
The following circuits will be found in the subsequent pages:

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MICROPHONE PREAMPLIFIER

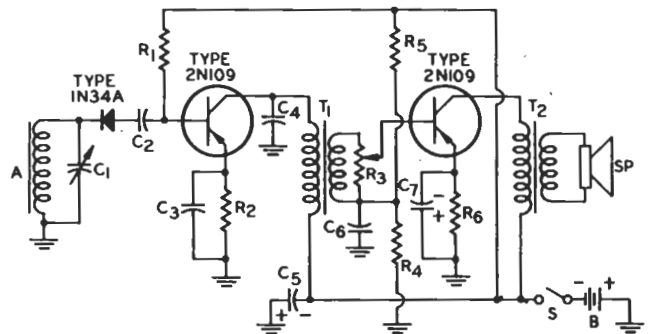


- B = 9 volts, VS300 or VS301
- C1 = 52 μ f, electrolytic, 12 v.
- C2 = 50 μ f, electrolytic, 12 v.
- C3 = 2.5 μ f, electrolytic, 25 v.
- M = RCA-239S1 2 $\frac{1}{2}$ " Speaker
- P = Plug to make connection to audio amplifier
- R1 = 10000 ohms, 0.5 watt
- R2 = 68000 ohms, 0.5 watt
- R3 = 1200 ohms, 0.5 watt
- R4 = 8200 ohms, 0.5 watt
- S = Switch, push-button, double-pole single throw, non-locking

NOTE:

The low-frequency characteristics of the RCA-239S1 2 $\frac{1}{2}$ " speaker can be improved for operation as a microphone by gluing a small disk of felt over each of the holes, except one, on the back of the speaker. The last hole should be covered with a piece of fibre or cardboard having a $\frac{1}{2}$ " hole. A baffle such as a case should be provided for the speaker. If an enclosed case is used, a $\frac{1}{8}$ " hole should be drilled in the case.

TWO-TRANSISTOR RECEIVER Standard AM Broadcast Band

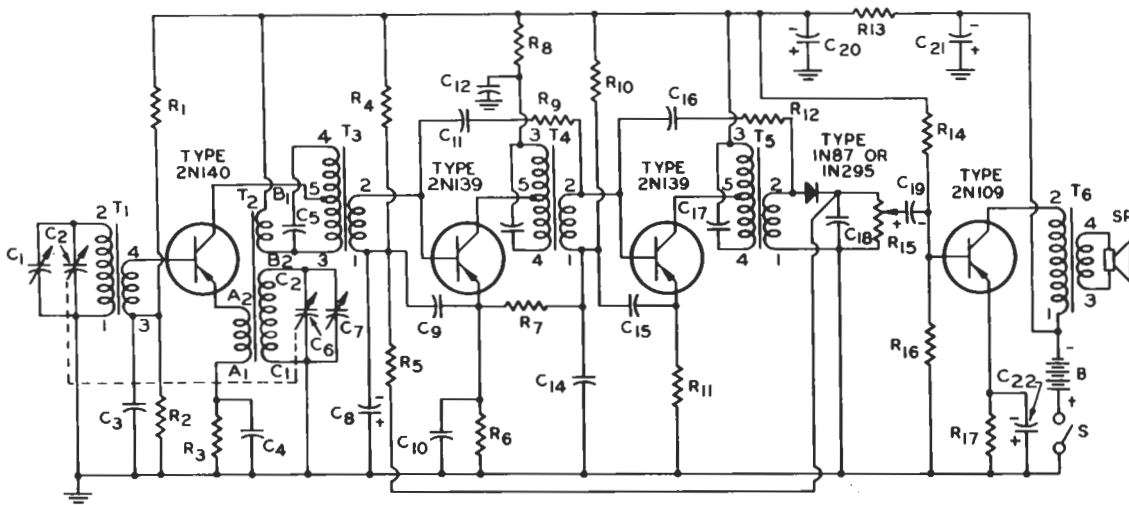


- A = Ferrite loop antenna, 540-1600 Kc
- B = 9 volts, VS300 or VS301
- C1 = Variable capacitor, 540-1600 Kc
- C2 = 0.01 μ f, paper, 150 v.
- C3 = 0.05 μ f, paper 150 v.
- C4 = 0.001 μ f, paper, 150 v.
- C5 = 10 μ f, paper, 150 v.
- C6 = 0.1 μ f, electrolytic, 3 v.
- C7 = 100 μ f, electrolytic, 6 v.
- R1 = 0.1 megohm, 0.5 watt
- R2 = 50 ohms, 0.5 watt
- R3 = Volume-control potentiometer, 5000 ohms, 0.5 watt
- R4 = 5600 ohms, 0.5 watt

- R5 = 10000 ohms, 0.5 watt
- R6 = 220 ohms, 0.5 watt
- SP = Speaker, high-sensitivity
- T1 = Interstage audio transformer to provide a 30000-ohm load at the primary terminals with a 1000-ohm load between the secondary terminals. DC primary current = 0.5 ma. DC primary resistance = 1000 ohms.
- T2 = Output transformer to provide a 500-ohm load at the primary terminals with the desired speaker connected to the secondary terminals.

CIRCUITS

FOUR-TRANSISTOR RECEIVER Standard AM Broadcast Band



B = 9 volts, VS300 or VS301
 C1 = Trimmer capacitor, 0-20 μf
 C2 = Variable capacitor, 12-230 μf
 C3 = 0.05 μf , paper, 150 v.
 C4 = 0.04 μf , mica, 150 v.
 C5 = 220 μf , mica, 150 v.
 C6 = Variable capacitor, 10-105 μf
 C7 = Trimmer capacitor, 0-20 μf

C8 = 10 μf , electrolytic, 15 v.
 C9 = 0.05 μf , paper, 150 v.
 C10 = 0.05 μf , paper, 150 v.
 C11 = 75 μf , mica, 150 v.
 C12 = 0.05 μf , paper, 150 v.
 C13 = 220 μf , mica, 150 v.
 C14 = 0.05 μf , paper, 150 v.
 C15 = 0.05 μf , paper, 150 v.
 C16 = 33 μf , mica, 150 v.
 C17 = 220 μf , mica, 150 v.
 C18 = 0.05 μf , paper, 150 v.
 C19 = 2 μf , electrolytic, 15 v.

C20 = 50 μf , electrolytic, 15 v.
 C21 = 50 μf , electrolytic, 15 v.
 C22 = 100 μf , electrolytic, 15 v.
 R1 = 33,000 ohms, $\frac{1}{8}$ watt
 R2 = 24,000 ohms, $\frac{1}{8}$ watt
 R3 = 820 ohms, $\frac{1}{8}$ watt
 R4 = 120,000 ohms, $\frac{1}{8}$ watt
 R5 = 18,000 ohms, $\frac{1}{8}$ watt
 R6 = 1200 ohms, $\frac{1}{8}$ watt
 R7 = 75,000 ohms, $\frac{1}{8}$ watt
 R8 = 560 ohms, $\frac{1}{8}$ watt

R9 = 560 ohms, $\frac{1}{8}$ watt
 R10 = 150,000 ohms, $\frac{1}{8}$ watt
 R11 = 1200 ohms, $\frac{1}{8}$ watt
 R12 = 2000 ohms, $\frac{1}{8}$ watt
 R13 = 100 ohms, $\frac{1}{8}$ watt
 R14 = 10,000 ohms, $\frac{1}{8}$ watt
 R15 = Volume control, 2500 ohms, $\frac{1}{8}$ watt
 R16 = 5600 ohms, $\frac{1}{8}$ watt
 R17 = 220 ohms, $\frac{1}{8}$ watt
 SP = Speaker

T1 = Antenna transformer wound on the largest feasible ferrite core to provide the following characteristics:
 Primary Inductance (With secondary open) 353 μh
 Primary Q at 1 Mc, mounted on chassis with secondary open 200
 Equivalent output resistance across secondary terminals, at 1 Mc with primary tuned 600 ohms

The primary should be wound on one end of the ferrite rod with spacing between turns equal to the thickness of the wire. The secondary should be wound on the opposite end of the ferrite rod with no spacing between turns (close wound). The end of the primary winding nearest the secondary is the ground end. Use #7/41 Litz wire. A ferrite rod about 8" long and $\frac{3}{8}$ " in diameter will provide excellent results.

T2 = Oscillator transformer. Wind as follows:

Using a threaded resinite coil form about 1" long with internal threads to match a $\frac{1}{4}$ "-diameter ferrite core about $\frac{3}{4}$ " in length, wind 4 turns near the center of the coil form. This winding is located between the terminals marked A1 and A2 on the circuit diagram. The winding between terminals B1 and B2 consists of a 10-turn winding and is wound on top of the 4-turn winding. The winding between terminals C1 and C2 is wound on top of the other two windings and is a multilayer 115-turn winding. This winding should extend over a length of about $\frac{3}{4}$ ". All the windings are universally wound with #5/44 Litz wire. A resinite collar with 6-terminals may be fitted over one end of the coil form to provide secure anchorage for the ends of the windings and convenient terminals for making connections to the windings. When the transformer is completed, connect a 120- μf capacitor across the 115-turn winding and tune the ferrite core to obtain resonance at 1.455 Mc with the other two windings open circuited. The Q of this winding under the same conditions should be 100 or greater.

