#### Errata

Title & Document Type: 4262A LCR Meter Operating and Service Manual

Manual Part Number: 04262-90007

Revision Date: October 1983

#### **HP References in this Manual**

This manual may contain references to HP or Hewlett-Packard. Please note that Hewlett-Packard's former test and measurement, semiconductor products and chemical analysis businesses are now part of Agilent Technologies. We have made no changes to this manual copy. The HP XXXX referred to in this document is now the Agilent XXXX. For example, model number HP8648A is now model number Agilent 8648A.

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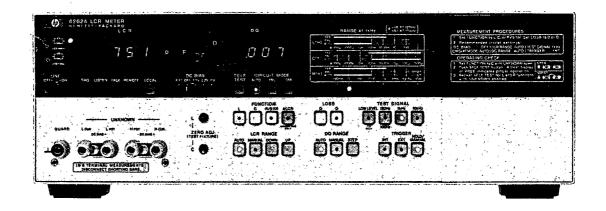
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# 4262A DIGITAL LCR METER





## MANUAL CHANGES

4262A

DIGITAL LCR METER

MANUAL IDENTIFICATION

Model Number: 4262A

Date Printed: OCT. 1983

Part Number: 04262-90007

This supplement contains important information for correcting manual errors and for adapting the manual to instruments containing improvements made after the printing of the manual.

To use this supplement:

Make all ERRATA corrections.

Make all appropriate serial number related changes indicated in the tables below.

SERIAL PREFIX OR NUMBER	MAKE MANUAL CHANGES	SERIAL PREFIX OR NUMBER	MAKE MANUAL CHANGES
A11	1		
			·
***************************************			

► NEW ITEM

## ERRATA

Page 2-8, Table 2-1:

Add the following item to the option 101 (HP-IB) components.

PN 2190-0577

2ea.

Spring Washer

Page 3-22, Figure 3-9:

Delete the following sentence from beneath the table in step 3: "\*Bias current when +40V is applied to DC BIAS connector."

#### NOTE

Manual change supplements are revised as often as necessary to keep manuals as current and accurate as possible. Hewlett-Packard recommends that you periodically request the latest edition of this supplement. Free copies are available from all HP offices. When requesting copies quote the manual identification information from your supplement, or the model number and print date from the title page of the manual.

Date/Div: Sep. 19, 1984/33

Page

1 of 4

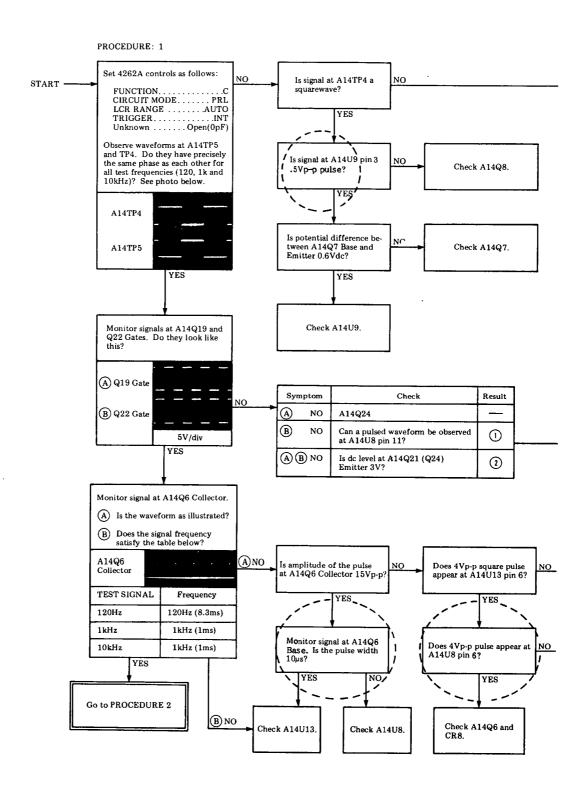


▶ Page 4-24, Figure 4-9: Correct the TTL outputs table in Figure 4-9 as follows:

Comparison	TTL	TTL output pins*				
LCR	19	20	45			
HIGH	0.C.	LOW	LOW			
IN	LOW	LOW	0.0.			
LOW	LOW	0.C.	LOW			
DQ	15	16	41			
HIGH	0.C.	LOW	LOW			
IN	LOW	LOW	0.0.			
LOW	LOW	0.C.	LOW			

- \* TTL low-level output is indicated as LOW, and open-collector turn-off state is indicated as O.C.
- ▶ Page 8-47, Figure 8-29: Change the part number for the power transformer to 9100-0865.

Page 8-53, Figure 8-39, Correct the flow diagram as shown below:



## CHANGE 1

## ▶ Page 1-10, Table 1-4:

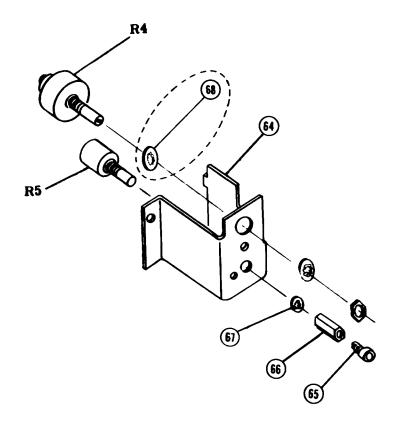
Change the recommended oscilloscope to the HP 1740A.

Table 6-3: Change the table as shown below:

Reference Designation	HP Part Number	Description
A14 Q4	1855-0570	TRANSISTOR J-FET N-CHAN SI
R5 (ZERO ADJ L)	2100-4086	RESISTOR-VAR 500 10%
66	04262-24004	NUT-HEX-DBL-CHAM 1/4-32-THD
68	2190-0016	WASHER-LK INTL T NO3/8

## Page 6-25:

Add a washer (reference designation (1)) to the illustration, as shown below:





## OPERATING AND SERVICE MANUAL

## MODEL 4262A LCR METER

(including Options 001, 004, 010, and 101)

#### SERIAL NUMBERS

This manual applies directly to instruments with serial numbers prefixed 2022J

With changes described in Section  $\mathbb{W}$ , this manual also applies to instruments with serial numbers prefixed 1710J, and 1739J

For additional important information about serial numbers, see INSTRUMENTS COVERED BY MAN-UAL in Section I.

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Printed: OCT, 1983

Manual Part No. 04262-90007 Microfiche Part No. 04262-90057

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#### CERTIFICATION

Hewlett-Packard Company certifies that this product met its published specifications at the time of shipment from the factory. Hewlett-Packard further certifies that its calibration measurements are traceable to the United States National Bureau of Standards, to the extent allowed by the Bureau's calibration facility, and to the calibration facilities of other International Standards Organization members.

#### WARRANTY

This Hewlett-Packard instrument product is warranted against defects in material and workmanship for a period of one year from date of shipment, except that in the case of certain components listed in Section 1 of this manual, the warranty shall be for the specified period. During the warranty period, Hewlett-Packard Company will, at its option, either repair or replace products which prove to be defective.

For warranty service or repair, this product must be returned to a service facility designated by HP. Buyer shall prepay shipping charges to HP and HP shall pay shipping charges to return the product to Buyer. However, Buyer shall pay all shipping charges, duties, and taxes for products returned to HP from another country.

HP warrants that its software and firmware designated by HP for use with an instrument will execute its programming instructions when properly installed on that instrument. HP does not warrant that the operation of the instrument, or software, or firmware will be uninterrupted or error free.

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 $\begin{tabular}{ll} Product & maintenance & agreements & and & other & customer & assistance & agreements & are available & for & Hewlett-Packard & products. \end{tabular}$ 

For any assistance, contact your nearest Hewlett-Packard Sales and Service Office. Addresses are provided at the back of this manual.

#### SAFETY SYMBOLS

General Definitions of Safety Symbols Used On Equipment or In Manuals.



Instruction manual symbol: the product will be marked with this symbol when it is necessary for the user to refer to the instruction manual in order to protect against damage to the instrument.



Indicates dangerous voltage (terminals fed from the interior by voltage exceeding 1000 volts must be so marked).



Protective conductor terminal. For protection against electrical shock in case of a fault. Used with field wiring terminals to indicate the terminal which must be connected to ground before operating equipment.



Low-noise or noiseless, clean ground (earth) terminal. Used for a signal common, as well as providing protection against electrical shock in case of fault. A terminal marked with this symbol must be connected to ground in the manner described in the installation (operating) manual, and before operating the equipment.



Frame or chassis terminal. A connection to the frame (chassis) of the equipment which normally includes all exposed metal structures.



Alternating current (power line).

\_\_\_

Direct current (power line).

\_\_\_

Alternating or direct current (power line).

#### WARNING

A WARNING denotes a hazard. It calls attention to a procedure, practice, condition or the like, which, if not correctly performed or adhered to, could result in injury or death to personnel.

#### **CAUTION**

The CAUTION sign denotes a hazard. It calls attention to an operating procedure, practice, condition or the like, which, if not correctly performed or adhered to, could result in damage to or destruction of part or all of the product.

Note

A Note denotes important information. It calls attention to a procedure, practice, condition or the like, which is essential to highlight.

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# SECTION I GENERAL INFORMATION

#### 1-1. INTRODUCTION.

1-2. This operating and service manual contains the information required to install, operate, test, adjust and service the Hewlett-Packard Model 4262A Digital LCR Meter. Figure 1-1 shows the instrument and supplied accessories. This section covers specifications, instrument identification, description, options, accessories, and other basic information.

1-3. Listed on the title page of this manual is a microfiche part number. This number can be used to order  $4 \times 6$  inch microfilm transparencies of the manual. Each microfiche contains up to 60 photoduplicates of the manual pages. The microfiche package also includes the latest manual changes supplement as well as all pertinent service notes. To order an additional manual, use the part number listed on the title page of this manual.

#### 1-4. DESCRIPTION.

1-5. The HP Model 4262A LCR Meter is a general

purpose, fully automatic test instrument designed to measure the parameters of an impedance element with high accuracy and speed. The 4262A measures capacitance, inductance, resistance (equivalent series resistance) and dissipation factor or quality factor over a wide range at test frequencies of 120Hz, 1kHz and 10kHz employing a five-terminal connection configuration between the component and the instrument. The measuring circuit for the device to be measured is capable of both parallel and series equivalent circuit measurements and the measured values are displayed by the two three-full digits LED displays on the front panel. A convenient diagnostic function, also featured in the 4262A, is actuated by a SELF TEST switch. This confirms functional operation of the instrument.

1-6. The measuring range for capacitance is from  $0.01 \mathrm{pF}$  to  $19.99 \mathrm{mF}$ , inductance from  $0.01 \mathrm{\mu H}$  to  $1999 \mathrm{H}$ , and resistance from  $1 \mathrm{m}\Omega$  to  $19.99 \mathrm{M}\Omega$ , which are measured with a basic accuracy of 0.2 to 0.3% depending on test signal level, frequency, and measuring equivalent circuit, and at typical measuring speeds of 220 to 260 milliseconds at

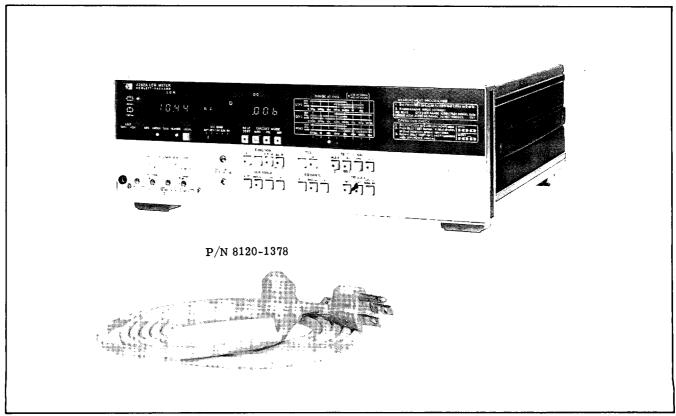


Figure 1-1. Model 4262A and Accessories.

#### Table 1-1. Specifications (Sheet 1 of 4).

#### COMMON SPECIFICATIONS

Parameters Measured: C - D or Q (1/D)

L - D or Q(1/D)

R (ESR) (Loss measurement can be negated by switch on internal board).

Display: 3-1/2 Digit, Maximum Display 1999
(When D value is more than 10, max-

mum display is 199).

Measurement Circuit Modes:
Auto, Parallel, and Series

Measurement Terminals: 5-terminal configuration (high and low terminals for both potential and current leads plus guard).

Range Modes: LCR - Auto and Manual

(up-down)

DQ - Auto and Manual (step)

Measurement Frequencies: 120(100)Hz, 1kHz

and  $10kHz \pm 3\%$ .

Test Signal Level: Normal level: 1Vrms.

Low level: 50mVrms (parallel

capacitance mode only)

Warm-up Time: 15 minutes

Deviation Measurement: When  $\triangle$ LCR key is depressed, the existing measured value is stored as a reference value and displayed value is offset to zero. The range is held and deviation is displayed as the difference between the referenced value and subsequent result. (Deviation spread in counts from -999 to 1999).

Offset Adjustment: Stray capacitance and residual inductance of test jig can be compensated for as follows:

C: up to 10pF L: up to 1 $\mu$ H

Self Test: Annunciates either Pass, or Fail for performance in each of the five basic ranges.

DC Bias:

Internal: 1.5V, 2.2V, 6V (Selectable at front

panel). Accuracy ±5%

External: External DC bias connector on rear panel. Maximum +40V.

Trigger: Internal, External, or Manual

#### **GENERAL**

Operating Temperature & Humidity:

0°C to 55°C at 95% RH(to 40°C)

Power Requirements:  $100/120/220V \pm 10\%$ , 240V + 5% - 10% 48 - 66Hz

Power Consumption: 55VA with any option

Dimensions:  $426(W) \times 147(H) \times 345(D)mm$ 

 $(16-3/4" \times 5-3/4" \times 13-3/4")$ 

Weight: Approximately 8kg (Std)

		C-D, C-Q MEASUREMENT							
Ranges	C 1kHz	$ \begin{array}{ c c c c c c c c c c c c c c c c c c c$							
ŭ	D	.001~19.9 (2 Ranges)							
	Q *1	0.05~1000 (4 Ranges)							
-	₩	1V or 50mV (LOW LEVEL)							
Test Signal Level *2	<b>⊣⊦</b> ₩-	10μA 100μA 1mA 10mA 40mA							
Level *2	AUTO	Same as - Mode Same as - Mode							
	-ch-	0.2% + 1 counts (Test signal level; 1V)							
		0.3% + 2 counts (Test signal level; 50mV)							
C Accuracy *3	-H-W-	(At 120Hz, 1kHz) 0.3% + 2 counts 0.5% 1% 4 2 counts 2 counts 2 counts							
	ATTE	(At 10kHz) 0.3% + 2 counts 1% + 2 5% + 2							
	AUTO	Same as - Mode Same as - Mode							
		0.2% + (2 + 200/Cx) counts At 120Hz, 1kHz (Test signal level; 1V)							
	_Hh_	0.5% + (2 + 200/Cx) counts (1est signal level, 1v) At 10kHz							
		0.3% + (2 + 1000/Cx) counts At 120Hz, 1kHz							
D(1/Q) Accuracy *3		1.0% + (2 + 1000/Cx) counts (Test signal level; 50mV) At 10kHz							
		(At 120Hz, 1kHz) $0.3\% + (2 + Cx/500)$ counts $\frac{C_x}{18 + (5 + \frac{C_x}{500})}$							
	-H-W-	(At 10kHz) $0.5\% + (2 + Cx/500) \text{ counts}$ $\left[1\% + (5 + \frac{Cx}{500})\right] \left[5\% + (5 + \frac{Cx}{500})\right]$							
	AUTO	Same as - Mode Same as - Mode							

Accuracy applies over a temperature range of  $23^{\circ}C \pm 5^{\circ}C$  (At  $0^{\circ}C$  to  $55^{\circ}C$ , error doubles).

Note: C accuracy for higher D values are unspecified.

<sup>\*1</sup> Calculated from D value as a reciprocal number.
\*2 Typical data, varies with value of D and number of counts.

<sup>\*3 ±(%</sup> of reading + counts). Cx is capacitance readout in counts. This accuracy only applies for D values to 1.999. (For higher D values, refer to General Information).

<sup>\*4</sup> (5% + 2 counts) at 1 kHz.

Table 1-1. Specifications (Sheet 3 of 4).

				L-D, L-Q N	1EASURE!	MENT			
Ranges	L	120Hz 1kHz 10kHz	1000μΗ 100.0μΗ 10.00μΗ	$1000 \mu H$	100.0mH 10.00mH 1000µH	1000mH 100.0mH 10.00mH	10.00H 1000mH 100.0mH	100.0H 10.00H 1000mH	1000H 100.0H 10.00H
	D			.001~19.9 (2 Ranges)					
	Q*1			0.05~100	00 (4 Range	es)			
	4	₹3					1	V	
Test Signal Level *2	-00	*	40mA	10mA	1mA	100μΑ	10μΑ		
	AU	TO		Same as	-380-₩ <del>-</del>	Mode	Same a	s -@-	Mode
	700			(At 120H	z, 1kHz)	0.3% + 2	2 counts	1% + 2	counts
	44	<b>W</b> -		(At 10	kHz)	0.3% + 2	counts	1% + 2	5% + 2
L Accuracy*3				0.	(At 120Hz,	1kHz)			
	-,00	<b>~~</b>	0.3% + 2 0.2% + 2 counts					(At 10kHz)	)
	AU	то		Same as	<b>-₩-</b>	Mode	Same a	s -(M)	Mode
	~	180		(At 120H	z, 1kHz)	0.3% + (3	+ Lx/500)	1% + (3 +	Lx/500)
7 (1)		<b>*</b> -		(At 10	kHz)	0.5% + (3 + Lx/500)		1% + (3 + Lx / 500 )	5% + (5 + Lx 500 )
D(1/Q) Accuracy	~~			0.2% + (	$3 + 200/L_2$	c) counts		(At 120Hz	, 1kHz)
	- 00	<b>\\\</b>		0.5% + (	(3 + 200/L	k) counts		(At 1kHz)	
	AU	то		Same as	-30·₩-	Mode	Same a	s -(CC)-	Mode

<sup>\*1</sup> Calculated from D value as a reciprocal number.

Accuracy applies over a temperature range of 23°C ± 5°C. (At 0°C to 55°C, error doubles).

### R/ESR MEASUREMENT

		<u> </u>	.,	LAJONE					
Ranges	120Hz R/ESR 1kHz 10kHz	$1000 \mathrm{m}\Omega$	10-00Ω	100.0Ω	1000Ω	10.00kΩ	$100.0$ k $\Omega$	1000kΩ	10.00Ms
,	4						1V	-	
Test Signal Level *1	-æ-₩- -IF₩-	40mA	10mA	1mA	100μΑ	10μΑ			
	AUTO	Sa	me as ⊣⊦	w-as-w-	Mode	Sam	eas C	₩- M	ode
	4					0.39	% + 2 cou	ınts *3	
Accuracy *2	-28		0.2	% + 2 cou	ınts			•	
	AUTO	Same as -I-w					ode		

<sup>\*1</sup> Typical data, varies with number of counts.

Accuracy applies over a temperature range of 23°C ± 5°C. (At 0°C to 55°C, error doubles.)

<sup>\*2</sup> Typical data, varies with value of D and number of counts.

<sup>\*3 ±(%</sup> of reading + counts). Lx is inductance readout in counts. This accuracy only applies for D values to 1.999.

<sup>\*2</sup>  $\pm$ (% of reading + counts).

<sup>\*3 (0.5% + 2</sup> counts) on 10.00M $\Omega$  range at 10kHz.

<sup>\*\*</sup> Measurement range for ESR (equivalent series resistance) is from  $1m\Omega$  to  $19.99k\Omega$  (typical), which varies with series capacitance and inductance value . . . . refer to "REFERENCE DATA".

#### **OPTIONS**

Option 001: Simultaneous BCD output of LCR and DQ data (positive true). Max. sink current 16mA. Mating connector (P/N 1251-0086). (Alternate BCD output of LCR and DQ data selectable by switch on internal board).

Option 004: Digital comparator (can not be used with OPT 101). Compares measured value with high and low limit settings for LCR or DQ and provides HIGH, IN, LOW comparison outputs.

Limit setting range: 0000 - 1999 for each limit switch.

Comparison output: Visual, relay contact, and TTL level.

Visual: 3 LED's indicate HIGH(red), IN (green), or LOW (red).

Relay contacts:

SPST contacts to circuit common for each HIGH, IN and LOW output. TTL level:

Open collector circuits to high level (open) for each HIGH, IN and LOW outputs (fanout max. 30mA).

Option 101: HP-IB data output & remote control.

Remotely controllable functions:

Function (L, C, R/ESR, △LCR)

Loss (D, Q)

LCR range

DQ range

Circuit mode

Test frequency & level

Trigger

Self test

Data output: C - D/Q, L - D/Q, R/ESR Internal function allowable subsets:

SH1, AH1, T5, L4, RL1, DCl, SRl and DTl.

Data output format: Either of two formats may be selected. Switchable at rear panel (no + sign outputs).

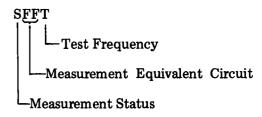
Format A.

SFFT±N.NNNE+NN, SF±N.NNCR(LF)

#### Format B.

SFFT±N.NNNE±NN(CR)(LF)

SF±N.NN(CR)(LF)



SF
Loss measurement D or Q
Measurement status

Option 010: 100Hz test frequency instead of 120Hz.

#### ACCESSORIES AVAILABLE

16061A: Test fixture, direct coupled, 5-terminal Two kinds of inserts are included for components with either axial or radial leads. Usable on all ranges of 4262A.

16062A: Test cable with alligator clips, 4-terminal. Useable for low impedance measurements. Measurement range at 1kHz is L  $\leq$  2H, C  $\geq$  200nF and R  $\leq$  10k $\Omega$ . [For L and C measurements, these ranges increase by x10 at 120 (100)Hz and decrease by same factor at 10kHz].

16063A: Test cable with alligator clips, 3-terminal. Useable for high impedance measurements. Measurement range at 1kHz is  $L \ge 3$ mH,  $C \le 10\mu$ F and  $R \ge 200\Omega$ . [For L and C measurement, these ranges increase by x10 at 120(100)Hz and decrease by same factor at 10kHz].

Table 1-2. General Information.

## Measurement Times (typical):

For a 1000 count measurement on a low loss component on a fixed range:

 Test Frequency
 Function
 Meas. Time

 1kHz, 10kHz
 C/L
 220-260ms

 R
 120-160ms

 120(100)Hz
 C/L
 900ms

 R
 700mS

When autorange is selected the following times per range step must be added to the above times:

1kHz, 10kHz 45ms/180ms 120(100)Hz 150ms/670ms

When U-CL is displayed, the faster ranging time is selected.

### Reading Rate:

Internal - Approx. 30ms between end of measurement and start of next cycle.

External - Measurement cycle is initiated by external trigger input.

## High D Factor Accuracies:

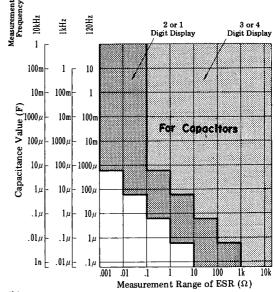
Typical

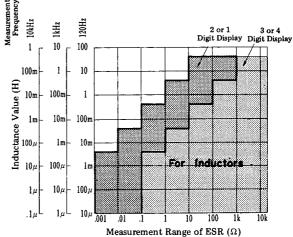
 $(\geq 2, \text{ on } 10.00 \text{ range}).$ 