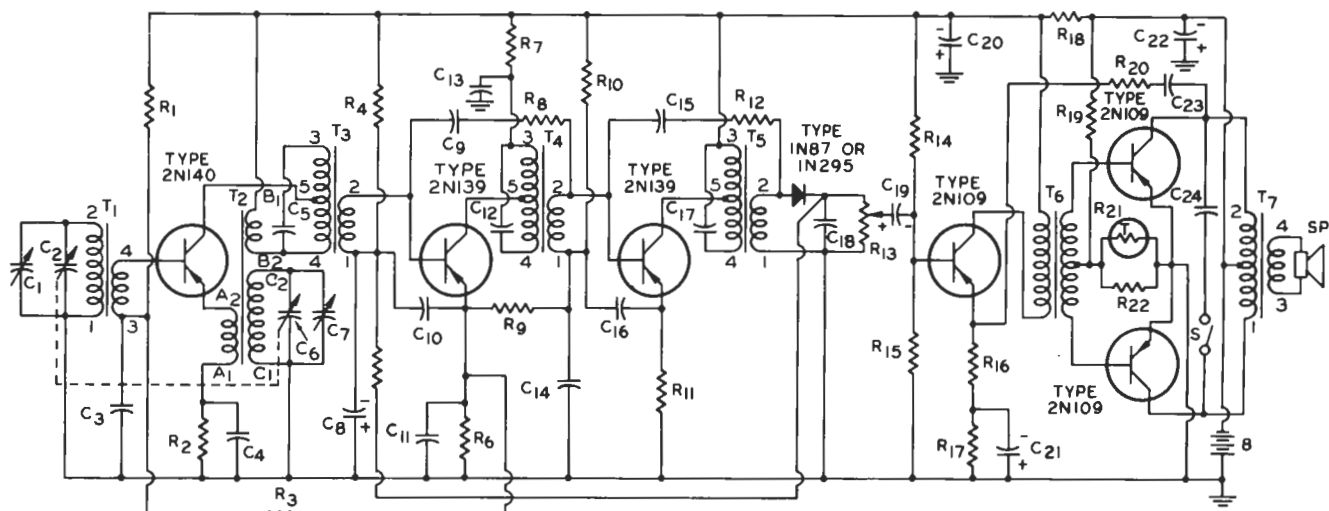
T3, T4, T5 = Intermediate-frequency transformers. Using materials like those described for the oscillator transformer T2, wind T3, T4, and T5 to meet the following requirements:

	T3	T4	T5
Tuned resistance at primary tap	118,000	15,300	17,200 ohms
Primary (terminals 3 and 5): Reflected resistance with secondary terminated	206,000	29,000	10,900 ohms
Secondary (terminals 1 and 2): Reflected resistance with primary terminated	1,000	500	1,000 ohms
Turns Ratios:			
Terminals 4 and 3 to terminals 5 and 3	1.17	2.48	3.16
Terminals 5 and 3 to terminals 2 and 1	14.35	7.62	3.3
Core (Ferrite):			
Unloaded Q (mounted in chassis)	110	61	110
Loaded Q (mounted in chassis)	35	35	35

T6 = Class A output transformer, primary impedance = 500 ohms, secondary impedance = 12 ohms.

CIRCUITS

SIX-TRANSISTOR RECEIVER Standard AM Broadcast Band



B = 9 volts, VS300 or VS301
 C1 = Trimmer capacitor, 2-20 μf
 C2 = Variable capacitor, 12-230 μf
 C3 = 0.01 μf , paper, 150 v.
 C4 = 0.04 μf , paper, 150 v.
 C5 = 220 μf , mica
 C6 = Variable capacitor, 10-105 μf
 C7 = Trimmer capacitor, 0-20 μf
 C8 = 10 μf , electrolytic, 3 v.
 C9 = 75 μf , mica
 C10 = 0.05 μf , paper, 150 v.
 C11 = 0.05 μf , paper, 150 v.
 C12 = 220 μf , mica

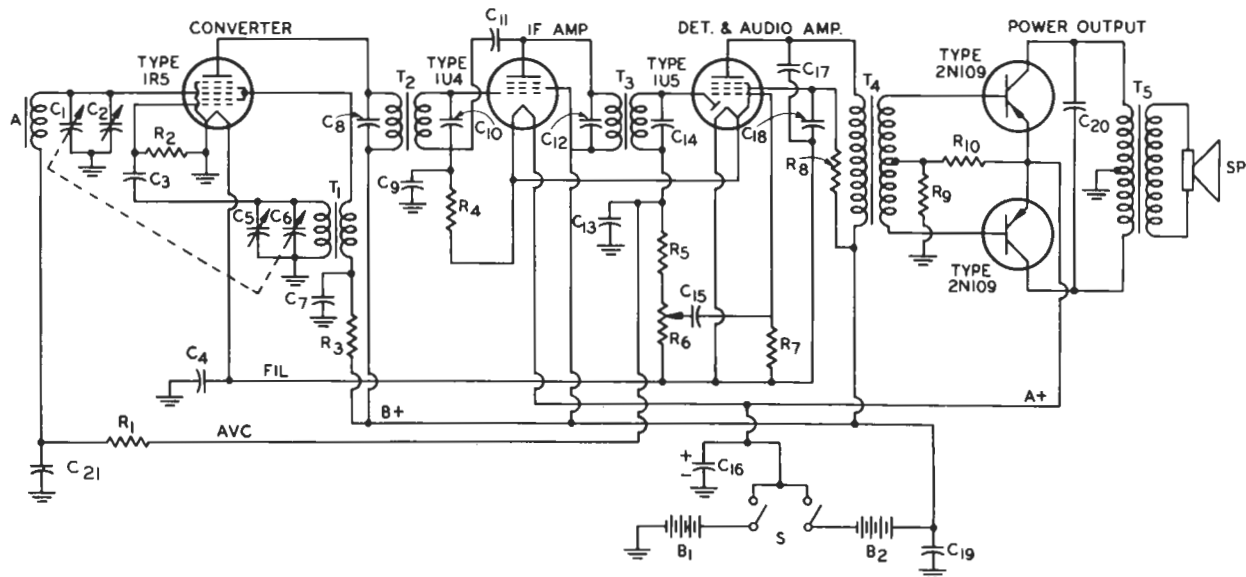
C13 = 0.05 μf , paper, 150 v.
 C14 = 0.05 μf , paper, 150 v.
 C15 = 33 μf , mica
 C16 = 0.05 μf , paper, 150 v.
 C17 = 220 μf , mica
 C18 = 0.05 μf , paper, 150 v.
 C19 = 2 μf , electrolytic, 12 v.
 C20 = 50 μf , electrolytic, 12 v.
 C21 = 100 μf , electrolytic, 12 v.
 C22 = 50 μf , electrolytic, 12 v.
 C23 = 0.01 μf , paper, 150 v.
 C24 = 0.03 μf , paper, 150 v.
 R1 = 33000 ohms, 0.5 watt
 R2 = 820 ohms, 0.5 watt
 R3 = 100 ohms, 0.5 watt
 R4 = 100000 ohms, 0.5 watt
 R5 = 8200 μf , 0.5 watt

R6 = 560 ohms, 0.5 watt
 R7 = 560 ohms, 0.5 watt
 R8 = 560 ohms, 0.5 watt
 R9 = 1800 ohms, 0.5 watt
 R10 = 47000 ohms, 0.5 watt
 R11 = 560 ohms, 0.5 watt
 R12 = 2000 ohms, 0.5 watt
 R13 = Potentiometer, 2500 ohms, 0.5 watt, volume-control
 R14 = 39000 ohms, 0.5 watt
 R15 = 51000 ohms, 0.5 watt
 R16 = 220 ohms, 0.5 watt
 R17 = 5100 ohms, 0.5 watt
 R18 = 100 ohms, 0.5 watt
 R19 = 10000 ohms, 0.5 watt
 R20 = 220000 ohms, 0.5 watt

R21 = Thermistor, 300 ohms at 25°C, 108 ohms at 50°C
 R22 = 270 ohms, 0.5 watt
 T1 = Antenna transformer
 T2 = Oscillator transformer
 T3, T4, T5 = Intermediate-frequency transformers
 T6 = Driver transformer with primary to secondary impedance (base-to-base, center tapped) of 20,000 ohms to 2,000 ohms.
 T7 = Class B output transformer with primary to secondary impedance (collector - to - collector, center tapped) of 750 to 3.2 ohms.

NOTE: Transformers T1, T2, T3, T4, and T5 are the same as those described for the 4-transistor receiver.