Circuit Mode	Accuracy
~~~~~	5% + (2 + 1000/Cx)
o-1⊢-Wo	5% + (5 + Cx/500)
٠٤٨٨٠	5% + (5 + Lx/500)
oww.	5% + (3 + 200/Lx)

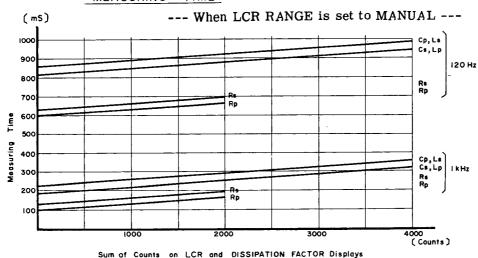
### ESR (Equivalent Series Resistance) Measurement:

Following tables show ESR measurement range for capacitors and inductors.









1kHz and 10kHz and about 900 milliseconds at 120Hz. The wide range capability of the 4262A enables a measurement range from small capacitances such as mica capacitors and the parasitic capacitance of a semiconductor device through high capacitances such as the measurement of electrolytic capacitors to be covered. A wide range of inductance measurements from the inductance of a high frequency transformer to that of a power transformer can be measured. The wide resistance range permits the measurement of wirewound resistors through the measurement of solid resistors. In parallel capacitance measurements, either a test signal level of 1Vrms, or 50mVrms can be selected.

1-7. The 4262A has the capability of making capacitance, inductance, and resistance deviation measurements. This function is enabled by pushing the  $\Delta$  LCR switch to display the deviation of a reference value. When the  $\Delta$  LCR switch is depressed the reference value is obtained and memorized from the preceding measurement. The practical use of this feature is evident when it is desired to make a measurement on a variable capacitor: First, the minimum value is measured, then the  $\triangle$ LCR button is pushed. Minimum to maximum capacitance is now displayed as the capacitor is rotated through its range. For parallel capacitance measurements, test signal levels of either 1Vrms or 50mVrms may be selected. Other versatile 4262A capabilities and features are, for example, the use of internal and external dc bias voltages, LC zero adjustment, and options providing BCD output, HP-IB interfacing capability, or a comparator function.

#### 1-8. SPECIFICATIONS.

1-9. Complete specifications of the Model 4262A LCR Meter are given in Table 1-1. These specifications are the performance standards or limits against which the instrument is tested. The test procedures for the specifications are covered in Section IV Performance Tests. Table 1-2 lists gen-

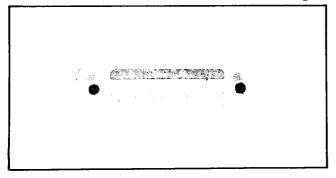


Figure 1-2. Serial Number Plate.

eral information. General information is not specifications but is typical characteristics included as additional information for the operator. When the 4262A LCR Meter is shipped from the factory, it meets the specifications listed in Table 1-1.

#### 1-10. SAFETY CONSIDERATIONS.

- 1-11. The Model 4262A LCR Meter has been designed to conform to the safety requirements of an IEC (International Electromechanical Committee) Safety Class I instrument and is shipped from the factory in a safe condition.
- 1-12. This operating and service manual contains information, cautions, and warnings which must be followed by the user to ensure safe operation and to maintain the instrument in a safe condition.

## 1-13. INSTRUMENTS COVERED BY MANUAL.

- 1-14. Hewlett-Packard uses a two-section nine character serial number which is marked on the serial number plate (Figure 1-2) attached to the instrument rear panel. The first four digits and the letter are the serial prefix and the last five digits are the suffix. The letter placed between the two sections identifies country where instrument was manufactured. The prefix is the same for all identical instruments; it changes only when a change is made to the instrument. The suffix, however, is assigned sequentially and is different for each instrument. The contents of this manual apply to instruments with the serial number prefix(es) listed under SERIAL NUMBERS on the title page.
- 1-15. An instrument manufactured after the printing of this manual may have a serial number prefix that is not listed on the title page. This unlisted serial number prefix indicates the instrument is different from those described in this manual. The manual for this new instrument may be accompanied by a yellow Manual Changes supplement or have a different manual part number. This supplement contains "change information" that explains how to adapt the manual to the newer instrument.
- 1-16. In addition to change information, the supplement may contain information for correcting errors (called Errata) in the manual. To keep this manual as current and accurate as possible, Hewlett-Packard recommends that you periodically request the latest Manual Changes supplement. The supplement for this manual is identified with this manual's print date and part number, both of which appear on the manual's title page. Complimentary copies of the supplement are available from Hewlett-Packard. If the serial prefix or number of an instrument is lower than that on title page of this manual, see Section VII Manual Changes.

1-17. For information concerning a serial number prefix that is not listed on the title page or in the Manual Changes supplement, contact your nearest Hewlett-Packard office.

#### 1-18. **OPTIONS.**

1-19. Options for the Model 4262A LCR Meter are available for adding the following capabilities:

Option 001: BCD Parallel Data Output.

Option 004: Comparator. A comparator function providing GO/NO-GO judgement with HIGH and LOW limits for LCR and D/Q.

Option 101: HP-IB Interface.

Option 010: 100Hz Test Frequency.

(instead of 120Hz)

Options 907, 908 or 909 are handle or rack mount kits. See paragraph 1-29 for details.

Option 910: Extra Manual.

#### 1-20. OPTION 001.

1-21. The 4262A option 001 provides separate BCD parallel data output for L, C, R/ESR and dissipation factor or quality factor simultaneously from the two rear panel connectors. With this option, external data processing devices such as a digital printer can be used with the 4262A.

### 1-22. OPTION 004.

1-23. The 4262A Option 004 provides for GO/NO-GO judgement by comparing L, C, R/ESR and D/Q values to HIGH and LOW limits. Three judgement outputs are provided: LED lamp display, relay contacts, or TTL level voltages (open collectors):

HIGH . .measured value is not less than HIGH limit.

IN . . . . measured value is less than HIGH limit and not less than LOW limit.

LOW ... measured value is less than LOW limit.

#### 1-24. OPTION 101.

1-25. The 4262A Option 101 provides interfacing functions to both transfer L, C, R/ESR and D/Q data to HP Interface Bus line and to receive remote control signals from HP Interface Bus line.

#### 1-26. OPTION 010.

1-27. The 4262A Option 010 provides test frequencies of 100Hz, 1kHz, and 10kHz (100Hz is used instead of standard 120Hz). All other electrical performance is the same as that of standard instrument.

#### 1-28. OTHER OPTIONS.

1-29. The following options provides mechanical parts necessary for rack mounting and hand carrying:

Option 907: Front Handle Kit. Option 908: Rack Flange Kit.

Option 909: Rack Flange and Front Handle

Kit.

The installation procedures for these options are detailed in section II.

1-30. The 4262A Option 910 provides an extra copy of the operating and service manual.

## 1-31. ACCESSORIES SUPPLIED.

1-32. Figure 1-1 shows the HP Model 4262A LCR Meter, power cord (HP Part No. 8120-1378), and fuses (HP Part No. 2110-0007 and 2110-0202).

## 1-33. ACCESSORIES AVAILABLE.

1-34. For effective and easy measurement, three styles of fixtures and leads for the measurement of various components are available. These are listed in Table 1-1. A brief description of each of these fixtures and leads is given in Table 1-3. Refer to Section III Figure 3-3 on page 3-8 for detailed information on these devices.

Table 1-3. Accessories Available.

Table 1-3. Access	Description
Model	Description
HP 16061A	Test Fixture (direct coupled type) for general measurement of both axial and vertical lead components.
HP 16062A	Test Leads (with alligator clips) useful for low inductance, high capacitance or low resistance (less than 10kΩ) measurements.
HP 16063A	Test Leads (with alligator clips) for general component measurement and especially useful for high impedance measurements.
HP P/N 5060-4017	Extender Board used for <b>4262A</b> troubleshooting.

Table 1-4. Recommended Test Equipment.

Instrument	Critical Specifications	Recommended Model	*Use
Frequency Counter	Frequency Range: 40Hz to 10kHz Sensitivity: 50mVrms min.	HP 5300A/ w 5306A	P
Capacitance Standard (See para. 4-3)	Capacitance Values: 100pF, 1000pF, 10nF, 100nF, 1000nF and 10µF	GR Type 1413 GR Type 1417	P, A
Resistance Standard (See para. 4-3)	Resistance Values: $1 k\Omega$ , $10 k\Omega$ , $100 k\Omega$ and $10 M\Omega$	GR Type 1443-Y	P, A
Inductance Standard (See Para. 4-3)	Inductance Value: 100mH	GR Type 1482-L	P
DC Voltmeter	Voltage Range: 1V to 10V Sensitivity: 10mV min.	HP 5300A/ w 5306A	P, A
Oscilloscope	Bandwidth: 10MHz min. Vertical Sensitivity: 5mV/div. Horizontal Sweep Rate: 1µs/div.	HP 180C/ w 1801A/ w 1821A	A, T
Signature Analyzer		HP 5004A	Т
Current Tracer		HP 547A	Т
Service Kit	Signature Analysis Test Board	HP P/N: 04262-87002	Т
DUT Box	Comprises L, C and R components whose values are calibrated at 120Hz and 1kHz.	HP 16361A	P, A
DUT Box  Comprises L, C and R components whose values are calibrated at 10kHz.  HP 16362A		P, A	
*P=Performan	ace Test A=Adjustments T=Troubleshootin	ng	<u> </u>

# SECTION II

#### 2-1. INTRODUCTION.

2-2. This section provides installation instructions for the Model 4262A LCR Meter. The section also includes information on initial inspection and damage claims, preparation for using the 4262A, packaging, storage, and shipment.

#### 2-3. INITIAL INSPECTION.

2-4. The 4262A LCR Meter, as shipped from the factory, meets all the specifications listed in Table 1-1. On receipt, inspect the shipping container for damage. If the shipping container or cushioning material is damaged, notify the carrier as well as the Hewlett-Packard office and be sure to keep the shipping materials for carrier's inspection until the contents of the shipment have been checked for completeness and the instrument has been checked mechanically and electrically. The contents of the shipment should be as shown in Figure 1-1. The procedures for checking the general electrical operation are given in Section III (Paragraph 3-5 Basic Operating Check) and the procedures for checking the 4262A LCR Meter against its specifications are given in Section IV. Firstly, do the self test. If the 4262A LCR Meter is electrically questionable, then do the Performance Tests to determine whether the 4262A has failed or not. If contents are incomplete, if there is mechanical damage or defects (scratches, dents, broken switches, etc.), or if the performance does not meet the self test or performance tests, notify the nearest Hewlett-Packard office (see list at back of this manual). The HP office will arrange for repair or replacement without waiting for claim settlement.

#### 2-5. PREPARATION FOR USE.

#### 2-6. POWER REQUIREMENTS.

2-7. The 4262A requires a power source of 100, 120, 220 Volts ac  $\pm 10\%$ , or 240 Volts ac  $\pm 5\%$ , -10%, 48 to 66Hz single phase. Power consumption is approximately 55 watts.

#### WARNING

IF THIS INSTRUMENT IS TO BE ENERGIZED VIA AN EXTERNAL AUTOTRANSFORMER FOR VOLTAGE REDUCTION, BE SURE THAT THE COMMON TERMINAL IS CONNECTED TO THE NEUTRAL POLE OF THE POWER SUPPLY.

## 2-8. LINE VOLTAGE AND FUSE SELECTION.

#### CAUTION

BEFORE TURNING THE 4262A LINE SWITCH TO ON, VERIFY THAT THE INSTRUMENT IS SET TO THE VOLTAGE OF THE POWER SUPPLIED.

2-9. Figure 2-1 provides instructions for line voltage and fuse selection. The line voltage selection card and the proper fuse are factory installed for the voltage appropriate to instrument destination.

#### **CAUTION**

USE PROPER FUSE FOR LINE VOLTAGE SELECTED.

#### CAUTION

MAKE SURE THAT ONLY FUSES FOR THE REQUIRED RATED CURRENT AND OF THE SPECIFIED TYPE ARE USED FOR REPLACEMENT. THE USE OF MENDED FUSES AND THE SHORT-CIRCUITING OF FUSEHOLDERS MUST BE AVOIDED.

### 2-10. POWER CABLE.

2-11. To protect operating personnel, the

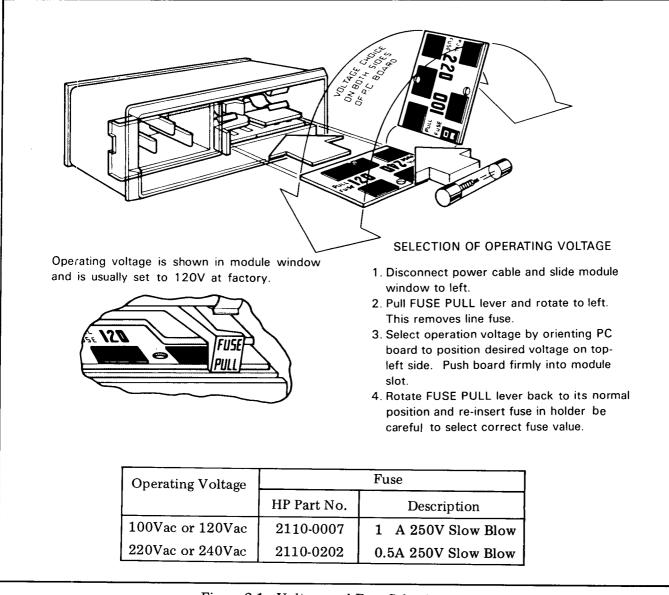


Figure 2-1. Voltage and Fuse Selection.

National Electrical Manufacturer's Association (NEMA) recommends that the instrument panel and cabinet be grounded. The Model 4262A is equipped with a three-conductor power cable which, when plugged into an appropriate receptacle, grounds the instrument. The offset pin on the power cable is the ground wire.

2-12. To preserve the protection feature when operating the instrument from a two contact outlet, use a three prong to two prong adapter (HP Part No. 1251-8196) and connect the green grounding tab on the adapter to power line ground.

## **CAUTION**

THE MAINS PLUG MUST ONLY BE INSERTED IN A SOCKET OUTLET PROVIDED WITH A PROTECTIVE EARTH CONTACT. THE PROTECTIVE ACTION MUST NOT BE NEGATED BY THE USE OF AN EXTENSION CORD (POWER CABLE) WITHOUT PROTECTIVE CONDUCTOR (GROUNDING).

2-13. Figure 2-2 shows the available power cords, which may be used in various countries including the standard power cord furnished with the instrument. HP Part number, applicable standards for power plug, power cord color, electrical characteristics and countries using each power cord are listed in the figure. If assistance is needed for selecting the correct power cable, contact nearest Hewlett-Packard office.

#### 2-14. Interconnections.

2-15. When an external bias is applied to the sample capacitor through DC BIAS input connectors on the 4262A rear panel, both plus and minus sides of the external power supply should be connected to the plus and minus sides of the 4262A EXT DC BIAS connector, respectively.

#### **CAUTION**

THE MAINS PLUG MUST BE INSERTED BEFORE EXTERNAL CONNECTIONS ARE MADE TO MEASURING AND/OR CONTROL CIRCUITS.

## 2-16. Operating Environment.

- 2-17. Temperature. The instrument may be operated in temperatures from 0°C to +55°C.
- 2-18. Humidity. The instrument may be operated in environments with relative humidities to 95% to 40°C. However, the instrument should be protected from temperature extremes which cause condensation within the instrument.

#### 2-19. Installation Instructions.

2-20. The HP Model 4262A can be operated on the bench or in a rack mount. The 4262A is ready for bench operation as shipped from the factory. For bench operation a two-leg instrument stand is used. For use, the instrument stands are designed to be pulled towards the front of instrument.

## 2-21. Installation of Options 907, 908 and 909.

2-22. The 4262A can be installed in a rack and be operated as a component of a measurement system. Rack mounting information for the 4262A is presented in Figure 2-3.

## 2-23. STORAGE AND SHIPMENT.

#### 2-24. Environment.

2-25. The instrument may be stored or shipped in environments within the following limits:

Temperature.			-4	$0^{\circ}$ C	to +75°C
Humidity	 				to 95%
Altitude					50,000ft

The instrument should be protected from temperature extremes which cause condensation inside the instrument.

## 2-26. Packaging.

- 2-27. Original Packaging. Containers and materials identical to those used in factory packaging are available through Hewlett-Packard offices. If the instrument is being returned to Hewlett-Packard for servicing, attach a tag indicating the type of service required, return address, model number, and full serial number. Also mark the container FRAGILE to assure careful handling. In any correspondence, refer to the instrument by model number and full serial number.
- 2-28. Other Packaging. The following general instructions should be used for re-packing with commercially available materials:
  - a. Wrap instrument in heavy paper or plastic. If shipping to Hewlett-Packard office or service center, attach tag indicating type of service required, return address, model number, and full serial number.
  - b. Use strong shipping container. A double-wall carton made of 350 pound test material is adequate.
  - c. Use enough shock absorbing material (3 to 4 inch layer) around all sides of instrument to provide firm cushion and prevent movement inside container. Protect control panel with cardboard.
  - d. Seal shipping container securely.
  - e. Mark shipping container FRAGILE to ensure careful handling.
  - f. In any correspondence, refer to instrument by model number and full serial number.

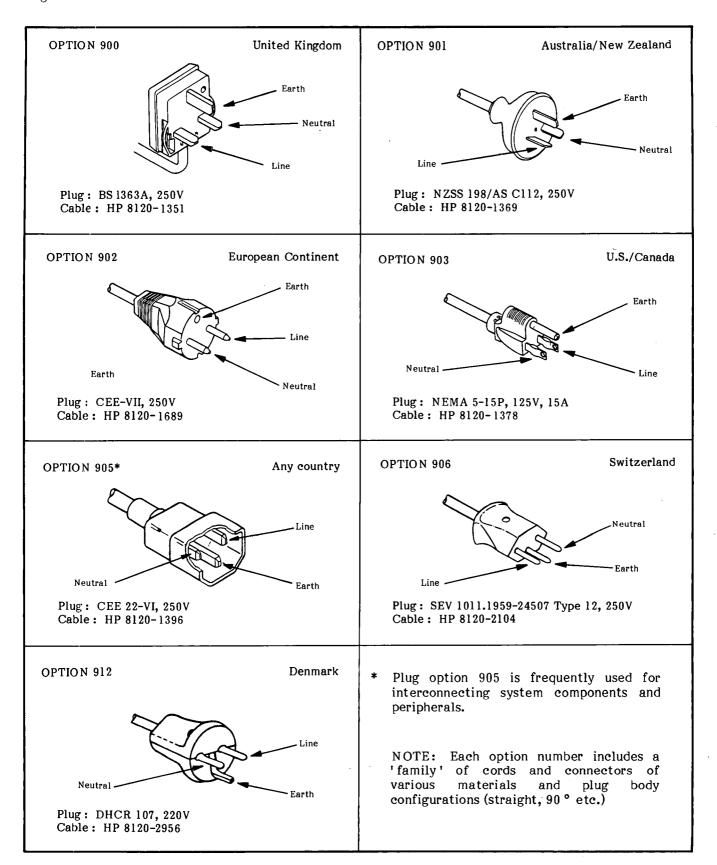
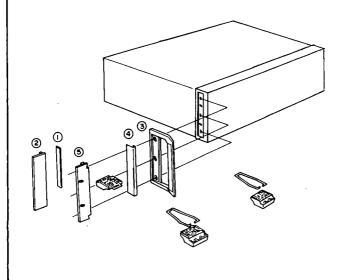


Figure 2-2. Power Cables Supplied.

Option	Kit Part Number	Parts Included	Part Number	Q'ty	Remarks
907	Handle Kit 5061-0089	Front Handle Trim Strip #8-32 x 3/8 Screw	3 5060-9899 4 5060-8896 2510-0195	2 2 6	9.525mm
908	Rack Flange Kit 5061-0077	Rack Mount Flange #8-32 x 3/8 Screw	② 5020-8862 2510-0193	2 6	9.525mm
909	Rack Flange & Handle Kit 5061-0083	Front Handle Rack Mount Flange #8-32 x 3/8 Screw	③ 5060-9899 ⑤ 5020-8874 2510-0194	2 2 6	15.875mm



- 1. Remove adhesive-backed trim strips (1) from side at right and left front of instrument.
- 2. HANDLE INSTALLATION: Attach front handle (3) to sides at right and left front of instrument with screws provided and attach trim (4) to handle.
- 3. RACK MOUNTING: Attach rack mount flange (2) to sides at right and left front of instrument with screws provided.
- 4. HANDLE AND RACK MOUNTING: Attach front handle 3 and rack mount flange 5 together to sides at right and left front of instrument with screws provided.
- 5. When rack mounting (3 and 4 above), remove all four feet (lift bar at inner side of foot, and slide foot toward the bar).

Figure 2-3. Rack Mount Kit

Section II Paragraphs 2-29 to 2-34

#### 2-29. OPTION INSTALLATION.

2-30. When it is desired to add one or two of the available optional features to a standard 4262A instrument, perform the installation as follows:

Refer to option installation illustrations on facing page.

- a. Push LINE switch to off.
- b. Remove instrument top cover.
- c. Follow the appropriate paragraph below.
- 2-31. OPTION 001 BCD DATA OUTPUT INSTALLATION.
  - a. Remove the left side middle and lower blind covers from the rear panel.
  - b. Install two 50-pin connector assemblies in the openings.
  - c. Set BCD switch of SW1 on A23 board assembly (RED/ORANGE GUIDE, P/N: 04262-66523 or 04262-66623) from OFF to opposite position. This board is located third from front on the right side.
  - d. Connect cable attached to A23 board (shown below) between A23 and A35 BCD Option board assemblies (P/N: 04262-66535). Install A35 in RED/GREEN GUIDE option receptacle.
  - e. Plug 2 each flat cable assemblies from A35 BCD Option board into connector boards of rear panel connector assemblies.
  - f. Install instrument top cover.

## 2-32. OPTION 004 COMPARATOR INSTALLATION.

Refer to Fig 2-4 for installation procedure.

- 2-33. COUPLING OPTION 004 COMPARATOR WITH OPTION 001 BCD DATA OUTPUT INSTALLATION.
  - a. Set CMP (comparator) and BCD option switches of SW1 ON A23 board assemblies (RED/ORANGE GUIDE, P/N: 04262-66523 or 04262-66623) from OFF to opposite position. This board is located third from front on the right side.
  - b. Connect cables attached to A23 board between A23 and A24 comparator option BCD board assembly. No other cable assembly change is necessary for this combination of options.
  - c. Refer to Paragraphs 2-31 and 2-32 for other installation procedures.
- 2-34. OPTION 101 HP-IB REMOTE CONTROL AND DATA OUTPUT INSTALLATION.
  - a. Remove right side blind covers from rear panel.
  - b. Install connector board assembly (P/N: 04262-66503) in the opening and mount with washers and nuts included with assembly.
  - c. Set the HP-IB switch of SW1 on A23 board assembly from OFF to opposite position. The A23 board is located on the right side third from front.
  - d. Connect cable assembly attached to A25 board between A23 and A25 HP-IB option board assemblies (P/N: 04262-66525). Install A25 in RED/GREEN GUIDE option receptacle.
  - e. Plug flat cable assembly from connector board assembly P/N: 04262-66503 into A25 board assembly (installed in RED/GREEN GUIDE receptacle).

OPTION 101 IS NOT COMPATIBLE WITH OPTIONS 001 AND 004.

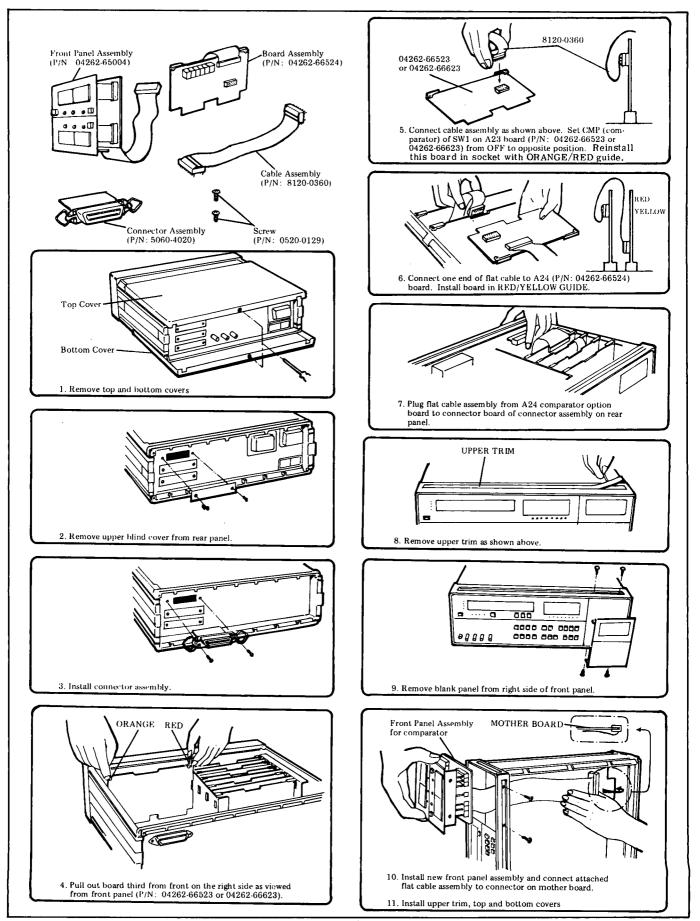


Figure 2-4. Option Installation Illustrations.

Table 2-1. Option Components

			Comp	ponents
Option	Function	HP Part No.	Q'ty	Description
001	BCD Data Output	04262-66535 5060-4020 8120-0360	1 2 1	A35 Board Assembly Connector Board Assembly Flat Cable Assembly
004	Comparator	04262-66544 04262-66505 04262-66524 3100-1201 5060-4020 8120-0360 04262-24003	1 1 1 2 1 1	A4 Board Assembly A5 Board Assembly A24 Board Assembly Thumbwheel Switch Connector Board Assembly Flat Cable Assembly Standoff
010	100Hz Test Frequency	04262-66911 04262-66914	1 1	All Board Assembly Al4 Board Assembly
101	HP-IB ·	04262-66525 04262-66503 8120-0360 0380-0644	1 1 1 2	A25 Board Assembly A3 Board Assembly Flat Cable Assembly Stud for A3 Board Assemby

Note: To mount Connector Board assemblies, use rear panel blank plate retaining screws (Part No. 0520-0129) removed for the option installation.

# SECTION III OPERATION

#### 3-1. INTRODUCTION.

3-2. This section provides the operating information to acquaint the user with the 4262A LCR Meter. Basic product features and characteristics, measurement procedures for various applications, an operational check of the fundamental electrical functions, and operator maintenance information is presented in this section. Operating cautions throughout the text should be carefully observed.

#### 3-3. PANEL FEATURES.

3-4. Front and rear panel features for the 4262A are described in Figures 3-1 and 3-2. Description numbers match the numbers on the photographs. Other detailed information for panel displays and controls are covered in the Operating Instructions (paragraph 3-7).

### 3-5. SELF TEST (Basic Operating Check).

## WARNING

ANY INTERRUPTION OF THE PROTECTIVE GROUNDING CONDUCTOR INSIDE OR OUTSIDE THE INSTRUMENT OR DISCONNECTION OF THE PROTECTIVE EARTH TERMINAL IS LIKELY TO CAUSE THE INSTRUMENT TO BE DANGEROUS. INTENTIONAL INTERRUPTION IS PROHIBITED.

## WARNING

WHENEVER IT IS LIKELY THAT THE PROTECTION OFFERED BY FUSES HAS BEEN IMPAIRED, THE INSTRUMENT MUST BE MADE INOPERATIVE AND BE SECURED AGAINST ANY UNINTENDED OPERATION.

## CAUTION

BEFORE ANY OTHER CONNECTION IS MADE, THE PROTECTIVE EARTH TERMINAL MUST BE CONNECTED TO A PROTECTIVE GROUNDING CONDUCTOR.

3-6. Functional operation of the Model 4262A should be confirmed by the SELF TEST switch before measuring samples of interest. This test can

be done under all conditions of FUNCTION and TEST SIGNAL settings. Tests under certain combined conditions of FUNCTION and TEST SIGNAL settings are done for five ranges. A test for a range ends with a display of PASS (normal operation) or FAIL (abnormal operation) and then next range test is started. Range shifting for this test is done automatically from lower to higher.

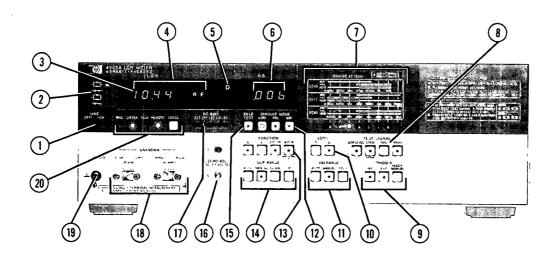


All the combinations of FUNCTION and TEST SIGNAL switch settings are listed below. Even if the FUNCTION or TEST SIGNAL switch settings are limited for proposed sample measurement, all combined conditions should be tested.

Pushbutton Switch Setting *	UNKNOWN** Connectors
(C), (120Hz), (SELF TEST)*** (C), (1kHz), (SELF TEST) (C), (10kHz), (SELF TEST) (C), (LOW LEVEL), (10kHz), (SELF TEST) (C), (LOW LEVEL), (1kHz), (SELF TEST) (C), (LOW LEVEL), (120Hz), (SELF TEST)	Open between HIGH side and Low side
(L), (120 Hz), (SELF TEST) (L), (1 kHz), (SELF TEST) (L), (10 kHz), (SELF TEST) (R/ESR), (10 kHz), (SELF TEST) (R/ESR), (1 kHz), (SELF TEST) (R/ESR), (120 Hz), (SELF TEST)	Short between HIGH side and LOW side.

\* When FUNCTION or TEST SIGNALS switch setting is changed, the SELF TEST switch is automatically disabled. Therefore, whenever a new setting is made, push the SELF TEST switch again.

For \*\* see page 3-5



- ① LINE ON/OFF switch: Turns instrument on and readies instrument for measurement
- 2 Circuit Mode Indicator: LED lamp, next to equivalent measuring circuit being used, lights. Sample connected to UNKNOWN terminals (18) is measured in an equivalent circuit selected by FUNCTION (13) and CIRCUIT MODE (12) switches and is indicated by appropriate LED lamp. Equivalent circuits are shown as electronic circuit symbols at the left of indicator lamps. Desired circuit parameter of component is measured in one of the following selected circuit modes:

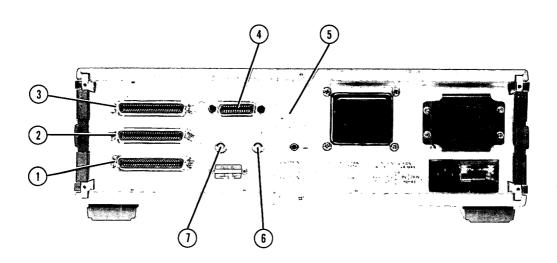
•	
Parallel capacitance Parallel resistance	
Series capacitance	
Series resistance	<del>-11-w-</del>
Parallel inductance	-[
Series inductance	
Series resistance	-30-W-

Trigger Lamp: Turns on during sample measuring period. Turns off during period when instrument is not taking measurement (or hold period). There is one turnon-and-off cycle per measurement. This lamp turns on and off repeatedly when TRIGGER (9) is set to INT.

- 4 LCR Display: Inductance, capacitance or resistance value including the decimal point and unit is displayed in 3-½ digit decimal number from 0000 to 1999. If the sample value exceeds 1999 in a selected range, O-F(Over-Flow) appears in this display. This display also shows PASS or FAIL when SELF TEST is performed.
- 5 D/Q Indicator: In a capacitance or inductance measurement, this indicator indicates which of D (dissipation factor) or Q (quality factor) is displayed in D/Q display 6. In resistance measurement, this indicator is also lit (however, D or Q indication has no meaning and D/Q display 6 is left blank).
- 6 D/Q Display: Value for dissipation factor or quality factor is displayed in capacitance and/or inductance measurement. In resistance measurement, this display is kept blank.
- (1) RANGE Indicator: The range automatically or manually selected is indicated by LED lamp. The table printed above the LED array shows the measurement ranges of the Model 4262A.
- (8) TEST SIGNAL These pushbuttons enable selection of measurement frequency—120Hz, 1kHz or 10kHz and that of low test voltage of the signal applied to sample to be tested. LOW LEVEL switch is effective only in parallel capacitance measurements, supplying a test voltage of 50mVrms. For units equipped with option 010, arrow on pushbutton (120Hz) points to 100Hz.

- TRIGGER: These pushbuttons select trigger mode, INT, EXT or HOLD/MANUAL. INT key provides internal trigger which enables instrument to make repeated automatic measurements. In external trigger mode (EXT), trigger signal should be applied to either of following two connectors: (1) EXT TRIGGER input connector on the rear panel (2) 50 pin connector of Option 001 or 004 on the rear panel. HOLD/MANUAL trigger mode provides trigger signal for one measurement cycle when this key is depressed.
- (10) LOSS: These pushbuttons select whether D or Q value is displayed in the D/Q display (6) in capacitance or inductance measurements.
- D/Q RANGE: These pushbuttons select ranging method for loss measurement. AUTO: Optimum D/Q range is selected by internal logic circuit. MANUAL: D/Q range is fixed to a range. Range change is done by depressing the STEP key on the right.
- (12) CIRCUIT MODE: Appropriate circuit mode for taking a measurement is selected and set with these pushbuttons. A parallel equivalent circuit is selected by PRL key and series equivalent circuit by SER key. When AUTO key is pushed, the instrument automatically selects the appropriate parallel or series equivalent circuit.
- (3) FUNCTION: These pushbuttons select electrical circuit parameter to be measured as follows:
  - C: Capacitance together with dissipation factor (D) or quality factor (Q).
  - L:Inductance with dissipation factor (D) or quality factor (Q).
  - R/ESR: Resistance or Equivalent Series Resistance.
  - △LCR: Difference in L, C, or R value between the value of the sample under test and the internally stored value obtained by a measurement just before △LCR key is depressed.
  - (4) LCR RANGE: These pushbuttons select ranging method for LCR measurement.
    - AUTO: Optimum range for the sample value is automatically selected.

- MANUAL: Measurement range is fixed (even when the sample connected to the UNKNOWN terminals is changed). Range change is done by depressing DOWN or UP key on the right.
- (15) SELF TEST: This pushbutton performs automatic check for checking the basic operation of Model 4262A. If normal operation is confirmed, "PASS" is displayed in LCR display 4. If wrong performance is detected, a display of "FAIL" appears. See paragraph 3-5 for details.
- (b) ZERO Adjustment Controls: These adjustments provide proper compensation for cancelling stray capacitance and residual inductance which are present when a test fixture is mounted on the UNKNOWN terminals. Connectors are kept open for cancelling stray capacitance and shorted for cancelling residual inductance.
- ① DC BIAS Selector Switch: This switch permits selection of internal DC bias voltage applied to sample (1.5Vdc, 2.2Vdc, or 6.0Vdc). When switch is set to EXT, it is used to apply external bias voltage from rear DC BIAS input connectors. OFF position is selected if no bias voltage is necessary.
- (B) UNKNOWN Terminals: Consist of four terminals: High current terminal (Hcur), High potential terminal (Hpot), Low potential terminal (Lpot) and Low current terminal (Lcur). A five-terminal configuration is constructed by adding the GUARD terminal (9). A three-terminal configuration is constructed by shorting High terminals and Low terminals together with shorting bars. Under DC Bias operation, the high terminals have a positive DC voltage with respect to LOW terminals.
- (9) GUARD Terminal: This is connected to chassis ground of instrument and can be used as Guard terminal for increasing accuracy in certain measurements.
- (20) HP-IB Status Indicator and LOCAL switch. LED lamps for SRQ, LISTEN, TALK, and REMOTE which indicate status of interface between the 4262A (Option 101) and HP-IB controller. LOCAL switch enables front panel controls instead of remote control signals from HP-IB line.



- (1) BCD D/Q DATA OUTPUT Connector: BCD parallel data of measured dissipation factor (D) or quality factor (Q) are outputted through this 50 pin connector installed on the 4262A Option 001.
- ② BCD LCR DATA OUTPUT Connector: With Option 001, BCD parallel data for inductance, capacitance and resistance measured values are outputted through this 50 pin connector.
- 3 COMPARATOR OUTPUT Connector: The 4262A Option 004 provides comparator decision outputs for LCR and D/Q through this 50 pin connector.
- 4 HP-IB Digital Bus Connector: This 24 pin connector conveys bus signals and remote programming instructions to the 4262A Option 101 and transmits data from the 4262A Option 101 to the bus.

- Address Switch: This seven section switch sets address code of 4262A Option 101 and TALK ONLY or ADDRESSABLE mode of operation.
- 6 EXT DC BIAS Connector: External dc bias voltage can be applied to the sample up to the maximum voltage of plus 40V through this connector.
- (i) EXT TRIGGER Connector: This connector is used for externally triggering the instrument by inputting an external trigger signal. TRIGGER SWITCH on front panel should be set to EXT.

Model 4262A Section III

\*\* Two HIGH side terminals and two LOW side terminals should be connected with the shorting strap, for each configuration of the UNKNOWN terminals. When the UNKNOWN terminal configuration is not appropriate, for example, shorted (C) or open (L), display will show FAIL 1 (because they result from different causes, FAIL 2 or FAIL 3 are rarely displayed).



\*\*\* Setting change required is only the underlined switch setting.

If FAIL is displayed, check the UNKNOWN terminal configurations as follows:

- (1) That the two HIGH side terminals (H<sub>CUR</sub> H<sub>POT</sub>) and the two LOW side terminals (L<sub>CUR</sub> L<sub>POT</sub>) are properly shorted.
- (2) That short or open conditions properly exist between HIGH and LOW side terminals.
- (3) That GUARD terminal is isolated (open) from both of HIGH and LOW terminals.

If FAIL is still displayed (under the above condition), notify the nearest Hewlett-Packard office with information detailing which combination of settings show FAIL.

During SELF TEST, other controls are automatically set as follows:

CIRCUIT MODE	SER when FUNCTION
	is set to L or R/ESR.
	PRL when FUNCTION
	is set to C.
LOSS	D
LCR RANGE	
D/Q RANGE	
TRICORD	TNIT

#### NOTE

TO ENSURE CORRECT RESULTS OF SELF-TEST OPERATION IN L AND R MEASUREMENT FUNCTIONS, CONNECT ALL (HIGH AND LOW SIDE) UNKNOWN TERMINALS TOGETHER WITH A LOW IMPEDANCE STRAP (IF THIS SHORT-CIRCUIT IS MADE AT THE ENDS OF THE TEST LEADS, CORRECT RESULTS MAY NOT OCCUR).

### 3-7. TEST SIGNALS.

3-8. Three test signal frequencies are available: these are 120Hz, 1kHz and 10kHz sinusoidal waveforms which have a frequency accuracy of 3%. The typical voltage applied to the sample or current flowing through the sample is specified in Table 3-1 for all test signal frequencies. A constant test voltage is supplied to the sample when measuring parallel parameters Lp, Cp, and Rp. The constant current method is adopted for the measurement of Ls, Cs, and Rs. The 50mVrms test voltage is used only for Cp measurement.

### 3-9. MEASUREMENT RANGE.

3-10. As given in Table 3-2, the 4262A has wide measurement ranges. Seven or eight ranges are available (depending upon measurement function) and the appropriate range is automatically selected for the value of sample connected to the 4262A UNKNOWN terminals. For applications which require a fixed measurement range (such applications are sometimes needed, for example, in inductance measurements), manual range control is pushbutton selectable. Four or five ranges, however, are used in the series and parallel equivalent circuit measurement modes. When the CIRCUIT MODE is set to AUTO, the 4262A will automatically select the appropriate circuit mode, range over the measurement ranges shadowed in Table 3-2, settle on the proper range, and measure the sample.

Table 3-1. Sample Voltage or Current.

DANCE	CIRCUIT MODE						
RANGE	Ls	Lp	Cs	Ср	Rs	Rp	
1	40mA rms			1Vrms (50mVrms)*	40mA rms		
2	10mA rms		<del></del>	1V rms (50mV rms)*	10mA rms		
3	1mA rms	l ——		1V rms (50mV rms)*	1mA rms		
4	$100~\mu A~{ m rms}$	1V rms	10 μA rms	1Vrms (50mVrms)*	$100~\mu \text{A rms}$	1V rms	
5	$10~\mu  ext{A}~ ext{rms}$	1V rms	100 μA rms	1Vrms (50mVrms)*	$10~\mu  ext{A}$ rms	1V rms	
6		1V rms	1 μA rms			1V rms	
7		1V rms	10mA rms			1V rms	
8			40mArms			1V rms	

<sup>\*</sup>When TEST SIGNAL is set to LOW LEVEL.

Table 3-2. Measurement Ranges.

CIRCUIT	TEST				Ra	inge			
MODE	SIGNAL Frequency	1	2	3	4	5	6	7	8
Lp	120 Hz 1 kHz 10 kHz				0000 mH 000.0 mH 00.00 mH	00.00 H 0000 mH 000.0 mH	H 0.000 H 00.00 Hm0000	H 00,000	
Ls	120 Hz 1 kHz 10 kHz	Ημ 0000 Ημ 0.000 Ημ 00.00	00,00 mH Hμ 0000 Hμ 0,000	000,0 mH 00,00 mH 0000 μH	0000 mH 000.0 mH 00.00 mH	00.00 H 0000 mH 000.0 mH			
Ср	120 Hz 1 kHz 10 kHz	0000 pF 000.0 pF 00.00 pF	00.00 nF 0000 pF 000.0 pF	000.0 nF 00.00 nF 0000 pF	0000 nF 000.0 nF 00.00 nF	00.00 μF 0000 nF 000.0 nF	·		
Cs	120 Hz 1 kHz 10 kHz				0000 nF 000.0 nF 00.00 nF	00,00 μF 0000 nF 000.0 nF	000.0 μF 00.00 μF 0000 nF	-0000 μF -000.0 μF -00.00 μF	00.00mF 0000 μF 000.0 μF
Rp	120 Hz 1 kHz 10 kHz				Ω 0000 Ω 0000 Ω	00.00 kΩ 00.00 kΩ 00.00 kΩ	000.0 kΩ 000.0 kΩ 000.0 kΩ	0000 kΩ 0000 kΩ 0000 kΩ	00,00 MΩ 00,00 MΩ 00,00 MΩ
Rs	120 Hz 1 kHz 10 kHz	Ωm 0000 Ωm 0000 Ωm 0000	Ω 00,00 Ω 00,00 Ω 00,00	Ω 0.000 Ω 0.000 Ω 0.000	Ω 0000 Ω 0000 Ω	00.00 kΩ 00.00 kΩ 00.00 kΩ			

Note:  $0000\mu$ H indicates a range of  $0001\mu$ H to  $1999\mu$ H (and similarly for F and  $\Omega$ ).

### 3-11. INITIAL DISPLAY TEST.

3-12. The Model 4262A automatically performs a front panel LED display test for a few seconds after instrument is tuned on (after LINE button is depressed). The display test sequence is:

- 1. All front panel indicator lamps, except numeric segments and multiplier indicator lamps will illuminate. (SRQ, LISTEN, TALK and REMOTE lamps illuminate only when HP-IB option is installed).
- 2. Front panel pushbutton LED's and indicator lamps indicate that automatic initial settings (see Paragraph 3-13 which follows) have been set. Simultaneously, the LCR DISPLAY and DQ DISPLAY readouts are tested. All numeric displays show figures of 8 (□) and multiplier indicators (p n μ m k M) light in turn.
- 3. Range indicator lamps step from right (upper range) to left (lower range). When steps 1, 2 and 3 have been completed, the trigger lamp begins to flash. Figures on numeric displays change to meaningful numbers showing that the 4262A is ready to take a measurement.

### 3-13. INITIAL CONTROL SETTINGS.

3-14. One of the sophisticated features of the 4262A is its automatic initial control setting function. After the instrument is turned on, the front panel control functions are automatically set as follows:

SELF TESTOFF	ř
CIRCUIT MODE AUTO	)
FUNCTION	3
LCR RANGE AUTO	)
LOSS	)
DQ RANGE AUTO	)
TEST SIGNAL 1kHz	Z
TRIGGER INT	1

As these initial settings provide the general capacitance measurement conditions applicable to a broad range of capacitance measurements, a capacitance can be usually measured by merely connecting the sample to the UNKNOWN terminals. Inductance or resistance can be measured by pressing the L FUNCTION or R/ESR FUNCTION buttons, as appropriate. When a different measurement is to be attempted, press appropriate pushbuttons and select desired functions.

### 3-15. D/Q MEASUREMENT.

3-16. The Model 4262A makes a loss measurement along with capacitance or inductance measurements on each measurement cycle. The measured loss factor is displayed in the form of the dissipation (D) or quality (Q) factor of the sample. The D or Q function is pushbutton selectable in both L and C measurements. D and Q measurement ranges are:

D:	2 ranges	.001 to 1.999
	_	0.01 to 19.9
Q:	4 ranges	.050 to 1.996
	_	0.05 to 19.61
		,00.1 to 166.7
		001 to 1000

The D range, appropriate to the value of the sample is automatically selected. Alternately, a manual D range control is pushbutton selectable. Quality factor (Q) is calculated as a reciprocal dissipation number from the measured D value. Hence, the Q readout display will skip some numbers when low dissipation samples are measured. For example, when the dissipation measured is .010, the quality factor display is 100. When dissipation is .009, the quality factor reading is 111 (Q readings of 101 to 110 are not obtained). On the high D measurement range, the readout is displayed in 3 digits.

### 3-17. △LCR MEASUREMENT.

3-18. When many components of similar value are to be tested, it is sometimes more practicable to measure the difference between the value of the sample and a predetermined reference value. The △LCR function permits repetitive calculation of the difference between the reference and each individual sample and to display the result on the LCR DISPLAY. When the \(^LCR\) FUNCTION button is pressed, the inductance, capacitance, or resistance value of the sample is stored in an internal memory. The 4262A will now display the difference between the stored value and the measured value of a sample connected to UNKNOWN. The LCR RANGE is automatically held in MANUAL for the duration of \(^LCR\) measurements. (if another pushbutton is inadvertently pressed, the \$\triangle LCR\$ measurement function will be reset and will require reactivating).

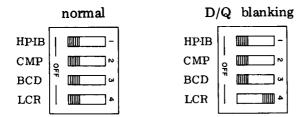
### Accessory Model Characteristics 16061A Test Fixture This fixture facilitates easy measurement of general type components with axial or vertical leads. To install fixture, disconnect shorting bars between high terminals and between low terminals. Insert fixture screws to firmly attach fixture to instrument. Two kinds of inserts are included (for components with either axial or vertical leads). DUT range (at 1kHz) pFnF $\mu F$ mΗ Η $\mu$ H 10 100 10 100 kΩ 10 100 MΩ C L Five terminal construction test fixture. R 16062A Test Leads The 16062A is especially useful when measuring low impedances. DUT values measurable with the 16062A are diagrammed below. If the measuring sample is more than approx. $300\mu F$ at 1kHz or less than approx. 100µH at 1kHz, it is recommended that the respective potential leads and current leads be twisted together. Measurable DUT ranges (at 1kHz) $\mu \mathbf{F}$ pF nF $\mu H$ mH H 100 $k\Omega$ $10 \ 100 \ M\Omega \ 10 \ 100$ Ω 10 C $\mathbf{L}$ Test Leads for four terminal measurement (does not contain guard conductor). $\mathbf{R}$ 16063A Test Leads The 16063A is particularly useful when measuring high impedances. DUT values measurable with the 16063A are diagrammed below. This test lead set is not intended to be used for the accurate measurement of small capacitances (less than approx. 100pF) due to the residual capacitance of the leads. Measurable DUT ranges (at 1kHz) nF рF $\mu F$ $\mu$ H mΗ Η $10 100 k\Omega$ 10 100 M $\Omega$ 10 100 $\mathbf{C}$ L Coaxial test leads with guard conductor for three terminal measurement. $\mathbf{R}$

Figure 3-3. Test Fixture and Leads.

# 3-19. D/Q Blanking Function (Switch selectable function inside cabinet).

3-20. The D/Q blanking function permits deactivating the D/Q measurement as desired. If operator has no need of D/Q measurement data, and alternatively desires to make higher speed LCR measurements, the switch for this function may be set. When the D/Q function is deactivated, measurement time is shortened to approximately 220 to 250 milliseconds (at 120Hz) and to 80 to 110 milliseconds (at 1kHz and 10kHz) as compared to standard measuring times (which includes a D/Q measurement). The D/Q deactivating switch is located on the A23 board assembly. To select this function, change setting of the switch as follows:

- a. Remove top cover.
- b. Take out A23 board (red and orange colored extractors).
- c. The selection switch is mounted near left edge of the A23 board.
- d. Change position of the switch as illustrated below.
- e. Reinstall the A23 board in its normal position,
- f. Replace top cover.



### 3-21. General Component Measurement.

3-22. Figure 3-7 shows the operating procedures for measuring an L, C or R (inductance, capacitance or resistance) circuit component. Almost all discrete circuit components (inductors, capacitors or resistors) except for components having special shapes or dimensions can be measured with this setup. Special components may be measured by using Test Leads 16062A or 16063A or by specially designed user built fixtures instead of 16061A Test Fixture.

### 3-23. Semiconductor Device Measurement.

3-24. The procedures for using the 4262A semiconductor device measurement capabilities are described in Figure 3-8. For example, the junction (interterminal) capacitance of diodes, collector output capacitance of transistors, etc., can easily and accurately be measured (with and without dc bias).

### 3-25. External DC Bias.

3-26. A special biasing circuit using external voltage or current bias, as needed for capacitor or inductor measurements, is illustrated in Figure 3-9. The figure shows sample circuitry appropriate to 4262A applications. Biasing circuits must avoid permitting dc current to flow into the 4262A as dc current increases the measurement error and the excess current sometimes may cause damage to the instrument. When applying a dc voltage to capacitors, be sure applied voltage does not exceed maximum working voltage and that you are observing polarity of capacitor. Note that the external bias voltage is present at Hcur and Hpot terminals.

3-27. Bias Voltage Settling Time. When a measurement with dc bias voltage superposed is performed, it takes some time for voltage across sample to reach a certain percentage of applied (desired) voltage. Figure 3-9 shows time for dc bias voltage to reach more than 99% of applied voltage and for 4262A to display a stable value. If the bias voltage across sample is not given sufficient time to settle, the displayed value may fluctuate or O-F may be displayed. Read measured value after display settles.

### 3-28. External Triggering.

3-29. For triggering the 4262A externally, connect an external triggering device to the rear panel EXT TRIGGER connector (BNC type) and press EXT TRIGGER button. The 4262A can be triggered by a TTL level signal that changes from low (0V) to high level (+5V). Triggering can be also done by alternately shorting and opening the center conductor of the EXT TRIGGER connector to ground (chassis).

### Note

The center conductor of the EXT TRIGGER connector is normally at high level (no input).

### 3-30. TERMINAL CONFIGURATION.

3-31. Connection of DUT. The 4262A Unknown terminals consists of five binding post (type) connectors:  $H_{\text{CUR}}$ ,  $H_{\text{POT}}$ ,  $L_{\text{CUR}}$ ,  $L_{\text{POT}}$  and GUARD. By connecting the stationary shorting straps to appropriate terminals, the UNKNOWN terminals can be adopted for the desired measurement terminal configuration: the two, three, four or five terminal method.

For measurements of samples having a medium order of impedance ( $100\Omega$  to  $10k\Omega$ ), the convenient two terminal method is suited to measurement requirements for good accuracy as well as for ease in connecting the sample. When converting to two terminals, shorting straps are attached to the UNKNOWN HCUR and HPOT terminals, and LCUR and LPOT terminals, respectively.

High impedance samples (greater than  $1k\Omega$ ) --which includes low capacitance, high inductance and high resistance -- should be measured by the three terminal method to eliminate the effects of stray capacitances on the measurements. For this purpose, the guard conductor of the sample is connected to the instrument GUARD terminal.

In the measurement of low impedance samples (less than  $1k\Omega$ ), efforts should be made to eliminate the effects of contact resistance, lead resistance, residual inductance and other residual parameters in the measuring apparatus. terminal configuration measurements allow stable, accurate measurement of high capacitance, low inductance and low resistance samples at minimum incremental errors in the measurement of low impedance samples. In the four terminal method, the shorting straps are disconnected to separate potential leads from current leads. Thereby, the characteristics of the sample can be precisely determined by the instrument irrespective of the various residual parameters present in the measuring signal current path. To ensure the best accuracy, the potential leads should be connected near to the sample.

The five terminal method, which adds the guard conductor to the four terminal configuration, expands the applicable measurement range into the higher impedance regions. Thus, this method covers a broad range of measurements from low to high impedance samples at the measuring frequency of the 4262A.

When test fixtures and test leads used have a shielding conductor and are designed to consider residual impedance, the measurement limitations described above for the individual terminal configurations can vary to some extent depending on the particular characteristics of the fixture and connections. Three accessories, the 16061A Test Fixture, the 16062A Test Leads, and the 16063A Test Leads are available. The characteristics of these accessories and applicable measurement ranges are outlined in Figure 3-3. These accessories make it easy to construct the desired terminal configuration.

### IMPORTANT!

FOR CERTAIN TERMINAL MEAS-UREMENT CONFIGURATIONS, THE HCUR TERMINAL MUST BE CON-NECTED TO HPOT TERMINAL AND THE LCUR TERMINAL CONNECTED TO THE LPOT TERMINAL. OTHER-WISE, THE DISPLAYS WILL HAVE NO MEANING AND THE LIFE OF THE RELAYS USED IN THE INSTRU-MENT WILL SOMETIMES BE SHORT-ENED.

### Note

The 4262A can not measure a sample which has one lead connected to earth (grounded).

### 3-32. OFFSET ADJUSTMENT.

3-33. Since test fixtures and test leads have different inherent stray capacitances and residual inductances, the measured value obtained with respect to the same sample may possibly differ depending on the test fixture (leads) used. These residual factors can be read from the 4262A display by properly terminating (short or open) the measurement terminals of the test jig. The front panel C ZERO ADJ and L ZERO ADJ controls permit compensation for these residual factors and can eliminate measurement errors due to the test jig. The capacitance or inductance readout can be set to zero for the particular test jig used with the instrument. In capacitance and inductance measurements, an incomplete offset adjustment causes two types errors:

### 1) Deviation from zero counts.

When a small capacity or a small inductance is measured, the measured capacitance (inductance) value becomes the sum of the capacitance (inductance) of sample and the stray capacitance (residual inductance) of test jig. The effects of the residual factors are:

Cm = Cx + CstLm = Lx + Lres

Where, subscripts are

m: measured value.x: value of sample.st: stray capacitance.res: residual inductance.

Both Cst and Lres cause the same measurement error and are independent of sample value. 2) Influence on high capacitance and high inductance measurements.

When a high inductance (a high capacitance) is measured, the residual factors in the test jig also contribute a measurement error. The affect of stray capacitance or residual inductance on measurement parameters are:

These measurement errors increase in proportional to the square of the test signal frequency. The effects of the residual factors can be expressed as follows:

$$Cm = \frac{Cx}{1 - \omega^2 CxLres}$$
or  $(\frac{Cm - Cx}{Cm} \approx \omega^2 CxLres)$ 

$$Lm = \frac{Lx}{1 - \omega^2 LxCst}$$
or  $(\frac{Lm - Lx}{Lm} \approx \omega^2 LxCst)$ 

In a 10kHz measurement, for the measurement error to be less than 0.1%, the product of Cx and Lres (Lx and Cst) should be less than 0.25 x  $10^{-12}$ . The relationship between the residual factors of the test jig and measurement accuracies are graphically shown in Figure 3-4.

The 4262A ZERO ADJ controls cover the following capacitance and inductance offset adjustment ranges:

C ZERO ADJ: up to 
$$10pF$$
 L ZERO ADJ: up to  $1\mu H$ 

An offset adjustment should always be performed before measurements are taken.

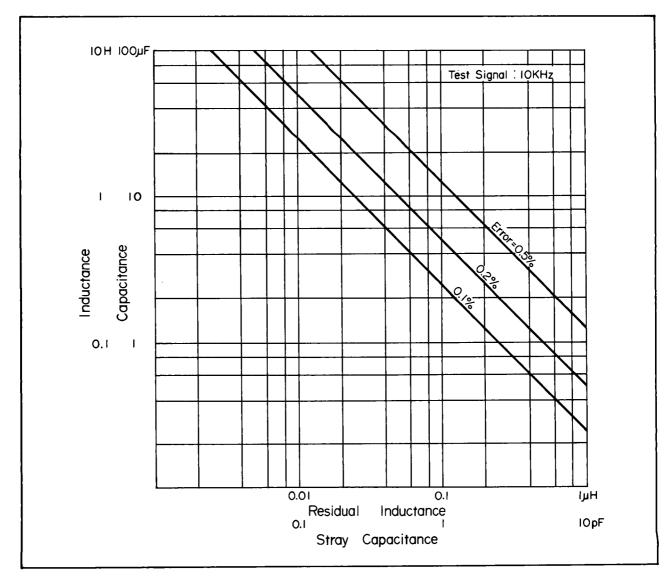


Figure 3-4. Measurement Error due to Misadjusted ZERO ADJ Controls.

### Measurement Parameter Conversions

Parameter values for a component measured in a parallel equivalent circuit and that measured in series equivalent circuit are different from each other. For example, the parallel capacitance of a given component is not equal to the series capacitance of that component. Figure A shows the relationships between parallel and series parameters for various values of D. Applicable diagrams and equations are given in the chart. For example, a parallel capacitance (Cp) of 1000pF with a dissipation factor of 0.5, is equivalent to a series capacitance (Cs) value of 1250pF at 1kHz. As shown in Figure A, inductance or capacitance values for parallel and series equivalents are almost identical when the dissipation factor is less than 0.01. The letter D in Figure A represents dissipation factor and is calculated by the equations presented in Table A for each circuit mode. The dissipation factor of a component always has the same dissipation factor at

a given frequency for both parallel equivalent and series equivalent circuits.

### Note

Dissipation factors displayed when CIRCUIT MODE is switched between PRL and SER may exhibit slight differences due to the measurement accuracy of the 4262A.

The reciprocal of the dissipation factor (D) is quality factor (Q) and D is often represented as  $\tan \delta$  which is the tangent of the dissipation angle ( $\delta$ ). Figure 3-6 is a graphical presentation of the equations in Table A. For example, a series inductance of  $1000\mu H$  which has a dissipation factor of 0.5 at 1kHz has a series resistance of 3.14 ohms.

Table A. Dissipation Factor Equations.

Circuit Mode		Dissipation Factor	Conversion to other modes	
Cp mode	Cp Rp	$D = \frac{1}{2\pi f C p R p} \left(= \frac{1}{Q}\right)$	$Cs = (1 + D^2)Cp, Rs = \frac{D^2}{1 + D^2} \cdot Rp$	
Cs mode	Cs Rs	$D = 2\pi f C s R s \left(= \frac{1}{Q}\right)$	$Cp = \frac{1}{1 + D^2} Cs, Rp = \frac{1 + D^2}{D^2} \cdot Rs$	
Lp mode	-Lp	$D = \frac{2\pi f Lp}{Rp} \ (= \frac{1}{Q})$	Ls = $\frac{1}{1 + D^2}$ Lp, Rs = $\frac{D^2}{1 + D^2}$ Rp	
Ls mode	-M-	$D = \frac{Rs}{2\pi f Ls} \ (= \frac{1}{Q})$	$Lp = (1 + D^2)Ls, Rp \approx \frac{1 + D^2}{D^2} \cdot Rs$	

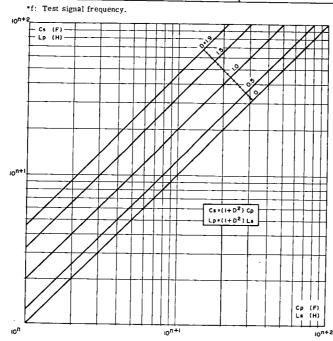


Figure A. Relationships between Parallel and Series Parameters.

Where n stands for a free integer.

Figure 3-5. Conversion Between Parallel and Series Equivalents.

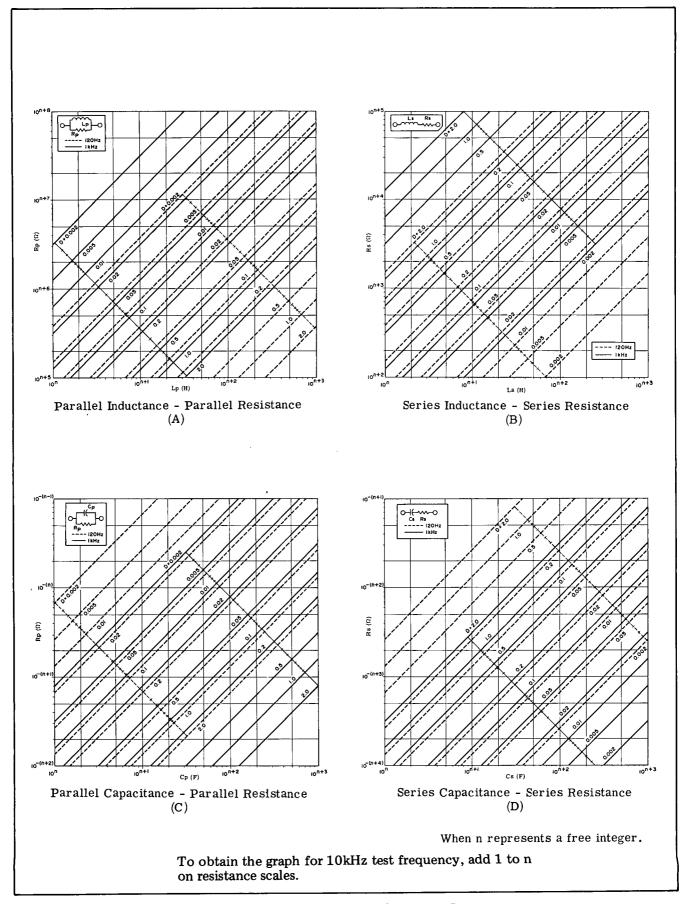
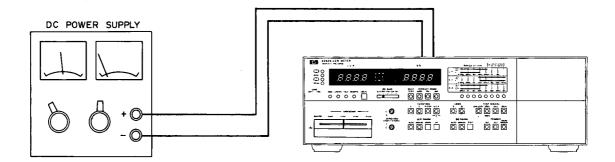


Figure 3-6. Relationship of Dissipation to Series and Parallel Resistance.

Table 3-3. Annunciation Display Meanings

DISPLAY	e 3-3. Annunciation Display Meaning  Indicated Condition	
LCR DISTRICT DQ		Action
0 - F ""	FUNCTION has been inappropriately set.	Change 4262A FUNCTION to L, C or R suitable for the sample being measured.
	Measured L or C value exceeds 1999 counts. DQ display indicates that DQ measurement has been omitted.	Set 4262A to: CIRCUIT MODE: AUTO LCR RANGE: AUTO
G - F	Measured R value exceeds 1999 counts.	Try changing TEST SIGNAL to 120, 1k or 10kHz.
(any LCR (overflowed) reading)	Measured D/Q value exceeds the upper range limit (1999 counts). Accuracy of LCR readings may not be within specifications.	Set 4262A DQ RANGE to AUTO. Try changing TEST SIGNAL to 120, 1k or 10kHz.
LI - E L	CIRCUIT MODE setting is not suitable for the sample being measured.	Set 4262A to: CIRCUIT MODE: AUTO LCR RANGE: AUTO
	Measured L, C or R value is extremely large or small compared with the selected range.	Try changing TEST SIGNAL to 120, 1k or 10kHz.
78 · (less than 80 counts)	When Measured L or C value is less than 80 counts, DQ measurement is omitted.	Set 4262A LCR RANGE to AUTO. Try changing TEST SIGNAL to 120, 1k or 10kHz.
(any DQ reading)	In ^LCR measurement, the difference between the preset value and the measured value of the sample exceeds -999 counts.	
	In ^LCR measurement, the calculated difference exceeds -999 counts. In addition, the value of measured sample is less than 80 counts.	<u> </u>
Minus (-) is displayed.	Minus display sometimes occurs when sample having a value around zero is measured.	Zero count display is meaning- ful when minus (-) display repeatedly turns on and off.
·	Sometimes a minus display occurs when a capacitor (or inductor) is measured in L (or C) FUNCTION.	Change to appropriate FUNCTION.
	Offset adjustment signal applied is too great (causes minus display).	Readjust offset signal for proper magnitude.

### MEASUREMENT PROCEDURE FOR GENERAL COMPONENTS



1. Remove shorting bar connections between high terminals and between low terminals (all terminals are now isolated from each other). Connect 16061A Test Fixture to 4262A UNKNOWN terminals.

### Note

User constructed test fixture may also be connected. Guard terminal is sometimes used in small capacitance measurements.

- 2. Depress LINE button to turn instrument on. An initial display test is automatically performed before measurement begins.
- 3. Check that 4262A trigger lamp begins to flash. The 4262A control functions are automatically set as follows (automatic initial settings):

DC BIASOF	F
SELF TESTOF	F
CIRCUIT MODE AUT	O'
FUNCTION	C
LCR RANGE AUT	O'
LOSS	D
DQ RANGE AUT	O.
TEST SIGNAL 1kH	Ιz
TRIGGER IN	1T

### Note

To check fundamental operating conditions of the instrument, perform SELF TEST (refer to Paragraph 3-5 for SELF TEST details). Press SELF TEST button again to release the function.

- 4. Rotate C ZERO ADJ control until capacitance readout is 000 counts on LCR DISPLAY (minus sign should not appear).
- 5. Connect a shorting lead to Test Fixture to short-circuit the Unknown terminals to zero ohms (zero microhenries).
- 6. Press L FUNCTION button.

Figure 3-7. General Component Measurements (Sheet 1 of 3).

7. Rotate L ZERO ADJ control until inductance readout is 000 counts on LCR DISPLAY.

### Note

To achieve more critical zero adjustments, when 10kHz test signal frequency is used, perform the capacitance and inductance zero offset adjustments (steps 4, 5, 6 and 7) at 10kHz.

- 8. Remove shorting lead from 16061A.
- 9. Select desired FUNCTION, either L, C or R/ESR.
- 10. Connect sample to be measured (L, C or R) to Test Fixture.
- 11. Model 4262A will automatically display value of unknown.

### Note

If O-F, U-CL, minus (-) or blank display occurs, see Table 3-3 for solution. Measured values for semiconductor devices are sometimes unreliable when TEST SIGNAL LOW LEVEL pushbutton is in its normal (1V) state (button lamp is not lit). In these instances, follow Figure 3-8 for semiconductor device measurement.

### Note

If manual triggering is required, press HOLD/MANUAL button. Each time the button is pressed, the instrument is triggered.

12. If internal DC bias is required, set DC BIAS switch to 1.5V, 2.2V or 6V: If not, OFF position should be selected.

### Note

DC bias application may only be used for capacitance measurements.

### CAUTION

POSITIVE POLE OF ELECTROLYTIC CAPA-CITOR MUST BE CONNECTED TO HIGH TERMINALS AS PLUS BIAS VOLTAGE IS APPLIED TO HIGH TERMINALS WITH RE-SPECT TO LOW TERMINALS.

### Note

An external bias voltage up to +40V may be applied to EXT DC BIAS rear panel connector. Connect DC power supply to EXT DC BIAS connector. Set DC BIAS switch to EXT.

### **CAUTION**

### EXTERNAL DC BIAS AT EXT BIAS CON-NECTOR MUST NEVER EXCEED +40V.

### 13. Read measured value on display.

### Note

It is usually recommended that the LCR RANGE be set to MANUAL and to hold the range when measuring multiple samples having almost the same value. Range hold operation will somewhat shorten measurement time.

### Note

Series resistance of electrolytic capacitors, inductors or transformers can be measured in series R/ESR measurement mode. In these cases, the number of digits is sometimes reduced. On the other hand, resistance can, of course, be indirectly measured with the C/L FUNCTION and calculated from one of the following equations:

Rs = D/ $\omega$ Cs (Cs-D measurement) Rs =  $\omega$ Ls·D (Ls-D measurement) Rs =  $\omega$ Lp· $\frac{D}{1 + D^2}$  (Lp-D measurement)

The above relationships are graphically shown in Figure 3-6.

— CAUTION —

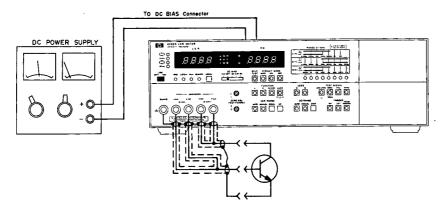
DO NOT CONNECT A CHARGED CAPACITOR (EXCEEDING 40V) DIRECTLY TO THE UNKNOWN TERMINALS AS A DUT.

– CAUTION –

NEVER APPLY A DC VOLTAGE DIRECTLY BETWEEN THE UNKNOWN H AND L TERMINALS WITHOUT PROPER PROTECTION AGAINST A POSSIBLE HARMFUL CURRENT. DC VOLTAGE MUST NOT BE APPLIED TO THE L TERMINAL WITH RESPECT TO GROUND.

Figure 3-7. General Component Measurements (Sheet 3 of 3).

### Junction Capacitance Measurement



Setup-

The figure above is a typical test setup used for measuring base-collector junction capacitance (Cob) of an NPN transistor. For this measurement, test leads or fixture may be user designed. If external DC bias is not necessary, arrangement and procedures associated with this function may be deleted from setup.

### Procedure -

1. Press LINE button to turn instrument on. After the initial display test, trigger lamp will begin to flash and the 4262A functions are automatically set as follows:

SELF TEST	OFF
CIRCUIT MODE	AUTO
FUNCTION	C
LCR RANGE	AUTO
LOSS	D
DQ RANGE	AUTO
TEST SIGNAL	1kHz
TRIGGER	INT

2. Press TEST SIGNAL LOW LEVEL and PRL CIRCUIT MODE buttons. The test signal level is now 50mV and the parallel equivalent circuit mode is selected.

### Note

A semiconductor junction capacitance measurement must be made with a low level test signal. If desired, TEST SIGNAL fequency may be set to 10kHz.

3. Adjust C ZERO ADJ control for zero counts on LCR DISPLAY.

### Note

If necessary, apply DC bias voltage internally or externally at rear panel EXT DC BIAS connector. External DC bias source should be stable with low noise. Set DC BIAS switch in EXT position during application of external DC bias.

Figure 3-8. Semiconductor Device Measurement (Sheet 1 of 2).

### **CAUTION**

## NEVER APPLY AN EXTERNAL DC BIAS OVER +40V.

4. Connect Semiconductor device to test lead or to fixture. To obtain reliable measurement results, observe the following:

### Note

- a. It is impossible to measure junction capacitance when bias current flows through sample.
- b. If lead length of device allows, it is recommended that the device be connected directly to UNKNOWN terminals.
- 5. Read displayed values. Loss factor of the sample will be simultaneously displayed on DQ DISPLAY.

### Note

When using manual trigger, press HOLD/MAN-UAL button. Each time the button is pressed, the instrument is triggered. When measuring multiple samples whose values are about the same, it is recommended that the LCR RANGE be set to MANUAL and that the range be held.

Parameter Measured	G
Farameter Measured	Connections to 4262A
Base-collector junction capacitance (Cob)- Emitter current = 0	High (+Bias)  Open High (+Bias)  Open Open
Base- collector junction capacitance (Cre)- Common emitter	High (+Blas) High (+Blas) OOK GUARD
FET gate capacitance	High (+ Bias) Copen S High (+ Bias) Copen S Open (+ Bias) Copen S Open (+ Bias) Copen S Open
Diode junction capacitance Note: Hot carrier diodes and germanium diodes sometimes cannot be measured.	High Low Low High (+ Bias)  Note: No bias should be applied.

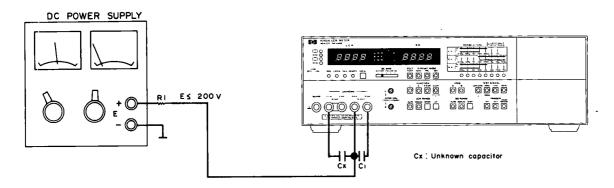
Figure 3-8. Semiconductor Device Measurement (Sheet 2 of 2).

### External DC Voltage Bias Circuits (40V \langle, \langle 200V)

1. Connect external dc bias source as shown in diagram.

### **CAUTION**

DO NOT APPLY DC VOLTAGE EXCEEDING 200VOLTS OR 4262A CIRCUITRY WILL BE DAMAGED.



### Note

+E voltage is applied to Cx in figure. -E voltage can be applied to Cx in this figure. In the above arrangement, the polarity of Cx and C1 must be taken into consideration.

### **CATUION**

NEVER SHORT BETWEEN HPOT AND LOW TERMINALS WHEN R1 IS SMALLER THAN  $1k\Omega$ . MAKE SURE THAT UNKNOWN CAPACITOR IS NOT DEFECTIVE BEFORE CONNECTING TO INSTURMENT.

TO AVOID HARMFUL SURGE CURRENT WHICH MAY FLOW THROUGH INTERNAL CIRCUITRY WHEN A HIGH VOLTAGE DC BIAS IS SUDDENLY APPLIED, IT IS RECOMMENDED THAT DC BIAS BE GRADUALLY INCREASED FROM A LOWER VOLTAGE.

### Note

Ripple or noise of external dc bias source should be as low as possible. The low frequency noise of bias source should be less than 1mVrms for a TEST SIGNAL level of 50mV (LOW LEVEL) and 30mVrms for 1V.

Figure 3-9. External DC Bias Circuit (Sheet 1 of 3).

2. Minimum values for both C1 (dc blocking capacitor) and R1 are given in table below:

### Note

Insulation resistance for Cx must be greater than a certain minimum value. Refer to Table 3-4 for unusual operating indications.

Range (at 120Hz)	1000pF	10.00nF	100.0nF	1000nF	10.00μF
Minimum C1	$0.01 \mu  \mathrm{F}$	$0.1 \mu  { m F}$	$1 \mu { m F}$	$10 \mu { m F}$	$10.00 \mu \mathrm{F}$
Minimum R1	300kΩ	100kΩ	<b>10k</b> Ω	1kΩ	100Ω

In 1kHz(10kHz) measurement, multiply both range value and value of C1 by 1/10 (1/100). If the calculated value of C1 is less than  $0.01\mu F$ , use  $0.01\mu F$  capacitor.

### Note

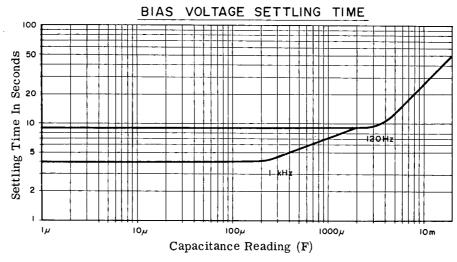
DC withstand voltage for C1 capacitor must be greater than dc applied voltage E. Also observe polarity of capacitor C1 with respect to applied voltage.

3. Set 4262A controls as follows:

SELF TEST	
FUNCTION	$\mathbf{C}$
CIRCUIT MODEPI	
Other controls any setti	ng

4. Read displayed value after allowing time for bias voltage to settle. Typical settling times are:

120Hz: 6 to 7 seconds. 1kHz/10kHz: 2 to 3 seconds.



If C1 and R1 which are larger than those given in table on above are connected, longer settling times are necessary.

Figure 3-9. External DC Bias Circuit (Sheet 2 of 3).

# Using Current Bias (for inductors). 1. Connect dc power supply as shown below: Note DC power supply should be floated from ground. If cable between low terminals of 4262A and power supply is relatively long, it should be shielded cable. The outer conductor is connected to GUARD terminal. To DC BIAS Connector DC POWER SUPPLY DC POWER SUPPLY

2. Set 4262A controls as follows:

DC BIASEXT
FUNCTIONL
CIRCUIT MODEPRL or SER
LCR RANGEMANUAL
Other controls any settings

### Note

First, determine appropriate range by connecting sample with no dc bias current applied. Then hold the range.

3. Recommended inductance ranges and maximum bias currents are:

Range (at 120Hz)	1000 дН	10.00 mH	100.0 mH	1000 mH	10.00 H	100.0 H
CIRCUIT MODE	SER			PARA		
Maximum Bias Current*	40m A	36m A	13mA	40m A	36m A	13mA

<sup>\*</sup>Bias current when +40V is applied to DC BIAS connector.

In 1kHz(10kHz) measurement, multiply range value by 1/10 (1/100).

### **CAUTION**

DC BIAS OVER +40 VOLTS MUST NOT BE APPLIED TO EXTERNAL DC BIAS INPUT CONNECTOR.

Figure 3-9. External DC Bias Circuit (Sheet 3 of 3).

Table 3-4. Unusual Operating Indications (Sheet 1 of 4).

# A. Same sample sometimes shows quite different values between PRL and SER CIRCUIT MODE measurements.

### B. The decimal point moves and measurement unit changes.

### Cause of trouble:

A and/or B may occur in the following cases:

Resistance of low loss inductor or capacitor being measured in R FUNCTION.

Inductance of lossy inductor or capacitance of lossy capacitor being measured in L or C FUNCTION.

### What to do:

- A. Do not set CIRCUIT MODE to AUTO. Set CIRCUIT MODE to a PRL or SER setting that shows a valid display.
- B. Set LCR RANGE to MANUAL.

  Manually settle the instrument
  on an appropriate range.

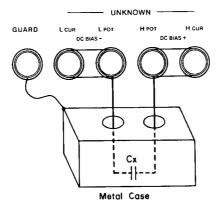
### Indication:

The displayed value fluctuates on minimum capacitance, maximum inductance or maximum resistance ranges in either PRL or SER circuit modes.

### Cause of trouble:

Here are some of the reasons why this happens:

- A. A large size sample is being measured.
- B. A high voltage power line or similar exists near the 4262A.
- C. The 4262A and sample are connected together with relatively long, non-shielded cable.



### What to do:

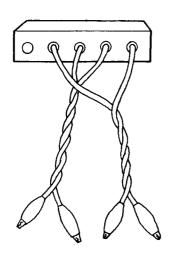
- Enclose sample in metal case.
   Connect case electrically to 4262A
   GUARD terminal as illustrated.
- 2. Use shielded cable for connection between sample and the instrument. Connect cable shield to GUARD.

Table 3-4. Unusual Operating Indications (Sheet 2 of 4).

### Cause of trouble:

When measuring a low impedance (small inductance, resistance or high capacitance), measurement error is excessive.

- Excessive residual impedance (inductance, capacitance or resistance) of test leads in a two terminal measurement.
- 2. Mutual test lead induction between current leads ( $H_{CUR}$  and  $L_{CUR}$ ) and potential leads ( $H_{POT}$  and  $L_{POT}$ ).



### What to do:

Use test leads in four-terminal configuration and measure.

Twist current leads ( $H_{CUR}$  and  $L_{CUR}$ ) together. Do the same with potential leads ( $H_{POT}$  and  $L_{POT}$ ).

Additional error is presented as  $\omega^2 Lr Cx \; X \; 100 \; (\%)$  for C measurement, where:

 $\omega = 2\pi f$ 

f = test frequency

Lr = residual inductance

Cx = unknown capacitance

### Indication:

### Cause of trouble:

Measurement error is excessive when high	Measurement	Cause of error
impedance (high inductance, small capacitance) is measured.	High Inductance	Stray capacitance between High and Low leads.
	Small Capacitance	Stray capacitance between High and Low leads.

### What to do:

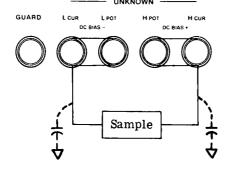
Use shielded cable for connection between sample and 4262A UNKNOWN terminals. Connect outer conductor to GUARD terminal.

Adjust C ZERO ADJ control properly to compensate for stray capacitance.

Table 3-4. Unusual Operating Indications (Sheet 3 of 4).

Excessive measurement error.

Measurement Frequency	Allowable Stray Capacitance Magnitude
120Hz	100nF
1kHz	1000pF
$10 \mathrm{kHz}$	200pF
	•



### Cause of trouble:

### Cause A.

Effect of Low terminal capacitance with respect to ground.

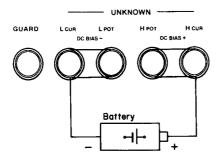
Sometimes the measurement can not be performed when a relatively large capacitance between  $L_{POT}$  terminal and ground exists. Allowable magnitudes for stray capacitance without additional error are given in figure at left.

### Cause B.

Effect of High terminal capacitance with respect to ground. The stray capacitance will reduce test signal level applied to the sample measured during capacitance measurement. This decrease in signal level will not produce an additional error even when measurement signal level is reduced to a third of its nominal level. It is neccessary, of course, that special care be taken to use the proper test signal level when a device is measured whose parameters may be affected by the test signal level. Display fluctuations may sometimes appear.

### Indication:

Internal resistance of a battery can not be measured.



### What to do:

- 1. Connect sample battery (observe polarity) as illustrated.
- Batteries up to 40V are measured under no load conditions.
- 3. If battery voltage exceeds 4V, set DC BIAS to EXT
- 4. Since the internal resistance of a battery is relatively low, use the four-terminal measurement configuration.

Table 3-4. Unusual Operating Indications (Sheet 4 of 4).

### Cause of trouble:

When a sample (for example, an iron core inductor) is measured in AUTO of CIRCUIT MODE, the instrument repeats range selection and does not complete the measurement depending upon level of test current used.

The measurement reading of sample depends on the level of measurement test signal applied.

### What to do:

Set LCR RANGE to MANUAL. Manually settle the instrument on an appropriate range.

### Indication:

When a capacitor is measured with dc bias voltage applied, an abnormal display occurs.

There are limitations to the permissible insulation resistance of a capacitor measured with dc bias. See table below.

MOD	Ε			RANGE		
1kHz	Ср	100.0pF	1000pF	10.00nF	100.0nF	1000nF
1 KHZ	Cs	100.0nF	1000nF	$10.00 \mu F$	100.0μ <b>F</b>	$1000 \mu  ext{F}$
Permissil insulation resistance		<b>30M</b> Ω	3000kΩ	300kΩ	30kΩ	3000Ω

### Note

In 120 Hz (10 kHz) measurement, multiply range value by 10 (1/10).

Ri given in above table is applicable for a dc bias of 40 V. When the bias voltage is less than 40 V, Ri limit is RiVb/40 ( $\Omega$ ) where Ri is value given in the table and Vb is applied dc bias voltage.

### 3-40. OPTION OPERATION.

3-41. Operating instructions for Options 001, 004, and 101 are described in the following paragraphs.

### 3-42. **OPTION 001: BCD PARALLEL DATA** OUTPUT.

The 4262A Option 001 provides parallel BCD outputs for LCR display, D/Q display and information for various control settings. These outputs are fed to two 50 pin connectors on the rear panel.

### 3-44. Output Data and Pin Assignment.

The 4262A Option 001 provides eight kinds of output data:

- (1) FUNCTION and CIRCUIT MODE.
- (2) Test Signal Frequency (LOW LEVEL or normal is excluded).
- (3) Annunciator: Normal, Overflow, Uncal, (LCR and D/Q are not annunciated).
- (4) Unit: p, n,  $\mu$ , m, k, M, D, Q (judgement whether capacitance, inductance or resistance depends on output of FUNCTION switch setting information).
- (5) Decimal Point.
- (6) Polarity.
- (7) Displayed value.
- (8) Other Input/Output Signals.

The signal pin assignments for the 50 pin connector are shown in Figure 3-40. When these signals are fed to digital printer, the print-out is given as a 10 digit decimal number.

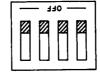
### 3-46. Alternate Output of LCR and D/Q Data.

BCD outputs for LCR and D/Q data of 4262A Option 001 can be alternately supplied through one 50 pin BCD LCR DATA OUTPUT connector on rear panel. This alternate output is enabled by changing slide switch setting on printed circuit board P/N 04262-66535. PC board 04262-66535 is located nearest to the rear panel in the right hand row of PC boards. Normal setting of the four section slide switch for parallel output and the setting for alternate output are illustrated below.

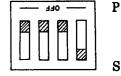
### Normal

Parallel output:

Alternate output:







S

### 3-47. **Output Timing.**

Timing charts for parallel (simultaneous) output and alternate output are shown in Figure 3-41.

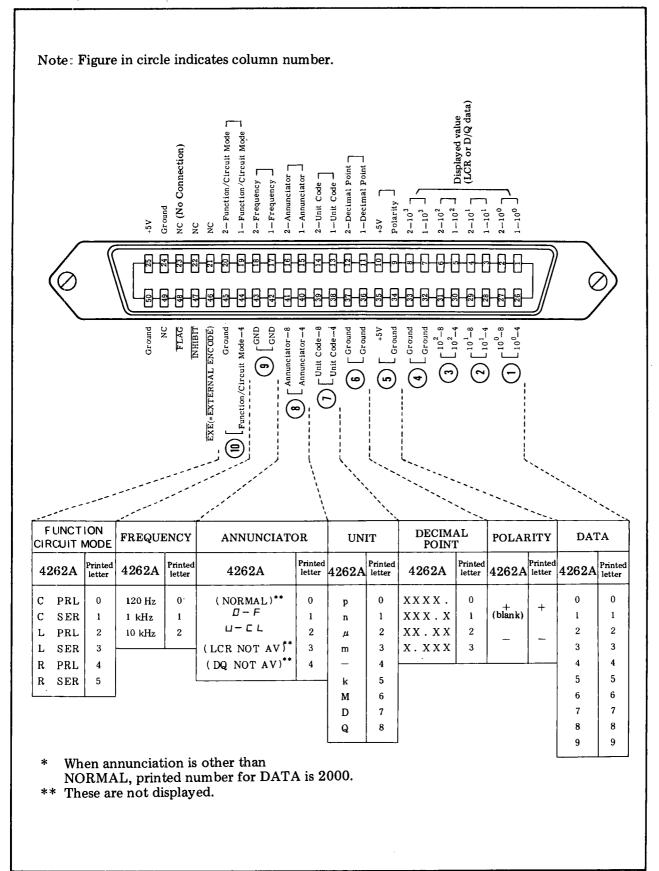


Figure 3-40. Pin Assignments of Output Connector and Output Format.

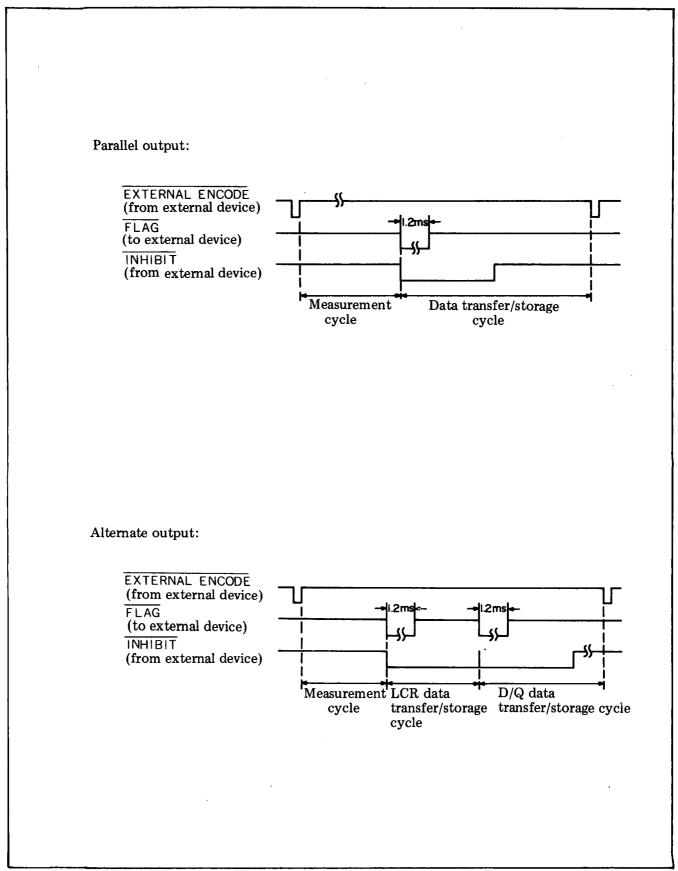


Figure 3-41. Timing Chart of BCD Data Output.

### 3-49. OPTION 004- COMPARATOR.

3-50. The 4262A Option 004 (shown in Figure 3-43) provides:

- (a) HIGH and LOW limits setting for comparison of LCR and D/Q measured data.
- (b) LED visual decision output lamps display of results of HIGH and LOW limit comparisons.
- (c) TTL outputs and relay outputs for HIGH, IN, and LOW decision outputs.

### 3-51. Front Panel Features (Figure 3-42).

- (1) LCR LIMIT Switch: Two four-digit switches provide HIGH and LOW limit values with which measured LCR value is compared. Setting range is from 0000 to 1999.
- (2) LCR Decision Output Lamp: Results of comparison are indicated by LED lamps as follows:

HIGH: (measured value ≥ High limit)
IN: (Low limit ≤ measured value < High limit)
LOW: (measured value < Low limit)

(3) LCR LIMIT CHECK Switch: While this switch is depressed, HIGH and LOW limit values set by LCR LIMIT switches (1) are displayed in LCR and D/Q displays. During this period, three LCR decision output lamps are lit. Comparator must be enabled display limits.

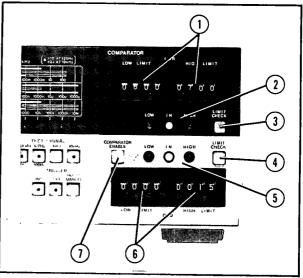


Figure 3-42. Front Panel Features

- (4) D/Q LIMIT CHECK Switch: While this switch is depressed, HIGH and LOW limit values set by D/Q LIMIT switches (6) are displayed in LCR and D/Q displays. During this period, three D/Q lamps of decision outputs are lit.
- (5) D/Q Decision Output Lamp: Results of comparison is indicated by LED lamps as follows:

(6) D/Q LIMIT Switch: Two four-digit switches provide HIGH and LOW limit values with which measured D/Q value is compared. Setting range is from 0000 to 1999.

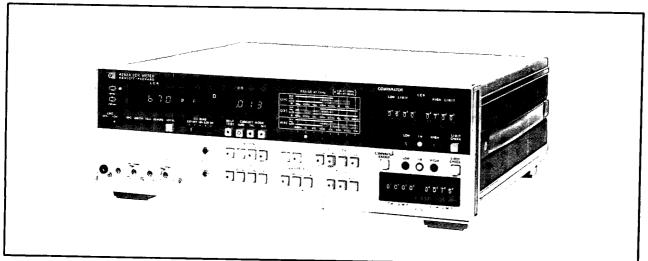


Figure 3-43. Option 004: COMPARATOR.

- (7) COMPARATOR ENABLE Switch: This switch enables the Option 004 to compare measured data with HIGH and LOW limits under a fixed range condition (LCR or D/Q RANGE switch set to MANUAL). If LCR RANGE switch or D/Q switch is set to AUTO, depressing COMPARATOR ENABLE switch changes LCR or D/Q RANGE switch setting to MANUAL.

  If AUTO key of LCR or D/Q RANGE
  - If AUTO key of LCR or D/Q RANGE switch is depressed while COMPARATOR ENABLE switch is ON, one measurement cycle is done in AUTO ranging and the range is fixed to that selected in this measurement cycle.
- 3-52. LIMIT Setting Warning: If HIGH LIMIT setting is lower than LOW LIMIT setting, HIGH and LOW lamps of decision output repeatedly turn ON and OFF to warn operator to change LIMIT setting.
- 3-53. DATA OUTPUT Connector Decision Output: Decision outputs in TTL open collector signal and in relay contact are supplied through COMPARATOR OUTPUT connector on the rear panel. Signal pin assignment is given in Figure 3-44.

### WARNING!

DO NOT APPLY AC LINE VOLTAGE TO RELAY OUTPUT CONNECTOR PIN TO SWITCH LINE CURRENT. For such relay applications, remotely control an external relay with relay output.

Relay Contact Ratings

	AC	DC
Contact Resistance	$100 \mathrm{m}\Omega$	100mΩ
Maximum Permissible Power	30VA	20W
Maximum Permissible Voltage	110V	30V
Maximum Permissible Current	0.3A	1A
Actuation Life	> 10 million	>1 million

### Decision Output Data Format

Decisions	Relay output pins			TTL output pins			
Decisions	DQ LCR 13 17	DQ LCR 14 18	DQ LCR 39 43	DQ LCR 15 19	DQ LCR 16 20	DQ LCR 41 45	
ні	s	0	0	Н	L	L	
IN	0	0	S	L	L	Н	
LO	0	S	0	L	Н	L	

S: Short O: Open

Referenced to common (pin 38 or 42). TTL Output sink current: 30mA max.

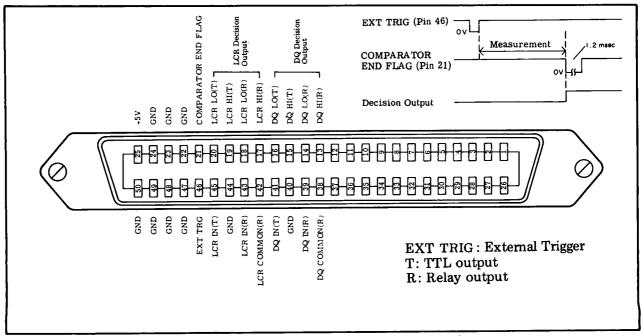


Figure 3-44. Comparator Data Output Pin Locations.

### 3-60. OPTION 101: HP-IB.

- 3-61. The 4262A Option 101 provides interface capabilities in accordance with IEEE-STD-488-1975 recommendations.
- 3-62. Connection to HP-IB Controller: The 4262A Option 101 can be connected to an HP-IB Controller (HP calculator) via HP-IB digital bus connector on the rear panel of the 4262A and the bus connector of the Bus I/O card installed in calculator.
- 3-63. HP-IB Status Indicator: The four LED lamps of the HP-IB Status Indicator (located below the LCR display) show which HP-IB condition the 4262A is in:

SRQ: SRQ signal put on HP-IB line from

4262A. See paragraph 3-70 for details.

LISTEN: 4262A is set to listen. See paragraph

3-69 for details.

TALK: The 4262A is set to talk. Se

paragraph 3-67 for details.

Remote: The 4262A is remotely controlled.

See paragraph 3-71 for details.

- 3-64. LOCAL Switch: This switch disables remote control and enables setting measurement conditions by front panel controls (pushbutton switches). REMOTE lamp of HP-IB status indicator turns off when LOCAL switch is depressed. (When Local Lock Out does not function).
- 3-65. HP-IB INTERFACE CAPABILITIES: The 4262A Opt 101 has the following eight bus interface functions:

SH1: Source Handshake Capability.

AH1: Acceptor Handshake Capability.

T5: Talker (the 4262A sends measurement data to the bus).

L4: Listener (the 4262A receives remote control signals from the bus).

SR1: Service Request Capability.

RL1: Remote/Local Capability.

DC1: Device Clear Capability.

DT1: Device Trigger Capability.

3-66. Source and Acceptor Handshake: SH1, AH1.

Three Bus handshake lines (DAV, NRFD and NDAC) perform Source and/or Acceptor handshake functions.

- (1) DAV (DAta Valid). DIO (Data Input Output) line is available.
- (2) NRFD (Not Ready For Data). Listener preparation for receiving data from Talker is not yet completed.

- (3) NDAC (Not Data Accepted). Listener has not yet received data from Talker.
- 3-67. Talker Capability: T5.

When set to Talker by MTA (My Talk Address) signal from controller, the 4262A sends measurement data to the Bus in one of three types of output formats:

Type A: Ordinary output format. Address switch on the rear panel set to FMT A.

$$\frac{S}{(1)} \frac{FC}{(2)} \frac{F}{(3)} \frac{-NN. NNE-NN}{(4)} , \frac{S}{(5)(1)(6)} \frac{N. NNN}{(7)} \frac{CRLF}{(8)}$$

Type B: Output format used for Model 5150A HP-IB Digital Recorder. Address switch on the rear panel set to FMT B.

$$\frac{S}{(1)} \frac{FC}{(2)} \frac{F}{(3)} \frac{-NN. NNE-NN}{(4)} \frac{CRLF}{(8)} \frac{S}{(1)(6)} \frac{N. NNN}{(7)} \frac{CRLF}{(8)}$$

Type C: Output format used in resistance measurement or LCR ONLY measurement when no D/Q data is to be outputted. Selection of this format is automatically done in accordance with FUNCTION switch setting.

$$\frac{S}{(1)} \frac{FC}{(2)} \frac{F}{(3)} \frac{-NN. NNE-NN}{(4)} \frac{CRLF}{(8)}$$

The numbered elements of output data are described below:

(1) Status:

 N
 Normal

 O
 Overflow

 U
 Uncal

 X
 LCRNA or DNA

(NA: Not Available)

(2) Function and Circuit Mode:

FUNCTION	MEASURE- MENT	CIRCUIT _MODE
<del></del>		
CP	C	$\mathtt{PRL}$
CS	$\mathbf{C}$	$\mathbf{SER}$
LP	${f L}$	$\mathtt{PRL}$
LS	L	$\mathbf{SER}$
$\mathbf{RP}$	$\mathbf{R}$	PRL
$\mathbf{RS}$	R/ESR	SER

(3) Frequency:

A..... 120Hz (100Hz)

B..... 1kHz

C..... 10kHz

- (4) LCR Data
- (5) Data Delimiter
- (6) Loss

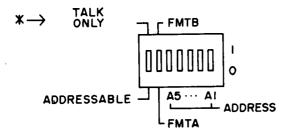
D..... Dissipation Factor measurement Q..... Quality Factor measurement

- (7) DQ Data
- (8) Data Terminator

### 3-68. Functions Related to Talker Capability.

EOI (End Or Identify): When multiple byte data of Source Handshake has been sent, the 4262A provides EOI to the bus.

Talk Only Mode: When ADDRESS switch is set to TALK ONLY "1" position, the 4262A is set to Talker regardless of address code.



Talk Address Disabled by Listen Address:

MTA (My Talk Address) is automatically disabled when MLA (My Listen Address) is set. MTA (My Talk Address) is otherwise disabled by IFC (Interface Clear) signal, OTA (Other Talk Address) signal or UTA (Untalk Address) signal.

### 3-69. Listener Capability: L4.

To receive Remote Program signal or Addressed Command signal, the 4262A is set to Listener by an MLA (My Listen Address) signal from the bus.

- Remote Program signal: Remote program codes for the 4262A are listed in Table 3-60.
- (2) Addressed Command signal: When the 4262A receives command signals GET, GTL, or SDC, it is set to Listener and controlled by command signals. These command signals are valid regardless of the status (remote or local).

GET (Group Execute Trigger): When the 4262A receives this command, it is triggered regardless of front panel TRIGGER switch setting.

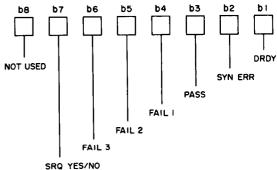
GTL (Go to Local). The 4262A is set to LOCAL by this command to enable front panel control.

SDC (Selected Device Clear): When this command is accepted, front panel controls are set to initial conditions (the same conditions that are automatically set after turn-on of power switch).

Listen status is automatically disabled when MTA (My Talk Address) is received. Listen status is otherwise disabled by IFC (Interface Clear) signal or ULA (Unlisten Address) signal.

### 3-70. Service Request Capability: SR1.

The 4262A sends an SRQ (Service Request) signal whenever it is set in one of the six possible RQS (Request Status) states. It does this by responding to a serial poll of the controller by setting an STB (Staus Byte) signal on the bus. The 7th bit of this 8 bit signal establishes whether or not a service request exists. The remainder of the 8-bit signal identifies the character of the SRQ.



SRQ (Service Request) is disabled when RQS (Request Status) or STB (Status Byte) is set to 000000000 or when STB (Status Byte) signal transfer is completed.

Request Statuses (RQS) of the 4262A:

- (1) DRDY (Data ReaDY): When the 4262A-completes a measurement cycle, this status bit is set. This status is set without serial polling if NOT DATA READY is set.
- (2) SYN ERR (SYNtax ERRor): When the 4262A receives an erroneous Remote Program Code which is not listed in Table 3-60, this status bit is set.
- (3) PASS (Self Test Pass): When PASS is displayed in Self Test done by remote control, this status bit is set.
- (4) FAIL 1 (Self Test Fail 1): When FAIL 1 is displayed in Self Test done by remote control, this status bit is set.
- (5) FAIL 2 (Self Test Fail 2): When FAIL 2 is displayed in Self Test done by remote control, this status bit is set.
- (6) FAIL 3 (Self Test Fail 3): When FAIL 3 is displayed in Self Test done by remote control, this status bit is set.

Table 3-60. Remote Program Codes.

	C	Program Code	
Function	L	F 1	
	C		F 2
	R/E	SR	F 3
Circuit Mode	AUT	)	C 1
	PRL		C 2
	SER		C 3
Loss	D		L 1
	Q	· · · · · · · · · · · · · · · · · · ·	L 2
Frequency	120 H	z	H 1
	1 kH2	3	Н 2
	10 kH	z	H 3
Trigger	INT		T 1
	EXT		T 2
	HOLE	MANUA L	Т 3
Self Test	OFF	S 0	
	ON	S 1	
△LCR	OFF	M 0	
	ON	M 1	
Cp Low Level	OFF	P 0	
	ON	P 1	
* Data Ready	OFF		D 0
RQS Mode	ON		D 1
	(C)	(L) (R)	
LCR Range	100 p	100 μ 1000 m	R 1
at 1 kHz	1000	1000 10	R 2
	10 n	10 m 100	R 3
i	100	100 1000	R 4
	1000	1000 10 k	.R 5
	10 μ	10 100 k	R 6
	100	100 1000 k	R 7
	1000	- 10 M	R 8
!		AUTO —	R 9
DQ Range	(D)	(Q)	
į		1000	N 1
		100.0	N 2
	10.00	10.00	N 3
	1.000	1.000	N 4
	— A	N 5	

Table 3-61. Remote Message Coding.

		CLASS	D D I O O S 7 6 5 4 3 2 1
DCL	device clear	UC	X 0 0 1 0 1 0 0
GET	group execute trigger	AC	X 0 0 0 1 0 0 0
GTL	go to local	AC	X 0 0 0 0 0 0 1
LLO	local lock out	UC	X 0 0 1 0 0 0 1
MLA	my listen address	AD	X 0 1 L L L L L L L 5 4 3 2 1
МТА	my talk address	AD	X 1 0 T T T T T T 5 4 3 2 1
ОТА	other talk address	AD	$(OTA = TAG \cap \overline{MTA})$
SDC	selected device clear	AC	X 0 0 0 0 1 0 0
SPD	serial poll disable	UC	X 0 0 1 1 0 0 1
SPE	serial poll enable	UC	X 0 0 1 1 0 0 0
STB	status byte	ST	s x s s s s s s
UNL	unlisten	AD	X 0 1 1 1 1 1 1
UNT	untalk	AD	X 1 0 1 1 1 1 1

CLASS UC: Universal Command

AC: Addressed Command

AD: Address
ST: Status Byte

### 3-71. Remote/Local Capability: RL1.

The 4262A goes to Remote Status only when it accepts Listen address with REN (Remote Enable) line in the Bus lines set to "1". Remote status is not obtained if REN line is set to "1" after Listen address is received. Remote status is returned to Local status when one of following conditions is present:

- (1) REN line is set to "0".
- (2) LOCAL switch on front panel is depressed.
- (3) GTL (Go To Local) command is received.

### Local Lock Out: LLO

Local Lock Out inhibits the function of LOCAL switch. This LLO command is a universal command and is valid when REN line is set to "1". LLO command is disabled when REN line is set to "0"

### 3-72. Device Clear Capability: DC1.

The 4262A is set to initial conditions (the same conditions that are automatically set after turn-on of power switch), when it accepts DCL (Device CLear) command—universal command—or SDC (Selected Device Clear)—addressed command.

### 3-73. Device Trigger Capability: DT1.

The 4262A is triggered regardless of TRIGGER switch setting when it accepts GET command—address command.

3-74. ADDRESS Switch: ADDRESS switch on the rear panel sets Listen/Talk address. Five section or five bit switch provides 30 settings from 00000 to 11110.

3-75. Remote Message Coding: Interface Bus Command signals for the 4262A are listed in Table 3-61.

# SECTION IV PERFORMANCE TESTS

### 4-1. INTRODUCTION.

4-2. This section provides the check procedures to verify the 4262A specifications listed in Table 1-1. All tests can be performed without access to the interior of the instrument. A simpler operational test is presented in Section III under Self Test (paragraph 3-5). The performance test procedures in this section can also be used to do an incoming inspection of the instrument and to verify whether the instrument meets its specified performance after troubleshooting or making adjustments. If specifications are found to be out of limits, check that controls are properly set, and then proceed to adjustments or troubleshooting.

### Note

Allow a 15-minute warm-up and stabilization period before conducting any performance test.

### 4-3. EQUIPMENT REQUIRED.

4-4. Equipment required for the performance tests is listed in Table 1-4 Recommended Test Equipment in Section I. Any equipment whose characteristics equal the critical specifications given in the table may be substituted for the recommended model(s).

Accuracy checks in this section use standard LCR components as the samples to be connected to the 4262A. Accessories 16361A and 16362A can be utilized for this purpose. These accessory models are DUT (device under test) boxes from which the desired component can be selected and connected to the 4262A through cables by use of a

rotary switch. If models 16361A/16362A are unavailable, use the discrete components recommended in Table 4-1.

### Note

All components used as standards should be calibrated by an instrument whose specifications are traceable to NBS, PTB, LNE, NRC, JEMIC, or equivalent standards group; or all components should be calibrated directly by an authorized calibration organization such as NBS. The calibration cycle should be determined by the stability specification for each component.

### 4-5. TEST RECORD.

4-6. Results of the performance tests may be tabulated on the Test Record at the end of these procedures. The Test Record lists all the tested specifications and their acceptable limits. Test results recorded at incoming inspection can be used for comparison in periodic maintenance and trouble-shooting and after repairs or adjustments.

### 4-7. CALIBRATION CYCLE.

4-8. This instrument requires periodic verification of performance. Depending on the use and environmental conditions, the instrument should be checked with the following performance tests at least once every year. To maximize the "up time" of the instrument, the recommended preventive maintenance frequency for the 4262A is twice a year.

### -PRELIMINARY OPERATIONS-

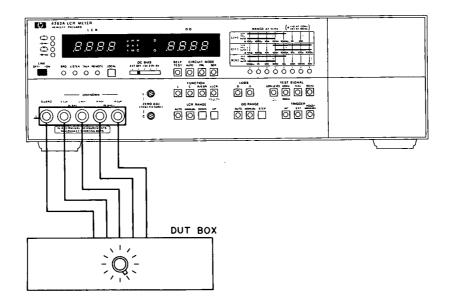
Before beginning performance test, adjustment, or calibration of 4262A, check fundamental operating conditions of the instrument and perform display ZERO adjustments in accord with the following procedures:

- 1) Confirm that power line power voltage in use is appropriate for the instrument operating power voltage.
- 2) Depress LINE pushbutton and confirm that all the front panel displays and indicators momentarily illuminate. The 4262A functions are automatically set to capacitance measurement mode.
- 3) ZERO offset adjustment should be made whenever a test fixture or DUT box is connected to 4262A UNKNOWN terminals. Adjust C ZERO ADJ and L ZERO ADJ controls so as to fully compensate for stray capacitance and residual inductance of equipment connected to UNKNOWN terminals. Adjustment procedures to adjust for individual test equipment used are provided in steps 3-a and 3-b which follow.
  - 3-a) 16361A/16362A or user built DUT box.
    - 1. Disconnect shorting bars from 4262A UNKNOWN terminals. Connect test leads between 4262A UNKNOWN terminals and DUT box.
    - 2. Set 4262A FUNCTION to C. Set TEST SIGNAL frequency as appropriate to DUT box being used.
    - 3. Set range control of DUT box to open-circuit position (2pF range on 16361A or 1pF range on 16362A). The 4262A is automatically set to its lowest capacitance measurement mode range.
    - 4. Adjust C ZERO ADJ control so that capacitance readout on 4262A LCR display is identical to calibrated value of DUT box range.
    - 5. Set 4262A FUNCTION to L.
      - 6. Set range control of DUT box to short-circuit position ( $20m\Omega$  range on 16361A or on 16362A).
      - 7. Adjust L ZERO ADJ control for 000 counts on LCR display.

### Note

To permit easy adjustment of ZERO ADJ controls for an individual DUT box, each DUT box should be equipped with short and open circuit ranges which provide  $0\mu$ H and 0pF (practical values), respectively.

### PRELIMINARY OPERATIONS-



### 3-b) 16061A or other test fixtures.

- 1. Disconnect shorting bars from 4262A UNKNOWN terminals and attach test fixture to UNKNOWN.
- 2. No DUT should be connected to the test fixture.
- 3. The 4262A is automatically set to lowest capacitance range in measurement mode. Set 4262A TEST SIGNAL frequency to 10kHz.
- 4. Adjust C ZERO ADJ control for 000 counts on LCR display.
- 5. Set 4262A FUNCTION to L.
- 6. Connect a shorting lead to test fixture to short-circuit the measurement terminals.
- 7. Adjust L ZERO ADJ control for 000 counts on LCR display.

### Note

When positions or mutual distance between Test Fixture contacts are changed, or contacts are changed to a different type, again perform ZERO adjustments.

### - CALIBRATION OF DUT'S -

Either user built DUT's or substitution standards with accuracies which satisfy the requirements may be used for performance testing and calibration of the 4262A. The DUT's recommended for making the tests and adjustments can be accuracy certified in accord with the calibration procedure detailed below. This calibration procedure applies to all alternate DUT's which do not carry public or testing laboratory certification.

### [CAPACITANCE CALIBRATION]

Measure the DUT or substitution standard capacity with a precision capacitance bridge that meets the calibration accuracy and frequency requirements. For testing or calibrating dissipation factor of DUT, use equipment with required dissipation measuring capability and verify the exact calibration frequency to permit compensating D value for the difference in measuring frequency between individual Model 4262A's and the calibration equipment. If the frequency error is less than 3%, compensation is not required for dissipation factors of 0.01 and below.

### [RESISTANCE CALIBRATION]

Use a metal film resistor of appropriate value for each DUT to maintain a constant resistance over a wide range of frequencies. Measure the resistance with a high accuracy DMM. When measuring  $1k\Omega$  and below, use a 4 terminal measurement configuration.

### [DISSIPATION FACTOR CALIBRATION]

DUT's used as D standards can be built with precisely measured components. The dissipation factor of the DUT is determined by an exact calculation from the calibrated values of each components in accord with the following equations:

Circuit Mode	Derivation of D
Cp Rp	$D = 1/\omega CpRp$
Cs Rs	$D = \omega CsRs$

Note

For easier calibration of dissipation, use accurately calibrated resistors rather than capacitors.

### - CALIBRATION OF DUT'S -

To minimize the calculation error, the inherent dissipation of the capacitor should be 0.001 or below. When using polystyrene or silvered mica type capacitors (dissipation factor is generally very low), the residual factors will not affect the derivation of accurate dissipation factors. If dissipation of capacitor alone is greater than 0.001, the effective value of the DUT is calculated in accord with the following equation:

$$Ds = Dc + Dr$$
 ( $Dr \leq Dc, Dr \leq 0.01$ )

where, Ds is actual dissipation factor of DUT.

Dc is calculated D value (excludes inherent dissipation).

Dr is inherent dissipation of capacitor.

Compensate the dissipation factor for the measuring frequencies of individual 4262A being tested or calibrated. Convert the D value of the calibration frequency to that of the actual 4262A measuring frequency in accord with the following equations:

	<del>-</del>	$x = \frac{fc}{fm}$	Dm: D value at 4262A measuring frequency. Ds: D value at calibration frequency.
Dm = X·Ds	- <b>I-</b> -W-	$x = \frac{fm}{fc}$	fm: 4262A measuring frequency fc: Calibration frequency.

Note

To accurately measure frequencies fm and fc, use a reciprocal counter or calculate reciprocal number of period.

### [CALIBRATION EQUIPMENT]

The recommended model and required performance of calibration equipment is listed below:

Instrument	nstrument Required Performance	
Capacitance Bridge	Capacitance Accuracy: 0.1% Dissipation Factor Accuracy: 0.1% (Resolution 0.0001)	GR 1620-A
DMM	Resistance Accuracy: 0.02%	HP 3490A HP 3455A
Freq. Counter	Reciprocal counter Resolution: 0.01Hz	HP 5300A/5307A HP 5323A

Table 4-1. Recommended Components for Accuracy Checks.

Compo	onent *1	HP Part Number	Alternate Source	Required Calibration Accuracy	
Capacitor	100pF 1000pF 10nF 100nF 1000nF 10\mu F 1000\mu F	0160-0336 0160-3766 0160-0408 0160-1571 0160-3645 0160-3563	HP Model 4440B GR Type 1413  SOSHIN TM-520C GR Type 1417	0.05% 0.2% 0.25%	
Resistor:	1kΩ 10kΩ 100kΩ 10MΩ	0698-3491 0698-6360 0698-4158 0698-8194	GR Type 1433-Y	0.05%	
Inductor:	100mH		GR Type 1482-L	0.05%	
Dissipation Factor: $1000 \text{nF in parallel with } 887\Omega$ $(D \approx 1.50 \text{ at } 120 \text{Hz})$ $100 \text{nF in parallel with } 887\Omega$ $(D \approx 1.79 \text{ at } 1 \text{kHz})$ $10 \text{nF in parallel with } 887\Omega$ $(D \approx 1.79 \text{ at } 10 \text{kHz})$		0160-3645 0698-4464 0160-1571 0698-4464 0160-3171 0698-4464	(D=1/ωCR)	**2 Capacitors · · · 0.1% Resistors · · · 0.02%	

<sup>\*1</sup> The components listed above or used as standards should be calibrated before they are utilized.

Proper method and procedure for calibrating the DUT's is given in "Calibration of DUT's" (Page 4-4).

<sup>\*\*2</sup> For easier calibration of dissipation to the required accuracy (0.1%), use accurately calibrated resistors rather than capacitors (use a high accuracy DMM to measure resistors).

#### 4-9. MEASUREMENT FREQUENCY TEST.

#### DESCRIPTION:

This test verifies the accuracy of the measurement frequencies that are applied to an unknown sample connected to the 4262A.

#### SPECIFICATIONS:

Measurement Frequencies:

120Hz ± 3% 1kHz ± 3% 10kHz ± 3%

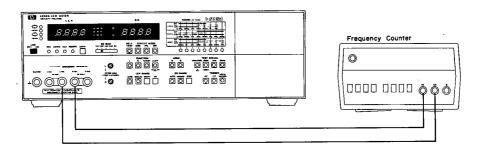


Figure 4-1. Measurement Frequency Test Setup.

### **EQUIPMENT:**

#### PROCEDURE:

- 1. Connect frequency counter to the 4262A UNKNOWN terminals as shown in Figure 4-1.
- 2. Set range of frequency counter as appropriate for measuring 4262A test frequencies of 120Hz, 1kHz and 10kHz.
- 3. Read display output of frequency counter when 4262A TEST SIGNAL is set to 120Hz, 1kHz or 10kHz.
- 4. Frequency readouts must be within the following limits (record measured frequency in table below as the data is used in paragraph 4-12):

TEST SIGNAL	Test Limits	Counter Readout
120Hz	116.4 - 123.6Hz	
1kHz	970 - 1030 Hz	
10kHz	9700 - 10300 Hz	

### Note

Test limits in table above do not take into account reading error caused by measurement error in test equipment.

#### Note

If this test fails, refer to Service Sheet 11 in Section VIII for troubleshooting.

## 4-10. CAPACITANCE ACCURACY TEST.

## **DESCRIPTION:**

This test checks capacitance measurement accuracy for zero and full scale displays at three test frequencies and at two signal levels. The test is made by connecting a stable capacitor more accurate than the 4262A to the instrument and reading the display to verify that the 4262A meets its measurement accuracy specifications. Check all ranges in Cp mode and one range in Cs mode at each frequency (120Hz, 1kHz and 10kHz) to guarantee C measurement accuracy since all variable elements (range resistors and detecting phases) needed for C measurement are thus checked. In this test, almost all ranges, from the lowest through the highest ranges, are being verified.

#### Note

If the following tests satisfy the accuracy specifications, all the accuracy specifications listed in Table 1-1 are guaranteed.

#### Capacitance Accuracy Test Ranges

TEST	TEST SIGNAL		RANGE								
Freq.	Level	CIRCUIT MODE	10.00pF	100.0pF	1000pF	10.00nF	100.0nF	1000nF	10.00µF		
	TOM TEAET	PRL	> <	> <							
120Hz	normal	PRL	> <	> <							
	norma	SER	> <	$\times$	> <	> <	> <				
	TOM FEART	PRL	${}$						$\overline{}$		
1kHz		PRL	$\supset \subset$						$\overline{}$		
	normal	SER	$\times$	> <	> <	> <					
	LOW LEVEL	PRL .						> <	$\searrow$		
10kHz	notmal	PRL						$\overline{\mathbf{x}}$	ightrightarrow		
		SER	$\times$	>	$\times$			$\overline{}$	_		

#### TEST SIGNAL level:

LOW LEVEL									.50 mV
normal			_						1 V

Tests for dissipation factor accuracy with above capacitance standards should be done at the same time as capacitance tests

Check all parallel (PRL) mode ranges. It is sufficient to check any one range in series (SER) mode.

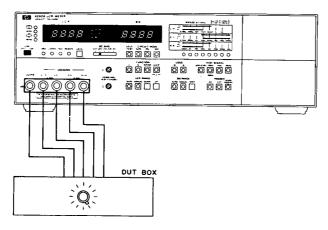


Figure 4-2. Capacitance Accuracy Test Setup.

### SPECIFICATIONS:

### C-D/Q MEASUREMENT ACCURACIES.

Range	120Hz 1kHz 1kHz 10kHz	$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$
-411-		
C Accuracy*1	-11	(At 120Hz, 1kHz) 0.3% + 2 counts 0.5% + $\frac{15. *2}{2 \text{ counts}}$ (At 10kHz) 0.3% + 2 counts 1% + 2 [5% + 2]
	AUTO	Same as - Mode Same as - Mode
D (1/Q) Accuracy *1	-₩-	0.2% + (2 + 200/Cx) counts  0.5% + (2 + 200/x) counts  1.0% + (2 + 1000/Cx) counts  At 120Hz, 1kHz  At 10kHz  At 10kHz  At 120Hz, 1kHz  (Test signal level: 50mV)  At 10kHz
riccuracy .	-I+-W-	(At 120Hz, 1kHz) $ \begin{array}{ c c c c c c c c c c c c c c c c c c c$
	AUTO	Same as - Mode Same as - Mode

<sup>\*1 ±(%</sup> of reading + counts). Cx is capacitance readout in counts. This accuracy only applies for D values to 1.999.

Accuracy applies over a temperature range of 23°C ±5°C (at 0°C to 55°C, error doubles).

### **EQUIPMENT:**

DUT Box..... HP 16361A/16362A Test Leads..... HP P/N 16361-61605

#### Note

User built test fixture or DUT box may be used instead of those HP provides. If user supplied, the residual impedance and stray capacitance of the fixture and box must be taken into account.

#### PROCEDURE:

- 1. Connect Test Leads (HP P/N 16361-61605) between 4262A UNKNOWN terminals and HP 16361A DUT Box (see Figure 4-2). When TEST SIGNAL frequency is 10kHz, use HP 16362A in place of HP 16361A.
- 2. Set 4262A controls as follows:

DC BIASOFF
FUNCTIONC
LCR RANGE AUTO
LOSSD
D/Q RANGE AUTO
TRIGGER INT

<sup>\*2</sup> (5% + 2 counts) at 1kHz.

3. Confirm that the table on page 4-11 is satisfied when the measurements are made by changing TEST SIGNAL, CIRCUIT MODE and DUT as given in the table. Record capacitance and dissipation factor readings in blank spaces provided in table.

### Note

Error caused by stability of standard component is not taken into account for test limits in the table.

Test limits in parentheses are those for dissipation factor measurement value.

If tests fail, proceed to Section V ADJUSTMENTS or Section VIII SERVICE.

TEST S	SIGNAL	CIRCUIT				16361	A/16362A	RANGE			
Freq.	level	MODE	10pF*1	100pF	1000pF	10nF	100nF	1000nF	10µF	1000µF	10mF
	LOW LEVEL	PRL		C. V. ±4 counts (———)		C. V. ±5 counts (±3 counts)	±5 counts				
120Hz		PRL		±2 counts		C. V. ±3 counts (±3 counts)		±3 counts			
	normal	SER						±5 counts	±5 counts	C. V. ±7 counts (±4 counts)	C. V. ±12 counts (±7 counts
	LOW LEVEL	PRL		±8 counts		C. V. ±5 counts (±3 counts)					
1kHz		PRL		•		C. V. ±3 counts (±3 counts)					
	normal	SER			•	3	1		ı	C. V. ±52 counts (±7 counts)	·
	LOW LEVEL	PRL	l			C. V. ±5 counts (±3 counts		1	<u> </u>		
10kHz		PRL				C. V. s±3 counts (±3 counts	l				
	normal	SER			1	C. V. s±5 counts ) (±4 counts	1	I .	1	1	

TEST SIGNAL level: LOW LEVEL . . . . . 50mV normal . . . . . . . . 1V

<sup>\*1</sup> HP 16362A Only \*\*2 C. V. = Calibrated Value of Standard Component.

## 4-11. RESISTANCE/\*\*ESR ACCURACY TEST.

#### DESCRIPTION:

This test verifies that resistance measurement accuracies for 4262A tested meets the specifications listed below. Although R measurement accuracies are actually guaranteed when C measurement accuracies meet the specifications, almost all ranges in Rp mode are checked in this test.

#### Note

Resistance accuracy has only to be proved for one resistor of about full scale value on any one range to verify specifications for 120Hz, 1kHz and 10kHz.

#### SPECIFICATION:

## RESISTANCE/ESR ACCURACY SPECIFICATIONS

		r			<del> </del>	<del>,</del> -				
Ranges	120Hz 1kHz 10kHz		10.00Ω	100.0Ω	1000Ω	10.00kΩ	100.0kΩ	1000kΩ	10.00Ms	
	4	0.3% + 2 counts *2								
Accuracy *1	-35-W- -11-W-	0.2% + 2 counts								
	AUTO				r ]		<b>.</b>		<del></del>	
	AUIU	Sam	e as 1PW		loge	Same a	<u>。</u> -🖏	Mo	de	

- \*1  $\pm$ (% of reading + counts).
- \*2 (5% +2 counts) on 10.00M $\Omega$  range at 10kHz.
- \*\* Measurement range for ESR (equivalent series resistance) is from  $1m\Omega$  to  $19.99k\Omega$  (typical), which varies with series capacitance or inductance value . . . . . refer to "REFERENCE DATA" on page 1-6.

Accuracy applies over a temperature range of 23°C ±5°C. (at 0°C to 55°C, error doubles).

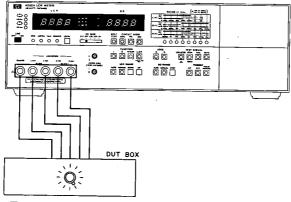


Figure 4-3. Resistance Accuracy Test Setup

### **EQUIPMENT:**

Note

User built fixture/leads or DUT box can be used. If user supplied, the residual resistance must be considered.

#### PROCEDURE:

- 1. Connect Test Leads (HP P/N 16361-61605) between 4262A UNKNOWN terminals and HP 16361A DUT Box (see Figure 4-3).
- 2. Set 4262A controls as follows:

DC BIASOFF
CIRCUIT MODEPRL
FUNCTION
LCR RANGE AUTO
TEST SIGNAL 1kHz
TRIGGER INT

3. Check that the resistance measurement accuracies meet specifications according to table below:

DUT	1kΩ	<b>10</b> kΩ	100kΩ	10ΜΩ
Test Limits	C. V. ±5 counts	C. V. ±5 counts	C. V. ±5 counts	C. V. ±5 counts
R Readout				

C. V. = Calibrated Value of Standard Component

### Note

Error caused by stability of standard component is not taken into account for test limits in table above.

### Note

If this test fails, go to Section V or Section VIII for the troubleshooting.

## 4-12. DISSIPATION FACTOR CONFIRMATION CHECK

#### DESCRIPTION:

This test verifies that a tested 4262A satisfies dissipation factor measurement accuracies. Only one Dissipation Factor (D = 1.8) is checked for 120Hz, 1kHz and 10kHz in this check because only one detecting phase needs to be checked. All other factors influencing D accuracy were checked in paragraph 4-10.

#### Note

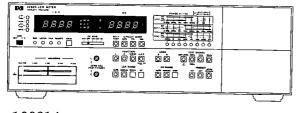
Dissipation factor accuracy for only one D standard which has a D value of approximately 1.8 need be proved to guarantee D accuracy. This test also verifies that 4262A correctly calculates Q factor as a reciprocal number of Dissipation Factor. Only one Q factor corresponding to a D value of approximately 1.8 is checked in this test. D accuracy in measuring inductance does not need to be checked because detecting phase accuracy is equated with that for capacitance measurement.

### C-D ACCURACY SPECIFICATIONS

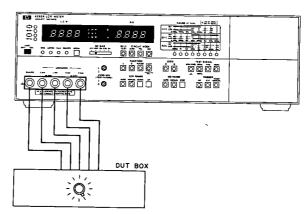
o b necessite i bi best teations										
Range	l kHz	1000pF 100.0pF 10.00pF	1000pF	10.00nF	100.0nF	l 1000nF	110.00uF	1100 Ou F	l 1000#F	
	dh			2 + 200/C 0.5% + (2		At 120Hz, 1kHz (Test signal level: 1V) At 10kHz				
D (1/Q) Accuracy *1	<b></b>			+ (2 + 10 + (2 + 10		At 120Hz, 1kHz (Test signal level: 50mV) At 10kHz				
1	- <del>11-W-</del>	(		z, 1kHz) ; 10kHz)		+ (2 + Cz	k/500) co		$1\% + (5 + \frac{Cx}{500})$	
	AUTO	San	ne as -	₩- N	1ode	Sam			ode	

<sup>\*1 ±(%</sup> of reading + counts). Cx is capacitance readout in counts.

Accuracy applies over temperature range of 23°C  $\pm 5$ °C. (At 0°C to 55°C, error doubles) This accuracy only applies for D values to 1.999.



16061A



(a) (b)

Figure 4-4. Dissipation Factor Accuracy Test Setups.

### **EQUIPMENT:**

#### Note

HP 16361A and HP 16362A DUT Boxes are equipped with D standards (D = 1.8) calibrated at 1kHz and 10kHz frequencies, respectively. For the test at 120Hz frequency or if DUT box is not available, it is recommended that the following DUT's be used as D standards:

DUT	Freq.	Values of components	Calculated D	Tolerance*
С	120Hz	C:1000nF(HP P/N 0160-3645) R:887Ω (HP P/N 0698-4464)	1.495	±0.030
-	1kHz	C: 100nF (HP P/N 0160-1571) R: 887Ω (HP P/N 0698-4464)	1.794	±0.036
R	10kHz	C : 10nF (HP P/N 0160-3171) R : 887Ω (HP P/N 0698-4464)	1.794	±0.036

<sup>\*</sup> After calibrating capacitance C  $\,$  to within 0.1% and resistance R  $\,$  to within 0.02%, the dissipation factor tolerance is  $\pm 0.002$  for each DUT.

#### PROCEDURE:

### 1. Connect DUT to 4262A.

### Note

To facilitate connecting recommended DUT's, attach HP 16061A Test Fixture to 4262A UNKNOWN terminals [see Figure 4-4 (a)]. When HP 16361A/16362A DUT Box is used for this test, connect Test Leads (HP P/N 16361-61605) between 4262A UNKNOWN terminals and DUT Box as shown in Figure 4-4 (b).

## 2. Set 4262A controls as follows:

DC BIASOFF
CIRCUIT MODEPRL
FUNCTIONC
LOSS
LCR RANGE AUTO
D/Q RANGE AUTO
TRIGGER INT

3. Check D accuracies according to following table:

Freq	Circuit Mode	Test Level	D Test Limits	D Reading
}		Low Level	Calibrated Value X ± 8 counts	
120Hz		normal	Calibrated Value X ± 6 counts	
	-I-W-	normal	Calibrated Value X ± 8 counts	
		Low Level	Calibrated Value X ± 8 counts	
1kHz		normal	Calibrated Value X ± 6 counts	
	-I-W-	normal	Calibrated Value X ± 9 counts	
		Low Level	Calibrated Value X ± 21 counts	
10kHz		normal	Calibrated Value X ± 11 counts	
	-I <del>-w-</del>	normal	Calibrated Value X ± 13 counts	

#### Note

X in above table is produced by test frequency error and may be determined from the following equations:

 $x = \frac{fn}{fx}$
 $x = \frac{fx}{fn}$

••• where fn is nominal measurement frequency and fx is measurement frequency from paragraph 4-9.

#### Note

Error caused by stability of standard component is not taken into account for test limits in table above.

- 4. Set 4262A TEST SIGNAL frequency to 1kHz and connect appropriate DUT to 4262A (Set 16361A LCR RANGE to D = 1.8). Note dissipation readout on D/Q display.
- 5. Push 4262A LOSS Q button.
- 6. Confirm that displayed Q factor is correct reciprocal number of dissipation.

#### Note

The 4262A rounds fractions of 5 or greater below the LSD to the next higher digit and drops any fractions of 4 or less. For example, if the actual dissipation is .0135, the display will read .014. If the actual dissipation is .0134, the display will read .013. If the test fails, refer to Section VIII Service.

## 4-13. INDUCTANCE ACCURACY TEST.

### DESCRIPTION:

This test verifies that inductance measurement accuracy satisfies the specifications listed below. L accuracy is proved to meet the specification when the results obtained in the accuracy checks of paragraphs 4-9 through 4-12 satisfy the specifications. This test is performed to confirm the L accuracy specification.

#### Note

Inductance accuracy has only to be proved for one inductor of about full scale value on any one range to verify specifications for all three test frequencies (120Hz, 1kHz and 10kHz).

#### SPECIFICATIONS:

#### INDUCTANCE ACCURACY SPECIFICATIONS

Range	120Hz 1kHz 10kHz	100.0μH		100.0mH 10.00mH 1000μH	100.0mH	10.00H 1000mH 100.0mH	100.0H 10.00H 1000mH	1000H 100.0H 10.00H
	魯	(At 120Hz, 1kHz) 0.3% + 2 counts (At 10kHz) 0.3% + 2 counts		1% + 2 counts 1% + 2 5% + 2				
L Accuracy			0.2	% + 2 cour	nts		(At 120Hz	, 1kHz)
*1		0.3% + 2 0.2% + 2 counts				(At 10kHz)		
	AUTO	Sa	me as -on	Mode		Same a	ıs <b>-(∰)</b> -	Mode

\*1  $\pm$ (% of reading + counts).

Accuracy applied over temperature range of 23°C ±5°C (at 0°C to 55°C, error doubles). This accuracy only applies for D values to 1.999.

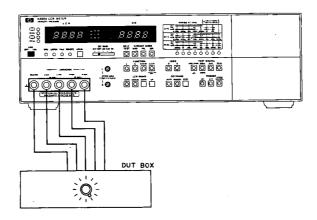


Figure 4-5 Inductance Accuracy Test Setup.

### **EQUIPMENT:**

#### Note

User built test fixture/leads or DUT box must take residual impedance into consideration.

#### PROCEDURE:

- 1. Connect Test Leads (HP P/N 16361-61605) between 4262A UNKNOWN terminals and HP 16361A DUT Box (see Figure 4-5). When TEST SIGNAL frequency is 10kHz, use HP 16362A in place of HP 16361A.
- 2. Set 4262A controls as follows:

DC BIASOF	F
FUNCTION	L
LOSS	O
LCR RANGE AUTO	)
D/Q RANGE AUTO	)
TRIGGER IN'	Г

- 3. Set HP 16361A/16362A LCR RANGE to 100mH.
- 4. Confirm that L accuracy is within the test limits shown in table below:

#### Note

Test limits below are given for 100mH inductance measurement. If another inductance value is measured, refer to SPECIFICATIONS above.

TEST SIG Freq.	CIRCUIT MODE	TEST Limits	L Readout
120Hz	PRL	Calibrated Value ± 3 counts	
120112	SER	Calibrated Value ± 4 counts	
1kHz	PRL	Calibrated Value ± 5 counts	
IRIIZ	SER	Calibrated Value ± 4 counts	
10kHz	PRL	Calibrated Value ± 5 counts	
TUKITZ	SER	Calibrated Value ± 4 counts	

#### Note

Error caused by stability of standard component is not taken into account for test limits in table above. If this test fails, refer to Section VIII, Service.

#### 4-14. INTERNAL DC BIAS SOURCE TEST.

#### DESCRIPTION:

This test verifies that the internal dc bias source will apply the specified bias values to the device under test.

### SPECIFICATIONS:

DC bias, Internal Source:

 $1.5V \pm 5\%$ ,  $2.2V \pm 5\%$ ,  $6V \pm 5\%$ 

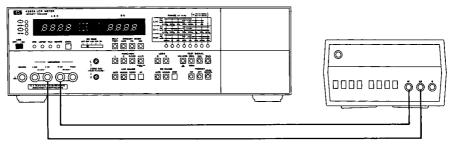


Figure 4-6. Internal DC Bias Source Test Setup.

### **EQUIPMENT:**

DC Voltmeter . . . . . . . . . . . . . . . . . HP 5300A/w5306A

#### PROCEDURE:

- 1. Connect DC Voltmeter to 4262A UNKNOWN terminals as shown in Figure 4-6.
- 2. Set 4262A controls as follows:

#### Note

Do not connect anything to UNKNOWN terminals.

3. Test limits are shown below. Read dc voltmeter output with DC BIAS switch set as follows:

DC BIAS Switch Setting	Test Limits	Voltmeter Readout
1.5V	1.425V thru 1.575V	
2.2V	2.09 V thru 2.31 V	
6 V	5.7 V thru 6.3 V	

### Note

Reading error caused by measurement error of test equipment is not taken into account for test limits in table above.

4. If tests fail, proceed to Troubleshooting in Section VIII.

## 4-15. OFFSET ADJUSTMENT TEST.

### DESCRIPTION:

This test checks that both C and L ZERO ADJ controls can be set (over their specified ranges) to respectively offset the stray capacitance and residual inductance of test jig.

### SPECIFICATIONS:

Offset Adjustment:

C:up to 10pF

L:up to  $1\mu H$ 

## EQUIPMENT:

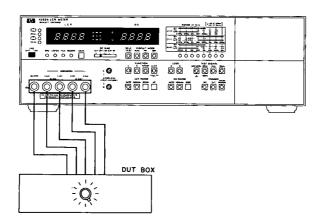


Figure 4-7. Offset Adjustment Test Setup.

#### PROCEDURE:

## (1) C ZERO ADJ test.

- 1. Connect shorting bars at 4262A UNKNOWN terminals for doing a two terminal measurement. Connect no DUT to unknown terminals (open).
- 2. Set 4262A controls as follows:

DC BIASOFF
CIRCUIT MODE AUTO
FUNCTION C
LOSS D
TEST SIGNAL 10kHz
LCR RANGEMANUAL
(Set to 10pF range)
DQ RANGE AUTO
TRIGGER INT

- 3. Rotate C ZERO ADJ control fully cw.
- 4. Verify that capacitance readout on 4262A LCR display is within 0.00 to 0.30 counts.
- 5. Disconnect shorting bars from 4262A UNKNOWN terminals and connect Test Leads (HP P/N 16361-61605) between 4262A UNKNOWN terminals and 16362A DUT Box as shown in Figure 4-7.

#### Note

If 16362A is not available, connect an 18pF capacitor (HP P/N 0160-2263) directly to UNKNOWN terminals (without disconnecting shorting bars).

- 6. Set 16362A LCR RANGE to 19pF.
- 7. Note capacitance readout on 4262A LCR display.
- 8. Rotate C ZERO ADJ control fully ccw.
- 9. Verify that capacitance readout on 4262A LCR display reduces count more than 10.30 counts as compared to count obtained in step 7.
- 10. Remove Test Leads (or DUT) from UNKNOWN terminals.
- (2) L ZERO ADJ test
  - 11. Set 4262A FUNCTION to L.
  - 12. Connect shorting bars on 4262A UNKNOWN terminals for doing a two terminal measurement. Connect a shorting lead to UNKNOWN terminals so that H and L terminals are short circuited.
  - 13. Rotate L ZERO ADJ control fully cw.
  - 14. Verify that inductance readout on 4262A LCR display is within 0.00 and 0.02 counts.
  - 15. Disconnect shorting lead from 4262A UNKNOWN terminals and connect a 5.6 $\mu$ H inductor (HP P/N 9100-1618) directly to UNKNOWN terminals as a DUT (without disconnecting shorting bars).
  - 16. Note inductance readout on 4262A LCR display.
  - 17. Rotate L ZERO ADJ control fully ccw.
  - 18. Verify that inductance readout on 4262A LCR display reduces count more than 1.02 counts as compared to count obtained in step 16.

### 4-16. COMPARATOR TEST (OPTION 004 ONLY).

#### DESCRIPTION:

This test verifies that the built-in 5 digit digital comparator makes the correct comparison between the digits set into the thumbwheel switch and the displayed counts. Comparison output data at COMPARATOR OUTPUT connector (rear panel) is also checked by this test.

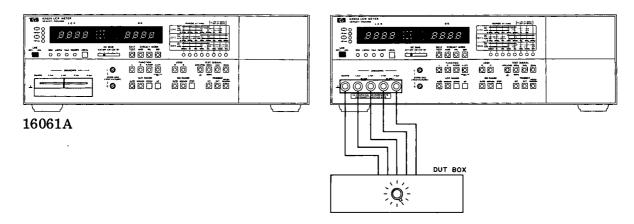


Figure 4-8. Comparator Test Setup.

### EQUIPMENT:

#### PROCEDURE:

- 1. Connect Test Leads (HP P/N 16361-61605) between 4262A UNKNOWN terminals and 16361A DUT Box as shown in Figure 4-8. If DUT Box is not available, attach 16061A Test Fixture to 4262A UNKNOWN terminals and use a 100pF capacitor as a DUT.
- 2. Set 4262A controls as follows:

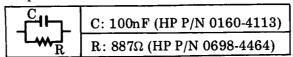
DC BIASOFI	7
CIRCUIT MODE AUTO	)
FUNCTION	
TEST SIGNAL 1kH	
LCR RANGE AUTO	)
TRIGGER INT	7

- 3. Set 16361A LCR RANGE to 100pF.
- 4. Push COMPARATOR ENABLE button (simultaneously, the LCR RANGE and DQ RANGE will be automatically changed to MANUAL).
- 5. Set LCR HIGH LIMIT switch to "1000" and LOW LIMIT switch to "0950".
- 6. Verify HIGH and LOW LIMIT settings by pushing and holding upper LIMIT CHECK pushbutton.
- 7. Adjust ZERO ADJ C control for a display reading of "949" (or less) counts.

- 8. LOW lamp should be lit. Verify circuit configuration on COMPARATOR OUT-PUT connector (J6) according to Figure 4-9.
- 9. Adjust ZERO ADJ C control cw for a display reading of "950" (up to "999").
- 10. IN lamp should be lit. Verify relay contact and TTL output as in step 8.
- 11. ADJUST ZERO ADJ C control cw for a display reading of "1000" or more.
- 12. HIGH lamp should be lit. Verify relay contact and TTL output as in step 8.
- 13. Set 16361A LCR RANGE to D = 1.8 and 4262A LCR RANGE manually to  $1\mu$ F.

### Note

If HP 16361A is not available, use a D factor sample as shown below.



14. Push D/Q RANGE AUTO button.

#### Note

The 4262A D/Q RANGE is automatically set to an appropriate range and successively reset to MANUAL.

15. Set appropriate numbers into D/Q LIMIT switches. Change the set numbers and check comparison outputs with Figure 4-9.

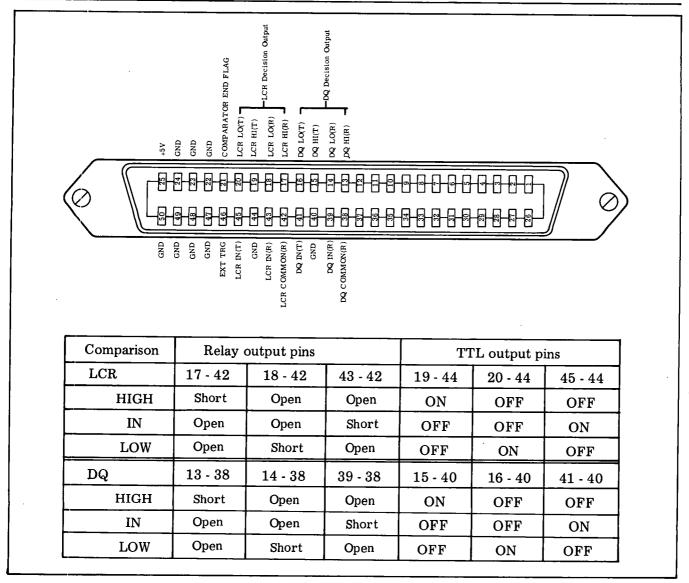


Figure 4-9. DATA OUTPUT (J6) comparator output data format.

### 4-17. HP-IB INTERFACE TEST (OPTION 101 ONLY).

#### **DESCRIPTION:**

This test verifies that the HP-IB circuitry has the capability to correctly communicate between external HP-IB devices and the 4262A through the interface bus cable.

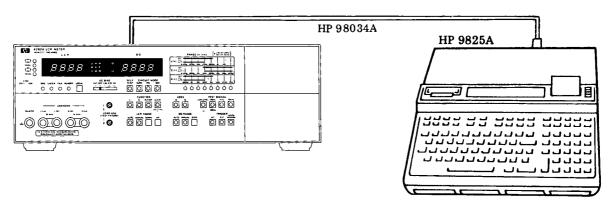


Figure 4-10. HP-IB Interface Test Setup.

#### **EQUIPMENT:**

#### PROCEDURE:

- 1. Connect 98034A Interface Card with cable between 9825A I/O slot and 4262A rear panel HP-IB connector. Install required ROM blocks in 9825A ROM slots.
- 2. Set 98034A Select Code Switch dial to select code 7 (using a screwdriver).
- 3. Set 4262A rear panel ADDRESS switch to address number 17 in binary code (refer to Paragraph 3-68).
- 4. Load test program (shown on Pages 4-26 through 4-35) in calculator.
- 5. Execute the program. Check that 4262A display, calculator display, and printed data are consistent with the results described for each program.
- 6. Perform steps 4 and 5 with respect to individual test programs and verify that 4262A and calculator correctly communicate through the HP-IB interface.

#### Note

Connect appropriate sample(s) to 4262A UNKNOWN terminals as necessary (and observe whether printout is correct).

#### **TEST PROGRAM 1**

#### [PURPOSE]

This test verifies that system controller remotely sets 4262A TEST SIGNAL and TRIGGER and successively accesses the measured data for printing.

## [PROGRAMMING]

- 0: prt "MEASURED DATA
   RECEIVED"; spc 3
  1: dev "4262A",717
  2: rem 7
  3: cli 7
  4: clr "4262A"
  5: wrt "4262A","H3T3"; wait 1000
  6: trg "4262A"
  7: red "4262A",A,B
  8: flt 3
  9: prt "LCR DATA=",A,
   "DQ DATA=",B
  10: spc 3
  11: end
  \*32657
  - 0) Commands calculator to print MEASURED DATA RECEIVED and successively to space three lines.
  - 1) Defines 717 (= Interface Select Code 7, address 17) as address code for 4262A in the programming.
  - 2) Sets REN (Remote Enable) line of the Bus line to "1". Enables remote control.
  - 3) Sets IFC (Interface Clear) line of Bus line to "1". Sets interface select code 7 to its initial conditions.
  - 4) Sets 4262A to its initial conditions. (Device Clear: ref to Para 3-72).
  - 5) Addresses calculator to talk and 4262A to listen. Program code string sets device: TEST SIGNAL 10kHz, and TRIGGER to HOLD/MANUAL (ref to Para 3-69).
  - 6) Triggers 4262A (ref to Para 3-73).
  - 7) Addresses calculator to listen and 4262A to talk. Takes incoming data and stores LCR measurement data in register A and DQ data in register B (ref to Para 3-67).
  - 8) Designates printer print format and floating decimal point (3 digits below decimal point).
  - 9) Prints LCR and DQ data.
  - 10) Commands printer to line space three vertical lines to put entire recording into proper cutting position.

#### [RESULTS]

The 4262A REMOTE lamp lights. LISTEN and TALK lamps alternately light once. Calculator prints measured LCR and DQ values.

#### **TEST PROGRAM 2**

### [PURPOSE]

This test verifies that system controller sets 4262A TEST SIGNAL and TRIGGER and prints the measured data along with the 4262A functional status codes.

### [PROGRAMMING]