THREE-TUBE, TWO-TRANSISTOR RECEIVER



A = Loop antenna, 540-1600 Kc
 B1 = 4.5 volts
 B2 = 67.5 volts
 C1 = Ganged tuning capacitor, 10-274 μf
 C2 = Trimmer capacitor, 2-15 μf
 C3 = 56 μf , ceramic
 C4 = 0.1 μf 100 v.
 C5 = Ganged tuning capacitor, 7.5-122.5 μf
 C6 = Trimmer capacitor, 2-15 μf
 C7 = 0.02 μf paper 100 v.
 C8 = Trimmer capacitor

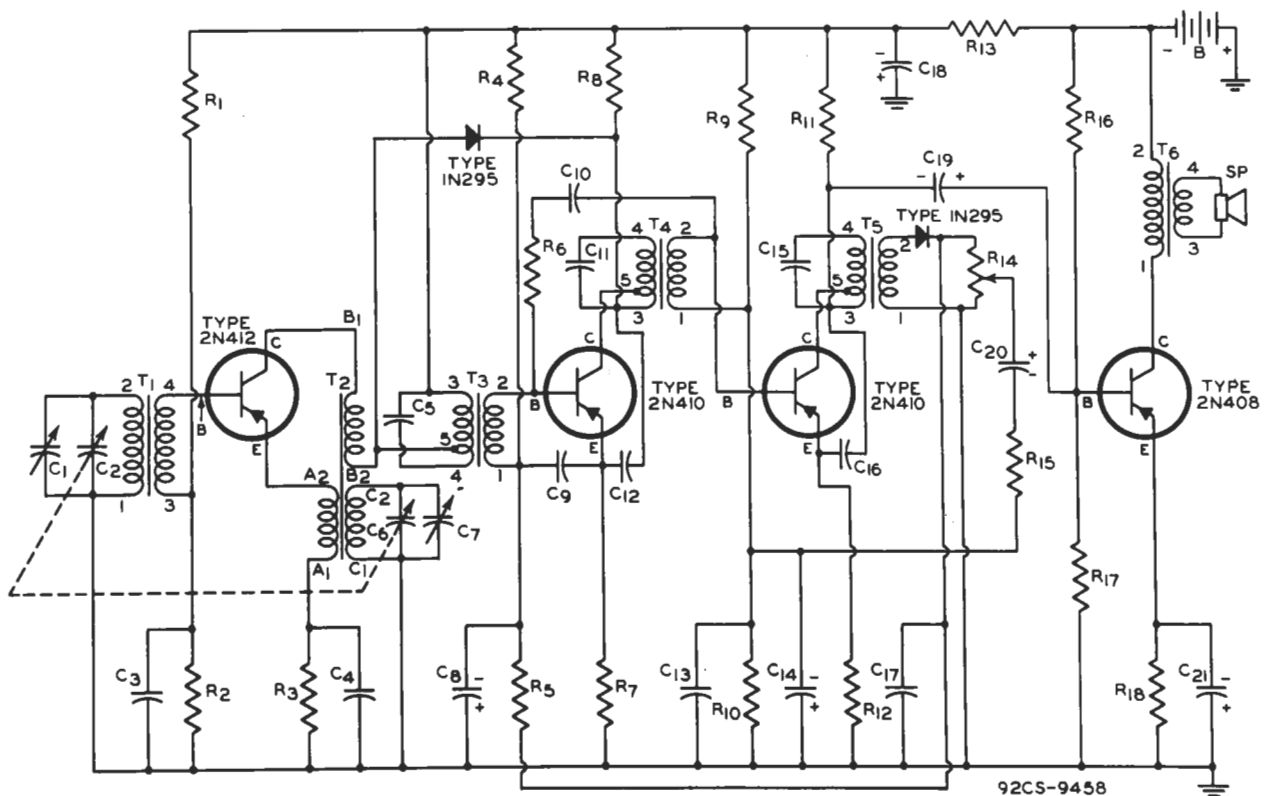
C9 = 0.05 μf , paper, 100 v.
 C10 = Trimmer capacitor
 C11 = 5 μf , silver mica
 C12 = Trimmer capacitor
 C13 = 82 μf , ceramic
 C14 = Trimmer capacitor
 C15 = 0.002 μf , paper, 150 v.
 C16 = 50 μf , electrolytic, 6 v.
 C17 = 0.005 μf , paper 200 v.
 C18 = 0.04 μf , paper, 100 v.
 C19 = 20 μf , electrolytic, 150 v.
 C20 = 0.04 μf , paper, 100 v.
 C21 = 0.02 μf , paper, 100 v.
 R1 = 3.3 megohms, 0.25 watt
 R2 = 100,000 ohms, 0.25 watt

R3 = 4700 ohms, 0.25 w.
 R4 = 10 megohms, $\frac{1}{4}$ w.
 R5 = 1000 ohms, $\frac{1}{4}$ w.
 R6 = 1 meg. $\frac{1}{2}$ w. pot. vol. control.
 R7 = 10 megohms $\frac{1}{4}$ w.
 R8 = 360,000 ohms $\frac{1}{4}$ w.
 R9 = 3000 ohms $\frac{1}{4}$ w.
 R10 = 100 ohms $\frac{1}{4}$ w.
 T1 = Oscillator coil for use with tuning capacitor of 7.5-122.5 μf , and 455 Kc IF transformer
 T2, T3 = 455 Kc IF transformers

T4 = Driver transformer, primary impedance = 100,000 ohms, primary resistance = 2900 ohms, secondary impedance (base-to-base) = 1500 ohms, secondary resistance = 30 ohms.
 T5 = Output transformer, primary impedance (collector-to-collector) = 400 ohms, primary resistance = 20 ohms, secondary impedance = 11 ohms, secondary resistance = 1 ohm.

CIRCUITS

FOUR-TRANSISTOR REFLEX RECEIVER Standard AM Broadcast Band



92CS-9458

B = 9 volts, VS300 or VS306
 C1 = Trimmer capacitor, 0-20 μf
 C2 = Variable capacitor, 12-230 μf
 C3 = 0.04 μf , paper, 150 v.
 C4 = 0.04 μf , paper, 150 v.
 C5 = Fixed tuner capacitor, 220 μf
 C6 = Variable capacitor, 10-105 μf
 C7 = Trimmer capacitor, 0-20 μf
 C8 = 10 μf , electrolytic, 15 v.
 C9 = 0.04 μf , paper, 150 v.
 C10 = 75 μf , mica, 150 v.
 C11 = Fixed tuner capacitor, 220 μf
 C12 = 0.04 μf , paper, 150 v.
 C13 = 0.02 μf , paper, 150 v.
 C14 = 10 μf , electrolytic, 15 v.
 C15 = Fixed tuner capacitor, 220 μf
 C16 = 0.05 μf , paper, 150 v.

C17 = 0.04 μf , paper, 150 v.
 C18 = 50 μf , electrolytic, 15 v.
 C19 = 10 μf , electrolytic, 15 v.
 C20 = 10 μf , electrolytic, 15 v.
 C21 = 50 μf , electrolytic, 15 v.
 R1 = 33,000 ohms, 0.5 watt
 R2 = 2,400 ohms, 0.5 watt
 R3 = 820 ohms, 0.5 watt
 R4 = 120,000 ohms, 0.5 watt
 R5 = 18,000 ohms, 0.5 watt
 R6 = 510 ohms, 0.5 watt
 R7 = 1,200 ohms, 0.5 watt
 R8 = 1,200 ohms, 0.5 watt
 R9 = 68,000 ohms, 0.5 watt
 R10 = 3,900 ohms, 0.5 watt
 R11 = 1,000 ohms, 0.5 watt
 R12 = 680 ohms, 0.5 watt
 R13 = 100 ohms, 0.5 watt
 R14 = Volume-control potentiometer, 2500 ohms, 0.5 watt
 R15 = 1200 ohms, 0.5 watt
 R16 = 4700 ohms, 0.5 watt
 R17 = 1500 ohms, 0.5 watt

on top of the 4-turn winding. The winding between terminals C1 and C2 is wound on top of the other two windings and is a multilayer 115-turn winding. This winding should extend over a length of about $\frac{3}{8}$ ". All the windings are universally wound with #5/44 Litz wire. A resinite collar with 6-terminals may be fitted over one end of the coil form to provide secure anchorage for the ends of the windings and convenient terminals for making connections to the windings. When the transformer is completed, connect a 120- μf capacitor across the 115-turn winding and tune the ferrite core to obtain resonance at 1.455 Mc with the other two windings open circuited. The Q of this winding under the same conditions should be 100 or greater.

T3, T4, T5 = Intermediate-frequency transformers. Using materials like those described for the oscillator transformer T2, wind T3, T4, and T5 to meet the following requirements:

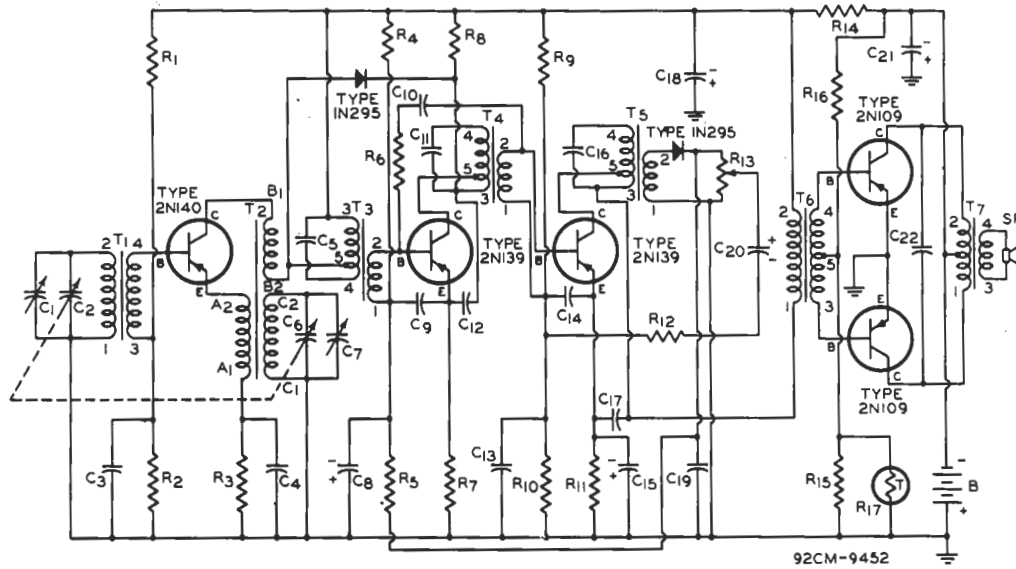
	T3	T4	T5
Tuned resistance at primary tap	118,000	15,300	2,260 ohms
Primary (terminals 3 and 5): Reflected resistance with secondary terminated	206,000	29,000	1,000 ohms
Secondary (terminals 1 and 2): Reflected resistance with secondary terminated	1,000	500	1,000 ohms
Turns Ratios: Terminals 4 and 3 to terminals 5 and 3	1.17	2.48	8.4
Terminals 5 and 3 to terminals 2 and 1	14.35	7.62	1
Core (Ferrite): Unloaded Q (mounted in chassis)	110	61	100
Loaded Q (mounted in chassis)	35	35	30

T6 = Class A output transformer, primary impedance = 400 ohms, secondary impedance = 12 ohms.

T1 = Antenna transformer wound on the largest feasible ferrite core to provide the following characteristics:
 Primary Inductance (With secondary open) 353 μh
 Primary Q at 1 Mc, mounted on chassis with secondary open 200
 Equivalent output resistance across secondary terminals, at 1 Mc with primary tuned 600 ohms
 The primary should be wound on one end of the ferrite rod with spacing between turns equal to the thickness of the wire.
 The secondary should be wound on the opposite end of the ferrite rod with no spacing between turns (close wound).
 The end of the primary winding nearest the secondary is the ground end. Use #7/41 Litz wire. A ferrite rod about 8" long and $\frac{3}{8}$ " in diameter will provide excellent results.
 T2 = Oscillator transformer. Wind as follows:
 Using a threaded resinite coil form about 1" long with internal threads to match a $\frac{1}{4}$ "-diameter ferrite core about $\frac{3}{8}$ " in length, wind 4 turns near the center of the coil form. This winding is located between the terminals marked A1 and A2 on the circuit diagram. The winding between terminals B1 and B2 consists of a 10-turn winding and is wound

CIRCUITS

FIVE-TRANSISTOR REFLEX RECEIVER Standard AM Broadcast Band



92CM-9452

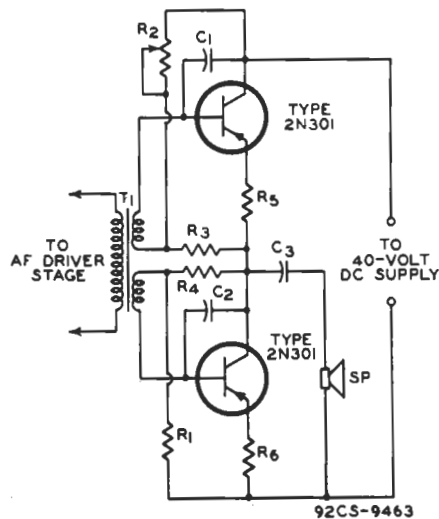
- B = 9 volts, VS300 or VS306
 C1 = Trimmer capacitor, 0-20 μf
 C2 = Variable capacitor, 12-230 μf
 C3 = 0.04 μf , paper, 150 v.
 C4 = 0.04 μf , paper, 150 v.
 C5 = Fixed tuner capacitor, 220 μf
 C6 = Variable capacitor, 10-105 μf
 C7 = Trimmer capacitor, 0-20 μf
 C8 = 10 μf , electrolytic, 15 v.
 C9 = 0.04 μf , paper, 150 v.
 C10 = 75 μf , mica, 150 v.
 C11 = Fixed tuner capacitor, 220 μf

- C12 = 0.04 μf , paper, 150 v.
 C13 = 0.04 μf , paper, 150 v.
 C14 = 0.04 μf , paper, 150 v.
 C15 = 50 μf , electrolytic, 15 v.
 C16 = Fixed tuner capacitor, 220 μf
 C17 = 0.02 μf , paper, 150 v.
 C18 = 50 μf , electrolytic, 15 v.
 C19 = 0.04 μf , paper, 150 v.
 C20 = 10 μf , electrolytic, 15 v.
 C21 = 50 μf , electrolytic, 15 v.
 C22 = 0.03, paper, 150 v.

- R1 = 33,000 ohms, 0.5 watt
 R2 = 2,400 ohms, 0.5 watt
 R3 = 820 ohms, 0.5 watt
 R4 = 120,000 ohms, 0.5 watt
 R5 = 18,000 ohms, 0.5 watt
 R6 = 510 ohms, 0.5 watt
 R7 = 1,200 ohms, 0.5 watt
 R8 = 1,200 ohms, 0.5 watt
 R9 = 68,000 ohms, 0.5 watt
 R10 = 10,000 ohms, 0.5 watt
 R11 = 680 ohms, 0.5 watt
 R12 = 820 ohms, 0.5 watt
 R13 = Volume-control potentiometer, 2500 ohms, 0.5 watt
 R14 = 100 ohms, 0.5 watt
 R15 = 180 ohms, 0.5 watt
 R16 = 6,200 ohms, 0.5 watt
 R17 = Thermistor, 495 ohms at 0°C., 150 ohms at 25°C., and 54 ohms at 50°C.
 SP = Speaker
 T1 = Antenna transformer (same as T1 of the four-transistor receiver)

- T2 = Oscillator transformer (same as T2 of the four-transistor receiver)
 T3, T4, T5 = Intermediate-frequency transformers (same as T3, T4, and T5 of the four-transistor reflex receiver).
 T6 = Driver transformer, primary impedance = 5000 ohms, secondary impedance (base-to-base, center tapped) = 2000 ohms.
 T7 = Class B output transformer, primary impedance (collector-to-collector, center tapped) = 800 ohms, secondary impedance = 12 ohms.

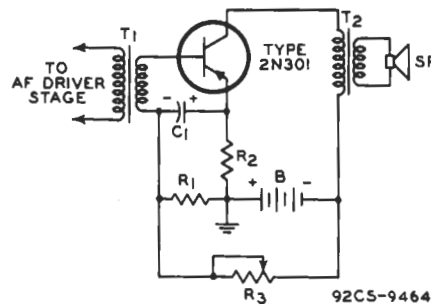
AUDIO POWER AMPLIFIER Class B; Output, 15 Watts



92CS-9463

- C1 = 0.01 μf , paper, 150 v.
 C2 = 0.01 μf , paper, 150 v.
 C3 = 1000 μf , electrolytic, 25 v.
 R1 = 2.7 ohms, 1 watt
 R2 = Bias-control potentiometer, 250 ohms, 1 watt
 R3 = 2.7 ohms, 1 watt
 R4 = 220 ohms, 1 watt
 R5 = 1 ohm, 1 watt
 R6 = 1 ohm, 1 watt
 SP = 15-watt, 8-ohm speaker
 T1 = Driver transformer

AUTOMOBILE RECEIVER AUDIO AMPLIFIER Class A; Output, 4 Watts

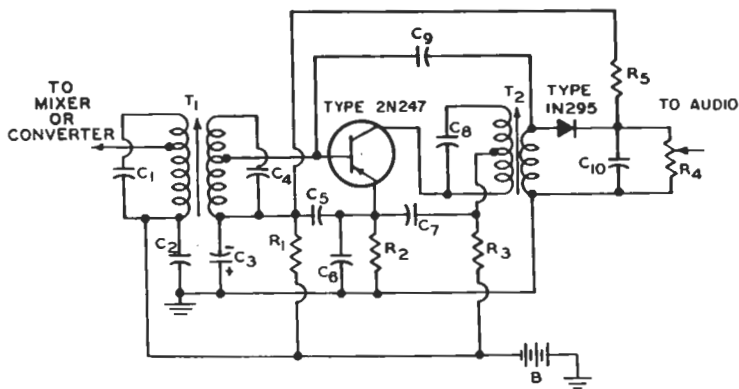


92CS-9464

- B = 14.4 volts (12-volt automobile battery source)
 C1 = 500 μf , electrolytic, 15 v.
 R1 = 27 ohms, 1 watt
 R2 = 1 ohm, 1 watt
 R3 = Bias-control potentiometer, 500 ohms, 1 watt
 T1 = Driver transformer
 T2 = Class A output transformer, primary impedance = 15 ohms, secondary impedance = 3.2 ohms

CIRCUITS

ONE-STAGE NEUTRALIZED 455-KC IF AMPLIFIER



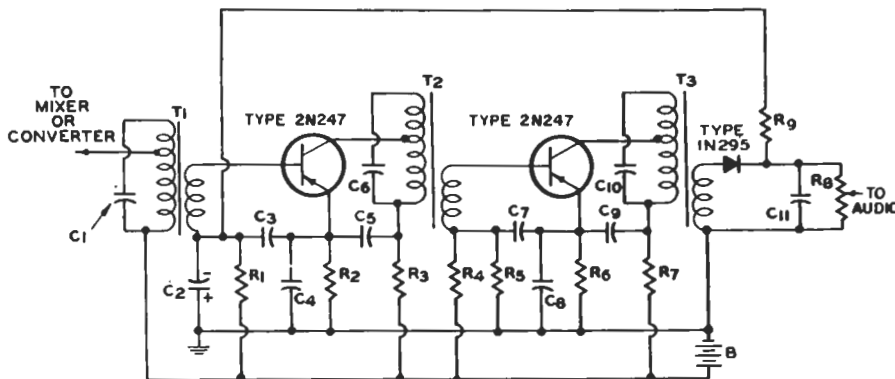
B = 9 volts, VS300 or VS301
 C1 = 220 μf , mica, 150 v.
 C2 = 0.05 μf , paper, 150 v.
 C3 = 10 μf , electrolytic, 15 v.
 C4 = 220 μf , mica, 150 v.
 C5 = 0.05 μf , paper, 150 v.
 C6 = 0.05 μf , paper, 150 v.
 C7 = 0.05 μf , paper, 150 v.
 C8 = 220 μf , mica, 150 v.
 C9 = 7 μf , mica, 150 v.
 C10 = 0.05 μf , paper, 150 v.
 R1 = 33000 ohms, 0.5 watt
 R2 = 1000 ohms, 0.5 watt
 R3 = 560 ohms, 0.5 watt
 R4 = 5000 ohms, volume control, 1 watt
 R5 = 10000 ohms, 0.5 watt

T1 = Doubled-tuned IF transformer, unloaded Q of pri-

mary and secondary = 120, loaded Q of primary = 41, tuned resistance of primary = 190,000 ohms, loaded Q of secondary = 50, tuned resistance of secondary = 190,000 ohms, reflected impedance of secondary = 138,000 ohms, impedance of secondary = 1700 ohms, coefficient of coupling = 82%.