```
0: prt "MEASURED DATA RECEIVED "; spc 3
1: rem 7
2: cli 7
3: clr 717
4: wrt 717,"H3PlT3";wait 1000
5: trg 717
6: fmt 4b, f, 2b, f
7: red 717,A,B,C,D,E,F,G,H
8: fxd 0;prt "S=",A, "F=",B,
  "C=",C,"F=",D
9: flt 3;prt "N=",E
10: fxd 0;prt "S=",F,"F=",G
ll: flt 3;prt "N=",H
12: spc 3
13: end
*15961
```

- 3) Sets device address code 717 (4262A) for initial conditions.
- 4) Addresses calculator to talk and device of address code 717 (4262A) to listen. Program code string sets device TEST SIGNAL to 10kHz, LOW LEVEL, and TRIGGER to HOLD/MAN-UAL (ref to Table 3-60).
- 6) Designates format for data in program step 7.
- 7) Addresses calculator to listen and 4262A to talk. Takes incoming data A, B, C, D, F and G in binary code and translates them into decimal code. Takes data E and H in free field format. Stores data items in the registers specified in the variable lists.
- 8-11) Prints data in fixed or floating decimal point format. Data items are:

A: Status,
C: Circuit Mode,
E: LCR Data,
G: DQ Function,
B: Function,
D: Frequency,
F: DQ Status,
H: DQ Data.

Refer to Paragraph 3-67 and Table 3-60.

#### [RESULTS]

The 4262A REMOTE lamp lights. LISTEN and TALK lamps alternately light once. Calculator prints 4262A functional codes along with the measured LCR and DQ data.

#### **TEST PROGRAM 3**

#### [PURPOSE]

This test verifies that 4262A notifies system controller of the Request Status (RQS) and that demands of the Service Request (SRQ) are processed according to programmed service routing.

### [PROGRAMMING]

- 0: prt "MEASURED DATA RECEIVED -DATA READY RQS MODE"; spc 3 1: oni 7, "SRQ" 2: rem 7 3: cli 7 4: clr 717 5: wrt 717,"H3D1T3";wait 1000 6: trq 717 7: "LOOP":eir 7,128 8: if bit(0,B)=1;gto "READ" 9: gto "LOOP" 10: "SRQ":rds(717)→B 11: if bit(6,B)=1; jmp 2 12: prt "OTHER DEVICE SRQ"; spc 3 13: "IRET":eir 7,128 14: iret 15: "READ":red 717,A,B 16: flt 3;prt "LCR DATA=",A. "DQ DATA=",B 17: spc 3 18: end \*22913
  - 1) Designates label (SRQ) for service routing to be performed when an interrupt is set by a device on select code 7 Bus Line.
  - 5) Addresses calculator to talk and 4262A to listen. Program code string set device: TEST SIGNAL 10kHz, Data Ready RQS Mode to ON (ref to Para 3-70), and TRIGGER to HOLD/MANUAL.
  - 7) Labels LOOP. Enables Service Request to be sent from device on select code 7 Bus Line. Checks status of SRQ line on the Bus Line.
  - 8) If the last bit of Status Byte (corresponding to Data Ready — ref to Para 3-70) is 1, goes to program step 15 labeled READ.

#### Note

When status of the SRQ line becomes 1, the programming sequence phase changes from cycling through steps 7, 8, and 9 and successively goes to step 10. Steps 10 through 14 comprise the service routing to process interrupt (Service Request) phase. See Figure 4-11 for programming flow diagram.

- 10) Labels SRQ. Takes Status Byte responding to serial poll of calculator and stores data in register B.
- 11) Verifies that SRQ YES/NO line of Status Byte is actually 1 (ref to Para 3-70).

- 13) Again enables acceptance of SRQ from device because SRQ is disabled when Status Byte signal transfer is completed (re to Para 3-70).
- 14) After service subroutine is completed, return to the step that follows step 7, 8, or 9 as appropriate to main programming sequence.

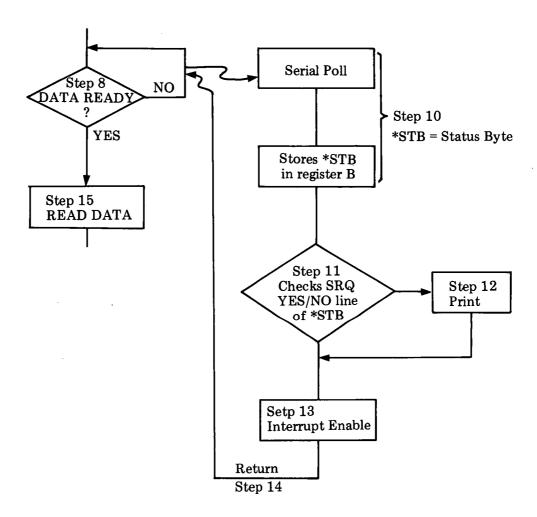


Figure 4-11 SRQ Service Routing.

## [RESULTS]

Calculator prints LCR and DQ values of the sample measured by 4262A (test frequency 10kHz). Verifies that 4262A SRQ lamp lights momentarily. Press calculator RUN button again to repeat checks. If calculator prints OTHER DEVICE SRQ, interface is faulty.

### **TEST PROGRAM 4**

### [PURPOSE]

This test confirms that 4262A FUNCTION, LOSS, and TEST SIGNAL functions are fully controlled by system controller.

## [PROGRAMMING]

#### Annotation is omitted.