T2 = Single-tuned IF transformer, unloaded Q = 120, loaded Q = 45.5, primary impedance = 17,000 ohms, secondary impedance = 1000 ohms, tuned resistance = 1000 ohms.

TWO-STAGE UNNEUTRALIZED 455-KC IF AMPLIFIER

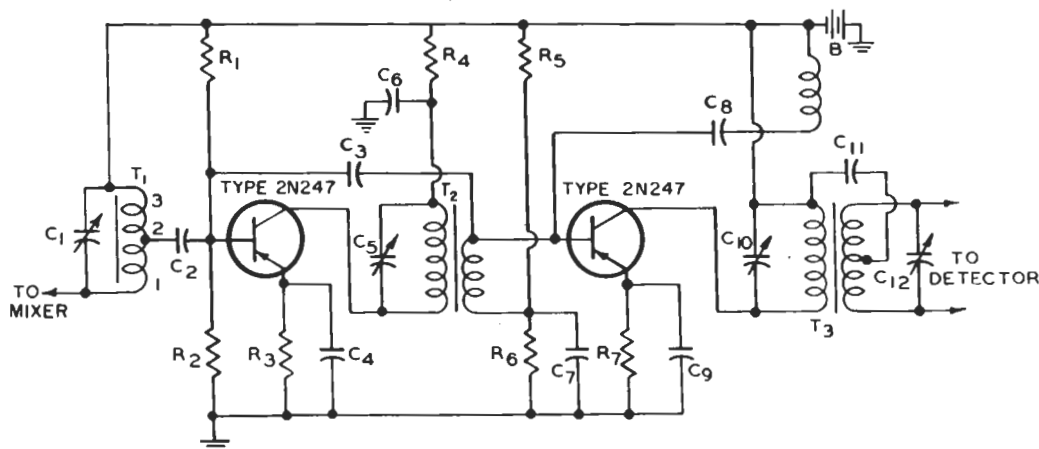


B = 9 volts, VS300 or VS301
 C1 = 220 μf , mica, 150 v.
 C2 = 10 μf , electrolytic, 15 v.
 C3 = 0.05 μf , paper, 150 v.
 C4 = 0.05 μf , paper, 150 v.
 C5 = 0.05 μf , paper, 150 v.
 C6 = 220 μf , mica, 150 v.
 C7 = 0.05 μf , paper, 150 v.
 C8 = 0.05 μf , paper, 150 v.
 C9 = 0.05 μf , paper, 150 v.
 C10 = 220 μf , mica, 150 v.
 C11 = 0.05 μf , paper, 150 v.

R1 = 33000 ohms, 0.5 watt
 R2 = 1000 ohms, 0.5 watt
 R3 = 560 ohms, 0.5 watt
 R4 = 39000 ohms, 0.5 watt
 R5 = 5600 ohms, 0.5 watt
 R6 = 1000 ohms, 0.5 watt
 R7 = 560 ohms, 0.5 watt
 R8 = 5000 ohms, volume control, 1 watt
 R9 = 10000 ohms, 0.5 watt
 T1 = IF transformer, primary impedance = 105000 ohms, secondary impedance = 1700

ohms, unloaded Q = 100, loaded Q = 35.
 T2 = IF transformer, primary impedance = 4600 ohms, secondary impedance = 1700 ohms, unloaded Q = 39, loaded Q = 35
 T3 = IF transformer, primary impedance = 30,000, secondary impedance = 1000 ohms, unloaded Q = 100, loaded Q = 35

TWO-STAGE, 10.7-MC IF AMPLIFIER



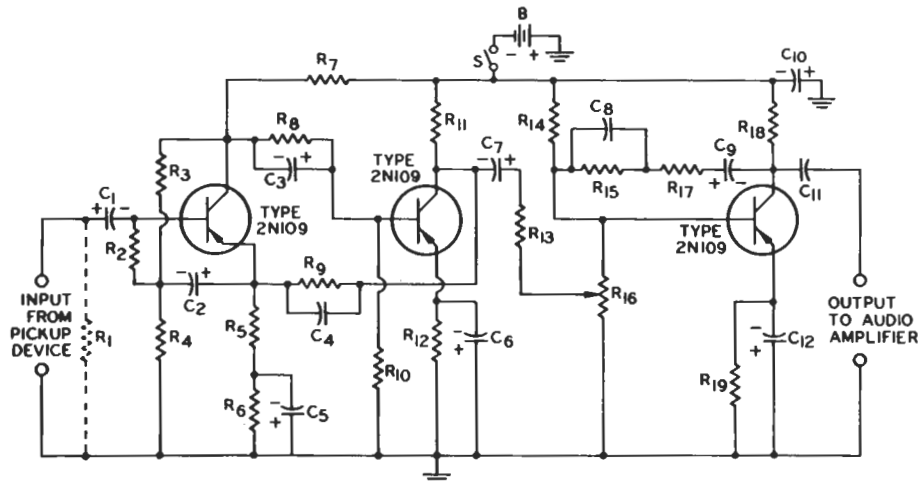
B = 9 volts, VS300 or VS301
 C1 = 30 μf , mica, 500 v.
 C2 = 1000 μf , mica, 500 v.
 C3 = 8.5 μf , mica, 500 v.
 C4 = 0.05 μf , paper, 0.5 watt
 C5 = Variable capacitor, 2-30 μf
 C6 = 0.05 μf , paper, 150 v.
 C7 = 0.05 μf , paper, 150 v.
 C8 = 0.05 μf , paper, 150 v.
 C9 = 0.05 μf , paper, 150 v.
 C10 = Variable capacitor, 2-30 μf

C11 = 0.05 μf , paper, 150 v.
 C12 = 30 μf , mica, 500 v.
 R1 = 39000 ohms, 0.5 watt
 R2 = 5600 ohms, 0.5 watt
 R3 = 1000 ohms, 0.5 watt
 R4 = 680 ohms, 0.5 watt
 R5 = 39000 ohms, 0.5 watt
 R6 = 5600 ohms, 0.5 watt
 R7 = 1000 ohms, 0.5 watt
 T1 = Tap transformer, primary impedance (terminal 1 to terminal 3) = 4250 ohms,

tap impedance (terminal 1 to terminal 2) = 170 ohms, unloaded Q = 150, loaded Q = 27.3, Turns Ratio = 5:1
 T2 = IF air-core transformer, inductance to tune with 30 μf , unloaded Q = 150, loaded Q = 27.3, Turns Ratio = 5:1
 T3 = IF air-core transformer, unloaded Q = 150, loaded Q = 27.3, Turns Ratio = 5:1

CIRCUITS

PHONOGRAPH PREAMPLIFIER



B = 12 volts, 9 cells from VS087 battery assembly or 8 VS035 batteries connected in series
 C1 = 15 μ f, electrolytic, 6 v.
 C2 = 5 μ f, electrolytic, 12 v.
 C3 = 1 μ f, electrolytic, 12 v.
 C4 = Equalization component. Refer to chart below.
 C5 = 10 μ f, electrolytic, 3 v.
 C6 = 100 μ f, electrolytic, 3 v.
 C7 = 5 μ f, electrolytic, 12 v.
 C8 = Equalization component. Refer to chart below.
 C9 = 1 μ f, electrolytic, 12 v.
 C10 = 100 μ f, electrolytic, 25 v.
 C11 = 0.02 μ f, paper, 150 v.