```
0: prt "ENTER REMOTE PROGRA1 CODE ";spc 3
1: fmt 1,4f1.0
2: rem 7
3: cli 7
4: clr 717
5: ent "FUNCTION?(1,2,3)",A
6: ent "LOSS?(1,2)",B
7: ent "FREQUENCY?(1,2,3)",C
8: wrt 717.1,"F",A,"L",B,"H",C,"T3";wait 1000
9: trg 717
10: red 717,D,E
11: flt 3;prt "LCR DATA=",D,"DQ DATA=",E
12: spc 3
13: end
*31495
```

## [RESULT]

The 4262A REMOTE lamp lights. LISTEN and TALK lamps alternately light once. Calculator prints LCR and DQ values. Confirms that 4262A functions were correctly set (check the printed data).

### **TEST PROGRAM 5**

### [PURPOSE]

This test verifies that 4262A self test function can be remotely controlled.

### [PROGRAMMING]

```
0: prt "REMOTE SELF TEST"; spc 3
1: .oni 7, "SRQ"
2: rem 7
3: cli 7
4: clr 717
5: wrt 717, "S1"
                                      5) Addresses calculator to talk and 4262A to listen.
6: "LOOP":eir 7,128
                                         Sets device to SELF TEST mode.
7: if bit(2,A)=1;dsp "PASS"
3: if bit(3,A)=1;dsp "FAIL 1"
                                      7, 8, 9, 10)
9: if bit(4,A)=1;dsp "FAIL 2"
                                         Checks status of the third through sixth bit of
10: if bit(5,A)=1;dsp "FAIL 3"
                                         Status Byte signal and displays its contents (ref
ll: gto "LOOP"
                                         to Para 3-70).
12: "SRQ":beep;rds(717) →A
13: if bit(6,A)=1;gto "IRET"
14: prt "OTHER DEVICE
                                     12) Labels SRQ. Takes Status Byte responding to
    SRQ"; spc 3
                                         serial poll of calculator and stores data in regis-
                                         ter A. Simultaneously beeps in announcement.
15: "IRET":eir 7,128
16: iret
17: end
*14058
[RESULT]
```

The 4262A performs self test. Letters "PASS" flash on both 4262A and calculator displays.

## **TEST PROGRAM 6**

### [PURPOSE]

This test verifies that system controller takes the incoming data in character (ASCII) code and prints the data in accord with the format shown in Paragraph 3-67.

## [PROGRAMMING]

```
0: prt "RECEIVING MEASURED DATA when using STRING-ADV. ROM"; spc 3
1: dim A$[25]
2: rem 7
3: cli 7
                                      1) Establish dimension of 25 character memory
4: clr 717
                                         capacity for using string variables.
5: wrt 717, "H3T3"; wait 1000
6: trg 717
7: red 717,A$
8: prt A$
9: spc 3
10: end
                                      7) Takes incoming data (measured data) in charac-
*671
                                        ter (ASCII) code.
                                      8) Prints data in character code.
```

## [RESULT]

The measured data and 4262A functional status code are printed in accord with the format shown in Paragraph 3-67.

#### **TEST PROGRAM 7**

### [PURPOSE]

This test verifies that 4262A FUNCTION, FREQUENCY and TRIGGER can be controlled in character (ASCII) code and that the measured data is printed in accord with the format shown in Paragraph 3-67.

## [PROGRAMMING]

#### Annotation is omitted.

```
0: prt "ENTER REMOTE PROGRAM CODE when using STRING-ADV ROM";spc 3
1: dim A$[20],B$[25]
2: rem 7
3: cli 7
4: ent "PROGRAM CODE ? (as F2H3T3)",A$
5: wrt 717,A$; wait 1000
6: trg 717
7: red 717,B$
8: prt B$
9: spc 3
10: end
*3337
```

## [RESULTS]

The 4262A REMOTE lamp lights. LISTEN and TALK lamps alternately light once. Calculator prints LCR and DQ values. Confirms that 4262A functions were correctly set (check the printed data).

#### **TEST PROGRAM 8**

### [PURPOSE]

This program checks function of 4262A ADDRESS switch (rear panel) and verifies that the address code set into the switch provides access to the 4262A by the system controller.

#### Note

To perform this test, set ADDRESS switch (ref to Para 3-68) according to calculator display and, after setting the switch, press calculator CONT button.

### [PROGRAMMING]

Annotation is omitted.

```
0: prt "REM ADDRESS TEST"; spc 3
1: dsp "Set up SW *ADDRESSABLE "; beep; stp
2: rem 7
3: cli 7; clr 7
4: dsp "Set up A5-A1=00000"; beep; stp
5: 700 → A; gsb "CHK"
6: dsp "Set up A5-Al=00001"; beep; stp
7: 701 → A; gsb "CHK"
8: dsp "Set up A5-A1=00010";beep;stp
9: 702→A;gsb "CHK"
10: dsp "Set up A5-A1=00100"; beep; stp
11: 704+A;gsb "CHK"
12: dsp "Set up A5-A1=01000"; beep; stp
13: 708 → A; gsb "CHK"
14: dsp "Set up A5-A1=10000"; beep; stp
15: 716→A;gsb "CHK"
16: dsp "Set up A5-A1=10001"; been; stp
17: 717+A;gsb "CHK"
18: prt "TEST END"; spc 3
19: end
20: "CHK":dsp "Check *LISTEN=1 *REMOTE=1"; beep; wrt A; wait 2000
21: dsp "Check *TALK=1 *REMOTE=1"; beep; red A; wait 2000
22: cli 7
23: ret
*11359
```

### [RESULT]

Both 4262A LISTEN and REMOTE lamps illuminate for two seconds. Successively, both TALK and REMOTE lamps light for two seconds. Calculator prints TEST END.

#### **TEST PROGRAM 9**

Checks that 4262A functions change at intervals of 1 second as follows:

```
0: prt "REMOTE/LOCAL TEST"; spc 3
1: cli 7
2: rem 7
3: 110 7
4: beep; clr 717; wrt 717, "FlH1"; 1) FUNCTION: L, TEST SIGNAL: 120Hz.
   wait 1000
                                     2) FUNCTION: C, CIRCUIT MODE: PRL, TEST
5: beep; lcl 717; wait 1000
                                      SIGNAL: 1kHz, LOSS: Q, TRIGGER: EXT.
6: beep; wrt 717, "F2C2H2L2T2";
   wait 1000
                                     3) FUNCTION: R/ESR, CIRCUIT MODE: SER,
7: beep; lcl 7; wait 1000
                                       TEST SIGNAL: 10kHz, TRIGGER: HOLD/
8: rem 7
                                      MANUAL.
9: beep; wrt 717, "F3C3H3T3";
                                      Calculator prints TEST END.
   wait 1000
10: clr 717
                                                       Note
11: cli 7
12: lcl 7
                                        llo in step 3: Local Lockout; causes 4262A
13: prt "TEST END"; spc 3
                                        LOCAL function to be invalid.
14: end
*15032
```

## **TEST PROGRAM 10**

Checks that 4262A range indicator lamps light (in turn) each for 1 second.