C12 = 100 μ f, electrolytic, 3 v.
 R1 = Refer to note below
 R2 = 0.01 megohm, 0.5 watt
 R3 = 3900 ohms, 0.5 watt
 R4 = 0.01 megohm, 0.5 watt
 R5 = Refer to note below
 R6 = 3300 ohms, 0.5 watt
 R7 = 22000 ohms, 0.5 watt
 R8 = 0.1 megohm, 0.5 watt
 R9 = 30000 ohms, 0.5 watt
 R10 = 0.01 megohm, 0.5 watt
 R11 = 15000 ohms, 0.5 watt
 R12 = 1000 ohms, 0.5 watt
 R13 = 3900 ohms, 0.5 watt
 R14 = 0.12 megohm, 0.5 watt
 R15 = Equalization component. Refer to chart below

R16 = Potentiometer, 5000 ohms, 0.5 watt, volume-control
 R17 = 15000 ohms, 0.5 watt
 R18 = 0.01 megohm, 0.5 watt
 R19 = 510 ohms, 0.5 watt

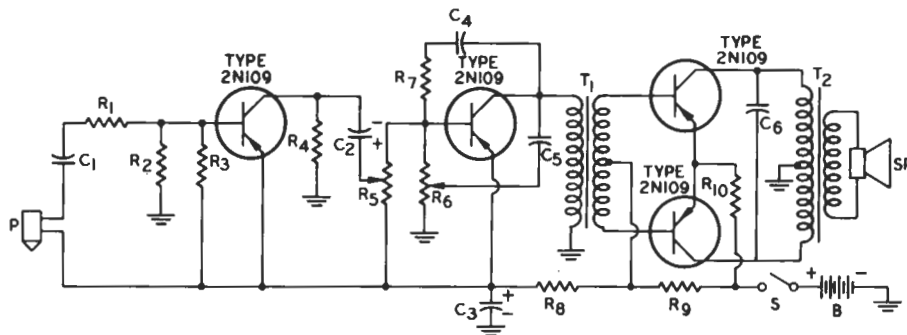
EQUALIZATION COMPONENTS

Curve	C4 μ f	C8 μ f	R15 ohms
R1AA	0.0045	0.02	300,000
AES	0.0033	0.038	omit
LP	0.0065	0.026	75000
NAB	0.0065	0.025	omit
Flat	omit	omit	short

NOTE: With low-inductance pickup, R1 = 2000 ohms, R5 = 200 ohms. With high-inductance pickup, omit R1, R5 = 1000 ohms. With piezoelectric-crystal pickup, R1 = 47000 ohms, R5 = 500 ohms, also insert a paper capacitor of 100 μ f in series with the pickup.

PHONOGRAPH AMPLIFIER

Class B, Output 200 mw



B = 9 volts, VS300 or VS301
 C1 = 0.01 μ f, paper, 150 v.
 C2 = 1 μ f, electrolytic, 12 v.
 C3 = 50 μ f, electrolytic, 12 v.
 C4 = 0.003 μ f, paper, 150 v.
 C5 = 0.002 μ f, paper, 150 v.
 C6 = 0.04 μ f, paper, 150 v.
 P = Phonograph cartridge, ceramic
 R1 = 1 megohm, 0.5 watt
 R2 = 0.22 megohm, 0.5 watt
 R3 = 4700 ohms, 0.5 watt
 R4 = 1500 ohms, 0.5 watt

R5 = Potentiometer, 5000 ohms, 0.5 watt, volume-control
 R6 = Potentiometer, 0.1 megohm, 0.5 watt, bass-boost
 R7 = 0.22 megohm, 0.5 watt
 R8 = 680 ohms, 0.5 watt
 R9 = 27 ohms, 0.5 watt
 R10 = 33 ohms, 0.5 watt
 SP = Speaker
 T1 = Interstage audio transformer with center-tapped secondary to provide a 3000-

ohm load at the primary terminals with a 5000-ohm load between the outer secondary terminals. DC primary current = 1.5 ma. DC primary resistance = 300 ohms.

T2 = Output transformer with center-tapped primary to provide a 550-ohm load between the outer primary terminals with the desired speaker connected to the

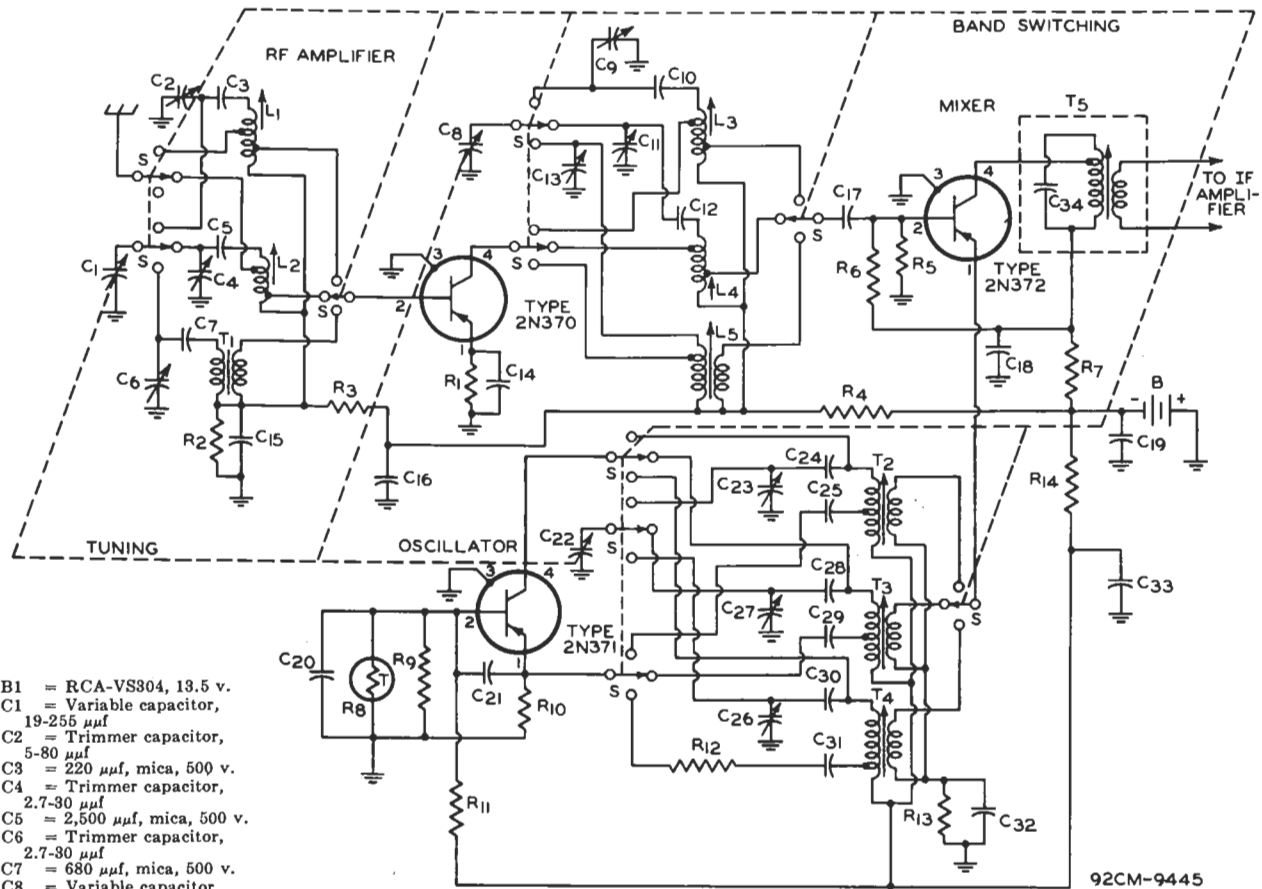
secondary terminals. DC current unbalance between halves of primary = 1 ma. DC primary resistance = 15 ohms per section.

Battery Current:
 No-signal current = 6 ma
 Average current = 26 ma
 Peak current = 42 ma

CIRCUITS

RF and MIXER-OSCILLATOR STAGES for 3-BAND RECEIVER

Utilizing Types 2N370, 2N371, and 2N372.



- B1 = RCA-VS304, 13.5 v.
- C1 = Variable capacitor, 19-255 μf
- C2 = Trimmer capacitor, 5-80 μf
- C3 = 220 μf , mica, 500 v.
- C4 = Trimmer capacitor, 2.7-30 μf
- C5 = 2,500 μf , mica, 500 v.
- C6 = Trimmer capacitor, 2.7-30 μf
- C7 = 680 μf , mica, 500 v.
- C8 = Variable capacitor, 19-255 μf
- C9 = Trimmer capacitor, 5-80 μf
- C10 = 39 μf , mica, 500 v.
- C11 = Trimmer capacitor, 2.7-30 μf
- C12 = 2,700 μf , mica, 500 v.
- C13 = Trimmer capacitor, 2.7-30 μf
- C14 = 0.05 μf , paper, 150 v.
- C15 = 0.05 μf , paper, 150 v.
- C16 = 0.05 μf , paper, 150 v.
- C17 = 2200 μf , mica, 500 v.
- C18 = 0.05 μf , paper, 150 v.
- C19 = .1 μf , paper, 150 v.
- C20 = .05 μf , paper, 150 v.

- C21 = 120 μf , mica, 500 v.
- C22 = Variable capacitor, 19-255 μf
- C23 = Trimmer capacitor, 5-80 μf
- C24 = 0.0068 μf , ceramic disc, 500 v.
- C25 = 56 μf , mica, 500 v.
- C26 = Trimmer capacitor, 2.7-30 μf
- C27 = Trimmer capacitor, 5-80 μf
- C28 = 0.0033 μf , ceramic disc, 500 v.
- C29 = 180 μf , mica, 500 v.

- C30 = 1,500 μf , mica, 500 v.
- C31 = 430 μf , mica, 500 v.
- C32 = 0.05 μf , paper, 150 v.
- C33 = 0.05 μf , paper, 150 v.
- C34 = Fixed tuner capacitor for T-5 Transformer
- R1 = 1,000 ohms, 0.5 watt
- R2 = 5,100 ohms, 0.5 watt
- R3 = 39,000 ohms, 0.5 watt
- R4 = 1,000 ohms, 0.5 watt
- R5 = 4,700 ohms, 0.5 watt
- R6 = 82,000 ohms, 0.5 watt
- R7 = 470 ohms, 0.5 watt
- R8 = Thermistor 20,000 ohms at 25° C, 78,600 ohms at 0° C, 6,520 ohms at 50° C

- R9 = 20,000 ohms, 0.5 watt
- R10 = 560 ohms, 0.5 watt
- R11 = 100,000 ohms, 0.5 watt
- R12 = 5,600 ohms, 0.5 watt
- R13 = 1,000 ohms, 0.5 watt
- R14 = 470 ohms, 0.5 watt
- T-5 = 455-Kc I.F. Transformer

NOTE: Standard Broadcast Band Antenna Coil (T1) should be wound on an 8" X .33" ferrite rod.