```
0: prt "REMOTE RANGING TEST"; spc 3
1: fmt 1,f1.0
2: rem 7
3: cli 7
4: clr 717
5: l+A
6: "LOOP":wrt 717.1,"R",A
7: beep; wait 1000
8: if (A+1+A) #9; gto "LOOP"
9: clr 717
10: prt "TEST END"; spc 3
11: end
*6328
```

Hewlett-Packard

Model 4262A

LCR METER

Serial No.\_\_\_\_\_

Paragraph	Test	Results		
Number	Test	Minimum	Actual	Maximum
4-9	MEASUREMENT FREQUENCY			
	TEST 120Hz	116.4		123.6
	1kHz	970		1030
	10kHz	9700		10300
4-10	CAPACITANCE ACCURACY TEST			
	120Hz PRL LOW LEVEL			
	100pF	C. V. * - 4 counts		C. V. + 4 cour
	1600pF	C. V 8 counts		C. V. + 8 cour
	10nF	C. V 5 counts		C. V. + 5 cour
	100nF	C. V 5 counts		C. V. + 5 cour
	1000nF	C. V 5 counts		C. V. + 5 cour
	10μF	C. V 5 counts		C. V. + 5 cour
	120Hz PRL 1V 100pF	C. V 2 counts		C. V. + 2 cou
	1000pF	C. V 3 counts	<del></del>	C. V. + 3 cour
	10nF	C. V 3 counts		C. V. + 3 cou
	100nF	C. V 3 counts	·	C. V. + 3 cou
	1000nF	C. V 3 counts		C. V. + 3 cou
	10μF	C. V 3 counts		C. V. + 3 cour
	120Hz SER 1V 100nF	C. V 3 counts		C. V. + 3 cour
	1000nF	C. V 5 counts		C. V. + 5 cour
	10μ F	C. V 5 counts		C. V. + 5 cour
	100μF	C. V 7 counts		C. V. + 7 cour
	10mF	C. V 12 counts	<del>.</del>	C. V. + 12 cou
	1kHz PRL LOW LEVEL			
	100pF	C. V8 counts		C. V. + 8 cour
	1000pF	C. V5 counts		C. V. + 5 cour
	10nF	C. V5 counts		C. V. + 5 cour
	100nF	C. V5 counts		C. V. + 5 cour
	1000nF	C. V5 counts		C. V. + 5 cour

<sup>\*</sup>C. V. = Calibrated Value.

Paragraph	m - 1	Results		
Number	Test	Minimum	Actual	Maximum
4-10	CAPACITANCE ACCURACY TEST (Continued)			
	1kHz PRL 1V 100pF	C. V 3 counts		C. V. + 3 counts
	1000pF	C. V 3 counts		C. V. + 3 counts
1	10nF	C. V 3 counts		C. V. + 3 counts
	100nF	C. V 3 counts		C. V. + 3 counts
	1000nF	C. V 3 counts		C. V. + 3 counts
	1kHz SER 1V 10nF	C. V 3 counts		C. V. + 3 counts
	100nF	C. V 5 counts		C. V. + 5 counts
	1000nF	C. V 5 counts		C. V. + 5 counts
	$10\mu\mathrm{F}$	C. V 5 counts		C. V. + 5 counts
	$1000 \mu  ext{F}$	C. V 52 counts		C. V. + 52 counts
	10kHz PRL LOW LEVEL			
	10pF	C. V 8 counts		C. V. + 8 counts
	100pF	C. V 5 counts		C. V. + 5 counts
	1000pF	C. V 5 counts		C. V. + 5 counts
	10nF	C. V 5 counts		C. V. + 5 counts
	100nF	C. V 5 counts		C. V. + 5 counts
	10kHz PRL 1V 10pF	C. V 3 counts	<u> </u>	C. V. + 3 counts
	100pF	C. V 3 counts		C. V. + 3 counts
]	1000pF	C. V 3 counts		C. V. + 3 counts
}	10nF	C. V 3 counts		C. V. + 3 counts
	100nF	C. V 3 counts		C. V. + 3 counts
	10kHz SER 1V 1000pF	C. V 3 counts		C. V. + 3 counts
	10nF	C. V 5 counts		C. V. + 5 counts
	100nF	C. V 5 counts		C. V. + 5 counts
	1000nF	C. V 5 counts		C. V. + 5 counts
	$10\mu\mathrm{F}$	C. V 12 counts		C. V. +12 counts

<sup>\*</sup>C. V. = Calibrated Value.

Dave manh				Results	
Paragraph Number	Test		Minimum	Actual	Maximum
4-11	RESISTANCE AC	CURACY TEST			
		$1 \mathrm{k} \Omega$	C. V.*- 5 counts		C. V. + 5 counts
		$10 \mathrm{k}\Omega$	C. V 5 counts		C. V. + 5 counts
		100kΩ	C. V 5 counts		C. V. + 5 counts
		<b>10M</b> Ω	C. V 5 counts		C. V. + 5 counts
4-13	INDUCTANCE AC	CCURACY TEST (100mH)			
	120Hz	PRL	C. V 3 counts		C. V. + 3 counts
		SER	C. V 4 counts		C. V. + 4 counts
	1kHz	PRL	C. V 5 counts		C. V. + 5 counts
		SER	C. V 4 counts	<del></del>	C. V. + 4 counts
	10kHz	PRL	C. V 5 counts		C. V. + 5 counts
		SER	C. V 4 counts		C. V. + 4 counts
4-14	INTERNAL DC B TEST	IAS SOURCE			
		1.5V	1.425		1.575
		2.2V	2.09		2.31
		6 V	5.7		6.3

<sup>\*</sup>C. V. = Calibrated Value.

# SECTION V ADJUSTMENT

#### 5-1. INTRODUCTION.

5-2. This section provides the information needed to adjust the 4262A to its specifications (listed in Table 1-1). Prime purpose of adjustment is to return the instrument to its peak operating capabilities after repairs have been made. The instrument should be tested and adjusted when a part or component has been replaced. Adjustments sometimes restore an instrument to its normal operating conditions without the necessity of repairs. Adjustment procedures can also be performed periodically to maintain top operating performance. Recommended adjustment schedule for the 4262A is every 12 months. All adjustable components referred to in individual tests are summarized in Table 5-1 and adjustments locations are identified pictorially on the foldout sheets in Section VIII. If proper performance cannot be achieved after adjustment procedures have been performed, refer to troubleshooting procedures beginning with paragraph 8-42.

#### Note

Before performing any adjustments, warm up instrument for more than 60 minutes to stabilize operating conditions.

#### 5-3. SAFETY REQUIREMENTS.

5-4. Although the instrument has been designed in accordance with international safety standards, this manual contains information, cautions, and warnings which must be followed to ensure safe operation and to keep the instrument in safe condition (see Sections II and III). Adjustments described in this section should be performed only by qualified service personnel.

### WARNING

ANY INTERRUPTION OF THE PROTECTIVE (GROUNDED) CONDUCTOR (INSIDE OR OUTSIDE THE INSTRUMENT) OR DISCONNECTION OF THE PROTECTIVE EARTH TERMINAL IS LIKELY TO MAKE THE INSTRUMENT DANGEROUS. INTENTIONAL INTERRUPTION IS PROHIBITED.

- 5-5. The opening of covers for removal of parts, except those to which access can be gained by hand, is likely to expose live parts. Accessible terminals may also be live.
- 5-6. Capacitors inside instrument may still be charged even if instrument has been disconnected from its source of supply.

#### WARNING

ADJUSTMENTS DESCRIBED HEREIN ARE PERFORMED WITH POWER SUPPLIED TO THE INSTRUMENT AFTER PROTECTIVE COVERS HAVE BEEN REMOVED. ENERGY EXISTING AT MANY POINTS MAY, IF CONTACTED, RESULT IN PERSONAL INJURY.

#### 5-7. EQUIPMENT REQUIRED.

5-8. The equipment needed to adjust the Model 4262A is listed in Table 1-4 (Page 1-6). This equipment should always be calibrated to satisfy its own specifications and those of the required characteristics. If the recommended model is not available, any instrument that has specifications equal to or better than required specifications may be substituted.

#### 5-9. FACTORY SELECTED COMPONENTS.

- 5-10. Factory selected components can be recognized by an asterisk near the reference designator on the schematic diagrams in Section VIII (a nominal value is shown). Section VI, Replaceable Parts, lists the part number of the nominal value component. If the nominal value of the selected component is changed, the Manual Changes supplement, supplied with this manual, will list the change to update the manual. Table 5-2 lists all factory selected components with their nominal value ranges and their influence on instrument performance.
- 5-11. Adjustable components, with reference designators, are listed in Table 5-1. The table gives the name of the control to be adjusted and the purpose of its adjustment.

#### 5-12. ADJUSTMENT RELATIONSHIPS.

5-13. The adjustment procedures, beginning with paragraph 5-20, should be performed in step sequence as they are interactive. Neglecting or changing procedures may make it impossible to gain best 4262A performance. Table 5-4 shows alignment procedures required when repairing the instrument (replacement of a component or board). The adjustments in Table 5-4 assume that no other adjustments were attempted prior to board or component replacement.

#### 5-14. ADJUSTMENT LOCATIONS.

5-15. For reference, overall adjustment location illustrations are given in Figure 8-22. The locations of individual board assemblies are denoted in board assembly component location illustrations included on each foldout service sheet.

Table 5-1. Adjustable Components.

Table 5-1. Adjustable Components.			
Reference Designator	Name of Control	Purpose	
A9R6 (Para. 5-20)	+12V	To set output of +12V dc power supply.	
A12R1 (Para. 5-22)		To eliminate any dc offset voltage in A12 Range Resistor Amplifier in order to maximize measurement accuracy on each range.	
A12C3 (Para. 5-25)		To eliminate measurement error due to stray capacitances on A12 board assembly. Maximizes measurement accuracies of 10kHz measurement.	
A12C11 (Para. 5-26)		To properly set C ZERO ADJ control range.	
A13C1 (Para. 5-25)		To eliminate measurement error due to phase error in A12 Range Resistor Amplifier output. Maximizes measurement accuracies of 10kHz measurement.	
A13R1 (Para 5-23)	OFS-1		
A13R2 (Para. 5-23)	OFS-2	To eliminate any dc offset voltage in A13 Process Amplifier in order to maximize measurement accuracies on each range.	
A13R66 (Para. 5-23)	OFS-3		
A13R67 (Para. 5-24)	OFS-4	To adjust reference phase of phase detector to minimize measurement errors.	
A14R1 (Para. 5-24)	ZOF	To adjust timing of integrator output zero detection in order to accurately set full scale display count.	
A14R15 (Para. 5-24)	APAO	To adjust auto phase adjustment circuit output level. Minimizes measurement errors due to phase detector error.	
A23R12 (Para 5-21)	VR1	To properly set operating power voltage to nanoprocessor integrated circuit.	

Table 5-2. Factory Selected Components.

Table 5-2. Factory Selected Components.				
Reference Designator	Nominal Value Range	Effect on Performance		
A11R16	HP P/N: 0757-0440, R:FXD 7.5kΩ ► HP P/N: 0698-3259, R:FXD 7.87kΩ HP P/N: 0757-0441, R:FXD 8.25kΩ	Changes test signal level. If signal level is too high, use less resistance; if too low, use more resistance.		
A12C1 (Para. 5-23.)	HP P/N: 0160-0159, C:FXD 6800pF ► HP P/N: 0160-0160, C:FXD 8200pF HP P/N: 0160-0161, C:FXD 10000pF	Minimizes dissipation measurement error on *100nF (100μF) and *10μH (10mH) ranges at 10kHz measurement. Refer to Paragraph 5-23 (2).		
A12C2 (Para. 5-23)	► HP P/N: 0140-0190, C:FXD 39pF HP P/N: 0160-2201, C:FXD 51pF	Minimizes dissipation measurement error on 100pF (100nF) and *10mH (10H) ranges at 10kHz measurement. Refer to Paragraph 5-23 (4).		
A12C3 (Para. 5-23)	► HP P/N: 0121-0059, C:VAR 2 - 8pF HP P/N: 0121-0036, C:VAR 5.5 - 18pF	Changes adjustment range for dissipation measurement error on *10pF (10nF) and 100mH ranges at 10kHz measurement. Refer to Paragraph 5-23 (3).		
A12C14	HP P/N: 0160-2199, C:FXD 30pF ► HP P/N: 0160-2307, C:FXD 47pF	Rejects parasitic oscillation of A12U2 OP AMP in measuring $10 m\Omega$ resistor at $10 kHz$ .		
A13C1 (Para. 5-23)	► HP P/N: 0121-0059, C:VAR 2 - 8pF HP P/N: 0121-0036, C:VAR 5.5 - 18pF	Changes adjustment range for dissipation measurement error on all ranges at 10kHz measurement. Refer to paragraph 5-23 (1).		
A13C5	► HP P/N: 0160-2251 5.6pF HP P/N: 0160-2253 6.8pF	Changes the phase delay of A13U3B OP AMP.		
A13C23	►HP P/N: 0160-0134 220pF	Changes the feedback signal amount of A13U5B OP AMP.		
A14C5	► HP P/N: 0160-2307, C:FXD 47pF HP P/N: 0140-0205, C:FXD 62pF HP P/N: 0160-2202, C:FXD 75pF HP P/N: 0160-2203, C:FXD 91pF	Eliminates switching transient noise from A14 phase detector output. Nominal value is usually used.		

Note: Component marked (  $\blacktriangleright\,$  ) in table is usually used.

<sup>\*</sup> Ranges in PRL mode for capacitance and in SER mode for inductance. Values in ( ) are ranges in SER mode for capacitance and in PRL mode for inductance.

# 5-16. DUT ADJUSTMENT RECOMMENDATIONS.

5-17. If HP 16361A/16362A DUT Boxes or substitute devices are not available, user built DUT's with required characteristics may be used to adjust or to calibrate the 4262A. When it is desired to adjust the 4262A to perform to its specifications, the recommended DUT may be selected from Table 5-3. To establish accuracies appropriate for comparing the 4262A performance to its specifications, calibrate the DUT's to the accuracies given in the table. Refer to "CALIBRATION OF DUT's" (Page 4-4) for proper DUT calibration methods.

Table 5-3. DUT's Recommended for making Adjustments.

Paragraph	DUT	Component	HP Part Number	Calibration Accuracy	Required Characteristics
5-24		C: 10nF	0160-0408	0.1%	D < 0.001 at 1kHz
	<b>→</b>	C: 1000pF	0160-3766	0.1%	D < 0.001 at 1kHz
	-L <sup>C</sup>	C: 10nF R: 10kΩ	0160-0408 0698-6360	*D:0.1% (at 1kHz)	
5-25		C: 100pF R: 100kΩ	0160-0336 0698-4158	*D: 0.1% (at 10kHz)	
	<u></u>	C: 1000pF R: 10kΩ	0160-3766 0698-6360	*D: 0.1% (at 10kHz)	
	-₩ <del>-</del>	C: 10nF R: 3kΩ	0160-0408 0698-6348	*D: 0.1% (at 10kHz)	
	Ţ.	C: 100nF R: 100Ω	0160-4113 0698-6323	*D: 0.1% (at 10kHz)	
	-1 <del>-</del>	C: 100nF R: 300Ω	0160-4113 0698-6346	*D: 0.1% (at 10kHz)	
5-26	-I-W-	C: 18pF R: 8.66kΩ	0160-2263 0698-3498	*D: 0.1% (at 10kHz)	

<sup>\*</sup> For easier calibration of dissipation to the required accuracy, use accurately calibrated resistors rather than capacitors (use a high accuracy DMM to measure resistors).

# 5-18. INITIAL OPERATING PROCEDURE.

5-19. Preparatory to adjusting the 4262A, do the following to locate and to gain access to the adjustment controls. This procedure facilitates a comprehensive adjustment of instrument.

# [FUNDAMENTAL OPERATING CHECKS]

Confirm that instrument power line module is set for local power line voltage. Check front panel displays using "PRELIMINARY OPERATIONS" on Page 4-2. Offset control should be individually set for "zero" display for DUT Boxes or Test Fixtures as they are connected to 4262A UNKNOWN terminals. After attaching or interchanging test equipment, adjust front panel ZERO ADJ controls in accord with the procedure in "PRELIMINARY OPERATIONS".

# [TOP COVER REMOVAL]

#### WARNING

WHEN TOP COVER IS REMOVED LIVE PARTS ARE EXPOSED.

Remove top cover as follows:

- a. Loosen the retaining screw at rear of top cover until screw is free.
- b. Pull top cover towards the rear and lift off.

#### WARNING

TO INSURE PERSONAL SAFETY FROM POSSIBLE ELECTRICAL SHOCK HAZARDS AND RESULTANT INJURY, USE INSULATED ADJUSTMENT TOOL.

Table 5-4. Adjustment Requirements.

Assembly Repaired or Replaced	Required Adjustments
A1 (04262-66501) A2 (04262-66502) A3 (04262-66503) A4 (04262-66504) A5 (04262-66505)	None
A9 (04261-77009)	Para. 5-18
A11(04262-66511)	None
A12(04262-66512)	Para. 5-20 and 5-22 thru 5-24
A13(04262-66513)	Para. 5-21 thru 5-23
A14(04262-66514)	Para. 5-22 and 5-23
A21(04262-66521) A22(04262-66522)	None
A23(04262-66623)	Para. 5-19 (only if A23U1 is replaced)
A24(04262-66524) A25(04262-66525) A35(04262-66535)	None

# 5-20. DC POWER SUPPLY ADJUSTMENT.

#### PURPOSE:

To adjust regulated +12V DC Supply (A9).

#### Note

Only +12V DC supply can be adjusted.

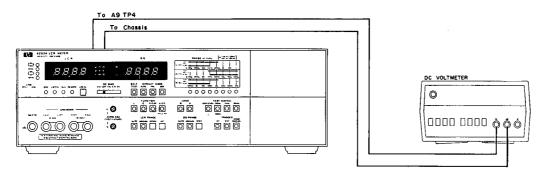


Figure 5-1. Power Supply Voltage Adjustment.

# **EQUIPMENT:**

#### PROCEDURE:

- a. Connect DC voltmeter plus input to test point A9TP4 (+12V) and minus input to 4262A chassis with dual banana plug to alligator clip cable. See Figure 5-1.
- b. Set DC Voltmeter range as appropriate for measuring +12 volts.
- c. Adjust "+12V" potentiometer A9R6 for +12 volts±0.05 volts (see Figure 8-22 for location).
- d. After adjustment of +12V, check dc voltages at test points listed below:

oltage Limits
12V ±0.15V +5V ±0.15V

e. Remove cables and DC voltmeter from 4262A.

#### Notes

1. DC supply voltage ripple should be equal to or less than the allowable limits given below.

DC supply voltage	Ripple voltage
+12V at A9TP4	< 30mVp-p
-12V at A9TP5	< 30mVp-p
+5V at A9TP6	< 50mVp-p

2. This adjustment is not affected by any other adjustment. If this adjustment fails to bring any of the output voltages to their specified values, refer to Section VIII Service Sheet No. 9 for troubleshooting.

# 5-21. NANOPROCESSOR OPERATING POWER VOLTAGE ADJUSTMENT.

#### PURPOSE:

This adjustment adjusts the operating power voltage to the nanoprocessor integrated circuit on A23 Nanoprocessor and ROM Assembly to its prescribed value.

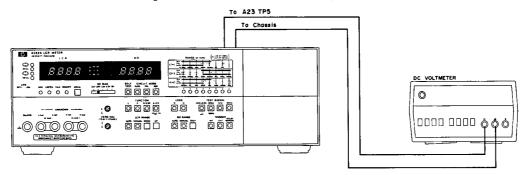
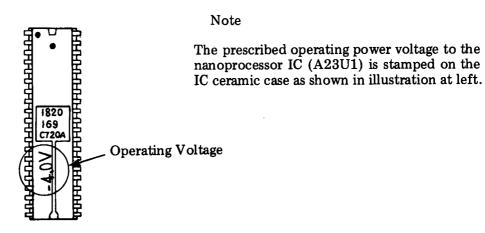


Figure 5-2. Nanoprocessor Operating Power Voltage Adjustment Location.

# **EQUIPMENT:**

# PROCEDURE:

a. Connect DC voltmeter plus input to test point A23TP4 and minus input to 4262A chassis with dual banana plug to alligator clip cable. See Figure 5-2.



- b. Set DC Voltmeter range as appropriate for measuring the prescribed operating voltage of A23U1 nanoprocessor.
- c. Adjust VR1 potentiometer A23R14 for the prescribed voltage to within ±0.1Vdc.
- d. Remove cables and DC voltmeter from 4262A.

# 5-22. A12 BOARD OFFSET ADJUSTMENT.

#### PURPOSE:

This adjustment eliminates any residual dc offset voltage from range resistor amplifier to maximize accuracy of measurement.

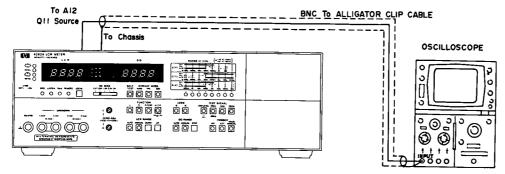


Figure 5-3. A12 Board Offset Adjustment.

# **EQUIPMENT:**

#### PROCEDURE:

a. Connect BNC to dual alligator clip cable between oscilloscope and transistor A12Q11\*source on the A12 Range Resistor Board Assembly (See Figure 5-3).

\*(Junction of A12R36 and R41)

b. Set 4262A controls as follows:

DC BIASOF	F
SELF TESTOF	F
FUNCTION	Ċ
CIRCUIT MODEPR	Ĺ
LOSS	D
TEST SIGNAL 1kH	_ [ 7.
LCR RANGE	T.
(Set to 100pF range	
DQ RANGE AUTO	ń
TRIGGERIN	Ť

c. Connect nothing (open,  $\infty$   $\Omega$ ) to UNKNOWN terminals.

#### Note

High terminals (HPOT and H  $_{\rm CUR}$ ) and Low terminals (L  $_{\rm CUR}$  and LPOT), respectively, must be connected together.

d. Set oscilloscope control as follows:

VOLTS/DIV		0.01V
TIME/DIV		. 0.5msec
TRIGGER	• • • • • • • • • • • • • • • • • • • •	INT
SWEEP MODE .	• • • • • • • • • • • • • • • • • • • •	AUTO
Input		GND

- e. Adjust position control of oscilloscope so that baseline is centered on the CRT.
- f. Set oscilloscope input mode to dc.
- g. Adjust potentiometer A12R1 until dc level of displayed waveform is 0mV ±10mV. Refer to Figure 5-4 which shows well-adjusted waveform.

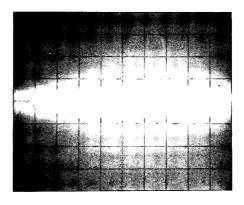


Figure 5-4. Waveform at A12Q11 Source.

Note

If adjustment is not successful, see Section VIII service sheet for troubleshooting.

# 5-23, A13 BOARD OFFSET ADJUSTMENT.

# PURPOSE:

This adjustment eliminates any residual dc offset voltage from the A13 Process Amplifier Board Assembly.

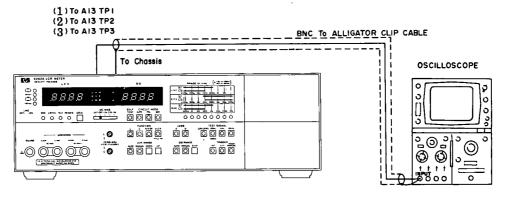


Figure 5-5. A13 Board Offset Adjustment.

# **EQUIPMENT:**

#### PROCEDURE:

#### Note

The A12 board offset adjustment (paragraph 5-22) must precede these adjustments. The adjustments in these steps can be performed separately, but steps (1) and (2) must be performed prior to step (3).

# (1) OFS - 1 ADJUSTMENT.

- a. Connect BNC to dual alligator clip cable between oscilloscope and 4262A test point A13TP1 and 4262A chassis (see Figure 5-5).
- b. Set 4262A controls as follows:

DC BIASOFF
SELF TESTOFF
FUNCTION L
CIRCUIT MODESER
LOSS
TEST SIGNAL 1kHz
LCR RANGE
(Set to 100mH range)
DQ RANGE AUTO
TRIGGER INT

- c. Short-circuit the four UNKNOWN terminals together.
- d. Set oscilloscope controls as follows:

VOLTS/DIV	0.005V
TIME/DIV	. 0.5msec
TRIGGER	INT
SWEEP MODE	AUTO
Input	GND

- e. Adjust position control of oscilloscope so that baseline is centered on the CRT.
- f. Set oscilloscope INPUT to DC.
- g. Adjust "OFS-1" potentiometer A13R1 until dc level of displayed waveform is 0mV ±1mV. Refer to Figure 5-6 which shows well adjusted waveform.

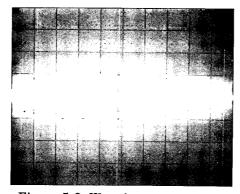


Figure 5-6. Waveform at A13TP1.

# (2) OFS - 2 ADJUSTMENT.

- a. Connect BNC to dual alligator clip cable (or 1:1 oscilloscope probe) between oscilloscope and 4262A test point A13TP2 and 4262A chassis (see Figure 5-5).
- b. Change 4262A controls as follows:

FUNCTION	C
CIRCUIT MODE	PRL
LCR RANGE	MANUAL
	(Set to 100pF range)

c. Connect nothing (open,  $\infty$   $\Omega$ ) to UNKNOWN terminals.

#### Note

High terminals (H<sub>POT</sub> and H<sub>CUR</sub>) and Low terminals (L<sub>CUR</sub> and L<sub>POT</sub>), respectively, must be connected together.

d. Adjust "OFS-2" potentiometer A13R2 until dc level of displayed waveform is within 0mV ±1mV. Refer to Figure 5-7 which shows well adjusted waveform.

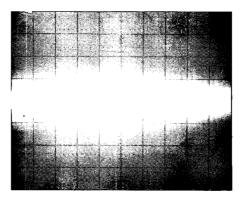


Figure 5-7. Waveform at A13TP2.

# (3) OFS -3 ADJUSTMENT.

- a. Use 10:1 oscilloscope probe for this adjustment. Connect oscilloscope probe to 4262A test point A13TP3 and ground clip lead of probe to 4262A chassis.
- b. Change 4262A controls as follows:

TEST SIGNAL	1kHz, LOW LEVEL
LCR RANGE	
	(set to 1000pF range)

c. Adjust "OFS-3" potentiometer A13R66 until dc level of displayed waveform is 0mV ±10mV. Refer to Figure 5-8 which shows well adjusted waveform.

# Note

Signal observed may be somewhat noisy. Adjust offset control so that signal is equally balanced around 0 volts dc.

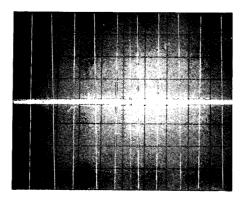


Figure 5-8. Waveform at A13TP3.

# 5-24. A14 PHASE DETECTOR & INTEGRATOR ADJUSTMENT.

#### PURPOSE:

These adjustments eliminate phase error in the phase detector and properly set timing of zero detector to minimize measurement error.

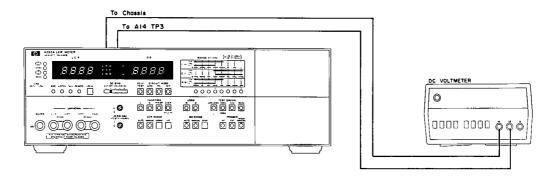


Figure 5-9. A14 Phase Detector & Integrator Adjustment.

# **EQUIPMENT:**

Note

If DUT box is not available, it is recommended that the following DUT's be used as standards:

DUT	Values of components	Calculated D (1kHz)	Required Calibration Accuracy
<u></u>	C: 10nF (HP P/N: 0160-0408)	D < 0.001	0.1%
	C: 1000pF(HP P/N: 0160-3766)	D < 0.001	0.1%
	C: 10nF (HP P/N: 0160-0408) R: 10kΩ (HP P/N: 0698-6360)	1.592	D: 0.1%

The components listed above should be calibrated before use. Refer to "Calibration of DUT's" on page 4-4 for proper DUT calibration method.

#### PROCEDURE:

# (1) OFS - 4 ADJUSTMENT.

- a. Connect DC voltmeter minus input to test point A14TP3 and plus input to 4262A chassis with dual banana plug to alligator clip cable. See Figure 5-9.
- b. Set DC voltmeter range as appropriate for measuring +3 volts.
- c. Set integrator test switch A22S1 (located at upper right on A22 Display Control and RAM Board Assembly) to TEST 1 position. See Figure 5-10 which shows location of switch S1.

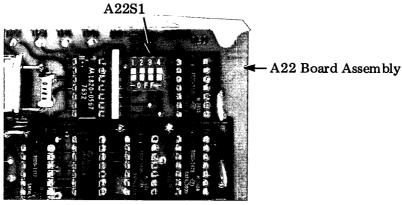


Figure 5-10. A22S1 Switch Setting.

d. Set 4262A controls as follows:

DC BIASOFF
SELF TEST
FUNCTION
CIRCUIT MODEPRL
LOSSD
TEST SIGNAL 1kHz
LCR RANGE AUTO
DQ RANGE AUTO
TRIGGER INT

e. Connect nothing (open,  $\infty \Omega$ ) to UNKNOWN terminals.

#### Note

High terminals (H POT and H CUR) and Low terminals (L CUR and L POT), respectively, must be connected together.

f. Adjust "OFS-4" potentiometer A13R67 for +2 volts ±0.5 volts (the voltage is actually negative).

# (2) ZERO DETECTOR & APAO ADJUSTMENT.

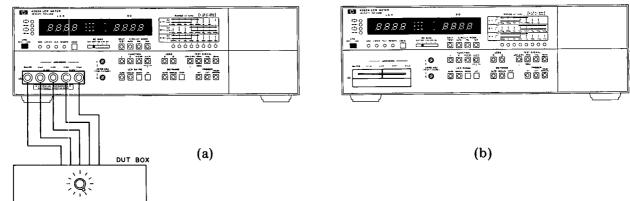


Figure 5-11. Zero Detector & APAO Adjustments.

#### Note

If DUT Box is available, use procedure A. If not, use procedure B.

#### PROCEDURE A.

- a. Adjust "ZOF" potentiometer A14R1 for 1000 counts ±1 count on 4262A LCR display.
- b. Adjust "APAO" potentiometer A14R15 for .000 to .001 count on 4262A DQ display.
- c. Set 4262A TEST SIGNAL control successively to each test frequency and test signal level shown in Table 5-5 and confirm that DC voltmeter readings are within 0 to +4 volts at each control setting. Also confirm that 4262A LCR display and DQ display are within the tolerances described in steps a and b.

Table 5-5. TEST SIGNAL Settings.

Frequency	Low Level
120Hz	off
1kHz	off
10kHz	off
120Hz	on
1kHz	on
10kHz	on

#### Note

If result of confirmation check is not satisfactory, readjust "OFS-4" potentiometer A13R67 for any voltage between +1 volt and +3 volts to satisfy the requirements of step c. If this adjustment fails to bring the voltage at A14TP3 to within its tolerance or to satisfy the confirmation check, refer to Section VIII for troubleshooting.

- d. Reset integrator test switch A22S1 to off.
- e. Connect Test Leads (HP P/N: 16361-61605) between 4262A UNKNOWN terminals and 16361A DUT Box as shown in Figure 5-11 (a).
- f. Set 16361A LCR RANGE to 1000pF.
- g. Note dissipation factor readout on DQ display.
- h. Manually change 4262A LCR RANGE to 10nF.
- i. The change in dissipation factor readout between that obtained in step g and that in step h should be less than ±1 count. If not satisfactory, readjust "ZOF" potentiometer A14R1 (step a).
- j. Set 4262A LCR RANGE to AUTO.
- k. Set 16361A LCR RANGE to D = 1.8.
- 1. Verify that DQ display count is the calibrated value of 16361A within ±3 counts. If this test fails, readjust "APAO" potentiometer A14R15 (step b).

#### PROCEDURE B.

- a. Set integrator test switch A22S1 to off.
- b. Attach HP 16061A Test Fixture to 4262A UNKNOWN terminals as shown in Figure 5-11 (b).
- c. Connect 10nF capacitor to the 16061A as DUT.
- d. Manually set 4262A LCR RANGE to 10nF.
- e. Adjust "ZOF" potentiometer A14R1 for the calibrated value of DUT ±1 count on 4262A LCR display.
- f. Adjust "APAO" potentiometer A14R15 for .000 count on 4262A DQ display.
- g. Connect a 1000pF capacitor in place of the 10nF capacitor as DUT.
- h. Adjust "ZOF" potentiometer A14R1 for 000 count on 4262A DQ display.
- i. Connect a 10nF capacitor with  $10k\Omega$  parallel resistance (D  $\approx$ 1.59) in place of the 1000pF capacitor.
- j. Adjust "APAO" potentiometer A14R15 for the calibrated D value of DUT ±2 counts on 4262A DQ display.

# 5-25. 10kHz MEASUREMENT ACCURACY ADJUSTMENT.

#### PURPOSE:

This adjustment eliminates measurement error due to stray capacitances on A12 and A13 board assemblies and maximizes measurement accuracies at 10kHz measurement.

#### Note

Each of the following adjustments are interrelated. To achieve correct adjustments, do not change adjustment procedure or sequence.

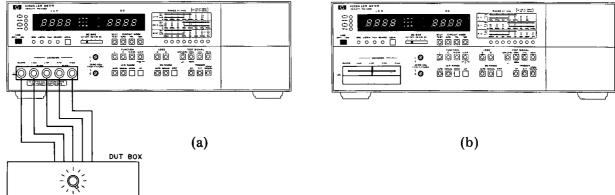


Figure 5-12. 10kHz Measurement Accuracy Adjustment.

# **EQUIPMENT:**

# Note

It is recommended that the following DUT's be used as dissipation factor standards. DUT's marked with a dot (•) in the table are included in the 16362A DUT Box.

DUT	Values of components	Calculated D (at 10kHz)	Required Calibration Accuracy
- L C1	•C1::100pF (HP P/N: 0160-0336) R1: 100kΩ (HP P/N: 0698-4158)	1.592	
	•C2: 1000pF (HP P/N: 0160-3766) R2: 10kΩ (HP P/N: 0698-6360)	1.592	
C3 R3	C3: 10nF (HP P/N: 0160-0408) R3: 3kΩ (HP P/N: 0698-6348)	1.885	D 0.1% [C 0.1%]* [R . 0.02%]
- H <sup>C4</sup> -	•C4: 100nF (HP P/N: 0160-4113) R4: 100Ω (HP P/N: 0698-6323)	1.592	
C5 R5	C5: 100nF (HP P/N: 0160-4113) R5: 300Ω (HP P/N: 0698-6346)	1.885	

<sup>\*</sup>After calibrating capacitances to within 0.1% and resistances to within 0.02%, the dissipation factor tolerance is ±0.002 for each DUT. Refer to "Calibration of DUT"s" on page 4-2 for the proper DUT calibration method.

# PROCEDURE:

# (1) A13C1 Adjustment.

- a. Connect Test Leads (HP P/N 16361-61605) between 4262A UNKNOWN terminals and 16362A DUT Box as shown in Figure 5-12 (a). If DUT Box is not available, attach 16061A Test Fixture to 4262A UNKNOWN terminals [see Figure 5-12 (b)].
- b. Set 4262A controls as follows:

DC BIASOFF
SELF TESTOFF
FUNCTION
CIRCUIT MODEPRL
LOSSD
TEST SIGNAL 10kHz
LCR RANGE AUTO
DQ RANGE AUTO
TRIGGER INT

- c. Rotate both C and L ZERO ADJ controls fully cw.
- d. Set 16362A LCR RANGE to 1000pF D = 1.8 or connect the following sample, as an alternate DUT, to 16061A:

DUT	Values of components		
r-1F-1	C: 1000pF (HP P/N: 0160-3766)		
	R: 10kΩ (HP P/N: 0698-6360)		

e. Adjust capacitor A13C1 for the calibrated value of the 16362A (or DUT) ±3 counts on 4262A DQ display.

#### Note

If this adjustment fails to bring dissipation factor readout to within the tolerance, change A13C1 to 5.5/18pF capacitor (HP P/N: 0121-0036) and try adjustment again.

# (Confirmation Check)

#### Note

If 16362A is available, perform the following check. If not, proceed to A12C1 adjustment which follows.

f. Verify that the table below is satisfied when the tests are made by changing DUT and CIRCUIT MODE (as given in table):

16362A LCR RANGE	4262A CIRCUIT MODE	Capacitance Readout	Dissipation Factor Readout
1000pF D=0.01	-THE PRI	*C. V. ± 2 counts	*C. V. ± 2 counts
1000pF D=1.8	PRL PRL	Approx. 1100 counts	*C. V. ± 3 counts
100nF D=1.8	→ SER	Approx. 500 counts	*C. V. ± 5 counts
1μF D=0.01	→ SER	*C. V. ± 2 counts	*C. V. ± 2 counts

\*C. V. = Calibrated Value of DUT.

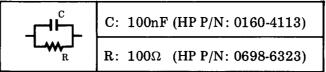
g. If table test fails, repeat step e.

# (2) A12C1 Adjustment.

#### Note

The following A12C1 Adjustment needs to be performed only when A12R4 is replaced.

a. Set 16362A LCR RANGE to 100nF D = 1.8 or connect the following sample, as an alternate DUT, to 16061A.



b. Verify that the dissipation factor readout on 4262A DQ display is the calibrated value of the DUT within a tolerance of ± 3 counts. If not within tolerance, change A12C1 to an appropriate value selected from the adjustment range below:

6800pF HP P/N: 0160-0159 8200pF HP P/N: 0160-0160 10000pF HP P/N: 0160-0161

#### Note

Nominal value is 6800pF. Increasing A12C1 by 1000pF increases display 2 counts.

# (3) A12C3 Adjustment.

- a. Remove Test Leads and attach 16061A Test Fixture to 4262A UNKNOWN terminals.
- b. Connect the following DUT to 16061A.

C R	C: 10nF (HP P/N: 0160-0408)		
71	R: 3kΩ (HP P/N: 0698-6348)		

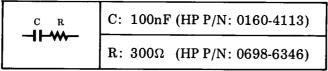
- c. Note dissipation factor readout on 4262A DQ display.
- d. Change 4262A CIRCUIT MODE to SER.
- e. Adjust A12C3 so that capacitance readout on 4262A CRL display is the calibrated value of DUT ±2 counts and the difference in dissipation factor readout between steps c and d is less than ±5 counts.

#### Note

If adjustment is not successful, change A12C3 to 5.5/18pF capacitor (HP P/N: 0121-0036) and try adjustment again.

# (4) A12C2 Adjustment.

a. Connect the following DUT to 16061A.



- b. Set 4262A CIRCUIT MODE to PRL.
- c. Note dissipation factor readout on 4262A DQ display.
- d. Change 4262A CIRCUIT MODE to SER.
- e. Verify that 4262A displays the following:
  - 1) Capacitance readout of CRL display should be the calibrated value of DUT ±2 counts.
  - 2) The difference in dissipation factor readout between steps c and d should be less than ±5 counts.
- f. If either 1) or 2) are not satisfied, change A12C2 to an appropriate value selected from the adjustment range below:

30pF	HP P/N: 0160-2139
39pF	HP P/N: 0140-0190
51pF	HP P/N: 0160-2201
62pF	HP P/N: 0140-0205

# Note

Nominal value is 39pF. Increasing A12C2 by 10pF decreases capacitance and dissipation factor readouts 2 and 3 counts respectively.

(Confirmation check)

#### Note

If 16362A DUT Box is available, use procedure A. If not, use procedure B.

#### PROCEDURE A.

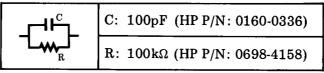
- g. Remove 16061A from 4262A UNKNOWN terminals and connect Test Leads (HP P/N: 16361-61605) between 4262A UNKNOWN terminals and 16362A DUT Box as shown in Figure 5-12 (a).
- h. Set 16362A LCR RANGE to 1pF position.
- i. Set 4262A CIRCUIT MODE to PRL.
- Adjust C ZERO ADJ potentiometer for calibrated value of 16362A on 4262A LCR display.
- k. Set 16362A LCR RANGE to 100pF D = 1.8.
- Verify that dissipation factor readout on 4262A DQ display is the calibrated value of 16362A ±5 counts.

#### Note

If this confirmation check fails, repeat A12C2 adjustment.

#### PROCEDURE B.

- g. Set 4262A CIRCUIT MODE to PRL.
- h. Connect nothing to 16061A Test Fixture.
- i. Adjust C ZERO ADJ potentiometer for 0.00 counts (10pF range) on 4262A LCR display.
- j. Connect the following DUT to 16061A.



k. Verify that dissipation factor readout on 4262A DQ display is the calibrated value of DUT ±5 counts.

#### Note

If this confirmation check fails, repeat A12C2 adjustment.

# 5-26. C ZERO ADJ CIRCUIT ADJUSTMENT (A12).

# PURPOSE:

To adjust C ZERO ADJ control range.

Note

No adjustment is required for L ZERO ADJ control.

# **EQUIPMENT:**

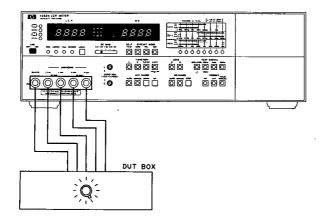


Figure 5-13. Offset Adjustment Setup.

#### PROCEDURE:

- 1. Connect Test Leads (HP P/N: 16361-61605) between 4262A UNKNOWN terminals and 16362A DUT Box as shown in Figure 5-13. If 16362A is not available, attach 16061A Test Fixture to UNKNOWN terminals.
- 2. Set 4262A controls as follows:

DC BIAS OF	F
SELF TESTOF	F
FUNCTION	C
CIRCUIT MODEPRI	
LOSS	)
TEST SIGNAL 10kH	z
LCR RANGE AUTO	)
DQ RANGE AUTO	)
TRIGGERIN	Γ

3. Set 16362A LCR RANGE to 19pF or connect the following DUT to 16061A:

<b>⊣ĭ⊢₩</b> −	C:	18pF	(HP P/N:	0160-2263)
-11- <del>-</del>	R:	8.66kΩ	(HP P/N:	0698-3498)

- 4. Note capacitance and dissipation factor readout on 4262A display.
- 5. Rotate 4262A C ZERO ADJ control ccw until capacitance readout on LCR display is half that obtained in step 4 within a tolerance of  $\pm 3$  counts.
- 6. Adjust A12C11 until dissipation factor readout becomes double that obtained in step 4 within a tolerance of ±2 counts.

#### Note

Because A12C11 and C ZERO ADJ controls interact with each other, maintain capacitance readout obtained in step 5 by controlling C ZERO ADJ until A12C11 is properly adjusted.

Section V

# SECTION VI REPLACEABLE PARTS

#### 6-1. INTRODUCTION.

6-2. This section contains information for ordering parts. Table 6-1 lists abbreviations used in the parts list and throughout the manual. Table 6-3 lists all replaceable parts in reference designator order. Table 6-2 contains the names and addresses that correspond to the manufacturer's code numbers.

#### 6-3. ABBREVIATIONS.

6-4. Table 6-1 lists abbreviations used in parts list, schematics and throughout the manual. In some cases, two forms of abbreviations are used, one in all capital letters, and one in partial capitals or no capitals. This occurs because the abbreviations in parts list are always all capitals. However, in the schematics and in other parts of the manual, other abbreviation forms with both lower case and upper case letters are used.

#### 6-5. REPLACEABLE PARTS LIST.

- 6-6. Table 6-3 is a list of replaceable parts and is organized as follows:
  - a. Electrical assemblies and their components in alphanumerical order by reference designation.
  - Chassis-mounted parts in alphanumerical order by reference designation.
  - c. Miscellaneous parts.
  - d. Illustrated parts breakdowns, if appropriate.

The information for each part includes:

- a. The Hewlett-Packard part number.
- b. The total quantity (Qty) in the instrument.

Table 6-1. List of Reference Designators and Abbreviations

			REFERENCE DESIG	GNATORS			
A	= assembly	E	= misc electronic part	P	= plug	U	= integrated circuit
В	= motor	F	= fuse	Q	= transistor	v	= vacuum, tube, neon
ВТ	= battery	FL	= filter	Ŕ	= resistor		bulb, photocell, etc.
2	= capacitor	J	= jack	RT	= thermistor	VR	= voltage regulator
СP	= coupler	к	= relay	S	= switch	W	= cable
CR	= diode	Ĺ	= inductor	T	= transformer	x	= socket
DL	= delay line	M	= meter	TB	= terminal board	Y	= crystal
DS	= device signaling (lamp)	MP	= mechanical part	TP	= test point		
			ABBREVIATI	ONS			
A	= amperes	н	= henries	NPN	= negative-positive-	RWV	= reverse working
	= automatic frequency control		= hexagonal	112.11	negative	2000	voltage
	= automatic frequency control	HG	= mercury	NRFR	= not recommended for		voltage
AMPL	= ampinier	HR	= hour(s)	NAFA	field replacement		
B. F. O.	= beat frequency oscillator	nk Hz	= hour(s) = hertz	NSR	= not separately	S-B	= slow-blow
BE CU	= beryllium copper	HZ	= nertz	NSK			= SIOW=DIOW = SCreW
ВН	= binder head	IF	= intermediate freg.		replaceable	SCR SE	
BP	= bandpass	IM PG	= impregnated			SECT	= selenium
BRS	= brass	INCD	= incandescent	OBD	= order by description		= section(s)
BWO	= backward wave oscillator	INCL	= include(s)	OH	= oval head	SEMICON	= semiconductor
		INS	= insulation(ed)	OX	= oxide	SI	= silicon
CCW	= counter-clockwise	INT	= internal		******	SIL	= silver
CER	= ceramic					SL	= slide
CMO	= cabinet mount only	k	= kilo = 1000	P	= peak	SPG	= spring
COEF	= coefficient	LH	= left hand	PC	= printed circuit	SPL	= special
COM	= common	LIN	= linear taper	p	= pico = 10 <sup>-12</sup>	SST	= stainless steel
COMP	= composition		= lock washer	PH BRZ	= phosphor bronze	SR	= split ring
COMPL	= complete	LOG		PHL	= phosphor bronze = Phillips	STL	= steel
CONN	= connector		= logarithmic taper				
CP	= cadmium plate	LPF	= low pass filter	PIV	= peak inverse voltage	TA	= tantalum
CRT	= cathode-ray tube		= milli = 10 <sup>-3</sup>	PNP	= positive-negative-	TD	= time delay
CW	= clockwise	m		- 10	positive	TGL	= toggle
		M	= meg = 10 <sup>6</sup>	P/O	= part of	THD	= thread
DEPC	= deposited carbon		= metal film	POLY	= polystyrene	TI	= titanium
DR	= drive	MET OX	= metallic oxide	PORC	= porcelain	TOL	= tolerance
FLECT	= electrolytic	MFR	= manufacturer	POS	= position(s)	TRIM	= trimmer
	= encapsulated	MINAT	= miniature	POT	= potentiometer	TWT	= traveling wave tube
EXT	= external	MOM	= momentary	PP	= peak-to-peak	• •	in accounting was a second
	= external	MTG	= mounting	PT	= point	и	$= micro = 10^{-6}$
F	= farads	MY	= ''mylar''	PWV	= peak working voltage	μ.	- 1111010 - 10
f	= femto = 10 <sup>-15</sup>		= nano = 10 <sup>-9</sup>		_	VAR	= variable
FH	= flat head	n N/C				VDCW	<ul><li>dc working volts</li></ul>
FIL H	= fillister head		= normally closed	RECT	= rectifier	w/	= with
FXD	= fixed	NE	= neon			w/ w	= with = watts
_	= giga = 10 <sup>9</sup>	NI PL	= nickel plate	RF	= radio frequency	WIV	= watts = working inverse
G		N/O	= normally open	RH	= round head or	MIA	
GE	= germanium	NPO	= negative positive zero	2110	right hand	ww	voltage = wirewound
GL	= glass		(zero temperature	RMO	= rack mount only		
GRD	= ground(ed)		coefficient)	RMS	= root-mean square	w/o	= without

- c. A description of the part.
- d. A typical manufacturer of the part in a five-digit code.
- e. The manufacturer's number for the part.

The total quantity for each part is given only once - at the first appearance of the part number in the list.

#### 6-7. ORDERING INFORMATION.

- 6-8. To order a part listed in the replaceable parts table, give the Hewlett-Packard part number, indicate the quantity required, and address the order to the nearest Hewlett-Packard office.
- 6-9. To order a part that is not listed in the replaceable parts table, state the full instrument model and serial number, the description and function of the part, and the number of parts required. Address your order to the nearest Hewlett-Packard office.

#### 6-12. DIRECT MAIL ORDER SYSTEM.

- 6-13. Within the USA, Hewlett-Packard can supply parts through a direct mail order system. Advantages of using the system are:
- a. Direct ordering and shipment from the HP Parts Center in Mountain View, California.
- b. No maximum or minimum on any mail order (there is a minimum order amount for parts ordered through a local HP Office when the orders require billing and invoicing).
- c. Prepaid transportation (there is a small handling charge for each order).
- d. No invoices to provide these advantages, a check or money order must accompany each order.
- 6-14. Mail order forms and specific ordering information is available through your local HP Office. Addresses and phone numbers are located at the back of this manual.

Table 6-2. Manufacturers Code List.

MFR NO.	MANUFACTURER NAME	ADDRESS		ZIP CODE
0024E	JERMYN INDUSTRIES			
0138Ј	AMP INC	HARRISBURG	PA	
0160G	ALLEN-BRADLEY CO	MILWAUKEE	WI	
0169H	TEXAS INSTR INC SEMICOND COMPNY DIV	DALLAS	TX	ļ
03888	KDI PYROFILM CORP	WHIPPANY	NJ	07981
0203G	MOTOROLA SEMICONDUCTOR PRODUCTS	PHOENIX	ΑZ	
0217B	AIRCO SPEER ELEK DIV AIR RDCN CO	NOGALES	ΑZ	
0223G	FAIRCHILD SEMICONDUCTOR DIV	MOUNTAIN VIEW	CA	i
07933	RAYTHEON CO SEMICONDUCTOR DIV HQ	MOUNTAIN VIEW C		94040
0248C	CTS OF BERNE INC	BERNE	IN	
0248D	CTS KEENE INC	PASO ROBLES	CA	
0291J	SIGNETICS CORP	SUNNYVALE	CA	
0299E	MEPCO/ELECTRA CORP	MINERAL WELLS	TX	1
03251	STANFORD APPLIED ENGINEERING INC	SANTA CLARA	CA	
0329B	CORNING GLASS WORKS (BRADFORD)	BRADFORD	PA	
0340F	NATIONAL SEMICONDUCTOR CORP	SANTA CLARA	CA	
0341B	CORNING GLASS WORKS (WILMINGTON)	WILMINGTON	NC	
28480	HP DIV 00 CORPORATE	PALO ALTO	CA	
0365A	MEPCO/ELECTRA CORP	SAN DIEGO	CA	
0374D	BOURNS INC TRIMPOT PROD DIV	RIVERSIDE	CA	
03 <b>7</b> 9D	ADVANCED MICRO DEVICES INC	SUNNYVALE	CA	
03791	HARRIS SEMICON DIV HARRIS-INTERTYPE	MELBOURNE	FL	
0420Ј	SPRAGUE ELECTRIC CO	NORTH ADAMS	MA	
0450G	TRW ELEK COMPONENTS CINCH DIV	ELK GROVE VLGE	IL	
72136	ELECTRO MOTIVE CORP SUB IEC	WILLIMANTIC CT		06226
73138	BECKMAN INSTRUMENTS INC HELIPOT DIV	FULLERTON	CA	92634
73899	J F D ELECTRONICS CORP	BROOKLYN	NY	11219
04678	TRW INC PHILADELPHIA DIV	PHILADELPHIA	PA	
76381	3M COMPANY	ST PAUL	MN	55101
0552D	DALE ELECTRONICS INC	COLUMBUS	NE	
28480	NO M/F DESCRIPTION FOR THIS MFG NUMBER			

Table 6-3. Replaceable Parts.

Reference Designation	HP Part Number	Qty	Description	Mfr Code	Mfr Part Number
A 1	04262-66501 04262-26501	i 1	MOTHER BOARD ASSEMBLY PC BOARD, BLANK	28480 28480	04262=66501 04262=26501
AlJi	1251+3004	1	CUNNECTOR 40-PIN M RECTANGULAR	76381	3432-2002
41 x 4 9 R 41 x 4 1 1 L 41 x 4 1 1 R 61 x 4 1 2 R 41 x 41 2 R	1251~1886 1251~1886 1251~1886 1251~1886 1251~1886		CONNECTOR=PC EDGE 15-CONT/ROW 2-ROWS	04506 04506 04506 04506 04506	252+15+30+340 252+15+30+340 252+15+30+340 252+15+30+340 252+15+30+340
A † X A 1 3 L A ‡ X A 1 5 R A 1 X A 1 4 L A ‡ X A 1 4 R A 1 X A 2 1 L A 1 X A 2 1 H	1251=1886 1251=1886 1251=1886 1251=1886 1251=1886 1251=1886		CONNECTOR-PC EDGE 15-CONT/ROW 2-ROWS	0450G 0450G 0450G 0450G 0450G	252-15-30-340 252-15-30-340 252-15-30-340 252-15-30-340 252-15-30-340 252-15-30-340
41 x 4 2 2 H 41 x 4 2 2 H 41 x 4 2 3 L 41 x 4 2 3 H 41 x 4 2 U 41 x 4 2 U R	1251-1886 1251-1886 1251-1886 1251-1886 1251-1886 1251-1886		CONNECTUR-PC EDGE 15-CONT/ROW 2-ROWS CONNECTOR-PC EDGE 15-CONT/ROW 2-ROWS	0450G 0450G 0450G 0450G 0450G	252-15-30-340 252-15-30-340 252-15-30-340 252-15-30-340 252-15-30-340 252-15-30-340
A1 XA25t. A1 XA25R	1251=1886 1251=1886		CONNECTOR-PC EDGE 15-CONT/ROW 2-ROWS CONNECTOR-PC EDGE 15-CONT/ROW 2-ROWS	0450G	252=15=30=340 252=15=30=340
4.5	04262+66502 04262+26502	1 1	KEYPOARD & DISPLAY ASSEMBLY . PC HOARD, RLANK	28480 25480	04262=66502 04262=26502
4201	0180-0291	o o	CAPACITOR=FXO 1UF+=10% 35VDC TA	0420J	1500105X9035AZ
2051 52054 2054 2050 2050 2055	1990-0#d6 1990-0486 1990-0486 1990-0486 1990-0452	37	LED-VISIBLE LUM-INT=1MCD IF=20MA=MAX LED-VISIBLE LUM-INT=1MCD IF=20MA=MAX LED-VISIBLE LUM-INT=1MCD IF=20MA=MAX LED-VISIBLE LUM-INT=1MCD IF=20MA=MAX DISPLAY-NUM SEG 1=CHAR .3-M	28480 28480 28480 28480 28480	1990-0486 1990-0486 1990-0486 1990-0486 1990-0482
42086 42087 42088 42089 42089	1990=0434 1990=0434 1990=0454 1990=0517 1990=0517	15	DISPLAY-NUM SEG 1-CHAR .3-H DISPLAY-NUM SEG 1-CHAR .3-H DISPLAY-NUM SEG 1-CHAR .3-H LED-VISIHEE LUM-INTESMED IF-20MA-MAX LED-VISIHEE LUM-INTESMED IF-20MA-MAX	28480 28480 28480 28480 28480	1990=0454 1990=0454 1990=0434 1990=0517 1990=0517
42US11 A2US12 A2US13 A2US14 A2US15	1990=0517 1990=0517 1990=0517 1990=0517 1990=0517		LED-VISIBLE LUM-INT=3MCD IF=20MA-MAX LEC-VISIBLE LUM-INT=3MCD IF=20MA-MAX LED-VISIBLE LUM-INT=3MCD IF=20MA-MAX LED-VISIBLE LUM-INT=3MCD IF=20MA-MAX LED-VISIBLE LUM-INT=3MCD IF=20MA-MAX	28480 28480 28480 28480 28480	1990-0517 1990-0517 1990-0517 1990-0517 1990-0517
A2DS16 A2DS17 A2DS18 A2DS19 A2DS20	1990 + 0517 1990 + 0517 1990 + 0517 1990 + 0517 1990 + 0434		LED-VISIBLE LUM-INTESMED IF=20MA-MAX LED-VISIBLE LUM-INTESMED IF=20MA-MAX LED-VISIBLE LUM-INTESMED IF=20MA-MAX DISPLAY-NUM SEG 1-CHAR 3-H	28480 28480 28480 28480 28480	1990-0517 1990-0517 1990-0517 1990-0517 1990-034
A2US21 A2US22 A2US23 42US24 42US24 A2US25	1990=0454 1990=0434 1990=0434 1990=0486 1990=0486		DISPLAY-NUM SEG 1-CHAR .3-H DISPLAY-NUM SEG 1-CHAR .3-H DISPLAY-NUM SEG 1-CHAR .3-H LED-VISIHLE LUM-IN1=1MCD IF=20MA-MAX LED-VISIHLE LUM-INT=1MCD IF=20MA-MAX	28480 28480 28480 28480 28480	1990-0434 1990-0434 1990-0434 1990-0486 1990-0486
A2US26 A2DS27 A2US28 A2US29 A2US30	1990+0486 1990+0486 1990-0665 1990-0665 1990-0665		LED-VISIBLE LUM-INT=IMCD IF=20MA=MAX LED-VISIBLE LUM-INT=IMCD IF=20MA=MAX LED-VISIBLE LUM-INT=IMCD IF=20MA=MAX LED-VISIBLE LUM-INT=IMCD IF=20MA=MAX LED-VISIBLE LUM-INT=IMCD IF=20MA=MAX	28480 28480 28480 28480	1990-0486 1990-0466 1990-0665 1990-0665 1990-0665
A20531 A20542 A20533 A20534 A20535	1990-0665 1990-0486 1990-0486 1990-0486 1990-0486		LED-VISIBLE LUM-INT=IMCD IF=20MA=MAX LED-VISIBLE LUM-INT=IMCD IF=20MA=MAX LED-VISIBLE LUM-INT=IMCD IF=20MA=MAX LED-VISIBLE LUM-INT=IMCD IF=20MA=MAX LED-VISIBLE LUM-INT=IMCD IF=20MA=MAX	\$8480 \$8480 \$8480	1990-0665 1990-0486 1990-0486 1990-0486 1990-0486
A2D536 A2D517 A2U538 A2U539 A2U540	1990-0486 1990-0486 1990-0486 1990-0486 1990-0665		LED-VISIBLE LUM-INTEIMCD IF=20M4-MAX LED-VISIBLE LUM-INTEIMCD IF=20M4-MAX LED-VISIBLE LUM-INTEIMCD IF=20M4-MAX LED-VISIBLE LUM-INTEIMCD IF=20M4-MAX LED-VISIBLE LUM-INTEIMCD IF=20M4-MAX	\$9480 \$9480 \$8480 \$8480 \$8480	1990-0486 1990-0486 1990-0486 1990-0486 1990-0665
1203u1 6203u3 6203u3 6203u4 6203u5	1990-0665 1990-0665 1990-0665 1990-0665 1990-0665		LED-VISIBLE LUM-INT=1MCD IF=20MA-MAX LED-VISIBLE LUM-INT=1MCD IF=20MA-MAX LED-VISIBLE LUM-INT=1MCD IF=20MA-MAX LED-VISIBLE LUM-INT=1MCD IF=20MA-MAX LED-VISIBLE LUM-INT=1MCD IF=20MA-MAX	28480 28480 28480 28480	1990-0665 1990-0665 1990-0665 1990-0665 1990-0665
		į			

Table 6-3. Replaceable Parts (Cont'd).

	T	Г -	Table 6-3. Replaceable Parts (Cont'o	1).	
Reference Designation	HP Part Number	Qty	Description	Mfr Code	Mfr Part Number
A20546 A20547 A20548 A20549 A20550	1990-0665 1990-0665 1990-0665 1990-0665 1990-0665		LED-VISIBLE LUM-INTEIMCD IFEROMA-MAX	28480 28480 28480 28480 28480	1990-0665 1990-0665 1990-0665 1990-0665 1990-0665
420551 420552 420553 420554 420555	1990-0665 1990-0665 1990-0665 1990-0665 1990-0665		LED-VISIBLE LUM-INTEIMCO IFE20MA-MAX LED-VISIBLE LUM-INTEIMCO IFE20MA-MAX LED-VISIBLE LUM-INTEIMCO IFE20MA-MAX LED-VISIBLE LUM-INTEIMCO IFE20MA-MAX LED-VISIBLE LUM-INTEIMCO IFE20MA-MAX	28480 28480 28480 28480 28480	1990-0665 1990-0665 1990-0665 1990-0665 1990-0665
420856	1990-0665		LED-VISIBLE LUM-INT#1MCD IF#20MA-MAX	28480	1990-0665
1 LSA 2 LSA 2 LSA 2 LSA 2 LSA	1200-0638 1200-0638 1200-0638 1200-0638 1200-0638	8	SOCKET-IC 14-CONT DIP-SLDR SOCKET-IC 14-CONT DIP-SLDR SUCKET-IC 14-CONT DIP-SLDR SOCKET-IC 14-CONT DIP-SLDR SUCKET-IC 14-CUNT DIP-SLDR	03251 03251 03251 03251 03251	
A2J6 A2J7 A2J8	1200-0638 1200-0638 1200-0638		SUCKET-IC 14-CONT DIP-SLOK SUCKET-IC 14-CONT DIP-SLOR SUCKET-IC 14-CONT DIP-SLOR	16250 16250 16250	
A2K1 A2R2 A2K3 A2R4 A2K5	0683-4715 0663-4715 0683-4715 0683-4715 0683-2715	37 20	RESISTOR 470 5% .25w FC TC==400/+600 FESISTOR 470 5% .25w FC TC==400/+600 RESISTOR 470 5% .25w FC TC==400/+600 RESISTOR 470 5% .25w FC TC==400/+600 RESISTOR 270 5% .25w FC TC==400/+600	0160G 0160G 0160G 0160G 0160G	CB4715 CB4715 CB4715 CB4715 CB4715 CB2715
4286 4287 4284 4289 42810	0663-2715 0663-2715 0663-2715 0663-2715 0663-2715		RESISTOR 270 5% ,25m FC TC==400/+600 RESISTOR 270 5% ,25m FC TC=-400/+600 MESISTOR 270 5% ,25m FC TC=-400/+600 RESISTOR 270 5% ,25m FC TC=-400/+600 RESISTOR 270 5% ,25m FC TC=-400/+600	0160G 0160G 0160G 0160G 0160G	CB2715 CB2715 CB2715 CB2715 CB2715
A2R11 A2H12 A2H13 A2H14 A2R15	0683-4715 0683-4715 0683-4715 0683-4715 0683-4715		##\$I\$TOM 470 5% .25% FC TC#=400/+600   #E\$I\$TOM 470 5% .25% FC TC#=400/+600   ##\$I\$TOM 470 5% .25% FC TC#=400/+600   ##\$I\$TOM 470 5% .25% FC TC#=400/+600   ##\$I\$TOM 470 5% .25% FC TC#=400/+600	01606 01606 01606 01606 01606	CB4715 CB4715 CB4715 CB4715 CB4715
A2816 A2817 A2816	0653+4715 0663+4715 0683+4715		RESISTOR 470 5% .25% FC TC==400/+600 RESISTOR 470 5% .25% FC TC==400/+600 RESISTOR 470 5% .25% FC TC==400/+600	0160G 0160G 0160G	C84715 C84715 C84715
4251 4252 4253	5060-9436 5041-0342 5060-4802 5080-3440 5060-9436 5041-0351	28 2 1 1 a	SWITCH, PUSHBUTTON  KEY CAP  SLIDE ASSEMBLY  SPRINGIPETENT  SWITCH, PUSHBUTTON  KEY CAP	28480 28480 28480 28480 28480 28480	5060-9436 5041-0342 5060-4802 5020-5440 5060-9436 5041-0351
4254 4255 4256	5000=9436 5041=0351 5060=9436 5041=0351 5060=9436 5041=0351		SWITCH, PUSHBUTTON KEY CAP SWITCH, PUSHBUTTON KEY CAP SWITCH, PUSHBUTTON KEY CAP	28480 26480 28480 28480 28480 28480	5060=9436 5041=0351 5060=9436 5041=0351 5060=9436 5041=0351
A 2 S 7 A 2 S R A 2 S 9	5060+9436 5041+0252 5060+9436 5041+0252 5060+9436	c	SWITCH, PUSHBUTTON KEY CAP SWITCH, PUSHBUTTON KEY CAP SWITCH, PUSHBUTTON	\$8480 \$8480 \$8480 \$8480	5060=9436 5041=0252 5060=9436 5041=0252 5060=9436
42510 42511	5041-0252 5060-9436 5041-0318 5060-9436 5041-0252	11	KEY CAP SWITCH, PUSHBUTTON KEY CAP SWITCH, PUSHBUTTON KEY CAP	28480 28480 28480 28480 28480	5041-0252 5060-9436 5041-0318 5060-9436
45515	5060-9436 5041-0252		SWITCH, PUSHBUTTON KLY CAP	28480 28480	5041-0252 5060-9436 5041-0252
A2513	5060+9436 5041=0318		SWITCH, PUSHBUTTON	28480	5060=9436
A2514 A2515	5060-9436 5041-0408 5041-0436 5041-0318	1	SWITCH, PUSHBUTTON KFY CAP SWITCH, PUSHBUTTON	28480 28480 28480 28480	5041-0318 5060-9436 5041-0408 5060-9436
A2516	5060-9436		KEY CAP SWITCH, PUSHBUTTON	28480 28480	5041-0316 5060-9436
A2S17 A2S18	5041-0316 5060-9436 5041-0318 5060-9436		KEY CAP SWITCH, PUSHBUTTON KEY CAP	28480 28480 28480	5041=0318 5060=9436 5041=0318
-6413	5041-0318		SWITCH, PUSHBUTTON KEY CAP	28480 28480	5060=9436 5041=0318
A2519 A2520	5060=9436 5041=0309 - 5060=9436	ч	SWITCH, PUSHBUTTON KEY CAP SWITCH, PUSHBUTTON	28480 28480 28480	5060=9436 5041=0309 5060=9436
A252)	5041=0309 5060=9436 5041=0318		MEY CAP SWITCH, PUSHBUTTON MEY CAP	28480 28480 28480	5041-0309 50-0-9436 5041-0318

Table 6-3. Replaceable Parts (Cont'd).

Reference Designation	HP Part Number	Qty	Description	Mfr Code	Mfr Part Number
A2523 A2523	5060-9436 5041-0318 5060-9436 5041-0309 5060-9436 5041-0318		SWITCH, PUSHBUTTON  KŁY CAP SWITCH, PUSHBUTTON  KŁY CAP SWITCH, PUSHBUTTON  KŁY CAP	28480 28480 28480 28480 28480 28480	5060 = 9436 5041 = 0318 5060 = 9436 5041 = 0309 5060 = 9436 5041 = 0318
42525 42526	5060~9436 5041-0318 5060-9436 5041-0318		SWITCH, PUSHBUTTON KEY CAP SWITCH, PUSHBUTTON KEY CAP	28480 28480 28480	5060=9436 5041=0318 5060=9436 5041=0318
A2U3 A2U3 A2U3	1820+1200 1820+0491 1820+0491 1820+0491	5 4	IC INV TTL LS HEX 1-INP IC DCDP TTL BCD-TO-DEC 4-TO-10-LINE IC DCDR TTL BCD-TO-DEC 4-TO-10-LINE IC DCDR TTL BCD-TO-DEC 4-TO-10-LINE	0169H 0169H 0169H 0169H	SN74L805N SN74145N SN74145N SN74145N
1 4 2 4 2 4	8120-0365 8120-0362	1 1	CABLE ASSEMBLY, 40-PIN Cable Assembly, 34-PIN	28480 28480	8120-0365 8120-0362
Δ 3	04262-66503 04262-26503	1 1	HP⇒IH CONNECTUR BOARD ASSEMBLY PC HOARD, BLANK	28480 28480	04262-66503 04262-26503
A3J1 A3J2	1251-3283 1200+0485	1 1	CUNNECTOR 24-PIN F MICRORIBBON SOCKET:IC 14-PIN PC MOUNTING	28480 28480	1251=3283 1200=0485
A351	3101-1973	1	Smitch-St 7-14-NS DIP-Stide-ASSY .14	05440	117-1028
43w1	04262-61609	1	CABLE ASSEMBLY		
A4 (OPTION 004)	04262-66544 04262-26544	1 1	THUMBWHEEL SWITCH HOARD ASSEMBLY PC HOARD, BLANK	28480 28480	
A4J1 A4J2 A4J3 A4J4 A4J5	1251-0923 1251-0923 1251-0923 1251-0923 1251-0923	16	CONNECTOR, PC 2 x 11 CONTACT	28480 28480 28480 28480 28480	
AUJ6 7 LUA 8 LUA 9 LUA 11 LUA	1251-0923 1251-0923 1251-0923 1251-0923 1251-0923		CONNECTOR, PC 2 x 11 CONTACT	28480 28480 28480 28480	
Auji; Auji; Auji; Auji; Auji;	1251-0923 1251-0923 1251-0923 1251-0923 1251-0923		CONNECTOR, PC 2 x 11 CONTACT	28480 08485 08485 28480 8480 8480	
A4J16 A4J17	1251-0923	5	CONNECTOR, PC 2 x 11 CONTACT SUCKET-IC 16-CONT DIP-SLOR	28480 0138J	
Auni	1200-0607 M120-0364	1	CABLE ASSEMBLY, FLAT	28480	8120-0364
A5 (OPTION 004)	04262-66505 04262-26505	1 1	CUMPANATOR KEYBUARD ASSEMBLY PC BOARD, BLANK	28480 28480	04262=66505 04262=26505
45051 45052 45053 45054 45084	1990-0517 1990-0521 1990-0517 1990-0517	5	LED-VISIBLE LUM-INTESMCD IF=20MA-MAX LED-VISIBLE LUM-INTE2,2MCD IF=50MA-MAX LED-VISIBLE LUM-INTE3MCD IF=20MA-MAX LED-VISIBLE LUM-INTE3MCD IF=20MA-MAX LED-VISIBLE LUM-INTE2,2MCD IF=50MA-MAX	28480 28480 28480 28480	1990-0517 1990-0521 1990-0517 1990-0517 1990-0521
ASDS 6 ASDS 7 ASS 1 ASS 2 ASS 3	1990-0517 1990-0665 5060-9436 5041-0342 5060-9436 5041-0309 5060-9436 5041-0252		LED-VISIBLE LUM-INT=3MCD IF=20MA-MAX LED-VISIBLE LUM-INT=1.0MCD IF=20MA-MAX SWITCH, PUSHBUTTON KEY CAP SWITCH, PUSHBUTTON KEY CAP SWITCH, PUSHBUTTON KEY CAP	28480 28480 28480 28480 28480 28480 28480	1990-0517 5060-9436 5041-0342 5060-9436 5041-0309 5060-9436 5041-0309
45*1	8120+0361	1	CABLE ASSEMBLY	28480	8120-0361
46			NOT ASSIGNED		
A 7			NUT ASSIGNED		
ΔR			NOT ASSIGNED		
				İ	

Table 6-3. Replaceable Parts (Cont'd).

Reference Designation	HP Part Number	Qty	Description	Mfr Code	Mfr Part Number
A 9	04261=77009 04261=87009	. !	POWER SUPPLY BOARD ASSEMBLY PC BOARD, BLANK	28480 28480	04261-77009 04261-87009
A 9C 1 A 9C 2 A 9C 3 A 9C 4 A 9C 5	0180-1057 0180-1057 0180-1057 0180-1056 0180-1056	2	CAPACITORIFXD 2200 UF 16VDCW AL ELECT CAPACITORIFXD 2200 UF 16VDCW AL ELECT CAPACITORIFXD 2200 UF 16VDCW AL ELECT CAPACITORIFXD 1000 UF 25VDC AL ELECT CAPACITORIFXD 1000 UF 25VDC AL ELECT	28480 28480 28480 28480 28480	0180=1057 0180=1057 0180=1057 0180=1056 0180=1056
4906 4907 4908 4909	0140+0200 0180-0814 0180-0814 0180+0814	3	CAPACITOR-FXD 390PF +-5% 300VDC MICAO+70 CAPACITURIFXD 100UF +100-10% 16VDCW AL CAPACITORIFXD 100UF +100-10% 16VDCW AL CAPACITORIFXD 100UF +100-10% 16VDCW AL	72136 28480 28480 28480	DM15f391J0300WV1CR 0180=0814 0180=0814 0180=0814
A9CR1 A9CR2	1901-0237 1901-0237	5	DIDDE:SI, RECTIFIER BRIDGE, 200V Didde:SI, Rectifier Bridge, 200V	28480 28480	1901-0237 1901-0237
4901 4902 4903 4904	1854-0039 5080-3078 5080-3078 5080-3078	1 20	TPANSISTOR NPN 2N3053S SI TO-39 PD=1W Transistor npn SI Transistor npn SI Transistor npn SI	0203G	2N5053
A 9R 1 A 9R 2 A 9R 3 A 9R 4 A 9R 4	0811-2771 0811-1746 0683-1025 0811-1746 0757-0436	1 20 1	RESISTOR .18 3% 3W PW fC=0+=90 RESISTOR .3 5% 2W PW fC=0+=800 RESISTOR 1K 5% .25W FC fC==400/+600 RESISTOR .36 5% 2W PW fC=0+=800 RESISTOR 5.11K 1% .125W F fC=0+=100	05520 04678 0160G 04678 03298	#S=28 hw#2-56/100=J CB1025 bw#2-36/100=J C4-1/8-T0=5111=F
49R6 49R7 49R8 49R9 49R10	2100-2521 0757-0440 0757-0289 0698-4020 0757-0442	1 1 1 1 4	RESISTOR-TRMR 2K 10% C SIDE-ADJ 1-TRN RESISTOR 7,5K 1% ,125W F TC=0+=100 RESISTOR 13,3K 1% ,125W F TC=0+=100 RESISTOR 9,53K 1% ,125W F TC=0+=100 RESISTOR 10K 1% ,125W F TC=0+=100	03654 03298 0299E 03298 03298	£150x202 C4-1/8-T0-7501-F MF4C1/8-T0-1352-F C4-1/8-T0-9531-F C4-1/8-T0-1002-F
AGR11 AGR12 AGR13 AGR14 AGR15	0757-0442 0698-3155 0698-3155 0698-3431 0757-0420	5 1 1	RESISTOR 10K 1% .125W F TC#0+=100 RESISTOR 4,64K 1% .125W F TC#0+=100 RESISTOR 4,64K 1% .125W F TC#0+=100 RESISTOR 25.7 1% .125W F TC#0+=100 RESISTOR 750 1% .125W F TC#0+=100	03298 05298 05298 03888 03298	C4-1/8-T0-1002-F C4-1/8-T0-4641-F C4-1/8-T0-4641-F PME55-1/8-T0-23R7-F C4-1/8-T0-751-F
49R16 49R17 A9R18 A9R19	0698=3427 0757=0317 0757-0159 0683-7529	1 2	RESISTOR 13.3 1% .125W F TC=0+-100 RESISTOR 1.33K 1% .125W F TC=0+-100 RESISTOR 1K 1% .5W RESISTOR 7.5K 5% .25W	888E0	PME55-1/8-10-1385-F C4-1/8-10-1331-F
A9U1 A9U3 A9U4	1826+0271 5080-3834 1826+0271 1826+0271	1	IC 741 OP AMP IC 723 V RGLTR IC 741 OP AMP IC 741 OP AMP	0340F 0340F 0340F	LM741CN LM741CN LM741CN
	5040+3304	9	A9 MISCELLANEOUS PARTS HOLDER, CAPACITOR	28480	5040 <b>~</b> \$304
A10	04261=50022	1	SUPPORTER, BOARD NOT ASSIGNED	28480	04261*50022
All	04262+66511 04262+26511	1 1	OSCILLATOR & SOURCE RESISTOR HOARD ASSY PC HOARD, BLANK	28480 28480	04202=06511 04202=26511
A11C1 A11C2 A11C3 A11C4 A11C5	0140-2396 0160-2200 0180-1051 0180-1051 0180-1052	1 1 20 4	CAPACITOR-FXD 1000UF+75=10% 75VDC AL CAPACITOR-FXD 43PF +=5% 300VDC CAPACITOR, FXD 100 UF 16V M CAPACITOR, FXD 100 UF 16V M CAPACITOR, FXD 220 UF 6.3V M	0420J 28480 28480 28480 28480	39D104G075JP4 0180-2200 0180-1051 0180-1051 0180-1052
A11Cb A11C7 A11C8 A11C9	0180-1051 0180-1051 0160-5819 0160-5819	3	CAPACITUR, FXO 100 UF 16V M CAPACITUR, FXD 100 UF 16V M CAPACITOR, FXD 3300°FF 50V CAPACITOR, FXD 3300°FF 50V	28480 28480	0160-1051 0160-1051
A11C10	0180-0228	5	CAPACITUR-FXD 22UF+-10% 15VDC TA  CAPACITUR-FXD 22UF+-10% 15VDC TA	0420J	1500226×901562 1500226×901562
Alicia	0180-1052	3	CAPACITOR, FXD 220 UF 6.3V M DIUDE-ZNR 53.6V 2x DD-15 PD=1w 1C=+.081%	28480 0203G	0180-1052
A11CR2 A11CR3 A11CR4 A11CR5	1901+0025 1901+0025 1901+0025 1901+0025	10	DIODE-GEN PRP 100V 200MA DO-7	58480 58480 58480 5480	1901-0025 1901-0025 1901-0025 1901-0025
A11CR6 A11CR7 A11CR8 A11CR9 A11CR10	1901-0040 1901-0040 1902-3037 1902-3149 1901-0040	1 6	DIODE-SWITCHING 30V 50MA 2NS D0-35 DIODE-SWITCHING 30V 50MA 2NS D0-35 DIODE-ZNR 3.16V 2% 00=7 PD=4W TC=064% DIODE-ZNR 9.09V 5% D0-7 PD=4W TC=+.057% DIODE-SWITCHING 30V 50MA 2NS D0-35	28480 28480 0203G 0223G 28480	1901-0040 1901-0040 FZ7256 1901-0040
Alicrii Alicriz Alicria Alicria Alicria	1901-0040 1901-0040 1901-0040 1902-0688 1902-0688		DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-ZNR 53.6V 2% DO-15 PD=1W	26460 26460 28460	1901-0040 1901-0040 1901-0040
WildWid	, , , , , , , , , , , , , , , , , , , ,				

Table 6-3. Replaceable Parts (Cont'd).

Reference Designation	HP Part Number	Qty	Description	Mfr Code	Mfr Part Number
AJIK1 AJIK2 AJIK3 AJIK4	0490-0234 0490-0234 0490-0234 0490-0276	3	RELAY, REED RELAY, REED RELAY, REED RELAY; REED	28480 28480 28480 28480	0490-0226
A1101 A1102 A1103 A1104 A1105	5080-3078 1453-0020 5080-3078 1855-0082 5080-3078	26 1	TRANSISTOR NPN SI PD#300MW FT#200MMZ TRANSISTOR PNP SI PD#300MW FT#150MMZ TRANSISTOR NPN SI PD#300MW FT#200MMZ TRANSISTOR MOSET P#CHAN D=MODE SI TRANSISTOR NPN SI PD#300MW FT#200MMZ	\$8480 \$8480 \$8480 \$8480 \$8480	1854-0071 1854-0020 1854-0071 1855-0082 1854-0071
A1106 A1107 A1108 A1109 A11010	5080 - 3078 5080 - 3078 5080 - 3078 5080 - 3830 5080 - 3830	22	TRANSISTOR NPN SI PD#300MW FT#200MMZ TRANSISTOR NPN SI PD#300MW FT#200MMZ TRANSISTOR NPN SI PD#300MW FT#200MMZ TRANSISTOR J=FET N=CHAN D=MODE SI TRANSISTOR J=FET N=CHAN D=MODE SI	28480 28480 28480	1854-0071 1854-0071 1854-0071
A11011 A11012 A11013 A11014 A11015	1855-0268 1855-0268 1853-0020 5080-3078 1853-0020	9	TRANSISTOR J=FET N=CHAN D=MODE SI TRANSISTOR J=FET N=CHAN D=MODE SI TRANSISTOR PNP SI PD=300Mw FT=150MHZ TRANSISTOR NPN SI PD=300Mw FT=270MHZ TRANSISTOR PNP SI PD=300Mw FT=150MHZ	28480 28480 28480 28480 28480	1885-0268 1885-0268 1855-0020 1854-0071 1853-0020
A11016	1853-0020		TRANSISTOR PNP SI PD#300MW FT#150MHZ	28480	1853-0020
A11R1 A11R2 A11R3 A11R4 A11R5	0768=0001 0683=3335 0698=4418 0683=5605 0683=5605	1 1 23	RESISTOR 1K 10% 3W MD TC=0+-250 RESISTOR 33K 5% ,25W FC TC=-400/+800 RESISTOR 205 1% ,125W F TC=0+-100 RESISTOR 56 5% ,25W FC TC=-400/+500 RESISTOR 56 5% ,25W FC TC=-400/+500	03418 0160G 03298 0160G 0160G	FP3-3-250-1001-K CH3335 C4-178-70-205R+F CH5605 CH5605
A11R6 A11R7 A11RA A11R0 A11R10	0757-0465 0757-0442 0698-0083 0698-0083 0757-0405	2	RESISTOR 100K 1% ,125W F TC=0+=100 RESISTOR 10K 1% ,125W F TC=0+=100 RESISTOR 1,96K 1% ,125W F TC=0+=100 RESISTOR 1,96K 1% ,125W F TC=0+=100 RESISTOR 162 1% ,125W F TC=0+=100	03298 03298 03298 03298 03298	C4-1/8-T0-1003-F C4-1/8-T0-1002-F C4-1/8-T0-1901-F C4-1/8-T1-1961-F C4-1/8-T0-102K-F
A11R11 A11R12 A11R13 A11R14 A11R15	0757-0405 0683-2705 0683-2705 0683-1535 0683-1535	2	RESISTOR 162 1% .125W F TC=0+-100 RESISTOR 27 5% .25W FC TC=-400/+500 RESISTOR 27 5% .25W FC TC=-400/+500 RESISTOR 15K 5% .25W FC TC=-400/+800 RESISTOR 15K 5% .25W FC TC=-400/+800	0329H 0160G 0160G 0160G 0160G	C4-1/8-T0-162A-F C82705 C82705 C81705 C81535 C81535
A11R164  A11R17  A11R18  A11R19	0698-4471 0757-0442 0698-4420 0698-4442	1 2	RESISTOR 7.87K 1% .125W F TC=0+=100 *FACTORY SELECTED PART RESISTOR 10K 1% .125W F TC=0+=100 HESISTOR 226 1% .125W F TC=0+=100 RESISTOR 4.42% 1% .125W F TC=0+=100	03298 03298 03298	C4-1/8-10-1002-F C4-1/8-10-220R-F C4-1/8-10-4421-F
A11R20 A11R21 A11R22 A11R23	0698-3155 0757-0276 0683-3335 0757-0281 0683-3335	1	RESISTOR 4,64k 1% .125W F TC=0+=100 RESISTOR 1.78K 1% .125W F TC=0++100 PESISTOR 33K 5% .25W FC TC==400/+800 RESISTOR 2.74K 1%.25W F TC=0+-100 RESISTOR 33K 5% .25W FC TC==400/+800	03298 03298 0160G	C4=1/8=10=4641=F C4=1/8=10=1781=F C83335 C83335
A11R24 A11R25 A11R26 A11R27 A11R28	0698#4498 0698=1427 0757-0437 0757-0459	5	PESISTOR 53.6% 1% 125% F TC=0+-100 RESISTOR 400% .5% .25W RESISTOR 4.75% 1% .125% F TC=0+-100 RESISTOR 56.2% 1% .125% F TC=0+-100 RESISTOR 400% .5% .25W	05298 25480 28480	C4-1/A-T0-5362-F 0698-1427
A11R30 A11R31 A11R31 A11R32 A11R33 A11R34	0698-1427 0698-4444 0683-8225 0683-4725 0683-3335 0757-0443	1 13 1	HESISTOR 4.87K 1% ,125W F TC=0+=100 RESISTOP 8.2K 5% ,25W FC TC=+400/+700 HESISTOR 4.7K 5% ,25W FC TC==400/+700 RESISTUM 33% 5% ,25W FC TC==400/+R00 RESISTOR 11K 1% ,125W F TC=0+-100	0160G 0160G 0160G 03298	C6H225 C64725 C63535 C4-1/8-T0-1102-F
A11R35 A11R36 A11R37 A11R38 A11R39	0757-0416 0698-3154 0663-5625 0683-3335 0683-7525	3 1 11	RESISTOR 511 1X ,125W F TC=0+-100  PESISTUR 4,22K 1X ,125W F TC=0+-100  RESISTOR 5,6K 5X ,25W FC TC=-400/+700  RESISTOR 35K 5X ,25W FC TC=-400/+00  RESISTOR 7,5K 5X ,25W FC TC=-400/+700	03298 03298 0160G 0160G 0160G	C4-1/6-10-511R-F C4-1/A-T0-4221-F C45625 C43335 C87525
A11R40 A11R41 A11R42 A11R43 A11R44	0643-3335 0683-3335 0683-3335 0683-3335 0757-0486	4	HESISTOR 33K 5% .25W FC TC==400/+800 PESISTOR 33K 5% .25W FC TC==400/+800 RESISTOP .3% 5% .25W FC TC==400/+800 RESISTOR 33K 5% .25W FC TC==400/+800 RESISTOR 33K 5% .25W FC TC==400/+800	0160G 0160G 0160G 0160G 05520	CB3335 CB3335 CB3335 CB3335 CMF=55=1
A11R45 A11R46 A11R47 A11R48 A11R49	0757-0486 0757-0486 0757-0486 0683-3335 0683-3335		RESISTOR 750K 1% ,125M F TC#0+=100 RESISTOR 750K 1% ,125W F TC#0+=100 RESISTOR 750K 1% ,125W F TC#0+=100 RESISTOR 33K 5% ,25W FC TC#=400/+800 RESISTOR 33K 5% ,25W FC TC#=400/+800	05520 05520 05520 0160G 0160G	CMF=55=1 CMF=55=1 CMF=55=1 CB3335 CB3335
411R50 411R51 411R52 A11R53	0683-3335 0683-3335 0683-3335 0683-5605		RESISTOR 33K 5% .25W FC TC==400/+800 RESISTOR 33K 5% .25W FC TC==400/+800 HESISTOR 33K 5% .25W FC TC==400/+800 RESISTOR 56 5% .25W	0160G 0160G 0160G	CB3335 CB3335 CR3335

Table 6-3. Replaceable Parts (Cont'd).

Reference Designation	HP Part Number	Qty	Description	Mfr Code	Mfr Part Number
Attri Attra	9100-0866 9100-0866	2	THANSFORMER, PULSE TOK412NG TRANSFORMER, PULSE TOK412NG	28480 28480	9100-0866 9100-0866
A11U1 A11U2 A11U3	1826-0043 1826-0319 1826-0326	5	IC OP AMP IC OP AMP IC OP AMP	0340F 07933	LF356H RC4558DN
A12	04262-66612 04262-26612	1 1	RANGE RESISTOR BOARD ASSEMBLY PC HOARD, BLANK	28480 28480	04262-66612 04262-26612
A12C3+ A12C3+	0160-0159 0140-0190 0121-0059	1 2	CAPACITOR-FXD 6R00PF +-10% 200VDC POLYE CAPACITOR-FXD 39PF +-5% 300VDC *FACTORY SELECTED PART CAPACITOR-V TRMR-CER 2-8PF 350V PC-MTG	0420J 72136 73899	292P68292 DM15E390J0300WV1CR DV11PR8A
A12C4 A12C5 A12C6 A12C7 A12C8	0180-1051 0180-1051 0150-0050 0150-0050 0150-0050	ь	*FACTORY SELECTED PART  CAPACITOR, FXD 100 UF 16V M CAPACITOR, FXD 100 UF 16V M CAPACITOR-FXD 1000PF +80-201 KVDC CEH CAPACITOR-FXD 1000PF +80-201 KVDC CER CAPACITOR-FXD 1000PF +80-201 KVDC CER	28480 28480 28480 28480 28480	0180-1051 0180-1051 0150-0050 0150-0050 0150-0050
415C13 415C11 415C10 415C10	0150-0050 0150-0050 0121-0105 0180-0269 0160-2150	1 1 1	CAPACITUR-FXD 1000PF +80=20% 1KVDC CER CAPACITOR-FXD 1000PF +80=20% 1KVDC CEP CAPACITOR-V TRMR-CER 9-35PF 200V PC-MTG CAPACITOR-FXD 1UF+75-10% 150VDC AL CAPACITUR-FXD 33PF +-5% 300VDC	28480 28480 73899 0420J 28480	0150-0050 0150-0050 DV11PR35D 300105G150BA2 0160-2150
A12C14* A12C15 A12C16 A12C17 A12C18	0160-2307 0180-1051 0180-1051 0180-1051 0180-1051	3	CAPACITOR-FXD 47pF +-5% 300VDC CAPACITOR, FXD 100 UF 16V M CAPACITUR, FXD 100 UF 16V M CAPACITOR, FXD 100 UF 16V M CAPACITOR, FXD 100 UF 16V M	28480 28480 28480 28480	0180-1051 0180-1051 0180-1051 0180-1051
053514 053514	0180-1051 0180-1051		CAPACITOR, FXD 100 UF 16V M CAPACITUR, FXD 100 UF 16V M	28480 28480	0180-1051 0180-1051
A12CR1 A12CR2 A12CR3 A12CR4 A12CR5	1901-0040 1901-0040 1901-0040 1901-0040 1901-0040	<b>6</b> 0	DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35	28480 28480 28480 28480 28480	1901-0040 1901-0040 1901-0040 1901-0040 1901-0040
A12CR6 A12CR7 A12CR8 A12CR9 A12CR10	1901-0040 1901-0040 1901-0040 1901-0040 1901-0040		DIODE-SWITCHING 30V 50MA 2NS DD-35 DIODE-SWITCHING 30V 50MA 2NS DD-35 DIODE-SWITCHING 30V 50MA 2NS DD-35 DIODE-SWITCHING 30V 50MA 2NS DD-35 DIODE-SWITCHING 30V 50MA 2NS DD-35	\$8480 \$8480 \$8480 \$8480	1901-0040 1901-0040 1901-0040 1901-0040 1901-0040
415CH15 415CH13 415CH13 415CH13 415CH11	1901-0040 1901-0040 1902-3149 1901-0040 1901-0040		DIODE-SWITCHING 30V 50MA 2NS DU-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-ZNR 9.09V 52 DO-7 PD#.4W TC++.057% DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35	28480 28480 02236 28480 28480	1901-0040 1901-0040 FZ7256 1901-0040 1901-0040
412CR16 412CR17 412CR18 412CR19 412CR20 A12CR21 A12CR21	1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 1901-0376 1901-0376		D10DE-SWITCHING 30V 50MA 2NS D0-35 D10DE-SWITCHING 30V 50MA 2NS D0-35 D10DE-SWITCHING 30V 50MA 2NS D0-35 D10DE-SWITCHING 30V 50MA 2NS D0-35 D10DE-SWITCHING 30V 50MA 2NS D0-35 D10DE-GEN PRP	28480 28480 28480 28480 28480	1901-0040 1901-0040 1901-0040 1901-0040 1901-0040
A12K1	0490+0237	1	RELAY, REED 2A	25480	0490-0237
A1202 A1203 A1204 A1205	1855-0223 1855-0223 1855-0223 1855-0128 1855-0223	i	TRANSISTOR JOFET NOCHAN DOMORE SI THANSISTOR JOFET NOCHAN DOMORE SI TRANSISTOR JOFET NOCHAN DOMORE SI TRANSISTOR JOFET NOCHAN SI THANSISTOR JOFET NOCHAN DOMORE SI	28480	
A1206 A1207 A1208 A1209 A12010	1855-0223 1855-0223 1855-0223 1855-0223 1855-0223		TRANSISTOR J-FET N-CHAN D-MODE SI		
A12011 A12012 A12013 A12014 A12015	1855-0223 5080-3078 5080-3078 5080-3835 1854-0013	6	TRANSISTOR J=FET N=CHAN D=MODE SI TPANSISTOR NPN SI PD=300MM FT=200MHZ TRANSISTOR NPN SI PD=300MW FT=200MHZ TRANSISTOR J=FET 2N5245 N=CHAN D=MODE SI TRANSISTOR NPN 2N22184 SI TU=5 PD=800MW	28480 28480 0169H 0203G	1854-0071 1854-0071 285245 2822184
A12016 A12017 A12018 A12019 A12020	1853-0012 1853-0020 1853-0020 1853-0020 5080-3078	2	TRANSISTOR PNP 2N2904A SI TU-39 PD=600MW TRANSISTOR PNP SI PD=300MW FT=150MMZ TRANSISTOR PNP SI PD=300MW FT=150MMZ TRANSISTOR PNP SI PD=300MW FT=150MMZ TRANSISTOR NPN SI PD=300MW FT=200MMZ	0169H 28480 28480 28480	2N2904A 1853-0020 1853-0020 1853-0020 1854-0071
A12021 A12022 A12023	1853-0020 1853-0020 1853-0020		TRANSISTOR PNP SI PD=300MW FT=150MHZ TRANSISTOR PNP SI PD=300MW FT=150MHZ TRANSISTOR PNP SI PD=300MW FT=150MHZ	28480 58480 58480	1853-0020 1853-0020 1853-0020

Table 6-3. Replaceable Parts (Cont'd).

Reference Designation	HP Part Number	Qty	Description	Mfr Code	Mfr Part Number
A12R1 A12R2 A12R3 A12R4 A12R4	2100-2514 0683-1055 0683-1055 0698-2298 0698-2294	1 35 1 1	RESISTOR-TRMR 20K 10% C SIDE-ADJ 1-TRN RESISTOR 1M 5% .25W FC TC=-800/+900 RESISTOR 1M 5% .25W FC TC=-800/+900 RESISTOR 10 .05% .33W RESISTOR 100 .1 .05%	0365A 0160G 0160G 28480 28480	ET50#203 CB1055 CB1055 0698-2298 0698-2294
A12R6 A12R7 A12R8 A12R9	0698+2296 0698-2214 0698-7847 0698-2225	1 1 1	RESISTOR 1010.1 .05%  RESISTOR: FXD 10.0K OHM 0.05% 1/8W MF RESISTOR: 1.111K .1% .125W F TC=0+-100 RESISTOR: 50 90.0K OHM 0.05% 1/8W MF	28480 28480 28480	0698=2296 0698=2214 0698=2225
A12R10	0698+3329 0683+3335	1	RESISTOR 10K .5% .125W F TC=0+=100 RESISTOR 33K 5% .25W FC TC==400/+800	03868 0160G	PME55-1/8-T0-1002-D CB3335
412R12 412R13 412R14 412R15	0683-4705 0683-4705 0683-1055 0683-1055	4	RESISTOR 47 5% ,25W FC TC==400/+500 RESISTOR 47 5% ,25W FC TC==400/+500 RESISTOR 1M 5% ,25W FC TC==800/+900 RESISTOR 1M 5% ,25W FC TC==800/+900	0160G 0160G 0160G 0160G	CB4705 CB4705 CB1055 CB1055
A12R16 A12R17 A12R18 A12R19 A12R20	0683-1055 0683-1055 0683-1055 0683-1055 0683-1055		RESISTOR 1M 5% .25W FC TC#+000/+900 RESISTOR 1M 5% .25W FC TC#+800/+900 RESISTOR 1M 5% .25W FC TC#+800/+900 RESISTOR 1M 5% .25W FC TC#+800/+900 RESISTOR 1M 5% .25W FC TC#+800/+900	0160G 0160G 0160G 0160G 0160G	CH1055 CH1055 CH1055 CH1055 CH1055
A12R21 A12R22 A12R23 A12R24 A12R25	0683-1055 0683-1055 0683-2225 0683-2225 0683-2225		RESISTOR 1M 5% .25% FC TC=-800/+900 RESISTOR 1M 5% .25% FC TC=-800/+900 RESISTOR 2.2K 5% .25% FC TC=-400/+800 RESISTOR 2.2K 5% .25% FC TC=-400/+800 RESISTOR 2.2K 5% .25% FC TC=-400/+800	0160G 0160G	C81055 C81055
A12R26 A12R27 A12R28 A12R29 A12R30	0683-2225 0683-2225 0683-1035 0683-5655 0757-0442	5 1	RESISTOR 2.2K 5% .25W FC TC=-400/+800 RESISTOR 2.2K 5% .25W FC TC=-400/+800 HESISTOR 10K 5% .25W FC TC=-400/+700 RESISTOR 5,6M 5% .25W FC TC=-900/+1100 RESISTOR 10K 1% .25W FC TC=-400/+700	0160G 0160G	CR1035 CR5655
A12R31 A12R37 A12R33 A12R34 A12R35	0757-0433 0683-1065 0683-1055 0757-0394 0683-1035	1 2	RESISTOR 3,32K 1% .125W FC TC=-400/+700 H&SISTOR 10M 5% .25w FC TC=+000/+1100 R&SISTOR 1M 5% .25w FC TC=+000/+900 H&SISTOR 51.1 1% .125w F TC=0+100 R&SISTOR 10M 5% .25w FC TC=-400/+700	0160G 0160G 03298 0160G	CB10B5 CR1055 C4-178-10+51k1-F CB1035
A12R36 A12R37 A12R38 A12R39 A12R40	0633-0275 0663-4705 0663-4705 0757-0394 0663-1035	5	PESISTOR 2,7 5% ,25w FC TC==400/+500 RESISTOR 47 5% ,25w FC TC==400/+500 RESISTOR 47 5% ,25w FC TC==400/+500 RESISTOR 51,1 1% ,125w F TC=0+=100 RESISTOR 10M 5% ,25w FC TC==400/+700	01606 01606 01606 03298 01606	CR2765 CB4705 CB4705 C4=1/8=T0=51R1=F CB1035
A12Ru1 A12Ru2 A12Ru3 A12Ru4 A12Ru5	0683-0275 0757-1090 0757-1090 0683-3335 0683-3335		RESISTOR 2,7 5% ,25W FC TC==400/+500 RESISTOR 261 1% ,5W F TC=0+=100 RESISTOR 261 1% ,5W F TC=0+=100 RESISTOR 33K 5% ,25W FC TC==400/+800 RESISTOR 33K 5% ,25W FC TC==400/+800	0160G 0299E 0299E 0160G 0160G	C82765 MF7C1/2=10=201H=F MF7C1/2=10=201H=F CB35355 CH3335
A12R46 A12R47 A12R48 A12R49 A12R50 A12R51 A12R52	0683-1035 0683-1035 0683-1035 0683-1035 0683-1035 0698-4105 0757-0401		HESISTUR 10K 5% .25W FC TC#+400/+800 RESISTOR 13.3 1% .25W RCSISTOR 10.0 1% .125W		
A12U1 A12U2 413	1826-0326 1826-0089 04262-66513 04262-26513	1 1 1	IC OP AMP IC 2525 OP AMP PROCESS AMPLIFIER BOARD ASSEMBLY PC HOARD, BLANK	07933 03791 28480 28480	RC4558DN HA2-2525-5 04262-66513 G4262-26513
A13C1+ A13C2 A13C3 A13C4	0121-0036 0160-1586 0160-2554 0160-1586	3	CAPACITOR-V TRMR-CER 5.5-18pF  +FACTORY SELECTED PART CIFXD MY 0.1 UF 10% 100VDCW CAPACITOR-FXD 7.5PF +25PF 500VPC CIFXD MY 0.1 UF 10% 100VDCW	26480 26480 28480	0160-1586 0160-2254 0160-1586
A13C5* A13C6 A13C7 A13C8 A13C9	0160-2261 0180-1051 0180-1051 0160-2055	8	CAPACITOR-FXD 15pF NOT ASSIGNED CAPACITOR, FXD 100 UF 16V M CAPACITUR, FXD 100 UF 16V M CAPACITUR-FXD ,01UF +80-20% 100VDC CER	58480 58480 58480	0180-1051 0160-1051 0160-2055
A13C10 A13C11 A13C12 A13C13 A13C14	0160=2055 0180=1051 0180=1051 0160=2055 0160=2055		CAPACITUR-FXD .01UF +80-20% 100VDC CER CAPACITUR, FXD 100 UF 16V M CAPACITUR, FXD 100 UF 16V M CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD .01UF +80-20% 100VDC CER	28480 28480 28480 28480 28480	0160-2055 0180-1051 0180-1051 0160-2055 0160-2055
A13C15 A13C16 A13C17 A13C18 A13C19	0150-0050 0140-0200 0160-2055 0160-2055 0180-1051		CAPACITOR-FXD 1000PF +80-20% 1KVDC CER CAPACITOR-FXD 300PF +-5% 300VDC MICA0+70 CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR, FXD 100 UF 16V M	28480 72136 28480 28480 28480	0150-0050 DM15F301J0300wv1CR 0100-2055 0160-2055 0180-1051

Table 6-3. Replaceable Parts (Cont'd).

Reference Designation	HP Part Number	Qty	Description	Mfr Code	Mfr Part Number
A13C20 A13C21 A13C22 A13C23* A13CR1 A13CR2	0180-1051 0160-2055 0160-2055 0760-0134 1901-0033	2	CAPACITOR, FXD 100 UF 16V M  CAPACITOR=FXD .01UF +80=20% 100VDC CER  CAPACITOR=FXD .01UF +80=20% 100VDC CER  CAPACITOR=FXD 220PF 5% 200V:  DIODE=GEN PRP 180V 200MA D0=7  DIDDE=GEN PRP 180V 200MA D0=7	28480 28480 28480 28480 28480	0180-1051 0160-2055 0160-2055
A13CR3 A13CR4 A13CR5 A13CR6	1901-0040 1901-0040 1901-0040		DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35	58480 58480 58480	1901-0033 1901-0040 1901-0040 1901-0040
A13CR7 A13CR8 A13CR9 A13CR10	1901-0040 1901-0040 1901-0040 1901-0040		DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35	28480 28480 28480 28480 28480	1901-0040 1901-0040 1901-0040 1901-0040 1901-0040
A13CR11 A13CR12 A13CR13 A13CR14 A13CR15	1901-0040 1901-0040 1901-0040 1901-0040 1902-0041	6	DIODE-SWITCHING 30V 50MA 2NS D0-35 DIODE-SWITCHING 30V 50MA 2NS D0-35 DIODE-SWITCHING 30V 50MA 2NS D0-35 DIODE-SWITCHING 30V 50MA 2NS D0-35 DIODE-SWITCHING 30V 50MA 2NS D0-35 DIODE-ZNR 5,11V 5% D0-7 PD=,4K TC=-,009%	28480 28480 28480 28480 0203G	1901-0040 1901-0040 1901-0040 1901-0040 SZ 10939-98
A13CH16 A13CH17 A13CH18 A13CH19 A13CH20	1902-0041 1902-0049 1901-0040 1901-0040 1902-3149	3	DIODE-ZNR 5, 119 5% DO-7 PD=.4W TC=009% DIUDE-ZNR 6,199 5% DO-7 PD=.4W TC=+.022% DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-ZNR 9,099 5% DO-7 PD=.4W TC=+.057%	0203G 0223G 28480 28480 0223G	SZ 10939-98 FZ7240 1901-0040 1901-0040 FZ7256
A1301 A1302 A1303 A1304 A1305	5080 - 3830 5080 - 3830 5080 - 3830 5080 - 3830 5080 - 3830		THANSISTOR JOFET NOCHAN DOMODE SI		
A1306 A1307 A1308 A1509 A13010	5080-3830 1853-0020 1853-0020 1853-0020 1853-0020	:	TRANSISTOR JEFET NECHAN DEMODE SI TRANSISTOR PNP SI PDE300MW FIE150MHZ TRANSISTOR PNP SI PDE300MW FIE150MHZ TRANSISTOR PNP SI PDE300MW FIE150MHZ TRANSISTOR PNP SI PDE300MW FIE150MHZ	\$#480 \$#480 \$9480 \$9480 \$8480	1855-0091 1853-0020 1853-0020 1853-0020
A13011 A13012 A13013 A13014 A13015	1853+0020 1853+0020 1853+0020 1853+0020 1853+0020		THANSISTOR PNP SI PD=300MW FT=150MHZ THANSISTOR PNP SI PD=300MW FT=150MHZ THANSISTOR PNP SI PD=300MW FT=150MHZ TRANSISTOR PNP SI PD=300MW FT=150MHZ TRANSISTOR PNP SI PD=300MW FT=150MHZ TRANSISTOR PNP SI PD=300MW FT=150MHZ	28480 28480 28480 28480 28480	1853-0020 1853-0020 1853-0020 1853-0020 1853-0020
A13016 A13017 A13018 A13019	1855-0062 1855-0062 1855-0062 1855-0062		TRANSISTOR J=FET N=CHAN D=MODE SI TRANSISTOR J=FET N=CHAN D=MUDE SI TRANSISTOR J=FET N=CHAN D=MUDE SI TRANSISTOR J=FET N=CHAN D=MODE SI	28480 28480 28480 28480	1855-0062 1855-0062 1855-0062 1855-0062
A13R1 A13R2 A13R3 A13R4 A13R5	2100-2516 2100-2516 0643-1035 0683-1035 0683-1055	ü	RESISTOR=TRMR 100K 10% C SIDE=ADJ 1=TRN RESISTUR=TRMR 100K 10% C SIDE=ADJ 1=TRN RESISTOR 10K 5% _25% FC TC==400/+700 RESISTOR 10K 5% _25% FC TC==400/+700 RESISTOR 10K 5% _25% FC TC==800/+900	73138 73138 0160G 0160G 0160G	62-231-1 62-231-1 CH1035 CH1035 CH1055
A13R6 A13R7 A13R8 A13R9 A13R10	0698-2206 0698-2207 0643-1055 0698-2206 0698-2207 :	\$	RESISTORIFXD 100 OHM 0.05% 1/HW MF RESISTORIFXD 900 OHM 0.05% 1/HW MF RESISTORIFXD 30 5% 75W FC 10==8007.4900 PESISTORIFXD 100 OHM 0.05% 1/HW MF RESISTORIFXD 900 OHM 0.05% 1/HW MF	28480 28480 0150G 28480 28480	0698-2206 0648-2207 CH1055 0698-2206 0648-2207
A13R11 A15R12 A13R13 A13R14 A13R15	0683-1055 0698-2297 0698-2297 0698-2297	8	RESISTOR JM 5% .25% FC 1C==800/+900 RESISTOR 3.01K .05% RESISTOR 3.01K .05% RESISTOR 3.01K .05% NOT ASSIGNED,	0160G 28480 28480 28480 03298	C81055 0698-2297 0698-2297 0698-2297 C4-1/8-T0-1333-F
A13R16 A15R17 A13R18 A13R19 A13R20	0698-2297 0683-1055 0698-2297 0698-2297 0698-2297	ļ	RESISTOR 3.01K .05% RESISTOR 1M 5% .25% FC TC==600/+900 RESISTOR 3.01K .05% RESISTOR 3.01K .05% RESISTOR 3.01K .05%	28480 0160G 28480 28480 28480	0696-2297 CB1055 0698-2297 0698-2297 0698-2297
A13R21 A13R22 A13R23 A13R24 A13R25 A13R25	0698-2297 0683-1035 0683-1035	6	NOT ASSIGNED NOT ASSIGNED RESISTOR 3.01K .05% RESISTOR 10K 5% .25W FC TC=-400/+700 RESISTOR 10K 5% .25W FC TC=-400/+700 NOT ASSIGNED	03298 28480 0160G 0160G	C4-1/8-T0-1333-F 0698-2297 CB1035 CB1035
A12R27 A13R28	0683-5605 0683-5605		RESISTOR 10 5% .25W FC TC=-400/+500 RESISTOR 10 5% .25W FC TC=-400/+500	0160G 0160G	CB1005 CH1005
A13R29 A13R30 A13R31	0683-1025 0683-2235 0683-5605	55	PESISTOR 1K 5% "25W FC TC==400/+600 RESISTOR 22K 5% "25W FC TC==400/+800 RESISTOR 10 5% "25W FC TC==400/+500	0160G 0160G 0160G	C81025 C82235 C81005

Table 6-3. Replaceable Parts (Cont'd).

r	Table 6-3. Replaceable Parts (Cont u).						
Reference Designation	HP Part Number	Qty	Description	Mfr Code	Mfr Part Number		
A13R32	0683-5605		RESISTOR 10 5% .25W FC TC==400/+500	0160G	C81005		
A13R33 A13R34 A13R35	0683-1055 0683-1055 0683-1055	ļ	RESISTOR 1M 5% .25W FC TC=+800/+900 RESISTOR 1M 5% .25W FC TC=-800/+900 RESISTOR 1M 5% .25W FC TC=-800/+900	0160G 0160G 0160G	C81055 C81055 C81055		
A13R36 A13R37 A13R3H A13R39 A13R40	0683-1055 0683-1055 0683-1055 0683-1055 0683-1055		RESISTOR 1M 5% .25% FC TC==800/+900 RESISTOR 1M 5% .25% FC TC==800/+900	0160G 0160G 0160G 0160G 0160G	C81055 C81055 C81055 C81055 C81055		
A13R41 A13R42 A13R43 A13R44 A13R45	0683-1025 0683-1035 0683-1235 0683-1235 0683-1235	4	RESISTOR 1K 5% .25% FC TC%=400/+600 RESISTOR 10K 5% .25% FC TC%=400/+700 RESISTOR 12K 5% .25% FC TC%=400/+800 RESISTOR 12K 5% .25% FC TC%=400/+800 RESISTOR 12K 5% .25% FC TC%=400/+800	0160G 0160G 0160G 0160G	CR1025 CR1035 CR1235 CR1235 CR1235		
A13R46 A13R47 A13R48 A13R49 A13R50	0683-1235 0683-1055 0683-2235 0683-2235		RESISTOR 12K 5% .25W FC TC==400/+600 RESISTOR 1M 5% .25W FC TC==800/+900 RESISTOR 22K 5% .25W FC TC==400/+800 RESISTOR 22K 5% .25W FC TC==400/+800 RESISTOR 22K 5% .25W FC TC==400/+800	0160G 0160G 0160G 0160G 0160G	CB1235 CB1055 CB2235 CB2235 CB2235		
413851 413852 413853 413854 413855	0663-2235 0663-2235 0663-2235 0663-2235 0663-2235		HESISTOR 22K 5% ,25% FC TC=-400/+800 HESISTOR 22K 5% ,25% FC TC=-400/+800 RESISTOR 22K 5% ,25% FC TC=-400/+800 HESISTOR 22K 5% ,25% FC TC=-400/+800 RESISTOR 22K 5% ,25% FC TC=-400/+800	0160G 0160G 0160G 0160G 0160G	CH2235 CH2235 CH2235 CH2235 CH2235		
A13R56 A13R57 A13R58 A13R59 A13R60	0683=2235 0683=2235 0683=2235 0683=2235 0683=2235		RESISTOR 22K 5% .25W FC TC==400/+800 RESISTOR 22K 5% .25W FC TC==400/+800	0160G 0160G 0160G 0160G 0160G	C#5532 C#5532 C#5532 C#5532 C#5532		
A13R61 A13R62 A13R63 A13R64 A13R65	0683-2235 0683-2235 0683-2235 0683-2235 0683-2235		RESISTOR 22K 5% ,25W FC TC==400/+800 RESISTOR 22K 5% ,25W FC TC==400/+800	0160G 0160G 0160G 0160G	CH2235 CH2235 CH2235 CH2235		
A13R66 A15R67 A13R68 A13R69 A13R70	2100-2516 2100-2516 06*3-1025 0683-1045 0683-1025	3	RESISTOR-TRMR 100k 10% C SIDE-ADJ 1-TRN RESISTOR-TRMR 100k 10% C SIDE-ADJ 1-TkN RESISTOR 1k 5%,25% FC TC=-400/+800 RESISTOR 10% 5%,25% FC TC=-400/+800 RESISTOR 1k 5%,25% FC TC=-400/+600	73138 73138 0160G 0160G	02-231-1 02-231-1 UB1025 CB1045 CR1025		
A13P71 A13P72 A13P73 A13P74 A13P75	0683-3935 0683-1035 0683-1045 0683-1035 0683-1025	\$	RESISTOR 39% 5% ,25% FC TC=-400/+800 RESISTOM 10K 5% ,25% FC TC=-400/+700 RESISTOR 10K 5% ,25% FC TC=-400/+800 RESISTOR 10K 5% ,25% FC TC=-400/+700 RESISTOR 3K 5% ,25% FC TC=-400/+600	0160G 0160G 0160G 0160G 0160G	CH5945 CH1045 CH1045 CH1045 CH1025		
A13R76 A13R77 A13R78 A13R79 A13R80	0683-1025 0683-1025 0683-2235 0683-2235 0683-1025	ē	RESISTOR 1K 5%, 25% FC TC==400/+600 RESISTOR 1K 5%, 25% FC TC==400/+600 RESISTOR 2,7% TS 32% FC TC==400/+600 RESISTOR 4,7% 5%, 25% FC TC==400/+700 RESISTOR 1K 5%, 25% FC TC==400/+600	0160G 0160G 0160G 0160G 0160G	CM1025 CB1025 CH2235 CH4725 CB1025		
A13RA1 A13RA2 A13RA3 A13RA4 A13RA4	0683-1055 0683-1825 0683-2235 0683-1825 0683-2235	5	RESISTOR 1M 5% ,25W FC TC=+000/+900 RESISTOR 1,8K 5% ,25W FC TC=+400/+700 RESISTOR 22K 5% ,25W FC TC=+400/+700 RESISTOR 1,8K 5% ,25W FC TC=+400/+700 RESISTOR 22K 5% ,25W FC TC=+400/+800	0160G 0160G 0160G 0160G 0160G	C81055 C81825 C42235 C81825 C82235		
A13R86 A13R87	0683+1055 0683+1025		RESISTOR 1M 5% .25W FC TC==800/+900 RESISTOR 1K 5% .25W FC TC==400/+600	0160G 0160G	C81055 C81025		
A13R88 A13R89	0683-1015 0683-1015	7	RESISTUR 100 5% ,25W FC TC==400/+500 RESISTOR 100 5% ,25W FC TC==400/+500	0160G 0160G	CH1015 CH1015		
A13U1 A13U2 A13U3 A13U4 A13U5	5080-3069 5080-3069 1826-0217 1826-0217 1826-0326	5	IC GP AMP IC OP AMP IC OP AMP IC OP AMP IC OP AMP	0340F 0340F 07933 07933	LF356H LF356H RC4558T RC4558T RC4558UN		
A13U6 A13U7 A13UR	1826-0326 1820-0321 1820-0125	2	IC OP AMP IC 710 COMPARATOR IC 711 COMPARATOR	07933 0223G 0223G	RC4558UN 710HC 711HC		
A14	04262-66514 04262-26514	1 1	PHASE DETECTOR & INTEGRATOR BOARD ASSY PC BOARD, BLANK	28480 28480	04545=54214 04545=44214		
A14C1 A14C2 A14C3 A14C4	0160-1603 0160-1674 0160-1605 0150-0075 0160-2307	2 1 1 1	C1FXD MY 1 UF 10X 100VDCW CAPACITOR .33 UF 5% 200VDCW C1FXD MY 1 UF 10X 100VDCW CAPACITOR=XD 4700PF +100-0X 500VDC CER CAPACITOR=XD 470PF +=5X 300VDC *FACTORY SELECTED PART	28480 28480 28480 28480 28480	0160-1603 0160-1674 0160-1603 0150-0075 0169-2307		

Table 6-3. Replaceable Parts (Cont'd).

Reference Designation	HP Part Number	Qty	Description	Mfr Code	Mfr Part Number
A14C6 A14C7 A14C8 A14C9 A14C10	0160-0207 0160-1587 0170-0040 0170-0040 0160-1586	2 1 2	CIFXD MY 0.01 UF 5% 200VDCW CAPACITOR, FXD POLY 0.33 UF 5% 200NVDC CIFXD MY 0.047 UF 5% 200VDCW CIFXD MY 0.047 UF 5% 200VDCW CIFXD MY 0.1 UF 10% 100VDCW	28480 28480 28480 28480 28480	0160-0207 0160-1567 0170-0040 0170-0040 0160-1586
A14C11 A14C12 A14C13 A14C14 A14C19	0160-0207 0160-5819 0160-0127 0160-1052 0160-2055	28	CIFXD MY 0.01 UF 5% 200VDCW  CAPACITOR 3300 PF 50V  CAPACITOR=FXD 1UF +=20% 25VDC CER  CAPACITOR =20 UF 6.5V M  CAPACITOR=FXD .01UF +80=20% 100VDC CER	28480 28480 28480	0160-0207 0160-0127 0180-1052
A14C16 A14C17 A14C18 A14C19 A14C20	0160-2055 0160-2055 0160-2055		CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD .01UF +80-20% 100VDC CER -NOT ASSIGNED NOT ASSIGNED		
V 1 4 C S 4 V 1 4 C S 5 V 1 4 C S 5 V 1 4 C S 5	0180=1051 0180=1051 0180=1052 9160=0127	:	CAPACITOR, FXD 100 UF 16V M CAPACITOR, FXD 100 UF 16V M CAPACITOR 220 UF 6,3V M CAPACITOR=FXD 1UF +=20X 25VDC CER NOT ASSIGNED	28480 28480 28480 28480	0180-1051 0180-1051 0180-1052 0160-0127
A14C25 A14C26 A14C27 A14C28 A14C29	0160-2261	1	C-FXD 15pF 5% 500V NOT ASSIGNED NOT ASSIGNED NOT ASSIGNED NOT ASSIGNED NOT ASSIGNED		
414031 414081 414083 414084 414085 414086	1901-0040 1901-0040 1902-3059 1902-0049 1901-0040	1	DIODE-SKITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-ZNR 3,83V 52 DO-7 PD=,4W TC=+,051% DIODE-ZNR 6,19V 5% DO-7 PD=,4W TC=+,022% DIODE-SWITCHING 30V 50MA 2NS DO-35	28480 28480 0203G 0223G 28480	1401-0040 1401-0040 SZ 10939-62 FZ7240 1901-0040
A14CR7 A14CR8 A14CR9 A14CR10 A14CR11	1901+0040 1902+3149 1902+3674 1901+0040 1901+0040	1	DIODE-SWITCHING 30V 50M4 2NS 00-35 DIODE-ZNR 9.09V 5% 00-7 PD#, 4W TC#+.057% DIODE-ZNR 4.32V 2% 00-7 PD#, 4W TC#+.055% DIODE-SWITCHING 30V 50M4 2NS 00-35 DIODE-SWITCHING 30V 50M4 2NS 00-35	28480 05036 05236 28480 28480	1901-6040 F27256 SZ 10939-78 1901-0040 1901-0040
A14CH12 A14CR13 A14CR14 A14CR15 A14CR16	1901-0040 1901-0040 1902-0048 1901-0040 1901-0040	?	D10DF=SWITCHING 30V 50MA 2NS D0=35 D10DE=SmITCHING 30V 50MA 2NS D0=35 D10DE=7NR 6,81V 5X D0=7 PD=,4W 1C=+,043% D10DE=5W1TCHING 30V 50MA 2NS D0=35 D10DE=SWITCHING 30V 50MA 2NS D0=35	28480 28480 28480 28480 28480	1901-0040 1901-0040 FZ7Z44 1901-0040 1901-0040
414CK17 414CK1K 414CK19 414CK20	1902-0049 1901-0040 1901-0040 1902-3149		DIODE-ZNR 6,19V 5% DO-7 PD#,4W TC#+,022% DIODE-SKITCHING 30V SOMA 2NS DO-35 DIODE-SWITCHING 30V SOMA 2NS DO-35 DIODE-ZNR 9,04V 5% DO-7 PD#,4W TC#+,057%	0223G 28480 28480 0223G	F27240 1901-0040 1901-0040 F27256
414CR22 414CR23	1902 <b>-3</b> 149 1902 <b>-3</b> 125	1	DIODE-ZNR 9,09V 5% 00-7 PD#.4% TC#+.057% DIODE-ZNR 6,98V 2% 00-7 PD#.4% TC#+.045%	05530 05530	F Z 7 2 5 6 F Z 7 4 4 5
A 1 401 A 1 402 A 1 403 A 1 404 A 1 405	1855-0062 5080-3830 5080-3830 1855-0119 5080-3835	1	TRANSISTOR J=FET N=CHAN D=MODE SI TRANSISTOR J=FET N=CHAN D=MODE SI TRANSISTOR J=FET N=CHAN D=MODE SI TRANSISTOR J=FET N=CHAN SI TRANSISTOR J=FET 2N5245 N=CHAN D=MODE SI	28480 28480 28480 28480 0169H	1855-0062 1855-0091 1855-0091 1855-0119 285245
A1406 A1407 A1408 A1459 A14510	1853-0020 1854-0023 5080-3078 5080-3830 1853-0020	1	TRANSISTOR PNP SI PD#300MW FT#150MH/ TRANSISTOR NPN SI TU#18 PD#360MW TRANSISTOR NPN SI PD#300MW FT#200MH/ TRANSISTOR J#FET N=CHAN D=MODE SI TRANSISTOR J#FET N=CHAN D=MODE SI TRANSISTOR PNP SI PD#300MW FT#150MH/	28480 28480 28480 28480 28480	1854-0020 1854-0023 1854-0071 1855-0091 1853-0020
A14011 A14012 A14013 A14014 A14015	5080-3078 1853-0020 5080-3078 1853-0020 1853-0020		THANSISTOR NPN SI PD#300MN FT#200MHZ THANSISTOR PNP SI PD#300MN FT#150MHZ THANSISTOR NPN SI PD#300MN FT#200MHZ THANSISTOR PNP SI PD#300MN FT#150MHZ THANSISTOR PNP SI PD#300MN FT#150MHZ THANSISTOR PNP SI PD#300MN FT#150MHZ	28480 28480 28480 28480 28480	1854-0071 1853-0020 1854-0071 1853-0020 1853-0020
A14016 A14017 A14018 A14019 A14020	1855-0062 1855-0062 5080-3830 5080-3835 5080-3835		TRANSISTOR J=FET N=CHAN D=MODE SI TRANSISTOR J=FET N=CHAN D=MODE SI TRANSISTOR J=FET N=CHAN D=MODE SI TRANSISTOR J=FET 2N5245 N=CHAN D=MODE SI TRANSISTOR J=FET 2N5245 N=CHAN D=MODE SI	28480 28480 28480 0169H 0169H	1855-0062 1855-0062 1855-0091 2N5245 2N5245
A14021 A14022 A14023 A14024 A14025	1853-0034 5080-3835 5080-3835 1853-0034 1853-0020	5	THANSISTOR PNP SI TU-18 PD=360MW TRANSISTOR J=FET 2N5245 N=CHAN D=MODE SI THANSISTOR J=FET 2N5245 N=CHAN D=MODE SI THANSISTOR PNP SI TO-18 PD=360MW THANSISTOR PNP SI PD=300MW FT=150MHZ	25480 0169H 0169H 28480 25480	1853-0034 2N5245 2N5245 1853-0034 1853-0020

Table 6-3. Replaceable Parts (Cont'd).

Designation   Number   Cry	Reference	HP Part	П	able 0-3. Replaceable Faits (Cont u)	Mfr	
			Qty	Description		Mfr Part Number
	A14026	1853-0020		TRANSISTOR PNP SI PD=300MW FT=150MHZ	28480	1853-0020
	414R2	0683-1525		RESISTOR 1.5K 5% .25W FC TC==400/+700	01606	CB1525
1	A14R4	0683-4725	1	RESISTOR 4.7K 5% .25W FC TC==400/+700	0160G	C84725
Authors	A14R8 A14R9	0683-1055 0683-1055		RESISTOR 1M 5% .25W FC TC#+800/+900 RESISTOR 1M 5% .25W FC TC#+800/+900	0160G 0160G	CB1055 CB1055
A   Sept   O   O   O   O   O   O   O   O   O	l					
	A14R12 A14R13	0757-0465	5	RESISTOR 19.6K 1% ,125W F TC=0+=100 RESISTOR 100K 1% ,125W F TC=0+=100	03298	C4-1/8-T0-1003-F
				RESISTOR -TRMR 10K 10X C SIDE-ADJ 1-TRN	03654	£750X103
	A14R17	0683-2225		RESISTOR 2.2K 5% .25W FC TC==400/+700	01606	C85552
Alumon   A	A14R19	0683-4745		RESISTOR 470K 5% .25W FC TC==800/+900	01606	C84745
1   1   1   1   1   1   1   1   1   1			,			C4-1/8-T0-StiR-F
A	A14R23 A14R24	0683-1055 0683-3335		RESISTOR IM 5% ,25W FC TC==800/+900 RESISTOR 33K 5% ,25W FC TC==400/+800	0160G	C83335
			1	•	01606	C83335
1	A14R28	0663-3335		RESISTOR 33K 5% .25W FC TC==400/+800 RESISTOR 33K 5% .25W FC TC==400/+800	0160G	CB3335
A   May   O   O   O   O   O   O   O   O   O				RESISTOR 6.49K 1% .125W F TC=0+=100	89520	C4=1/8=70=6491=F
A	A14832	0683=1025	,	RESISTOR 1K 5% ,25% FC TC=+400/+600	01606	CB1025
A	A14R34	0683-1035		RESISTOR 10k 5% .25W FC 1C#=400/+700	0160G	C81035
A			1	RESISTOR 100K 1% .125W F TC=0++100	05298	C4-1/8-10-1005-F
A14RA1	A14R39	0698-3155	٫	RESISTOR 4.64K 1% .125W F TC=0+-100	03298	E4-1/R-T0-4641-F
A	A14R41	0757-0401		RESISTOR 100 1% .125W F TC=0+=100	03298	
A14R46 A14R47 A14R48 A14R47 A14R48 A14R48 A14R48 A14R48 A14R48 A14R48 A14R59 A1	A14R43	0683-1055		RESISTOR 1M 5% .25W FC TC==800/+900 RESISTOR 19.6K 1% .125W F TC=0+=100	0160G 03298	CB1055 C4+178+T0+1962+F
A   4   4   4   4   4   4   4   4   4	1					
A14R50	A14R47 A14R48	0683-1035		RESISTOR 10K 5% ,25W FC TC==400/+700 RESISTOR 10K 5% ,25W FC TC==400/+700	0160G 0160G	CH1035 CB1035
A14R54				RESISTOR 3.3K 5% .25W FC TC==400/+700 RESISTOR 3.3K 5% .25W FC TC==400/+700		
A 1 4 8 5	A14R52	0685-3335		RESISTOR 33% 5% .25% FC TC#=400/+800	0160G	C83335
A14R56	A14R54	0683-3335		RESISTOR 33k 5% .25W FC TC#-400/+800	01606	C83555
A14R58	A14R56			HESISTON 33K 5% ,25W FC TC==400/+800 RESISTON 4.7K 5% ,25W FC TC==400/+700		
A14R61 0698-6943 2 RESISTOR 20K .1% .125W F TC=0+=50 0329H NC55  A14R62 0698-0083 2 RESISTOR 1.96K 1% .25W FC TC=-400/+700 A14R64 0698-0083 A14R65 0757-0401 2 RESISTOR 100 1% .25W FC TC=-400/+700 RESISTOR 100 1% .25W FC TC=-400/+700 A14R66 0683-3335 A14R67 0683-1245 1 RESISTOR 100 1% .25W FC TC=-400/+700 NESISTOR 100 1% .25W FC TC=-400/+700 NESISTOR 100 1% .25W FC TC=-400/+700 NESISTOR 100 1% .25W FC TC=-400/+900 0160G CB1345 NESISTOR 120K 5% .25W FC TC==000/+900 0160G CB1345 NESISTOR 31K 5% .25W FC TC==000/+900 0160G CB1345 NESISTOR 33K 5% .25W FC TC==000/+900 0160G CB13355	A14R58 A14R59	0698-4157 0698-4157		RESISTOR 10K .1% .125W F TC=0++50 RESISTOR 10K .1% .125W F TC=0++50	0329B 0329B	NC55 NC55
A14R62				RESISTOR ZOK .1% .125W F TC#0+-50	1	
A14R65 0757-0401 RESISTOR 100 1% .25W FC TC=-400/+700  A14R66 0683=3335 RESISTOR 33K 5% .25m FC TC==400/+800 0160G C83535  A14R67 0683=1245 1 RESISTOR 120K 5% .25W FC TC==800/+900 0160G C81245  A14R68 0683=4735 1 RESISTOR 47K 5% .25W FC TC==400/+800 0160G C81245  A14R69 0683=3355 RESISTOR 33K 5% .25W FC TC==400/+800 0160G C81335	A14R62 A14R63	0757-0401		RESISTOR 100 1% .25W FC TC=-400/+700		
414867 0683-4735 1 RESISTOR 120K 5% ,25W FC TC==800/+900 0160G CB1245 414868 0683-4735 1 RESISTOR 47K 5% ,25W FC TC==400/+800 0160G CB4735 414869 0683-3355 RESISTOR 33K 5% ,25W FC TC==400/+800 0160G CB3335	A14R65	0757-0401	]	RESISTOR 100 1% .25W FC TC=-400/+700	01.05	CHITIE
	A14R67 A14R68	0683-1245 0683-4735		RESISTOR 120K 5% .25W FC TC=-800/+900 RESISTOR 47K 5% .25W FC 1C=-400/+800	0160G 0160G	CB1245 CB4735
A14R70 0083-4725 RESISTOR 4.7K 5% .25W FC TC==400/+700 016UG CB4725 A14R71 0683-2265 RESISTOR 22M 5% .25W	A14R70	0683-4725		RESISTOR 4.7K 5% .25W FC TC==400/+700	0160G 0160G	CH \$335 CB 4725
A1477 0757-1094 RESISTOR 1.47K 1% .125W  A14U1 1826-1071 2 Icilin Op. AMPL. FET-INPT 28480 LF411CH	A14R72	0757-1094	,	RESISTOR 1.47K 1% .125W	28480	F411CH
A14U2 1826-0271 IC 741 OP AMP' 0340F LM741CN 1820-0321 IC 710 COMPARATOR 0223G 710MC	A14U2 A14U3	1826-0271 1820-0321		IC 741 OP AMP' IC 710 COMPARATOR	0340F 0223G	LM741CN 710HC
A1404 1826-1071 ICILIN OP, AMPL, FET-INPT 28480 LF411CH RC4558DN						

Table 6-3. Replaceable Parts (Cont'd).

Reference Designation	HP Part Number	Qty	Description	Mfr Code	Mfr Part Number
A1406 A1407 A1408 A1409 A14010	1826-0319 1826-0326 1820-0054 5080-3832 1826-0180	1 1 1	IC OP AMP IC OF AMP IC GATE TIL NAND QUAD 2-INP IC MISC TIL IC 555	0340F 07933 0223G 0203G 0291J	LF 356H RC 4558DH 7405PC MC 4044P NE 555V
A14U11 A14U12 A14U13 A14U14 A14U15	1820-0379 1820-0075 1820-1210 1820-1210 1820-1490	1 1 2	IC GATE TIL H AND-OR IC FF TIL J-K PULSE CLEAR DUAL IL GATE TIL LS AND-OR-INV DUAL 2-INP IC GATE TIL LS AND-OR-INV DUAL 2-INP IC CNIR ITL LS DECD ASYNCHRO	0223G 0223G 0169H 0169H 0169H	74H52PC 7473PC SN74L551N SN74L551N SN74L590N
A15			NUT ASSIGNED		
Alb			NUT ASSIGNED		
A 1 7			NOT ASSIGNED		
A18			NOT ASSIGNED		
A19			NUT ASSIGNED		
050			NOT ASSIGNED		
A21	04262=66521 04262=26521	1 1	KEYBOARO & DISPLAY BOARD ASSEMBLY PC HOARD, BLANK	28480 28480	04262=66521 04262=26521
A21C1 A21C2 A21C3 A21Cu	01H0+029t 0160-2055 0160-2055 0160-2055		CAPACITOR-FXD 1UF++10% 35VDC TA CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD .01UF +80-20% 100VDC CER	0420J	1500105X903542
A21C6 A21C6 A21C7 A21C8 A21C9 A21C10	0180-0376 0150-0197 0180-0197 0180-0197 0180-0197 0140-0198	1 6	CAPACITOR=FXD ,47uF+=10x 35v0C TA  CAPACITOR=FXD 2,2UF+=10x 20vDC TA  CAPACITOR=FXD 2,2UF+=10x 20vDC TA  CAPACITOR=FXD 2,2UF+=10x 20vDC TA  CAPACITOR=FXD 2,2UF+=10x 20vDC TA  CAPACITOR=FXD 20vPF+=5x 30vDC MICA	0420J 0420J 0420J 0420J 0420J 72136	1500474x903542 1500225x902042 1500225x902042 1500225x902042 0M15F201J030UWV1CR
AZICRI AZICRZ AZICRZ AZICRZ AZICRZ	1901-0040 1901-0040 1901-0040 1901-0040 1901-0040		DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35	28480 28480 28480 28480 28480	1901-0040 1901-0040 1901-0040 1901-0040 1901-0040
A21CR6 A21CR7	1901-0040 1901-0040		DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35	28480 28480	1901-0040 1901-0040
A21J1	1251+0541	ē	CONNECTOR 34-PIN M RECTANGULAR	76381	3431=1002
42101 421R1 421R2 421R3 421R4 421R5	1654-0019 0685-4715 0683-4715 0683-4715 0683-4715 0683-4715	1	TRANSISTOR NPN S] TO-18 PD=360MM  HESISTOR 470 5% .25W FC TC=-400/+600  RESISTOR 470 5% .25W FC TC=-400/+600	28480 0160G 0160G 0160G 0160G 0160G	1854-0019 CR4715 CR4715 CB4715 CB4715 CR4715
A21Rb A21R7 A21RB A21R9 A21R10	0683-3305 0683-1015 0683-1015 0683-1015 0683-4715	t	RESISTUR 33 5% ,25% FC TC==400/+500 PESISTOP 100 5% ,25% FC TC==400/+500 RESISTOR 100 5% ,25% FC TC==400/+500 PESISTUR 100 5% ,25% FC TC==400/+500 RESISTOR 470 5% ,25% FC TC==400/+600	01606 01606 01606 01606 01606	C83505 CB1015 CB1015 CB1015 CB4715
A21R11 A21R12 A21R13 A21R14 A21R14	0683-4715 0685-4715 0683-4715 0683-4715 0683-4715		RESISTOR 470 5% .25% FC TC=+400/+600 RESISTOR 470 5% .25% FC TC=+400/+600	0160G 0160G 0160G 0160G 0160G	C84715 C84715 C84715 C84715 C84715
A21R16 A21R17 A21R18 A21R19 A21R20	0683-1015 0683-1015 0683-4715 0683-4715 0683-4715		RESISTOR 100 5% .25w FC TC*-400/+500 RESISTOR 100 5% .25w FC TC*-400/+500 RESISTOP 470 5% .25w FC TC*-400/+600 RESISTOR 470 5% .25w FC TC*-400/+600 RESISTOR 470 5% .25w FC TC*-400/+600	0160G 0160G 0160G 0160G 0160G	CH1015 CB1015 CH4715 CB4715 CB4715
A21R21 A21R22 A21R23 A21R24 A21R29	0663-4715 0683-4715 0683-4715 0683-4715 0683-4715		HESISTOR 470 5% .25W FC TC==400/+600 PESISTOR 470 5% .25W FC TC==400/+600 PESISTOR 470 5% .25W FC TC==400/+600 RESISTOR 470 5% .25W FC TC==400/+600 PESISTOR 470 5% .25W FC TC==400/+600	0160G 0160G 0160G 0160G 0160G	C84715 C84715 C84715 C84715 C84715

Table 6-3. Replaceable Parts (Cont'd).

Reference Designation	HP Part Number	Qty	Description	Mfr Code	Mfr Part Number
A21R26 A21R27 A21R28 A21R28 A21R29 A21R30	0683-4715 0683-3935 0683-1035 0683-1035 0683-1035		RESISTOR 470 5% .25W FC TC=-400/+600 RESISTOR 39K 5% .25W FC TC=-400/+800 RESISTOR 10K 5% .25W FC TC=-400/+700 RESISTOR 10K 5% .25W FC TC=-400/+700 RESISTOR 10K 5% .25W FC TC=-400/+700	0160G 0160G 0160G 0160G 0160G	C84715 C83935 C81035 CB1035 CB1035
A21R31 A21R32	0683*1035 1810=0164	5	RESISTOR 10K 5% ,25W FC TC#=400/+700 NETWORK-RES 9-PIN-SIP ,15-PIN-SPCG	0160G 28480	C81035 1810-0164
42181 42182 42184 42184 42184 42185	1:820-1415 1820-1279 1820-0261 1820-1200 1820-1200	2 1 1	IC SCHMITT-TRIG TTL LS NAND DUAL 4-INP IC CNTR TTL LS DECD UP/DOWN SYNCHRO IC MY TTL MONOSTBL IC INV TTL LS MEX 1-INP IC INV TTL LS MEX 1-INP	-0169H -0169H -0169H -0169H	SN74L513N SN74L3190N SN74121N SN74L505N SN74L505N
A2106 A2107 A2108 A2109 A21010	1820-1200 1820-1195 1820-1195 1820-1198 1820-1197	15 1 8	IC INV ITL LS MEX 1-INP IC FF TIL LS D-TYPE POS-EDGE-TRIG COM IC FF TIL LS D-TYPE POS-EDGE-TRIG COM IC GATE TIL LS NAND GUAD 2-INP IC GATE TIL LS NAND GUAD 2-INP	0169H 0379D 0379D 0169H 0169H	5N74L805N AM74L8175A AM74L8175A 5N74L803N SN74L800N
A21U11 A21U12 A21U13 A21U14 A21U15	1820-1081 1820-1470 1820-1197 1820-1112 1820-1195	18 8 7	IC DRVR TIL BUS DRVR GUAD 1=TNP IC MUXR/DATA=SEL TIL LS 2=TO=1=LINE GUAD IC GATE TIL LS NAND GUAD Z=INP IC FF TIL LS D=TYPE POS=EDGE=TRIG IC FF TIL LS D=TYPE POS=EDGE=TRIG COM	0379D 0379D 0169H 0169H 0379D	AMRT26 SN74L8157N SN74L800N SN74L874N AM74L8175A
A21U16 A21U17 A21U18 A21U19 A21U20	1820+1195 1820+1195 1820+1195 1820+1195 1820+1245	2	IC FF ITL LS D-TYPE POS-EDGE-TRIG COM IC DCDR ITL LS 2-TO-4-LINE DUAL 2-INP	03790 03790 03790 03790 0169H	AM74LS175A AM74LS175A AM74LS175A AM74LS175A SN74LS175A
A21021 A21022 A21023 A21024 A21025	1820-1195 1820-1081 1820-1470 1820-1473 1820-1201	1 5	TO FF TIL LS D-TYPE PUS-EDGE-TRIG COM TO DRYR TIL BUS DRYR GUAD 1-INP TO MUXR/DATA-SEL TIL LS 2-TO-1-LINE GUAD TO ENCOR TIL H-INP TO GATE TIL LS AND GUAD 2-INP	0379D 0379D 0379D 0169H 0169H	AM74L5175A AM8T26 SN74L5157N SN74148N SN74L508N
5 S A	04262~66522 04262~26522	s	DISPLAY CONTROL & RAM BOARD ASSEMBLY PC BOARD, BLANK	28480 28480	04262 <b>~</b> 66522 04262 <b>~</b> 26522
A22C1 A22C2 A22C3 A22C4 A22C5	0180-0291 0160-2055 0160-2055 0160-2055 0160-2055		CAPACITUR-FXD 1UF+=10% 35VDC TA  CAPACITOR-FXD .01UF +80-20% 100VDC CER	04507	150D105X9035A2
A22C6 A22C7 A22C8 A22C9 A22C10	0160-2204 0160-2261 0160-0939 0180-0291 0160-0939	5	CAPACITOR-FXD 100PF +-5% 300VDC MICA0+70 CAPACITOR-FXD 15PF +-5% 500VDC CER0+-30 CAPACITOR-FXD 450PF +-5% 300VDC MICA0+70 CAPACITOR-FXD 1UF4-10% 35VDC TA CAPACITOR-FXD 450PF +-5% 300VDC MICA0+70	28480 28480 28480 0420J 28480	0160=2204 0160=2201 0160=0939 150D105×9035A2 0160=0939
A22C11 A22C13 A22C13 A22C14 A22C15	0160-0939 0160-2205 0150-0121 0150-0121 0150-0121		CAPACITOR-FXD 430PF +-5% 300VDC MICAO+70 CAPACITOR-FXD 120PF +-5% 300VDC MICAO+70 CAPACITOR-FXD 11F +80-20% 50VDC CER CAPACITOR-FXD 11F +80-20% 50VDC CER CAPACITOR-FXD 11F +80-20% 50VDC CER	28480 28480 28480 28480 28480	0160-0939 0160-2205 0150-0121 0150-0121 0150-0121
A22C16 A22C17 A22C18 A22C19 A22C20 A22C21 A22C22	0150-0121 0150-0121 0150-0121 0150-0121 0150-0121 0150-0121 0180-1743 0160-2205		CAPACITOR-FXD .1UF +80-20% 50VDC CER CAPACITOR-FXD 0.1UF 35VDC TA CAPACITOR-FXD 120DF 5% 300VDC MICA	28480 28480 28480 28480 28480	0150-0121 0150-0121 0150-0121 0150-0121 0150-0121
A22CR1 A22J1 A22Q1 A22Q2 A22Q3 A22Q4 A22Q5	1902-0041 1200-0541 1853-0084 1853-0084 1853-0084 1853-0084	8	DIODE=ZNR 5.11V 5% D0=7 PD=.4W TC=009% SOCKET=IC 24=CONT DIP=SLOH  TRANSISTOR PNP 2N4918 SI PD=30W FT=3MHZ	0203G 0203G 0203G 0203G 0203G	5W4318 5W4318 5W4318 5W4318
A2206 A2207	1853=0084 1853=0084 1853=6084		THANSISTOR PNP 2N4918 SI PD=30# FT=3MHZ THANSISTOR PNP 2N4918 SI PD=30# FT=3MHZ TRANSISTOR PNP 2N4918 SI PD=30# FT=3MHZ	0203G 0203G 0203G	5/4418 5/4418 5/4418
A2208 A22R1 A22R2 A22R3 A22R4 A22R5	0603-2715 0683-2715 0683-2715 0683-2715 0683-2715		RESISTOR 270 5% .25W FC TC=-400/+600	0160G 0160G 0160G 0160G	CB2715 CB2715 CB2715 CB2715 CB2715
A22R6 A22R7 A22R8 A22R9 A22R10	0683-2715 0683-2715 0683-2715 0683-2715 0683-6805 0683-6805		RESISTOR 270 5% 25% FC TC==400/+600 RESISTOR 270 5% 25% FC TC==400/+600 RESISTOR 270 5% 25% FC TC==400/+600 RESISTOR 68: 5% 25% FC TC==400/+500 RESISTOR 68: 5% 25% FC TC==400/+500	0160G 0160G 0160G 0160G	CH2715 CB2715 CH2715 CH6805 CH6805

Table 6-3. Replaceable Parts (Cont'd).

Reference Designation	HP Part Number	Qty	Description	Mfr Code	Mfr Part Number
422R11 422R12 422R13 422R14 422R15	0683-6805 0683-6805 0683-6805 0683-6805 0683-6805		RESISTOR 68 5% .25W FC TC=-400/+500 RESISTOR 68 5% .25W FC TC=-400/+500	0160G 0160G 0160G 0160G 0160G	C#68 05 C#68 05 C#68 05 C#68 05 C#68 05
A 2 2 R 1 6 A 2 2 R 1 7 A 2 2 R 1 R A 2 2 R 1 P A 2 2 R 2 O	0683-6805 0683-2725 0683-1825 0683-4725 1810-0121	2	RESISTOR 68 5% .25W FC TC==400/+500 RESISTOR 2,7K 5% .25W FC TC==400/+700 PESISTOR 1,8K 5% .25W FC TC==400/+700 RESISTOR 4,7K 5% .25W FC TC==400/+700 NETWORK=RES 9=PIN=SIP .15=PIN=SPCG	0160G 0160G 0160G 0160G 28480	C86805 C82725 C81825 C84725 1810-0121
A 22 P 2 1 A 22 P 2 2 A 22 P 2 2 A 22 P 2 2 A 25 P 2 5	1810-0205 1810-0206 0693-1025 0683-1025 0683-1025	2	NETWORK-RES 8-PIN-SIP .1-PIN-SPCG NETWORK-RES 8-PIN-SIP .1-PIN-SPCG RESISTOR 1K 5% .25% FC TC=-400/+600 RESISTOR 1K 5% .25% FC TC=-400/+600 RESISTOR 1K 5% .25% FC TC=-400/+600	0248C 0374D 0160G 0160G 0160G	750-81-R4,7K 4308R-101-1058 CB1025 CB1025 CB1025
A 22 R 26 A 22 R 28 A 22 R 29 A 22 R 20 A 22 R 30	0683-1025 0683-1025 0683-1025 0683-1025 0683-1025		RESISTOR 1K 5% .25W FC TC=+400/+600 RESISTOR 1K 5% .25W FC TC=+400/+600	01606 01606 01606 01606 01606	CB1025 CB1025 CB1025 CB1025 CB1025
422R31 422R32 427R33 422R34 422R35		8	NOT ASSIGNED NOT ASSIGNED NOT ASSIGNED NOT ASSIGNED NOT ASSIGNED NOT ASSIGNED		
A22R36 A22R37 A22R38 A22R39 A22S1	1810-0164		NOT ASSIGNED NOT ASSIGNED NOT ASSIGNED NETWORK-RES 9-PIN-SIP .15-PIN-SPCG	28480	1810-0164
A22U4 A22U4 A22U3 A22U3	3101-0299  1820-0738  1820-1194  1820-1199  1820-1201  1820-1688	1 2 7	SWITCH, SLIDE 4-SPST  IC DCDR TIL 2-TO-4-LINE DUAL 2-INP IC CNTR TIL LS BIN UP/DOWN SYNCHRO IC INV TIL LS HEX 1-INP IC GATE TIL LS AND QUAD 2-INP IC DCDR TIL HCD-TO-7-SEG	28480 0203G 0379D 0169H 0169H 0169H	3101-0299 MC74155P AM74L8193PC SN74L504N SN74L504N SN74LS241N
422116 422117 422118 422119 4221110	5080-3068! 1820-1490 1858-0033 1820-0628 1820-1470	s :	IC MV TTL DUAL IC CNTR TTL LS DECD ASYNCHRU TRANSISTOR IC SN7489N 64-BIT RAM ITL IC MUXR/DATA-SEL TTL LS 2-TU-1-LINE QUAD	0169H 28480 0340F 0379D	SN74L590N 1858-0033 DM7489N SN74L5157N
A22011 A22012 A22013 A22014 A22014	1820-1425 1820-1112 1820-1197 1820-1490 1820-1478	5	IC SCHMITT-TRIG TIL LS NAND QUAD 2-INP IC FF TIL LS D-TYPE PUS-EDGE-TRIG IC GATE TIL LS NAND QUAD 2-INP IC CNIR TIL LS DECD ASYNCHRO IC ENTR TIL LS BIN ASYNCHRO	0169H 0169H 0169H 0169H 0169H	5N7/1L 51 3.2N 5N7/4L 57 4N 5N7/4L 50 0N 5N7/4L 59 0N 5N7/4L 59 3N
A22016 A22017 A22018 A22019 A2200	1858-0033 1820-0628 1820-1470 1820-1081 1820-1081		TRANSISTOR IC SN7489N 64-BIT RAM TTL IC MUXR/DATA-SEL TTL LS Z-TU-1-LINE QUAD IC DRVK TTL HUS DRVR QUAD 1-INP IC DRVR TTL HUS DRVR QUAD 1-INP	28460 0340F 0379D 03790 03790	1950-0033 DM7489N SN74LS157N AM8126 AM8126
1425A1 75713 75713	1820-1196 1818-0135 0410-0209	5	IC FF TIL LS D-TYPE PUS-EDGE-TRIG COM IC MC 6610L-1 1K RAM NMOS CRYSTAL, GUARTZ	0374D 0203G 28480	AM74L5174N MC6810L=1 0410=0209

Table 6-3. Replaceable Parts (Cont'd).

Reference Designation	HP Part Number	Qty	Description	Mfr Code	Mfr Part Number
A23	04262-66623 04262-26623		PROCESSOR & ROM BOARD ASSEMBLY PC BOARD, BLANK	28480 28480	04262-66 <b>5</b> 23 04262-26623
A23C1 A23C2 A23C3 A23C4 A23C5	0160-2202 0180-2141 0180-0291 0180-0197 0180-0197		CAPACITOR-FXD 75pF 5% 300VDC CAPACITOR-FXD 3.3µF +-10% 50VDC TA CAPACITOR-FXD 1UF +-10% 35VDC TA CAPACITOR-FXD 2.2UF +-10% 20VDC TA CAPACITOR-FXD 2.2UF +-10% 20VDC TA	0420J 0420J 0420J	150D105X9035A2 150D225X9020A2 150D225X9020A2
A23C6 A23C7 A23C8 A23C9 A23C10	0180-0229 0160-2055 0160-2055 0160-2055 0160-2055		CAPACITOR-FXD 33UF +-10% 10VDC TA CAPACITOR-FXD .01UF +80-20% 100VDC CER	0420J	150D336X9010B2
A23CR1 A23CR2 A23CR3 A23CR4	1901-0040 1901-0040 1902-3158 1902-0048		DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE, ZENER, 9.76V DIODE, ZENER, 6.81V	28480 28480 0223G 0223G	1901-0040 1901-0040 FZ7459 FZ7244
A23J1 A23J2 A23J3 A23J4	1200-0853 1200-0541 1200-0541 1200-0654		SOCKET-IC 16-CONT DIP-SLDR SOCKET-IC 24-CONT DIP-SLDR SOCKET-IC 24-CONT DIP-SLDR SOCKET-IC 40-CONT DIP-SLDR	28480 28480 28480	1200-0541 1200-0541
A23Q1 A23Q2 A23Q3 A23Q4	1853-0089 5080-3078 1854-0477 1854-0215		TRANSISTOR PNP 2N4917 SI PD=200MW FT=450MHz TRANSISTOR NPN SI PD=300MW FT=200MHz TRANSISTOR NPN 2222A SI TO=18 PD=500MW TRANSISTOR NPN SI PD=350MW FT=300MHz	0223G 0203G	2N4917 2N2222A SPS3611
A23R1 A23R2 A23R3 A23R4 A23R5	0683-4725 0683-4725 0683-1025 0683-1025 0683-1035		RESISTOR 4.7K 5% .25W FC TC=-400/+700 RESISTOR 4.7K 5% .25W FC TC=-400/+700 RESISTOR 1K 5% .25W FC TC=-400/+600 RESISTOR 1K 5% .25W FC TC=-400/+600 RESISTOR 1K 5% .25W FC TC=-400/+700	0160G 0160G 0160G 0160G 0160G	CB4725 CB4725 CB1025 CB1025 CB1035