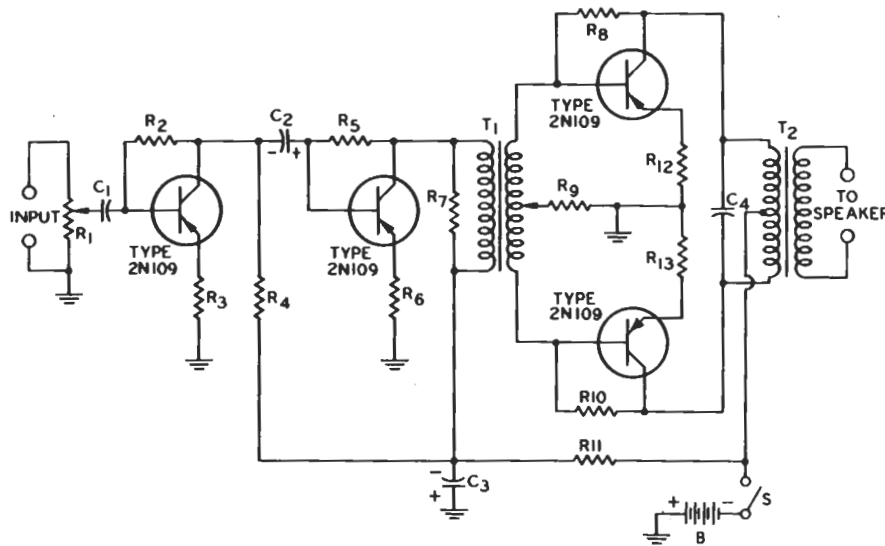
Radio-Frequency Tuner Circuit Coil Data

COIL	ANTENNA			INTERSTAGE			OSCILLATOR		
	T1 (Standard Broadcast Band)	L2 (4.5 to 11.5 Mc)	L1 (10.5 to 23 Mc)	L5 (Standard Broadcast Band)	L4 (4.5 to 11.5 Mc)	L3 (10.5 to 23 Mc)	T4 (Standard Broadcast Band)	T3 (4.5 to 11.5 Mc)	T2 (10.5 to 23 Mc)
Total primary turns	127	21	13	225	21	12	120	18	10
1st primary tap-turns from bottom	-	2	1	13	2	1	12	2	1
2nd primary tap-turns from bottom	-	7	4	-	6	6	-	-	-
Secondary turns	7	-	-	12	-	-	2	1	1
Wire size	10/38 Litz	#22	#24	7/41 Litz	#22	#24	7/41 Litz	#28	#28
Coil diameter (inside) - inches	.33	3/8	3/8	1/4	3/8	3/8	1/4	1/4	1/4
Turns per inch	24	24	24	Universal	24	24	Universal	60	11

CIRCUITS

HIGH-GAIN, LOW-DISTORTION, LOW-DRAIN AUDIO AMPLIFIER

Class B; Output, 250 mw; Power Gain, approx. 90 db
Input Impedance, approx. 7500 ohms



B = 12 volts, 9 cells from VS087 battery assembly or 8 VS035 batteries connected in series
C1 = 1 μ f, paper, 150 v.
C2 = 2 μ f, electrolytic, 12 v.
C3 = 100 μ f, electrolytic, 12 v.
C4 = 0.05 μ f, paper, 150 v.
R1 = Volume-control potentiometer 10000 ohms, 0.5 watt
R2 = 0.15 megohm, 0.5 watt
R3 = 10 ohms, 0.5 watt
R4 = 10000 ohms, 0.5 watt
R5 = 0.27 megohm, 0.5 watt
R6 = 10 ohms, 0.5 watt
R7 = 10000 ohms, 0.5 watt

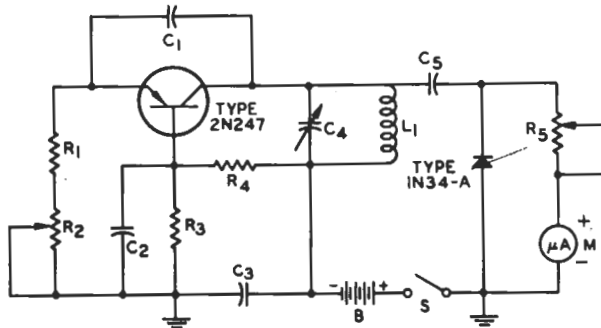
R8 = 15000 ohms, 0.5 watt
R9 = 60 ohms, 0.5 watt
R10 = 15000 ohms, 0.5 watt
R11 = 1000 ohms, 0.5 watt
R12, R13 = 4.7 ohms, 0.5 watt
T1 = Interstage audio transformer with center-tapped secondary to provide a 25000-ohm load at the primary terminals with a 3000-ohm load between the outer secondary terminals. DC primary resistance = 300 ohms.

T2 = Output transformer with center-tapped primary to provide a 300-ohm load between the outer primary terminals with the desired speaker connected to the secondary terminals. DC current unbalance of primary = 5 ma. DC primary resistance = 20 ohms per section.

Battery Current:
No-signal current = 3.8 ma
Average current = 18 ma
Peak current = 50 ma

GRID-DIP METER

For Measuring Resonant Frequencies from 2 to 50 Mc (approx.)

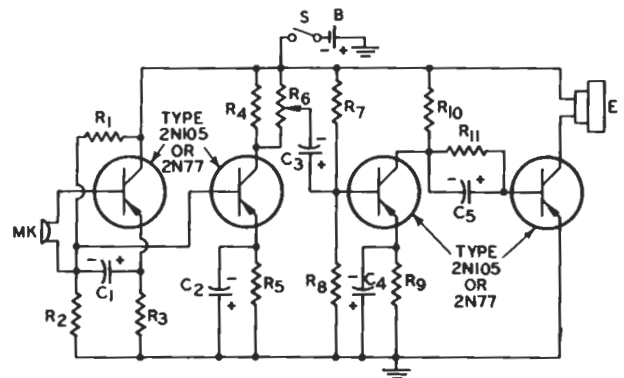


B = 9 volts, VS300
C1 = 10 μ f, mica, 500 v.
C2 = 0.05 μ f, paper, 150 v.
C3 = 0.05 μ f, paper, 150 v.
C4 = Variable capacitor, 10-100 μ f
C5 = 5 μ f, mica, 500 v.
L1 = Air-core plug-in coil to resonate with C4

M = DC Microammeter, range: 0 to 50 μ A
R1 = 220 ohms, 0.5 watt
R2 = Potentiometer, 3000 ohms, 0.5 watt
R3 = 3900 ohms, 0.5 watt
R4 = 39000 ohms, 0.5 watt
R5 = Potentiometer, 0.25 megohms, 0.5 watt

NOTE: R1 and R2 may be replaced with a single 1000-ohm resistor, but there will be a small decrease in output at the lower and higher frequencies.

HEARING AID

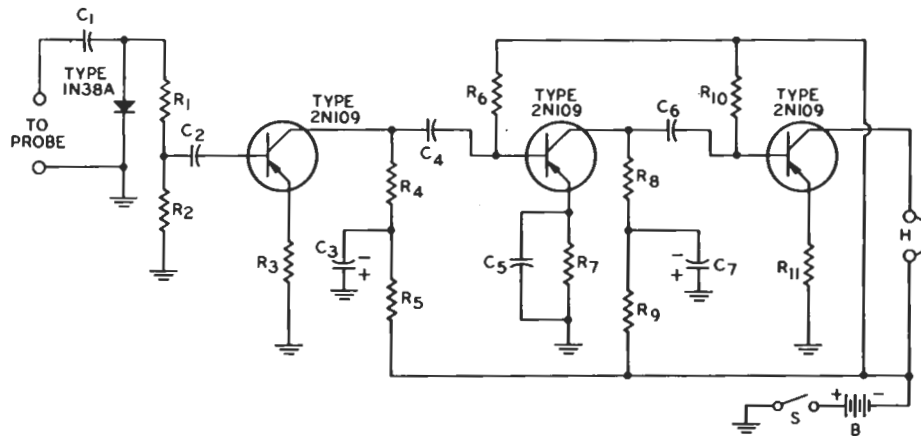


B = 1.5 volts, VS034
C1 = 4 μ f, electrolytic, 3 v.
C2 = 10 μ f, electrolytic, 3 v.
C3 = 4 μ f, electrolytic, 3 v.
C4 = 10 μ f, electrolytic, 3 v.
C5 = 1 μ f, electrolytic, 3 v.
MK = Microphone, 1000-ohm, magnetic
R1 = 12000 ohms, 1/8 watt
R2 = 6800 ohms, 1/8 watt

R3, R4, R5 = 1500 ohms, 1/8 watt
R6 = Potentiometer, 0.1 megohm, 1/8 watt, volume-control
R7 = 15000 ohms, 1/8 watt
R8 = 6800 ohms, 1/8 watt
R9 = 1200 ohms, 1/8 watt
R10 = 1500 ohms, 1/8 watt
R11 = 56000 ohms, 1/8 watt

CIRCUITS

HIGH-GAIN TRANSFORMERLESS SIGNAL-TRACING PROBE

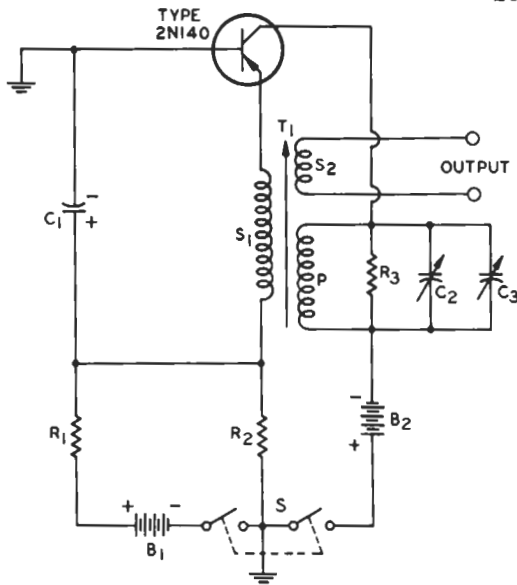


B = 6 volts, 4 VS035 batteries connected in series
 C1 = 0.001 μ f, mica, voltage rating as required
 C2 = 0.5 μ f, paper
 C3 = 25 μ f, electrolytic, 6 v.
 C4, C5, C6 = 0.5 μ f, paper

C7 = 25 μ f, electrolytic, 6 v.
 H = Headphones, 2000-ohm
 R1 = 51000 ohms, 0.5 watt
 R2 = 0.5 megohm, 0.5 watt
 R3 = 5100 ohms, 0.5 watt
 R4 = 0.01 megohm, 0.5 watt
 R5 = 1000 ohms, 0.5 watt

R6 = 1 megohm, 0.5 watt
 R7 = 1000 ohms, 0.5 watt
 R8 = 0.01 megohm, 0.5 watt
 R9 = 1000 ohms, 0.5 watt
 R10 = 0.7 megohm, 0.5 watt
 R11 = 47 ohms, 0.5 watt

IF AND BROADCAST-BAND SIGNAL TRACER (0.4-Mc to 1.7-Mc) Self-Quenching Oscillator

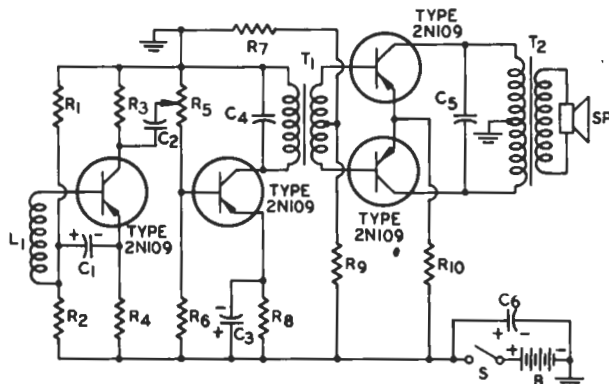


B1 = 4.5 volts, 3 VS034 batteries connected in series
 B2 = 4.5 volts, 3 VS034 batteries connected in series
 C1 = 0.25 μ f, electrolytic, 6 v.
 C2 = Variable capacitor, 10-335 μ f
 C3 = Trimmer capacitor, 2-20 μ f
 R1 = 47000 ohms, $\frac{1}{8}$ watt
 R2 = 56000 ohms, $\frac{1}{8}$ watt
 R3 = 33000 ohms, $\frac{1}{8}$ watt
 P = Primary winding of T1
 S = Switch
 S1, S2 = secondary windings of T1

T1 = Transformer wound as follows: Using a threaded resinite coil form and a $\frac{1}{2}$ " ferrite core with matching threads, wind 180 turns of #7/41 Litz wire over a length of about $\frac{1}{2}$ ". This winding will have more than one layer and is the primary winding P. The secondary windings, S1 and S2, are wound on top of the primary winding with #34 SSE wire. Winding S1 has 10 turns, and winding S2, 3 turns. A resinite collar with terminals may be fitted over one end of the coil form to provide connections to all the windings.

NOTE: The fundamental frequency of oscillation is about 0.4 to 0.85 megacycles per second. The second harmonic frequencies from 0.8 to 1.7 megacycles per second may be used but will provide a smaller output. The modulation frequency is about 500 cycles per second.

TELEPHONE PICKUP AMPLIFIER



B = 12 volts, 9 cells from VS087 battery assembly or 8 VS035 batteries connected in series
 C1 = 10 μ f, electrolytic, 6 v.
 C2 = 1 μ f, paper, 150 v.
 C3 = 10 μ f, electrolytic, 6 v.
 C4 = 0.01 μ f, paper, 150 v.
 C5 = 0.04 μ f, paper, 150 v.
 C6 = 50 μ f, electrolytic, 12 v.
 L1 = Telephone pickup coil
 R1 = 0.1 megohms, 0.5 watt
 R2, R3 = 10000 ohms, 0.5 watt
 R4 = 1500 ohms, 0.5 watt
 R5 = Potentiometer, 0.1 megohm, 0.5 watt, volume-control
 R6 = 10000 ohms, 0.5 watt
 R7 = 5100 ohms, 0.5 watt
 R8 = 680 ohms, 0.5 watt
 R9 = 75 ohms, 0.5 watt

R10 = 10 ohms, 0.5 watt

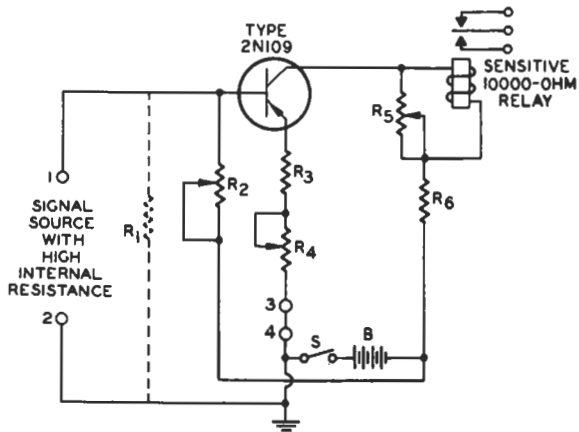
SP = Speaker
 T1 = Interstage audio transformer with a center-tapped secondary to provide a 9000-ohm load at the primary terminals with a 9000-ohm load between the outer secondary terminals.

T2 = Output transformer with a center-tapped primary to provide a 750-ohm load at the outer primary terminals with the desired speaker connected to the secondary terminals.

Battery Current:
 No-signal current = 6 ma
 Average current = 10.3 ma
 Peak current = 23 ma

CIRCUITS

SENSITIVE RELAY



B = 22.5 volts, VS084

R1 = 1000 ohms, 0.5 watt

R2 = Bias-control potentiometer, 0.1 megohm, 0.5 watt

R3 = 10 ohms, 0.5 watt

R4 = Input-impedance-control potentiometer, 1000 ohms, 0.5 watt

R5 = Sensitivity-control potentiometer, 0.1 megohm, 0.5 watt

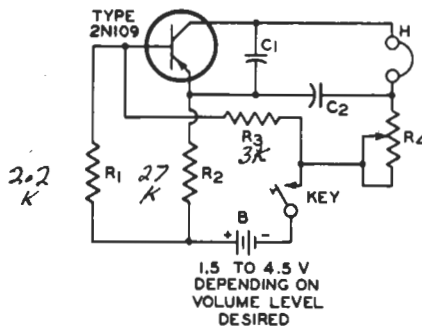
R6 = 1000 ohms, 0.5 watt

NOTES: (1) If a signal source with internal resistance is used, omit R1 and connect a jumper between terminals 3 and 4 as shown. Adjust R4 for best performance.

(2) If a signal source with a low internal resistance is used, connect the source between terminals 3 and 4, omitting the jumper. Connect R1 as shown. Adjust R4 to provide a direct connection from R3 to terminal 3.

(3) Relays having a dc coil resistance of less than 10000 ohms may be used, provided the battery voltage is proportionately reduced. In such event, circuit sensitivity will be reduced.

CODE-PRACTICE OSCILLATOR



B = VS036 (see note)

C1, C2 = 0.01 μ f, paper, 150 v.

H = Headphones, 2000-ohm, magnetic

R1 = 2200 ohms, 0.5 watt

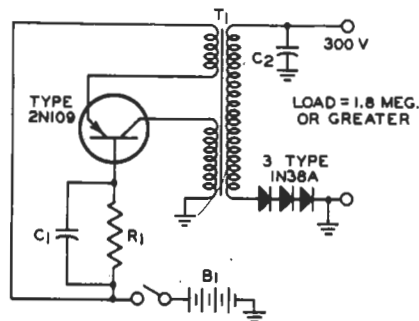
R2 = 27000 ohms, 0.5 watt

R3 = 3000 ohms, 0.5 watt

R4 = Volume control potentiometer, 50000 ohms, 0.5 watt

NOTE: One to three series-connected RCA-VS036 dry cells may be used, depending upon the volume level desired.

PORTABLE LOW-DRAIN HIGH-VOLTAGE POWER SUPPLY



B = 12 volts, 9 cells from VS087 battery assembly or 8 VS035 batteries connected in series

C1 = 0.01 μ f, paper, 150 v.

C2 = 0.1 μ f, paper, 600 v.

R1 = 22000 ohms, 0.5 watt

T1 = Transformer with 15-turn primary, 5-turn tickler and 530-turn secondary