A23R6 A23R7 A23R8 A23R9 A23R10	0683-1055 0683-1845 0683-1035 0698-3430 0683-5615		RESISTOR 1M 5% .25W FC TC=-800/+900 RESISTOR 180K 5% .25W FC TC=-800/+900 RESISTOR 10K 5% .25W FC TC=-400/+700 RESISTOR 21.5 1% .125W F TC=0+-100 RESISTOR 560 5% .25W FC TC400/+600	0160G 0160G 0160G 03888 0160G	CB1055 CB1845 CB1035 RME 55-1/8-TO-21R5-F CB5615
A23R11 A23R12 A23R13	0683-5625 1810-0164		RESISTOR 5.6K 5% .25W FC TC=-400/+700 NETWORK-RES 9-PIN-SIP .15-PIN-SPCG NOT ASSIGNED	0160G 28480	CB5625 1810-0164
A23R14	2100-2633		RESISTOR-TRMR 1k 10% C SIDE-ADJ 1-TRN	0365A	ET50X102
A23S1	3101-0299		SWITCH SLIDE 4-SPST	28480	3101-0299
A23U1 A23U2 A23U3 A23U4 A23U5	1820-1691 1820-1197 1820-2053 1820-2053 1820-1081	:	IC MICROPROC MOS IC GATE TTL LS NAND QUAD 2-INP IC DCDR TTL LS 4-TO-16-LINE 4-INP IC DCDR TTL LS 4-TO-16-LINE 4-INP IC DCDR TTL LS 4-TO-16-LINE 4-INP IC DRVR TTL BUS DRVR QUAD 1-INP	28480 0169H 0379D	1820-1691 SN74LSOON 74LS154N 74LS154N AM8T26
A23U6 A23U7 A23U8 A23U9 A23U10	1820-1081 1820-1195 1820-1196 1820-1112 1820-0471		IC DRVR TTL BUS DRVR QUAD 1-INP IC FF TTL LS D-TYPE POS-EDGE-TRIG COM IC FF TTL LS D-TYPE POS-EDGE-TRIG COM IC FF TTL LS D-TYPE POS-EDGE-TRIG IC INV TTL HEX 1-INP	0379D 0379D 0379D 0169H 0223G	AM8T26 AM74LS175A AM74LS174N SN74LS74N 7406PC
A23U11 A23U12 A23U13 A23U14 A23U15 A23U16	1820-1195 1820-1201 1820-197 1820-1199 04262-85009 04262-85010		IC FF TTL LS D-TYPE POS-EDGE-TRIG COM IC GATE TTL LS AND QUAD 2-INP IC GATE TTL LS NAND QUAD 2-INP IC INV TTL LS HEX I-INP IC, ROM MOS IC, ROM MOS	0379D 0169H 0169H 0169H	AM74LS175A SN74LS08N SN74LS00N SN74LS04N
<b>424</b> (OPTION: 004)	04262=66524 04262=26524	1 1	CUMPARATOR CONTROL BOARD ASSEMBLY PC BOARD, BLANK	28480 28480	04262-66524 04262-26524
A24C1 A24C2 A24C3	0180-0229 0180-0229 0160-2055		CAPACITOR-FXD 33UF+=10% 10VDC TA CAPACITOR-FXD 33UF+=10% 10VDC TA CAPACITUR-FXD .01UF +80=20% 100VDC CER	04507 04507	150D336x9010H2 150D336x9010H2

Table 6-3. Replaceable Parts (Cont'd).

Reference Designation	HP Part Number	Qty	Description	Mfr Code	Mfr Part Number
A24CR1 A24CR3 A24CR4 A24CR4 A24CR5	1901-0040 1901-0040 1901-0040 1901-0040 1901-0040		DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35	28480 28480 28480 28480	1901-0040 1901-0040 1901-0040 1901-0040 1901-0040
A24CH6	1901-0040		CINDE-SWITCHING 30V 50MA 2NS DO-35	28480	1901-0040
A24J1	1200-0853		SOCKET-IC 16-CONT DIP-SLUR		
\$54K2 \$54K3 \$54K5 \$54K5	0490+0235 0490+0235 0490+0235 0490+0235 0490+0235	6	RELAY, REED RELAY, REED RELAY, REED RELAY, REED RELAY, REED RELAY, REED	\$8480 \$8480 \$8480 \$8480	0490=0235 0490=0235 0490=0235 0490=0235 0490=0235
A24M6	0490-0235		RELAY, REED	28480	0490-0235
42461	9100-1616	1	COTE-MED 5. OUH 10% G#45 .1550x.375LG	02178	15-4435-1K
10428 10428	5080-3078 5080-3078		TRANSISTOR NPN SI PD#300MW FT#200MHZ Transistor npn si PD#300MW FT#200MHZ		
A 2 4 R 7 A 2 4 R 3 A 2 4 R 4 A 2 4 R 5	0663-4715 0663-4725 0663-4725 0663-4725 0663-4725	:	RESISTOR 470 5% .25W FC TC=+400/+600 RESISTOR 4.7K 5% .25W FC TC=+400/+700	0160G 0160G 0160G 0160G 0160G	CB4715 CB4725 CB4725 CB4725 CB4725
A24R6 A24R7 A24RR A24RR A24R1	0683-2715 0683-2715 0683-2715 0683-2715 0683-2715		RESISTOR 270 5% .25% FC TC=-400/+600 RESISTOR 270 5% .25% FC TC=-400/+600 RESISTOR 270 5% .25% FC TC=-400/+600 HESISTOR 270 5% .25% FC TC=-400/+600 RESISTOR 270 5% .25% FC TC=-400/+600	0160G 0160G 0160G 0160G 0160G	CB2715 CB2715 CB2715 CB2715 CB2715
424R11 424R12	0683-2715 1810-0164		RESISTOR 270 5% .25W FC TC==400/+600 NETWORK=RES 9=PIN=SIP .15=PIN=SPCG	0160G 28480	CH2715 1810-0164
A24111 A2442 A2443 A2444 A2445	1820-1112 1820-1200 1820-1196 1820-1199 1820-1199		IC FF TIL LS D=TYPE POS=EDGE=TRIG IC INV TIL LS HEX I=INP IC FF TIL LS D=TYPE POS=EDGE=TRIG COM IC INV TIL LS HEX I=INP IC INV TIL LS HEX I=INP	0169H 0169H 0379D 0169H 0169H	SN74LS74N SN74LS05N AM74LS174N SN74LS04N SN74LS04N
A24U6 A24U7 A24U8 A24U9 A24U9	1820-1415 1820-1081 1820-0471 1820-0668 1820-0491	2	IC SCHMITT-TRIG TIL LS NAND DUAL 4-INP IC DRVR TIL BUS DRVR QUAD 1-INP IC INV TIL HEX 1-INP IC RFP ITL NON-INV HEX 1-INP IC DCDW TIL RCD-TO-DEC 4-TO-10-LINE	0169H 0379D 0223G 0223G 0169H	SN74L513N Ambt26 7406PC 7407PC SN74145N
A24111 A24U12 A24U13	1820-1195 1820-1081 1820-1081		IC FF ITE LS D-TYPE POS-EDGE-TRIG COM IC DRVR ITE BUS DRVR QUAD 1-INP IC DRVR ITE BUS DRVR QUAD 1-INP	03790 03790 03790	AM74L5175A AM8126 AM8126
454M1	04261-72009	3	CAHLE ASSEMBLY	28480	04261-72009
#25 (OPTION 101)	04262+66525 04262 <b>-</b> 26525	1 1	HP-IR INTERFACE BUARD ASSEMBLY PC BOARD, BLANK	28480 28480	04262=66525 04262=26525
425C1 425C2 425C3 425C4 425C5	0140~0291 0160-2055 0160-2055 0160-2055 0160-2704		CAPACITOR-FXD 1UF++10% 35VDC TA CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD 100PF +-5% 300VDC MICAO+70	0450J	150D105x903542
A25C6 A25C7	0160-2204 0160-0153	1	CAPACITOR-FXD 100PF +-5% 300VDC MICAO+70 CAPACITOR-FXD 1000PF +-10% 200VDC POLYE	28460 0420J	4055-0410 585014585
1254 5758	1 <b>251-0541</b> 1200-0853		CONNECTOR 34-PIN M RECTANGULAR SUCKET-IC 16-CUNT DIP-SLDR	70381	3451-1002
A2501	5080-3078		TRANSISTUR NPN SI PD#300Mm FT#200MHZ	.	
A25R1 A25R2 A25R3 A25R4 A25R5	0683-4715 0683-4715 0683-4715 0683-4715 0683-4715		RESISTOR 470 5% .25w FC TC=-400/+600 RESISTOR 470 5% .25w FC TC=-400/+600 RESISTOR 470 5% .25w FC TC=-400/+600 RESISTUK 470 5% .25w FC TC=-400/+600 RESISTOK 470 5% .25w FC TC=-400/+700	0160G 0160G 0160G 0160G 0160G	CB4715 CB4715 CB4715 CB4715 CB4715 CB1825
A25R6 A25R7	1810-0136 1810-0125	2	NETWORK-RES 10-PIN-SIP .1-PIN-SPCG NETWORK-RES 8-PIN-SIP .125-PIN-SPCG	28480 0248C	1810=0136 750
A25U1 A25U2 A25U3 A25U4 A25U5	1820-1197 1820-1558 1820-1558 1820-1199 1820-0269	2	IC GATE TIL LS NAND QUAD 2-INP IC MISC TIL+ QUAD IC MISC TIL+ QUAD IC INV TIL LS HEX 1-INP IC GATE TIL NAND QUAD 2-INP	0169H 0203G 0203G 0169H 0223G	5N74L300N MC3441P MC3441P 5N74L504N 7403PC

Table 6-3. Replaceable Parts (Cont'd).

Reference Designation	HP Part Number	Qty	Description	Mfr Code	Mfr Part Number
A25U6 A25U7 A25U8 A25U9 A25U10	1820-1199 1820-1201 1820-1195 1820-1195 1820-1195		IC INV TTL LS MEX 1-INP IC GATE TTL LS AND QUAD 2-INP IC FF TTL LS D-TYPE PUS-EDGE-TRIG COM IC FF TTL LS D-TYPE POS-EDGE-TRIG COM IC MUXR/DATA-SEL TTL LS 2-TO-1-LINE QUAD	0169H 0169H 0379D 0379D 0379D	SN74LS04N SN74LS08N AM74LS175A AM74LS175A SN74LS157N
A25U11 A25U12 A25U13 A25U14 A25U15	1820=1470 1820=1195 1820=1195 1820=1081 1820=1081		IC MUXR/DATA=SEL TIL LS 2=TO=1=LINE QUAD IC FF TIL LS D=TYPE POS=EDGE=TRIG COM IC FF TIL LS D=TYPE POS=EDGE=TRIG COM IC DRVR TIL BUS DRVR QUAD 1=INP IC DRVR TIL BUS DRVR QUAD 1=INP	0379D 0379D 0379D 0379D 0379D	SN74LS157N AM74LS175A AM74LS175A AM6126 AM8126
A25U16 A25U17 A25U18 A25U19 A25U20	1820-1081 1820-1081 1820-1081 1820-1081 1820-0328	1	IC DRVR TIL BUS DRVR QUAD 1-INP IC GATE TIL NOR QUAD 2-INP	0379D 0379D 0379D 0379D 0223G	AM8126 AM8126 AM8126 AM8126 7402PC
15055 452055	1820-1112 1820-1112		IC FF TIL LS DOTYPE POS-EDGE-TRIG IC FF TIL LS DOTYPE POS-EDGE-TRIG	0169H 0169H	3N74L374N 5N74L374N
A26			NOT ASSIGNED		
427			NOT ASSIGNED		
858		:	NOT ASSIGNED		
429		<b>.</b>	NOT ASSIGNED		
A 30			NOT ASSIGNED		
A 3 1			NOT ASSIGNED		
432			NGT ASSIGNED		
a 3 3			NUT ASSIGNED		
A 3 4			NOT ASSIGNED		
435 (OPTION 001)	04262=66535 04262=26535	1 1	BCD DUTPUT CONTROL BOARD ASSEMBLY PC BOARD, BLANK	28480 28480	04262=66535 04262=26535
A35C1 A35C2 A35C3 A35C4 A35C5	0160-2199 0160-2199 0180-0229 0160-2055 0160-2055		CAPACITON=FXD 30PF +=5% 300VDC CAPACITOR=FXD 30PF +=5% 300VDC CAPACITOR=FXD 33UF+=10% 10VDC TA CAPACITOR=FXD 01UF +80=20% 100VDC CER CAPACITOR=FXD 01UF +80=20% 100VDC CER	28480 28480 0420J	120 <u>032</u> 0x40108S 0100=5144 0100=5144
A35C6 A35C7 A35C8	0160-2055 0160-2055 0160-2055		CAPACITUR-FXD .01UF +80-20X 100VDC CER CAPACITUR-FXD .01UF +80-20X 100VDC CER CAPACITUR-FXD .01UF +80-20X 100VDC CER		
A35CR1 A35CR2	1902-0041 1902-0041		DIODE-ZNR 5.11V 54 DO-7 PD=.4W TC=0094 DIDDE-ZNR 5.11V 54 DO-7 PD=.4W TC=0094	0203G 0203G	32 10939-98 32 10939-98
A35J1	1200-0853		SOCKET-IC 16-CONT DIP-SLDR		
A35L1	9100=1611	1	COIL-MLD 220NH 20% 9#50 ,1550%,375LG	02178 0160G	15-4415-2M CB5025
A35R1 A35R2 A35R3 A35R4 A35R5	0683-5625 0683-5625 0683-5625 0683-5625 0683-5625		RESISTOR 5.6K 5% .25W FC TC==400/+700 RESISTOR 5.6K 5% .25W FC TC=+400/+700	0160G 0160G 0160G 0160G	CB5025 CB5025 CB5025 CB5025 CB5025
A35R6 A35R7 A35R8 A35R9 A35P10	0683=5625 0683=5625 0683=2225 0683=2225 0683=5625		RESISTOR 5.6K 5% .25W FC TC==400/+700 RESISTOR 5.6K 5% .25W FC TC==400/+700 RESISTOR 2.2K 5% .25W FC TC==400/+700 RESISTOR 2.2K 5% .25W FC TC==400/+700 RESISTOR 5.6K 5% .25W FC TC==400/+700	0160G 0160G 0160G 0160G	CH5625 CH5625 CH2225 CH2225 CH5625
A35R11 A35R12	0663-5625 1810-0136		RESISTOR 5.6K 5% 25W FC TC==400/+700 NETWORK=RES 10-PIN-SIP .1-PIN-SPCG	0160G 28480	C85625 1810=0136
A 35 S 1 A 35 S 2 A 35 U 1 A 35 U 2 A 35 U 3 A 35 U 4 A 35 U 5	3101-0299 3101-1273 1820-1423 1820-0077 1820-1197 1820-0294 1820-0294	1 1	SWITCH, SLIDE 4-SPST SWITCH, DPDT-NS IC MV TIL LS MONOSTBL RETRIG DUAL IC FF TIL D-TYPE POS-EDGE-TRIG CLEAR IC GATE TIL LS NAND GUAD 2-INP IC SHF-RGTR TIL R-S SERIAL-IN PHL OUT IC SHF-RGTR TIL R-S SERIAL-IN PHL OUT	28480 -0169H 0223G 0169H 0340F 0340F	3101-0299 SN74LS123N 7474PC 3874LS00N DM8570N DM8570N

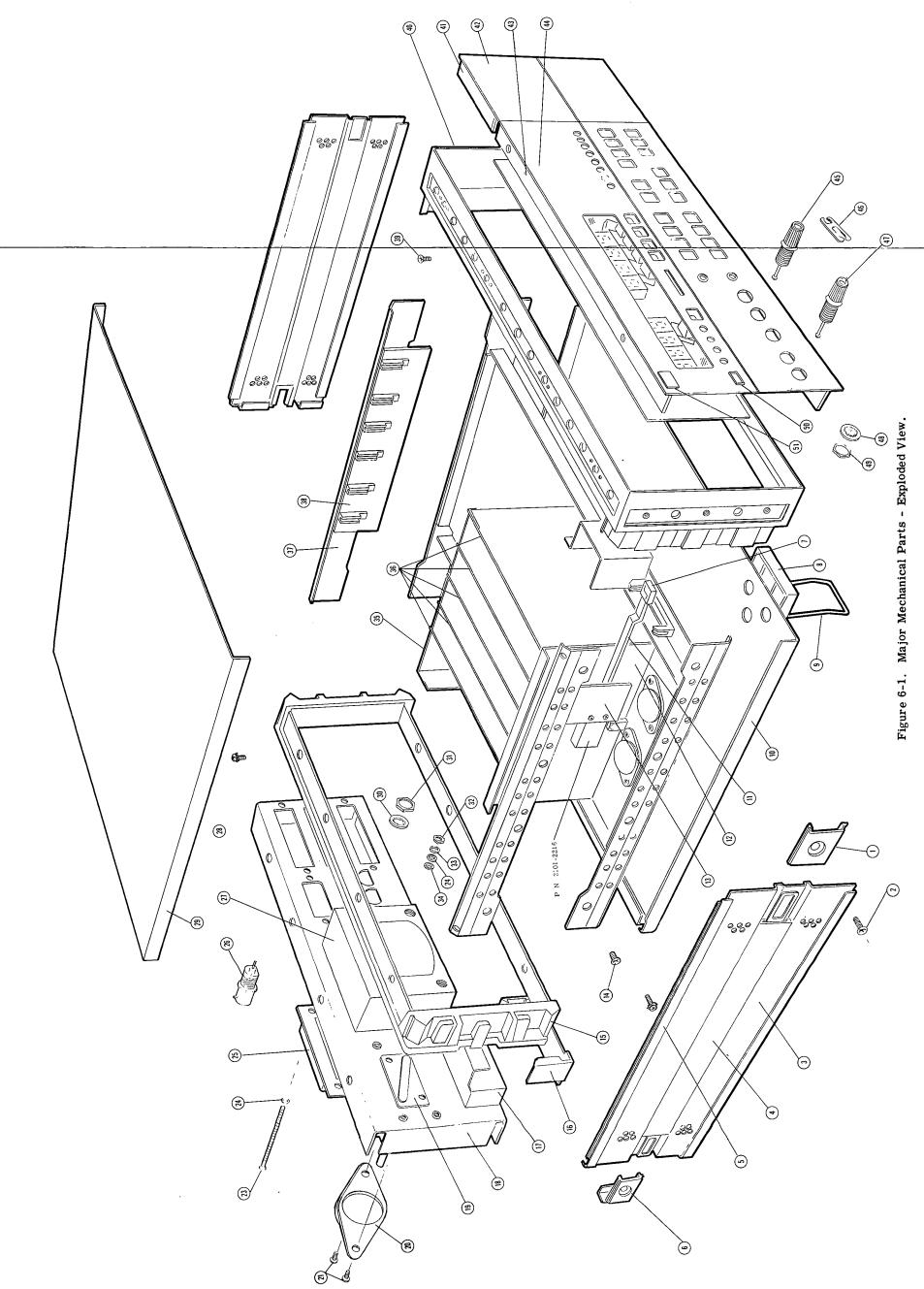
Table 6-3. Replaceable Parts (Cont'd).

Reference Designation	HP Part Number	Qty	Description	Mfr Code	Mfr Part Number
A35116 A3517 A3508 A3509 A3509	1820-0294 1820-0294 1820-0668 1820-0694 1820-1081		IC SHF-RGTR TIL R-S SERIAL-IN PRL OUT IC SHF-RGTR TTL R-S SERIAL-IN PRL OUT IC BFR TTL NON-INV HEX 1-INP IC SHF-PGTR TIL R-S SERIAL-IN PRL OUT IC OFF-PGTR TIL R-S SERIAL-IN PRL OUT IC OFF TTL BUS DRVR QUAD 1-INP	0340F 0340F 0223G 0340F 0379D	DM8570N DM8570N 7407PC DM8570N AMHT26
A35U11 A35U12 A35U13	1820-0294 1820-0294 1820-0294		IC SHF-RGTR TIL R-S SERIAL+IN PRL OUT IC SHF-RGTR TIL R-S SERIAL+IN PRL OUT IC SHF-RGTR TIL R-S SERIAL-IN PRL OUT	0340F 0340F 0340F	DM8570N DM8570N DM8570N
A 35W1 A 35W2	04261-72009 04261-72009		CABLE ASSEMBLY CABLE ASSEMBLY	28480 28480	04261-72009 04261-72009
			·		
			•		
	II.		CHASSIS MOUNTED COMPONENTS		
C1 C2 C3	0160-4259 0160-1586 0160-1586	1 2	CAPACITOR FXD .22UF 10% CAPACITOR FXD .1UF 200VDC CAPACITOR FXD .1UF 200VDC		
CR1, CR2 CR3 CR4 ~ CR7 F1	1901-0496 1902-1232 1901-0033 2110-0007 2110-0202	2 1 4 1	DIODE:RECTIFIER POWER DIODE:ZNR IN3997AR 5.6V PD = 10W DIODE Ge 180V 200mA FUSE 1A 250V FUSE .5A 250V	j	
J6, J7, J8	5060-4020	3	CONNECTOR ASSEMBLY,50 CONTACTS (OPT. 001/004)	:	
A3 Q1, Q2, Q3	04262-66503 0380-0644 2190-0034 1854-0063	1 2 2 3	CONNECTOR BOARD ASSEMBLY, HP-IB (OPT. 101) SCREW, STAND OFF WASHER SP WASHER SP TRANSISTOR NPN 2N3055		
R1 R2, R3 R4 R5 S1 S2, S4	0683-1025 0698-3391 2100-1250 2100-1832 3101-2216 3100-1201	2 1 1 1 2	RESISTOR 1k 5% .25W RESISTOR 21.5 1% .5W RESISTOR-VAR 500 20% RESISTOR-VAR 500 10% SWITCH:LINE SWITCH:THUMBWHEEL (OPT. 004)		
			CABLE ASSEMBLIES		
	8120-0360 04262-61601 04262-61602 04262-61603 04262-61604 04262-61605 04262-61901	] ] ] ]	FLAT CABLE ASSY (OPT. 001, 004, 101) CABLE ASSEMBLY, Lc, 19cm CABLE ASSEMBLY, Lp, 19cm CABLE ASSEMBLY, Hc, 16cm CABLE ASSEMBLY, Hp, 22cm CABLE ASSEMBLY, Hp, 18cm CABLE ASSEMBLY, LINE SWITCH		
			MISCELLANEOUS		
	5001-0439 5040-7202 04261-40024 04262-40002 04262-85001	2 1 1 1	TRIM, SIDE TRIM, TOP LAMP HOUSE, UNIT INDICATOR WINDOW ANNUNCIATOR FILM, UNIT		
TOOL	8710-0340		SCREWDRIVER (FURNISHED)		
			•		
	!				

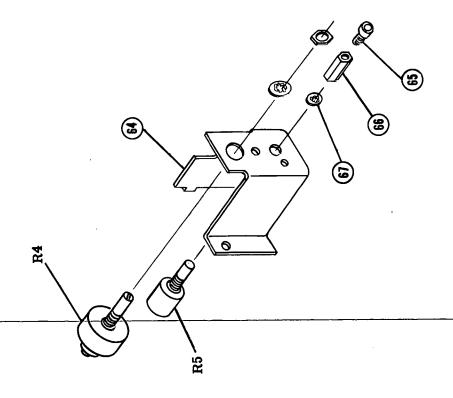
Model 4262A Section VI

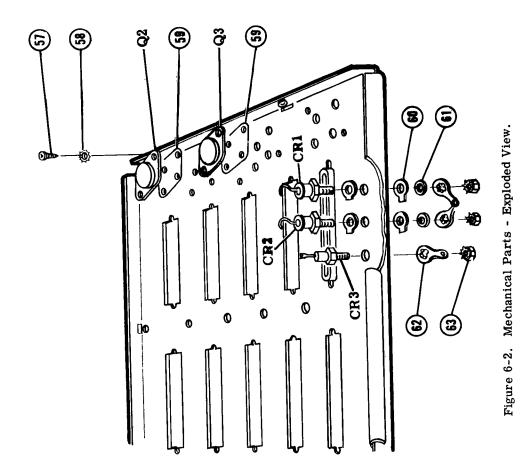
Table 6-3. Replaceable Parts (Cont'd).

Reference Designation	HP Part Number	Qty	Description	Mfr Code	Mfr Part Number
CHASSIS PARTS					***
1 2 3 4 5	5040-7219 2680-0172 5060-9935 5060-9802 2360-0115	2 4 2 2 2 6	CAP HANDLE FRONT SCREW-MACH 10-32 .375-IN-LG COVER. SIDE HANDLE SREW-MACH 6-32 .312-IN-LG		
6 7 8 9	5040-7220 0370-2159 5040-7201 1460-1345 5060-9845	2 1 4 2 1	CAP HANDLE REAR KNOB:PUSHBUTTON LINE FOOT, FULL/HALF MODULE STAND TILT COVER, BOTTOM		
11 12 13 14	5040-7023 04262-00602 04262-00606 2510-0192 5020-8804	1 1 1 16 1	ROD, PUSHBUTTON DECK, LEFT PLATE, LINE SWITCH SCREW-MACH 8-32 .25-IN-LG FRAME, REAR		
16 17 18 19 20	5040-3318 0960-0443 04262-00205 1200-0041 .0340-0833	1 1 1 3 1	COVER, L MODULE LINE MODULE PANEL, REAR SOCKET, TRANSISTOR COVER, TRANSISTOR		
21 22	2190-0020	2	SCREW		
23 24 25	2510-0135 3050-0139 7100-0129	4 8 1	SCREW-MACH 8-32 2.25-IN-LG WASHER FL MTLC NO8 COVER, POWER TRANSFORMER		
26(J9, J10) 27 28 29 30	1250-0118 9100-0865 2360-0113 5060-9833 2190-0016	2 1 8 1 3	CONNECTOR, BNC TRANSFORMER, POWER SCREW-MACH 6-32 .25-IN-LG . COVER, TOP WASHER-LK INTL T NO3/8		
31 32 33 34 35	2950-0001 2580-0004 2190-0087 3050-0239 04262-00603	2 4 4 4 1	NUT-HEX-DBL-CHAM 3/8-32-THD NUT-HEX-DBL-CHAM 8-32-THD WASHER-LK HLCL NO8 WASHER-FL NM NO8 DECK, CENTER		
·36 37 38 39 40	04262-00605 5020-8835 04262-00604 2360-0333 5020-8803	5 4 1 1	PLATE, SHIELD STRUT CORNER DECK, RIGHT SCREW-MACH 6-32 .25-IN-LG FRAME, FRONT		
41 41 42 42 43	04262-00204 04262-00214 04262-00202 04262-00212 04262-00203	] ] ]	SUB PANEL, FRONT (STD) SUB PANEL, FRONT (OPT. 004) PANEL, FRONT (STD) PANEL, FRONT (OPT. 004) SUB PANEL, FRONT		
44 44 45 (J2 - J5) 46 47 (J1)	04262-00201 04262-00211 1510-0090 5000-4206 1510-0107	1 1 4 2 1	PANEL, FRONT (HP) PANEL, FRONT (YHP) BINDING POST GRAY SHORTING LINK BINDING POST BLK		
48 49 50 51 51	2190-0016 2950-0043 0370-0451 7120-1254 7120-0478	2 5 1 1 1	WASHER-LK INTL T NO3/8 NUT-HEX-DBL-CHAM 3/8-32-THD BEZEL, PUSHBUTTON LINE TRADE MARK (HP) TRADE MARK (YHP)		
52 53 54 55 56	04262-00607 2360-0115 0520-0129 04262-00608 2420-0006	1 2 6 3 2	PLATE, BLIND SCREW-MACH 6-32 .312-IN-LG SCREW-MACH 2-56 .312-IN-LG PLATE, BLIND NUT-HEX-W/LKWR 6-32-THD		
57 58 59 60 61	0624-0045 2190-0008 0340-0458 1200-0080 3050-0226	6 6 3 4 2	SCREW-TPG 6-20 .375-IN-LG WASER-LK EXT T NO6 INSULATOR, TRANSISTOR INSULATOR, DIODE WASHER-FL MTLC NO10		;
62 63 64 65 66 67	0360-0270 2740-0003 04262-01201 1490-0848 0590-0061 2190-0060	3 3 1 1 1	SOLDER LUG NUT-HEX-W/LKWR 10-32-THD PLATE, ANGLE BUSHING NUT-HEX-DBL-CHAM 1/4-32-THD WASHER-LK INTL T NO1/4		
			•		



Model 4262A





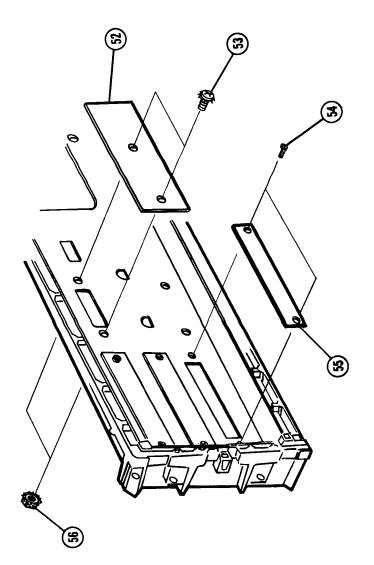


Table 6-4. Option 010 Modification

Reference Designation	HP Part Number	Qty	Description
All (OPTION 010)	04262-66911		OSCILLATOR & SOURCE RESISTOR BOARD ASSY
Al 1C8 Al 1C9	0160-5821 0160-5821		CAPACITOR, FXD 5000pF .25% 50V CAPACITOR, FXD 5000pF .25% 50V
A11R25 A11R26 A11R27 A11R28 A11R29 A11R30	0698-4494 0698-2228 0757-0279 0757-0123 0698-2228 0757-0279		RESISTOR 35.7K 1% RESISTOR 318.3K .5% RESISTOR 3.16K 1% RESISTOR 34.8K 1% RESISTOR 318.3K .5% RESISTOR 3.16K 1%
			other parts are same as 04262-66511
A14 (OPTION 010)	04262-66914		PHASE DETECTOR & INTEGRATOR BOARD ASSY
A14C7	0160-1554		CAPACITOR, FXD .47UF 200V
A14R33	0698-4511		RESISTOR 86.6K 1%
		i	other parts are same as 04262-66514
	-		
	1		

# SECTION VII MANUAL CHANGES

#### 7-1. INTRODUCTION.

7-2. This section contains information for adapting this manual to instruments to which the contents do not directly apply. The following paragraphs explain how to adapt this manual to apply to older instruments with a lower serial prefix.

## 7-3. MANUAL CHANGES.

- 7-4. To adapt this manual to your particular instrument, refer to Table 7-1 and make all of the manual changes listed opposite your instrument serial number. Perform these changes in the summary by assembly.
- 7-5. If your instrument serial number is not listed on the title page of this manual or in Table 7-1 to the right, it may be documented in a yellow MANUAL CHANGES supplement. For additional information about serial number coverage, refer to INSTRUMENT COVERED BY MANUAL in Section I.

Table 7-1. Manual Changes by Serial Number.

Serial Prefix or Number	Make Manual Changes
1710J00260 and below	1, 2, 3
1710J00340 and below	2, 3
1739J00600 and below	3, 5
1739J02280 and below	4, 5
2022J03750 and below	5

Table 7-2. Summary of Changes by Assembly (Continued on Page 7-2).

	Assembly											
A1	A2	A3	A4	A5	A9	A11	A12					
					-		04262- 26512					
							04262- 66612					
	,											
							-					
	Al			A1 A2 A3 A4	A1 A2 A3 A4 A5	A1 A2 A3 A4 A5 A9	A1 A2 A3 A4 A5 A9 A11					

Table 7-2. Summary of Changes by Assembly (Continued).

CHANGE					Assembly	7			
GHANGE	A13	A14	A21	A22	A23	A24	A25	A35	No Prefix
1				04262- 66522					
2	_				04262- 66623				
3									
4									
5					U15 U16				

Pages 6-16 and 6-17, Table 6-3, Replaceable Parts, Change A22 board parts list to Table A.

Page 8-61, Figure 8-46, A22 schematic diagram,
Partially change Figure 8-46 as shown in Figure A.

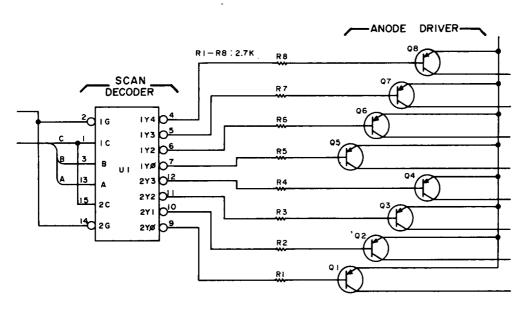


Figure A.

Table A.

Reference Designation	HP Part Number	Qty	Description	Mfr Code	Mfr Part Number
55A	04262-66522 04262-26522	2	DISPLAY CONTROL & RAM BOAKO ASSEMBLY PC HOARD, BLANK	28480 28480	04565-5625 04565-66255
A 2 2 C C A 2 2 C C C A 2 2 C C C A 2 2 C C C A 2 2 C C C A 2 2 C C C A 2 2 C C C A 2 2 C C C A 2 2 C C C A 2 C C C A 2 C C C A 2 C C C A 2 C C C C	0180-0291 0160-2055 0160-2055 0160-2055 0160-2055		CAPACITOR=FXO 1UF+=10% 35VDC TA CAPACITOR=FXD .01UF +80-20% 100VDC CER	04507	1500105%903542
42264 42367 42368 42369 423610	0160-2204 0160-2261 0160-0939 0180-0291 0160-0939	i)	CAPACITUR-FXD 100PF +=5% 300VDC MICA0+70 CAPACITUR-FXD 15PF +=5% 500VDC CER0+=30 CAPACITUR-FXD 430PF +=5% 300VDC MICA0+70 CAPACITOR-FXD 10F++10% 35VDC TA CAPACITOR-FXD 450PF +=5% 300VDC MICA0+70	28480 0420J 0420J 28480 28480	0160=2204 0160=2261 0160=0939 1500105×9035A2 0160=0939
455C11 455C15	0169-0939 0169-2205	۶	CAPACITUR-FXD 430PF +-5% 300VDC MICAO+70 CAPACITUR-FXD 120PF +-5% 300VDC MICAO+70	28480 28480	9390-0410 2055-0410
422CR1	1902-0041		DIODE-ZNR 5.11V 5% 00-7 PD#,4W TC==,009%	05036	S7 10939-9H
15521	1200-0541	1	SUCKET-IC 24-CONT DIP-SEDR		
A2201 A2202 A2203 A2204 A2205	1853-0107 1853-0107 1853-0107 1853-0107 1853-0107		TRANSISTOR, PNP SI	\$8480 \$8480 \$8480 \$8480 \$8480	1853-0107 1853-0107 1853-0107 1853-0107 1853-0107
42846 70854 40554	1853-0107 1853-0107 1853-0107		TRANSISTOR, PNP SI TRANSISTOR, PNP SI TRANSISTOR, PNP SI	28480 28480 26460	1855-0107 1853-0107 1855-0107
A22H1 A22H7 A22K3 A22K4 A22H5	0683-2735 0683-2735 0683-2735 0683-2735 0683-2735	8	RESISTOR 27K 5% .25W FC TC==400/+600 RESISTOR 27K 5% .25W FC TC==400/+600	0160G 0160G 0160G 0160G 0160G	CH2735 CH2735 CH2735 CH2735 CH2735
42246 42247 42244 42249 422410	0643-2735 0643-2735 0643-2735 0643-5605 0643-5605		PESISTOM 27K 5% .25W FC TC=-400/+800 RESISTOM 27K 5% .25W FC TC=-400/+800 HESISTOM 27K 5% .25W FC TC=-400/+800 RESISTOM 50 5% .25W FC TC=-400/+500 RESISTOM 50 5% .25W FC TC=-400/+500	01606 01606 01606 01606	CH2735 CH2735 CH2735 CH3605 CH5605
A22H11 A22R12 A22H13 A22H14 A22H15	0643-5605 0643-5605 0643-5605 0643-5605 0663-5605		RESISTOR 56 5% .25% FC TC==400/+500 RESISTOR 56 5% .25% FC TC==400/+500	01606 01606 01606 01606 01606	L #3605 C #3605 C #3605 C #3605 C #3605
A22R16 A22R17 A22R18 A22R19 A22K20	06#3=5605 06#3=2725 06#3=1#25 06#3=4725 1810=0121		HESISTON 56 5% .25% FC TC==400/+500 RESISTOP 2,7% 5% .25% FC TC==400/+700 RESISTOR 1.6% 5% .25% FC TC==400/+700 RESISTOW 4,7% 5% .25% FC TC==400/+700 NETWORK=PES 9=PIN=SIP .15=PIN=SPC/	0160G 0160G 0160G 0160G 28480	(85005 CH2725 CH34725 CH4725 1810~0121
422821 422822 A22839	1810-0205 1810-0206 1810-0164	2	NETWORK-RES H-PIN-SIP .I-PIN-SPCG NETWORK-HES R-PIN-SIP .I-PIN-SPCG NETWORK-RES 9-PIN-SIP .IS-PIN-SPCG	0248C 0374D 28480	750-61-R4.7K 4508R-101-103S 1810-0164
42251	3101-0299		SWITCH, SLIDE 4-SPST	28480	3101-0299
A22U1 A22U2 A22U3 A22U4 A22U5	1820-1245 1820-1194 1820-1199 1820-1201 1820-1201		IC DCDR TIL LS 2-TD-4-LINE DUAL 2-INP IC CNTR TIL LS HIN UP/DDWN SYNCHHO IC INV TIL LS HEX 1-INP IC GATE TIL LS AND GUAD 2-INP IC DCDR TIL HCD-TD-7-SLG	0169H 0379D 0169H 0169H 0169H	5N74L8155N AM74LS193PU SN74L504N SN74L508N SN74L5247N
A22H6 A22U7 A22UA A22U9 A22U10	1820-9567 1820-1490 1858-0033 1820-0628 1820-1470	8	IC MV TIL DUAL IC CNTR TIL LS DECD ASYNCHRO TRANSISTOR FT5712M IC SN7489N 64-BIT RAM TIL IC MUXRZDATA-SEL TIL LS 2-T0-1-LINE QUAD	0203G 0169H 28480 0340F 03790	MC4024P SN74L590H DM7489N SN74L5157N
A22U11 A22U12 A22U13 A22U14 A22U14	1820+1425 1820+1112 1820+1197 1820+1490 1820+1478		IC SCHMITT-TRIG TIL LS NAND QUAD 2-INP IC FF TIL LS D-TYPE POS-EDGE-TRIG IC CATE TIL LS NAND QUAD 2-INP IC CNTR TIL LS DECD ASYNCHRO IC CNTR TIL LS BIN ASYNCHRO	0169H 0169H 0169H 0169H 0169H	SN74LS132N SN74LS74N SN74LS96N SN74LS96N SN74LS93N
A22016 A22017 A22018 A22019 A22020	1858-0033 1820-0628 1820-1470 1820-1081 1820-1081		TRANSISTOR FT5712M  IC SN7489N 64-81T RAM TTL  IC MUXR/DATA-SEL TTL LS 2-10-1-LINE GUAD  IC DRYR TTL BUS DRYR GUAD 1-INP  IC DRYR TTL BUS DRYR GUAD 1-INP	28480 0340F 03790 03790 03790	DM7459N 5N74L5157N AM5726 AM5726
150554 550554	1820-1196 1818-0135	4	IC FF TTL LS D-TYPE POS-EDGE-TRIG COM IC MC 0810L-1 1K RAM NMUS	0379D n203G	AM74LS174N MCG810L=1
A22Y1	0410-0209	2	CRYSTAL, QUARTZ	28480	0410-0209

Page 6-17, Table 6-3, Replaceable Parts, Change A23 board parts list to Table B.

Page 8-63, Figure 8-47, A23 Component Locations, Change Figure 8-47 to Figure B.

Page 8-63, Figure 8-48, A23 schematic diagram, Change Figure 8-48 to Figure C.

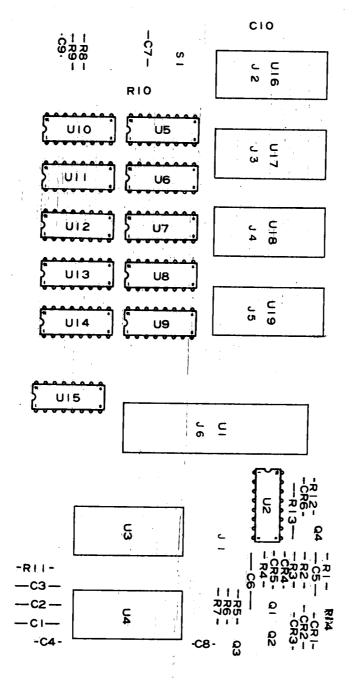


Figure B.

Table B.

Designation	HP Part Number	Qty	Description	Mfr Code	Mfr Part Number
		,	FROCESSOR & ROM BUARD ASSEMBLY	28480	04262+66523
453	04262-66523 04262-26523	1 1	PC HOARD, BLANK	28480	04262-26523
A23C2 A23C3 A23C4 A23C5	0150-0291 0180-0197 0180-0197 0160-2055 0180-0291		CAPACITUR-FXD 1UF++10X 35VDC TA CAPACITUR-FXD 2.2UF++10X 20VDC TA CAPACITUR-FXD 2.2UF++10X 20VDC TA CAPACITUR-FXD 3.0UF +80-20X 100VDC CEN CAPACITUR-FXD 1UF++10X 35VDC TA	10200 10500 10500 10500	1500105x9035A2 1500225x9020A2 1500225x9020A2
A23C6 A23C7 A23C8 A23C9 A23C9	0180-2141 0180-0229 0160-2055 0160-2055 0160-2055	1 4	CAPACITOR-FXD 3.3UF+-10% 6VDC TA CAPACITOR-FXD 33UF+-10% 10VDC TA CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD .01UF +80-20% 100VDC CEP CAPACITOR-FXD .01UF +80-20% 100VDC CER	CuSuO	1500336x9010h2
A23CP2 A23CP2 A23CR3 A23CR4 A23CR4	1902-315A 1902-315A 1902-1299 1902-0048 1901-0040	1	DIODE, ZENER, 9.76V CIODE, ZENER, 3.3V DIODE-ZNE 6,81V 5% DD-7 PDE,4W TC=+,043% DIODE-SWITCHING 30V 50MA 2NS DD-35 DIODE-SWITCHING 30V 50MA 2NS DD-35	0223G 0203G 0223G 28480 28480	F77459 SZ11213-1 F27244 1901-0040 1901-0040
A23CR6	1902-3107	1	DIODE-ZNK 5.76V 2% DO-7 PD=.4% 1C=+,017%	02036	87 10939-114
1.625 2.625 2.625 2.625 2.625 2.625	1200-0853 1200-0541 1200-0541 1200-0541 1200-0541	1	SOCKET-IC 16-CONT DIP-SLDR SOCKET-IC 24-CONT DIP-SLDR SOCKET-IC 24-CONT DIP-SLDK SUCKET-IC 24-CONT DIP-SLDR SOCKET-IC 24-CONT DIP-SLDR	\$8480 \$8480 \$8480 \$8480	1200-0541 1200-0541 1200-0541 1200-0541
A23J6	1200-0654		SOCKET-IC 40-CONT DIP-SLDR		
A2301 A2302 A2303 A2304	5080-3078 1854-0215 1854-0477 1853-0012	i i	THANSISTOR MPN SI PD=300M, FT=200PH/ THENSISTOR MPN SI PD=350Mm FT=360MH/ THANSISTOR MPN 2M2222A SI TU=18 PD=500MM THANSISTOR PMP 2M2290MA SI TU=39 FU=600MM	02036 02236 01698	245557 545557 5456447
A23R1 A23R2 A23R3 A23R4 A23R4	06-3-1035 06-43-16-45 06-43-10-55 06-43-10-35 06-43-56-25	1	#ESISTON 10K 5% .25% FL 1C==#00/+700 #ESISTOP 180K 5% .25% FC 1C==600/+700 #ESISTON 10K 5% .25% FC 1C==600/+700 #ESISTOM 10K 5% .25% FC 1C==400/+700 #ESISTOM 5,6K 5% .25% FC 1C==400/+700	01606 01606 01606 01606 01606	CM1035 CM1M45 CM1055 CM1035 CB5625
A23R6 A23R7 A23RR A23R9 A23R10	0698-5430 0643-5615 0683-4725 0683-4725 1819-0164	1	RESISTOR 21,5 1% .125% F TC=0++100  RESISTOR 4,7% 5% .25% FC TC=+400/+000  RESISTOR 4,7% 5% .25% FC TC=+400/+700  RESISTOR 4,7% 5% .25% FC TC=+400/+700  NETWORK-PES 9-PIN-SIP .15-PIN-SPCG	03858 01606 01606 01606 28480	FMESS=170=10=1185=1 CMS015 CM4725 CM4725 1810=0164
A23R11 A23R12 A23R13 A23R14	0643-1025 0757-0418 0698-3391 2100-2033	1 1	RESISTOR 1K 5% ,25% FC TC==400/+660 RESISTOR 619 1% ,125% F TC=0+=100 RESISTOR 21,5 1% ,5% F TC=0+=100 RESISTOR=TRMK 1K 10% C 510E=ADJ 1=TRN	20610 18580 18680 18680	(8)025 (4=1/8=10=6)9x=F (MF=65=2 ET50X102
42351	3101-0299	1	SWITCH, SLIDE 4-SPST	28480 28480	3101-0299
42301 42302 42303 42304 42305	1820=1691 1820=1197 1820=0702 1820=0702 1820=1081	2	IC MICPROC MOS IC CATE TIL LS HAND QUAD 2-INP IC DCDR TIL L 4-TU-16-LINE 4-INP IC DCDR TIL L 4-T0-16-LINE 4-INP IC DRVR TIL BUS DRVR HUAD 1-INP IC DRVR TIL BUS DRVR HUAD 1-INP	0169H 0223G 0223G 0273G	91.11PC 93.11PC AMOTES
A2306 A2307 A2308 A2309 A23010	1820-1081 1820-1195 1820-1196 1820-1418 1820-0471	1 2	IC DRVR TIL BUS DRVR UUAD 1-INP IC FF TIL LS D-TYPE PUS-EDGE-TRIG COM IC FF TIL LS D-TYPE POS-EDGE-TPIG COM IC DCDR TIL LS BCD-TO-DEC 4-TO-10-LINE IC INV ITL HEX 1-INP	03790 03790 03790 0169H 02236	AM6126 AM74LS1754 AM74LS174N SN74LS42N 7406PC
A23011 A23012 A23013 A23014 A23015	1620-1195 1820-1201 1820-1197 1820-1199 1820-1112		IC FF TTL LS D-TYPE PUS-EDGE-TPIG COM IC GATE TIL LS AND QUAD 2-INP IC GATE TIL LS NAND QUAD 2-INP IC INV TIL LS HEX 1-INP IL FF TTL LS G-TYPE PUS-EDGE-TPIG	0379D 0169H 0169H 0169H 0169H	AM74LS175A SN74LS08N SN74LS08N SN74LS08N SN74LS78N
	04262-85002 04262-65003 04262-85004	1 1	IC. RUM INTEL 2708 IC. RUM INTEL 2708	28480 28480 28480 28480	04262-65002 04262-65003 04262-65005

Page 8-51, Figure 8-34. Al2 Component Locations, Change Figure 8-34 to Figure D.

		* * ***	-
—C12 ÷R2	CIE	-R12+	
÷R2 -R3 -R3	0-   ***	.C5 (C4) QF (R)	
	<u></u> 	Q2 + R2 + - CR2 +	
C <sub>1</sub> 15	-R32-	-R14- Q3 -R3-	
-R37- -R38-	U2	-R15- Q4 -R4-	
	C14	-RI6- Q5 -CR6- Q6 —CI—	
C16	Q14	-R17	
CR   7 - R39	-R33 - -R34 -	-RI8- Q8	
CR   8	CR 14 - R35-	-CR8 R7-	
CR19	CR 15:	-CR9-	
CR20	CR16	-R20-	
-R41-	-R36-	CR10 -R21- K1	
Q16	Q15	Source	
CIB	C17	-R22- QIIR9-	
		С3	
	R42—	-R10-	
-R44-	R43— Q19	-CR3-	
-R45- -R46-	Q20	Q12 C6 -R23-	
-R47-	Q21 Q17	C7 -R24- Q13	
-R48 <i>-</i> -R49 <i>-</i>	Q22 Q18	C8 -R25RII- C9 -R26-	
-R50-	Q23	CIO -R27- CRI3	
C20	C19		

Figure D.

Page 6-8, Table 6-3, Replace Parts, Change Al2 board parts list to Table C.

Page 8-51, Figure 8-34. Al2 Compornent Locations, Change Figure 8-34 to Figure E.

Page 8-51, Figure 8-35. Al2 Schematic diagram, Change Figure 8-35 to Figure F.

Table C.

Reference HP Part Number		Qty	Description	Mfr Code	Mfr Part Number	
A12	04262-66512 04262-26512	1 1	RANGE RESISTOR BOARD ASSEMBLY PC BOARD, BLANK	28480 28480	04262-66612 04262-26612	
A12C3*	0160-0159 0140-0190 0121-0059	1 1	CAPACITOR-FXD 6800PF +-10X 200VDC POLYE CAPACITOR-FXD 39PF +-5X 300VDC *FACTORY SELECTED PART	0420J 72136	292P68292 DM15E390J0300WV1CR	
<b>412C</b> 4	0180-1051		CAPACITOR-V TRMR-CER 2-8PF 350V PC-MTG *FACTORY SELECTED PART  CAPACITOR, FXD 100 UF 16V M	73899	0180-1051	
412C5 412C6 412C7 412C8	0180=1051 0150=0050 0150=0050 0150=0050	6	CAPACITUR, FXD 100 UF 16V M CAPACITUR-FXD 1000PF +80=20% 1KVDC CEH CAPACITUR-FXD 1000PF +80=20% 1KVDC CEH CAPACITUR-FXD 1000PF +80=20% 1KVDC CEH	28480 28480 28480 28480	0180-1051 0150-0050 0150-0050 0150-0050	
A12C9 A12C11 A12C12 A12C13	0150-0050 0150-0050 0121-0105 0180-0269 0160-2150	1 1 1	CAPACITUR-FXD 1000PF +80-20% 1KVDC CER CAPACITOH-FXD 1000PF +80-20% 1KVDC CER CAPACITOR-V TRMH-CER 9-35PF 200V FC-MTG CAPACITOR-FXD 1UF+75-10% 150VDC AL CAPACITOR-FXD 33PF +-5% 300VDC	28480 28480 73899 0420J 28480	0150-0050 0150-0050 0111PR350 3001050150842 0100-2150	
A12C14* A12C15 A12C16 A12C17 A12C18	0160-2307 0180-1051 0180-1051 0180-1051 0180-1051	3	CAPACITOR-FXD 47pF +-5% 300VDC CAPACITOR, FXD 100 UF 16V M	58480 58480 58480 58480	0180-1051 0180-1051 0180-1051 0180-1051	
412C10	0180-1051 0180-1051		CAPACITOR, FXD 100 UF 16V M CAPACITUR, FXD 100 UF 16V M	28480 28480	0180-1051 0180-1051	
A12CR1 A12CR3 A12CR3 A12CR4 A12CR5	1901-0040 1901-0040 1901-0040 1901-0040 1901-0040	60	DIODE-SMITCHING 30V 50MA 2NS DO-35 DIODE-SMITCHING 30V 50MA 2NS DO-35 DIODE-SMITCHING 30V 50MA 2NS DO-35 DIODE-SMITCHING 30V 50MA 2NS DO-35 DIODE-SMITCHING 30V 50MA 2NS DO-35	28480 28480 28480 28480 28480	1901-0040 1901-0040 1901-0040 1901-0040 1901-0040	
A12CH6 A12CH7 A12CH9 A12CH9 A12CH10	1901-0040 1901-0040 1901-0040 1901-0040 1901-0040		DIODE-SWITCHING 30V 50MA 2NS 00=35 DIODE-SWITCHING 30V 50MA 2NS DO=35 DIODE-SWITCHING 30V 50MA 2NS DO=35 DIODE-SWITCHING 30V 50MA 2NS DO=35 DIODE-SWITCHING 30V 50MA 2NS 00=35	28480 28480 28480 28480 28480	1901-0040 1901-0040 1901-0040 1901-0040 1901-0040	
A12CR11 A12CR12 A12CR13 A12CR14 A12CR15	1901-0040 1901-0040 1902-3149 1901-0040 1901-0040		DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-ZNP 9,09V 52 DO-7 PDE,4W 1C=+.057X DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35	28480 28480 02236 28480 28480	1901-0040 1901-0040 FZ7256 1901-0040 1901-0040	
412CR16 412CR17 412CR1H 412CR19 412CR20	1901-0040 1901-0040 1901-0040 1901-0040 1901-0040		D10DE-SWITCHING 30V 50MA 2NS D0-35 D10DE-SWITCHING 30V 50MA 2NS D0-35 D10DE-SWITCHING 30V 50MA 2NS D0-35 D10DE-SWITCHING 30V 50MA 2NS D0-35 D10DE-SWITCHING 30V 50MA 2NS D0-35	28480 28480 28480 28480 28480	1901-0040 1901-0040 1901-0040 1901-0040 1901-0040	
A12K1	0490-0237	1	RELAY, REED 2A	26480	0490-0237	
A1201 A1202 A1203 A1204 A1205	5080-3830 5080-3830 5080-3830 1855-0128 5080-3830	1	TRANSISTOR J=FET N=CHAN D=MODE SI THANSISTOR J=FET N=CHAN D=MODE SI THANSISTOR J=FET N=CHAN D=MODE SI TRANSISTOR J=FET N=CHAN SI THANSISTOR J=FET N=CHAN D=MODE SI	28480		
A1206 A1207 A1208 A1209 A12010	5080 - 3830 5080 - 3830 5080 - 3830 5080 - 3830 5080 - 3830		TRANSISTOR J-FET N-CHAN D-MODE SI			
A12011 A12012 A12013 A12014 A12015	5080 - 3830 5080 - 3078 5080 - 3078 5080 - 3078 5080 - 3835 1854 - 0013	6	TRANSISTOR J=FET N=CHAN D=MODE SI TRANSISTOR NPN SI PD=300MM FT=200MHZ TRANSISTOR NPN SI PD=300MW FT=200MHZ TRANSISTOR J=FET 2N5245 N=CHAN D=MODE SI TRANSISTOR NPN 2N2218A SI TO=5 PD=800MW	0169H 0203G	2N5245 2N2218A	
A12016 A12017 A12018 A12019 A12020	1853-0012 1853-0020 1853-0020 1853-0020 5080-3078	5	TRANSISTOR PNP 2N2904A SI TU=39 PD#600MW TRANSISTOR PNP SI PD#300Mm FT#150MMZ TRANSISTOR PNP SI PD#300Mm FT#150MMZ TRANSISTOR PNP SI PD#300Mm FT#150MMZ TRANSISTOR NPN SI PD#300Mm FT#200MMZ	0169H 28480 28480 28480	2N2904A 1853-0020 1853-0020 1853-0020	
12021 A12022 A12023	1853-0020 1853-0020 1853-0020		TRANSISTOR PNP ST PD=300MW FT=150MMZ TRANSISTOR PNP SI PD=300MW FT=150MMZ TRANSISTOR PNP SI PD=300MW FT=150MMZ	28480 28480 28480	1853-0020 1853-0020 1855-0020	

Table C. (Cont'd).

Reference Designation	HP Part Number	Qty	Description	Mfr Code	Mfr Part Number
A12R1 A12R2 A12R3 A12R4 A12R5	2100-2514 0683-1055 0683-1055 0698-2298 0698-2294	1 35 1 1	RESISTOR-TRMR 20K 10% C SIDE-ANJ 1-TRN RESISTOR 1M 5% ,25% FC TC=-800/+900 RESISTOR 1M 5% ,25% FC TC=-800/+900 RESISTOR 10 .05% .33W RESISTOR 100 .1 .05%	0365A 0160G 0160G 28480 28480	£150m203 CB1055 CH1055 0698-2298 0698-2294
A12R6 A12R7 A12R8 A12R9 A12R10	0698+2296 0698-2214 0698-5408 0698-2225 0698-3329	1 1 1 1	RESISTOR 1010.1 .05%  RESISTOR: FXD .10.0K .0HM 0.05% 1/8K MF  RESISTOR 1.111K .1% .125W F TC=0+-100  RESISTOR: FXD .00.0K .0HM 0.05% 1/8W MF  RESISTOR 10K .5% .125W F TC#0+=100	28480 28480 28480 03888	0698-2296 0698-2214 0698-2225 PME55-178-10-1002-D
A12R11 A12R12 A12R13 A12R14 A12R15	0683-3335 0683-4705 0683-4705 0683-1055 0683-1055	4	RESISTOR 33K 5% ,25% FC TC=-400/+900 RESISTOR 47 5% ,25% FC TC=-400/+500 RESISTOR 47 5% ,25% FC TC=-400/+500 RESISTOR 1M 5% ,25% FC TC=-800/+900 RESISTOR 1M 5% ,25% FC TC=-800/+900	0160G 0160G 0160G 0160G 0160G	CB3335 CH4705 CB4705 CB1055 CH1055
A12R16 A12R17 A12R18 A12R19 A12R20	0683-1055 0683-1055 0683-1055 0683-1055 0683-1055		PESISTOR IM 5% .25W FC TC==000/+900 PESISTOR IM 5% .25W FC TC==000/+900 RESISTOR IM 5% .25W FC TC==000/+900 RESISTOR IM 5% .25W FC TC==000/+900 RESISTOR IM 5% .25W FC TC==000/+900	0160G 0160G 0160G 0160G 0160G	CH1055 CH1055 CH1055 CH1055 CH1055
A12R21 A12R22 A12R23 A12R24 A12R25	0683-1055 0683-1055 0683-3335 0683-3335 0683-3335		RESISTOR 1M 5% .25% FC TC==800/+900 HESISTOR 1M 5% .25% FC TC==800/+900 RESISTOR 33K 5% .25% FC TC=-400/+800 RESISTOR 33K 5% .25% FC TC=-400/+800 RESISTOR 33K 5% .25% FC TC=-400/+800	0160G 0160G	CB1055 CB1055
A12R26 A12R27 A12R28 A12R29 A12R30	0683-3335 0683-3335 0683-1035 0683-5055 0683-1035	51	RESISTOR 33K 5% .25W FC TC=-400/+800 RESISTOR 33K 5% .25W FC TC=-400/+800 ht SISTOP 10K 5% .25W FC TC=-400/+700 PESISTOP 5.6W 5% .25W FC TC=-900/+1100 PESISTOR 10K 5% .25% FC TC=-400/+700	0160G 0160G	CH1035 CH5655
A12R31 A12R32 A12R33 A12R34 A12R35	0683-3325 0683-1065 0683-1055 0757-0394 0683-1035	4 1 2	RESISTOR 3.3K 5% .25W FC TC=-400/+700 HESISTOR 10M 5% .25W FC TC=-900/+1100 RESISTOR 1M 5% .25W FC TC=-800/+900 HESISTOR 51.1% .12% F TC=6+=100 FESISTOR 10% 5% .25W FC TC=+400/+700	0160G 0160G 0329B 0160G	C81065 CH1055 C4-1/8-T0-51k1-f CK1035
A12R36 A12R37 A12R38 A12R39 A12R40	0683-0275 0683-4705 0663-4705 0757-0594 0683-1035	2	PESISTOR 2,7 5% ,25W FC TC==400/+500  PESISTOR 47 5% ,25W FC TC==400/+500  RESISTOR 47 5% ,25W FC TC==400/+500  RESISTOR 51,1 1% ,125W F TC=0+=100  RESISTOR 10% 5% ,25W FC TC==400/+700	0160G 0160G 0160G 0329B 0160G	CR2765 CR4705 CR4705 C4-1/8-T0-51R1-F CB1035
A12R41 A12R42 A12R43 A12R44 A12R45	0683-0275 0757-1090 0757-1090 0683-3335 0683-3335	5	RESISTOR 2,7 5% ,25% FC TC==400/+500 RESISTOR 261 1% ,5% F TC=0+=100 RESISTOR 261 1% ,5% F TC=0+=100 RESISTOR 33% 5% ,25% FC TC==400/+800 RESISTOR 33K 5% ,25% FC TC==400/+800	0160G 0299E 0160G 0160G	CB2765 MF7C1/2=10=261R=F MF7C1/2=10=261R=F CB3335 CB3335
A12R46 A12R47 A12R48 A12R49 A12R50	0683-3335 0683-3335 0683-3335 0683-3335 0683-3335	l la	#ESISTUM 33K 5% 25W FC TC#=400/+800 RESISTOM 33K 5% 25W FC TC#=400/+800 #ESISTOM 33K 5% 25W FC TC#=400/+800 #ESISTOM 33K 5% 25W FC TC#=400/+800 RESISTOM 33K 5% 25W FC TC#=400/+800		
A12U1 A12U2	1826-0326 1826-0089	1	IC OP AMP IC 2525 OP AMP	07933 03791	RC4558DN HA2-2525-5
	į				

Page 6-17, Table 6-3, Replaceable Parts
Change part numbers for A23 U15 and U16 to 1818-0423 and 1818-0424, respectively.

Page 8-63, Figure 8-47, A23 Component Locations, Change Figure 8-47 to Figure G.

Page 8-63, Figure 8-48, A23 schematic diagram, Partially change Figure 8-48 as shown in Figure H.

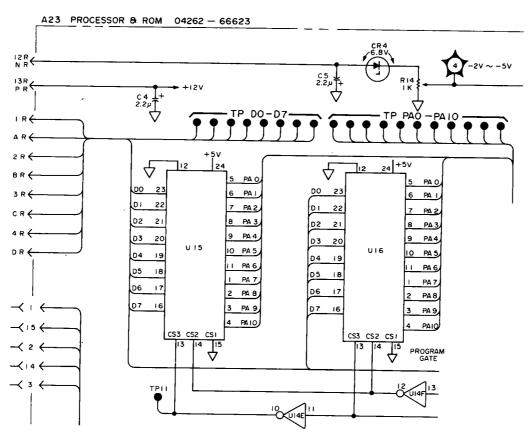
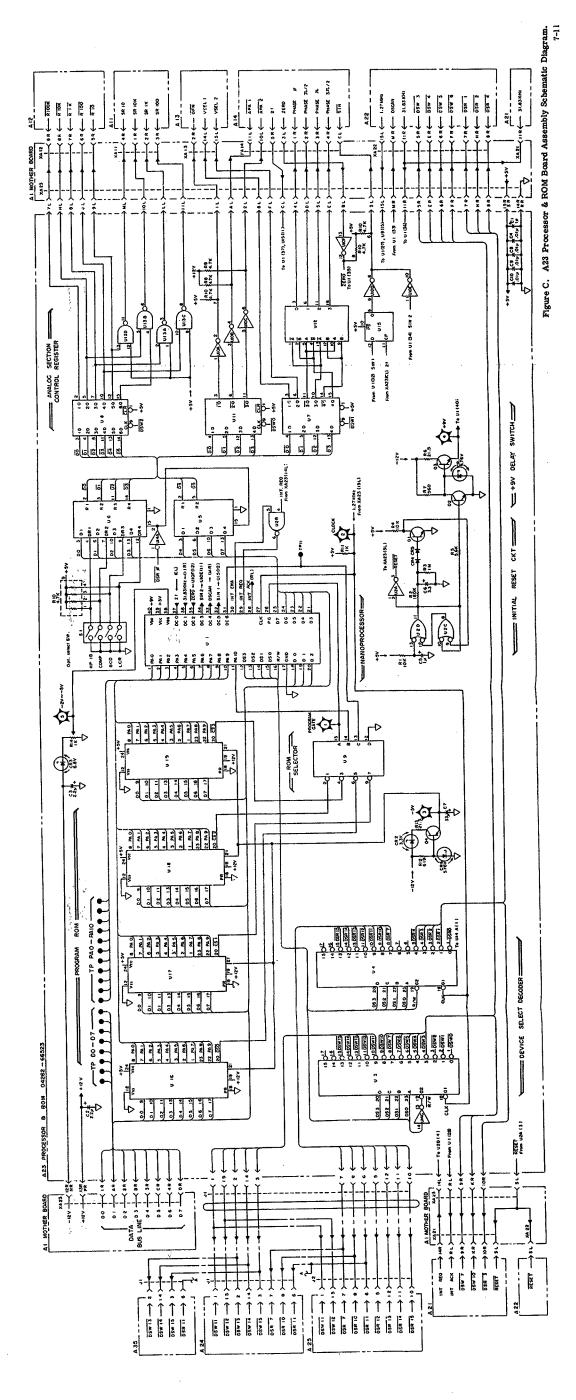


Figure H.



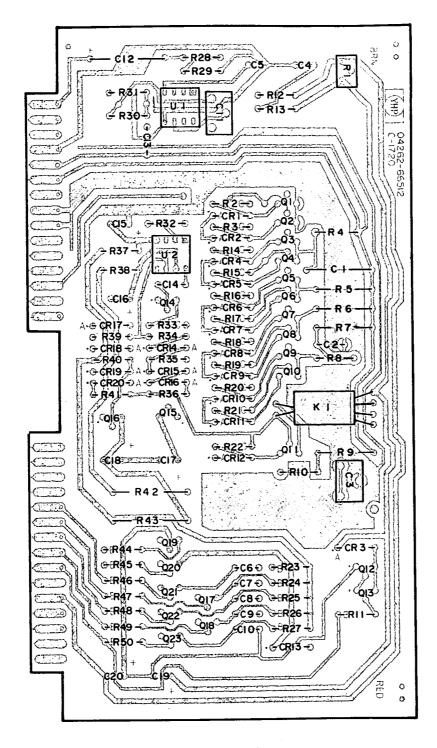
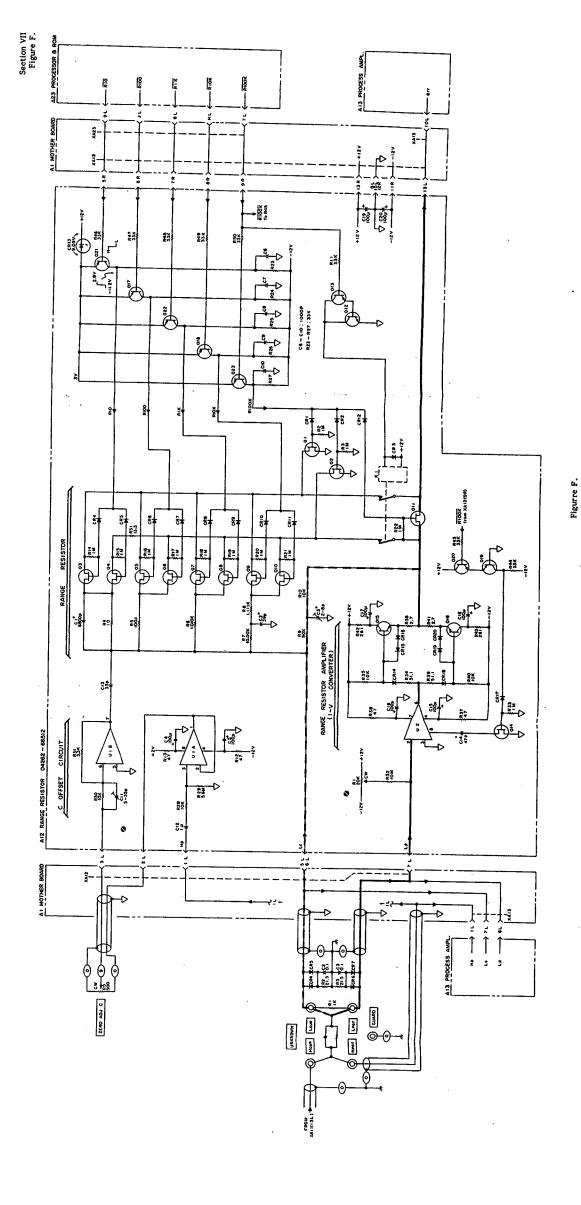


Figure E.



Model 4262A

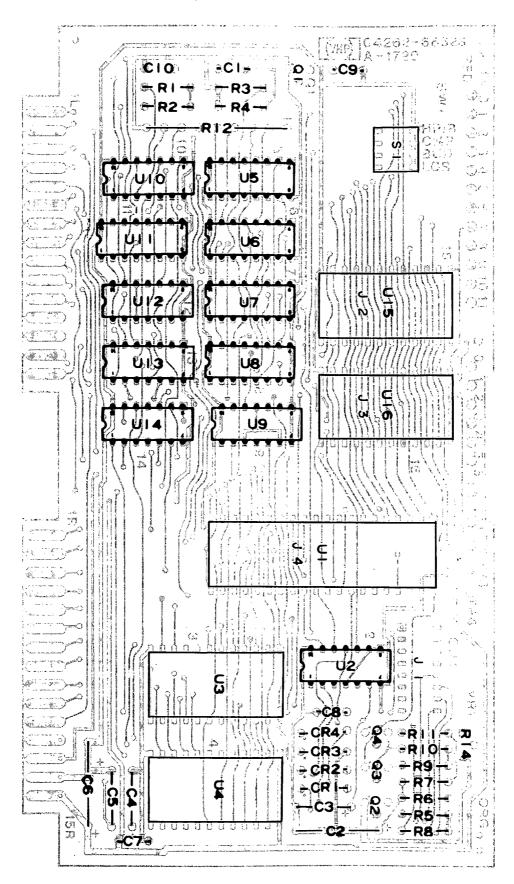


Figure G.

## SERVICE

#### 8-1. INTRODUCTION.

8-2. This manual section provides the information and instructions required for servicing the HP Model 4262A LCR Meter. Included are Theory of Operation and Troubleshooting Guide with Circuit Schematics. The Theory of Operation describes fundamental principles and circuit operating theory of the 4262A with block diagrams. Circuit schematics, locator illustrations, troubleshooting guide, circuit analysis and other technical data necessary for repairs are integrated into the service sheet foldouts. An illustration of the instrument interior is shown in Figure 8-21.

#### Note

When the instrument circuitry includes expanded capabilities provided by optional equipment, refer to paragraphs entitled OPTIONS for specific option service information.

#### WARNING

TROUBLESHOOTING AND REPAIR ARE ALLOWED FOR QUALIFIED TECHNICAL PERSONNEL ONLY. IF YOUR INSTRUMENT FAILS, REFER INSTRUMENT TO SERVICE PERSONNEL. H-P SERVICE OFFICES OFFER YOU THE BEST ANSWER TO YOUR PROBLEM. A GUIDE TO YOUR LOCAL H-P SERVICE OFFICES MAY BE FOUND ON THE BACK COVER OF THIS MANUAL.

#### 8-3. THEORY OF OPERATION.

8-4. This theory of operation has been organized into three sections: basic theory, a block diagram discussion, and circuit analysis. The basic theory, beginning with paragraph 8-11, explains the concepts and fundamental theory of the 4262A instrument technique adapted for accurately measuring the DUT and for fully achieving automated measurement performance. The block diagram discussion describes the overall circuit operating theory of the 4262A with block-to-block signal flow. Included are block and timing diagrams. The

circuit analysis provides a detailed description of how the circuit on each board functions. For reference convenience, when servicing the instrument, a circuit description is included in the service sheets.

### 8-5. TROUBLESHOOTING.

8-6. This troubleshooting guide provides instructions and information for locating a faulty circuit instrument component that requires service. instructions consider the safety of service personnel who will perform the procedures. These diagnostic guides are in the form of step-by-step procedures with flow diagrams. The board level troubleshooting diagrams are the procedures for isolating the problem to an individual malfunctioning circuit board assembly. The guides for locating a defective component are given on the individual board service sheets and integrate service test point locations, waveform support data: illustrations, voltage data, timing digrams, and other technical information in addition to providing schematic diagrams for each board. To facilitate easy troubleshooting of the 4262A digital section, the troubleshooting guide for the logic circuit employs a signature analysis technique incorporating the concept of data stream analysis. A guideline to signature analysis is provided in Figure 8-12.

## 8-7. RECOMMENDED TEST EQUIPMENT.

8-8. The test equipment required to perform operations outlined in this section is listed in Table 1-4 (Section I). The table includes: type of instrument required, critical specifications, use, and recommended model. If the recommended model is not available, equipment which meets or exceeds critical specifications listed may be substituted.

## 8-9. REPAIR.

8-10. Repair explanations tell how to replace defective circuit components. The recommended replacement procedures for components and parts which require special repair, replacement tools, or test equipment should be observed. Correct disassembly and the exchange procedures for such special parts are outlined in Paragraphs 8-46 through 8-52. To prevent damage from improper repair procedure, refer to the appropriate manual section before proceeding with repair.

#### 8-11. BASIC THEORY.

8-12. Figure 8-1 is the basic block diagram of the 4262A showing mainly the analog measurement section. It illustrates how the 4262A measures inductance L, capacitance C, resistance R and/or dissipation factor D. In this figure, the dotted lines denote the directions of control signals to and from the nanoprocessor centered control circuit. A measuring test signal from the oscillator is applied (at level E1) through the source resistor to both the unknown device and the range resistor Rr. Amplifier Rr causes the same current that flows through the unknown device to flow through Rr and operates as a current to voltage converter. The effect of the Rr amplifier is to produce a voltage (E2) equal in phase to and exactly proportional to the current that flows through the unknown device. This amplifier drives the junction of the unknown device and Rr to zero volts (virtual thus Rr does not affect the unknown device current. The voltage E2 represents the vector current which flows through unknown device at test signal level E1. E1 and E2 completely define the electrical characteristics of the DUT (Device Under Test) at a given test level and frequency. The details of how the measured values are derived from the ratio of E1 and E2 are discussed in Paragraph 8-16.

8-13. Voltages E1 and E2, across the unknown device and Rr, respectively, are connected to selector switches S1 and S2. These switches have two

important functions: first, S1 selects either E1 or E2 as the voltage to drive the four phase generator [this also establishes the measurement mode-either series or parallel which is automatically or manually set (PARA or SER - as selected at the front panel)] and, secondly, S2 selects either E1 or E2 as the measurement voltage to charge or discharge the integrator (as appropriate to the measurement function and mode - i. e. Cp, Cs, Lp, Ls, Rp or Rs) in the Vector Voltage-Ratio Measurement Section.

The Vector Voltage-Ratio Measurement Section calculates the measured value for L, C, R or D by ascertaining the voltage ratio between E1 and E2 through a dual-slope (type) analog to digital conversion technique. (This technique is popularly used in digital voltmeters). The section also processes the E1 and E2 signal flow to make the desired measurement. Selection of either an L, C, R or D measurement and an appropriate equivalent measuring circuit is established by setting detector phase reference and by S1 and S2 switch operation timing. The analog section receives its measurement instructions from the digital section. A detailed operating description of the Vector Voltage-Ratio Measurement Section is given in Paragraph 8-15.

8-14. Appropriate values for the source and range resistors, Ro and Rr, are selected with respect to the impedance of unknown device. In a series equivalent circuit measurement (Ls, Cs or Rs), the

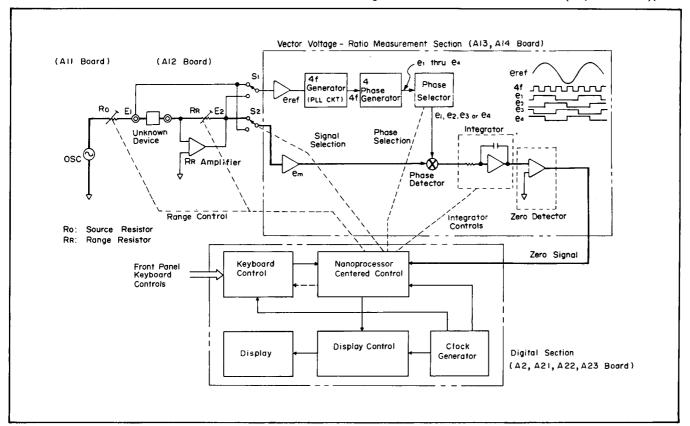


Figure 8-1. Basic Block Diagram.

impedance of the unknown is usually low and Ro is set to a value much greater than the impedance of the unknown device to achieve a constant current drive. On the other hand, for a parallel equivalent circuit measurement (Lp, Cp or Rp), the impedance of the unknown device is usually high so Ro is set to a much smaller value than the impedance of the unknown. Thus, a constant voltage drive is realized. The resistance values for Ro and Rr are always equal.

Here is a brief discussion of Vector Voltage-Ratio Measurement Section operation. The em signal selected by S2 (from either E1 or E2) is detected by a phase detector that outputs the rectangular component or in-phase component to an integrator. Phase detector drive signals e1 through **e**4 are produced in the following manner: a 4f signal is generated from an eref signal (at a frequency of f) as selected by switch S1. This creates signals **e**1 through **e**4, each being different by 90 degrees in phase from one another (a 4 phase generator). As a PLL (Phase Lock Loop) circuit is used for generating the reference phase signal to minimize measurement error, the phase of signals E1 through E4 is very accurate. One of these signals. as directed by the digital circuitry, detects the em measurement signal. Phase detector output is a vector component signal representing the capacitive, reactive, or other characteristic of unknown to be measured.

8-16. This paragraph discusses the parallel capacitance Cp measurement principle. To simplify the explanation, the example used here is that of measuring an ideal capacitor. See Figure 8-2, Cp Measurement. During time T1, Switch S2 selects E2 and the integrator is charged by that portion of the E2 sinusoidal waveform which is synchronously phase detected by the **e**2 pulse train. Both S1 and S2

switches select the E1 signal that is fed to discharge the integrator after being phase-detected by the **e**1 signal. Since time period T2, for the integrator to discharge to zero volts, is proportional to the value of Cx, Cx can be directly obtained from the contents of a counter if the values for Rr and T1 are properly and accurately set. A zero detector signals the digital section to establish a counted number corresponding to Cx each time the integrator output crosses the zero level. Other measurements are done similar to the Cp measurement.

8-17. The analog section of the 4262A is controlled by nanoprocessor centered control which manages the various sequences required to perform the desired measurements. Range control, selection of measurement mode, and timing of the A-D conversion processes are governed by the nanoprocessor. The nanoprocessor also acts as a computing device and calculates deviation  $\triangle$ LCR and the quality factor of sample (mathematical operation) as well as counting the L, C, R and D values converted into time periods.

8-18. The functions set by pushing front panel pushbuttons are inputted to the nanoprocessor through the keyboard control. The keyboard switches are assigned individual addresses for discrimination. When a panel control pushbutton is depressed, the keyboard control identifies the address of switch and causes the nanoprocessor to treat the "interruption" of the function it recognizes by the address code. The nanoprocessor gives priority to specific pushbutton functions so as to be able to restrict improper control settings. Keyboard operation is monitored by and in-part managed by nanoprocessor programming. This is partly to assist the operator and partly to prevent misoperation.

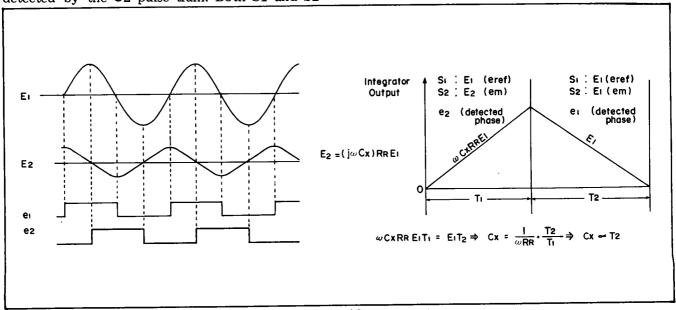


Figure 8-2. Cp Measurement.

## PRINCIPLES OF OPERATION

The following outlines 4262A measurement principles using some equations to aid and acquaint you with the basic concepts of the unit. To simplify explanation in general, only the principles for C-D (capacitance and dissipation factor) measurements are discussed here. The measurement principles for other impedance paramters can be deduced by a similar course of reasoning.

In Cp - D measurements, since a constant test voltage is applied to the unknown, the DUT generally presents a high impedance to the test signal. The following equation shows the relationship beteen voltage E1 at the "H" terminal (voltage across the DUT) and range resistor amplifier output voltage E2 (voltage across range resistor):

$$-E2 = (Gp + j\omega Cp) Rr \cdot E1 \dots eq. 8-1$$

where, Gp is parallel conductance Cp is unknown capacitance Rr is value of range resistor  $\omega$  is angular frequency of test signal

The phase detector separately extracts the real and the imaginary voltage components of E2 (represented by formulas GpRrE1 and j $\omega$ CpRrE1, respectively). Figure A is a vector diagram of phase detector output voltage.

During the charging cycle T1, the phase detector detects the 90 degree phase component of the E2 signal. Thus, the integrator output voltage becomes:

$$k1\omega$$
 CpRrE1T1 . . . . . . eq. 8-2

where, k1 is a constant value determined by 4262A circuitry.

Following the E2 signal, the E1 signal is applied to the phase detector and the discharge cycle begins. The phase detector detects a signal whose magnitude is E1/10 (that is, the E1 signal is attenuated to 1/10 to develop the appropriate time T2 for discharging the integrator) by phase detection of the signal in phase with E1. The resulting change in integrator output voltage developed by the E1/10 signal is:

$$-k1 - \frac{E1}{10}$$
 T2 . . . . . . eq. 8-3

The integrator output eventually reaches zero volts (as a result of the charge and discharge cycle). Thus, the sum of the voltages given in equations 8-2 and 8-3 is zero. And,

$$k\omega CpRrE1T1 = k1 - \frac{E1}{10} T2 \dots eq. 8-4$$

Cp is derived from equation 8-4 as follows:

$$Cp = \frac{T2}{10\omega RrT1} \dots eq. 8-5$$

$$(\omega = 2\pi fm)$$

To eliminate  $\omega$  from equation 8-5, the 4262A establishes a constant charging time T1 as follows:

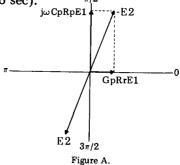
$$T1 = k2 \frac{1}{fm} \dots eq. 8-6$$

where k2 is a constant value (for each test signal frequency).

Equation 8-5 then becomes:

$$Cp = \frac{T2}{20k2\pi Rr} \dots eq. 8-7$$

This is how the measurement frequency is cancelled out of the equation for the measured capacitance value. The discharge period, T2, is measured by counting clock fc whose frequency is constant at 31.83 kHz (its period is  $31.4 \mu \text{sec} = 10\pi \times 10\text{-}6 \text{ sec}$ ).



Thus, if n is the number of counts for fc, T2 can be expressed as follows:

$$T2 = n \cdot 10\pi \times 10^{-6} \text{ (seconds)} \dots eq. 8-8$$

And, if equation 8-8 is substituted in equation 8-7,

$$Cp = n \cdot \frac{10^{-6}}{2k^2Rr} \cdot \dots eq. 8-9$$

(Sheet 1 of 2)

This equation means that discharge period T2 (number of counts for fc) is directly equal to the mantissa of a measured Cp value (note that  $Rr = 10^{m}$ ; and m is an integer).

For example, if a 1200pF capacitor is measured at a measurement frequency of 1kHz, the 4262A automatically selects  $10k\Omega$  as the Rr and constant k2 is 50. Therefore, equation 8-9 may be written as:

$$Cp = n \cdot \frac{10^{-6}}{2k \, 2Rr} = n \cdot \frac{10^{-6}}{2 \times 50 \times 10 \times 10^{3}} = n \cdot 10^{-12}$$

Consequently,

$$n = Cp \times 10^{12} = (1200 \times 10^{-12}) \times 10^{-12} = 1200$$

The 4262A will display 1200 counts and the "pF" unit lamp will light.

In a D measurement cycle, the integrator is charged for period T3 by the E2 signal as detected by a detection phase in phase with E2. Integrator output voltage rises to k1GpRrE1T3. During the discharge cycle T4, the detection phase is different by 90 degrees as referred to E2. The discharge voltage becomes  $k1\omega$ CpRrE1T4. From these integrator voltage changes in the D measurement cycle, the following equation may be composed:

$$k1GpRrE1T3 = k1\omega CpRrE1T4 \dots eq. 8-10$$

Dissipation factor D is derived as follows:

$$D = \frac{Gp}{\omega Cp} = \frac{T4}{T3} \dots eq. 8-11$$

The period T3 is constant and is equal to  $1000 \frac{1}{\text{fc}}$  (fc = 31.83kHz). If n stands for number of 1 counts for fc during period T4, T4 is equal to  $n \cdot \frac{1}{\text{fc}}$  Thus, equation 8-11 may be converted to:

$$D = \frac{T4}{T3} = \frac{n \frac{1}{fc}}{1000 \frac{1}{fc}} = \frac{n}{1000}$$

Therefore, n = 1000D.

If D value for the unknown is 1.2, n will become 1200 which will be displayed at the front panel with the decimal point. Figure 8-3 shows the expanded forms of calculations for impedance parameters.

As shown in Figure 8-3, two kinds of integrator waveforms exist. These two distinctive integrator operations may be examined with respect to Cp and Cs measurement modes. For a Cs - D measurement, a constant current drive is applied to the unknown. Voltage E2 is a constant value drop across Rr and E1 is a variable voltage produced by DUT. The following equation shows the relationship between voltages E1 and E2:

$$-E1 = \left(\frac{Rs}{Rr} + \frac{1}{j\omega CsRr}\right) \cdot E2 \quad \dots \quad eq. \ 8-12$$

The reference phase for the phase detector is now taken from E2 signal. During charging cycle T1, the phase detector detects input voltage E1/10 by a detection phase in phase with E2. The integrator output voltage becomes:

$$k1 \cdot \frac{E2}{10} \cdot T1 \cdot \dots \cdot eq. 8-13$$

The integrator charges to a constant voltage regardless the value of the DUT. During integrator discharge cycle, the phase detector detects E1 signals with a detection signal that is different in phase by 90 degrees with respect to the E2 signal. The resulting integrator output voltage change is:

$$-k1 \cdot \frac{E2}{\omega \, CsRr} \cdot T2 \, \dots \, eq. \, 8\text{-}14$$

Therefore,

$$k1 \frac{E2}{10} T1 = k1 \frac{E2}{\omega CsRr} T2 \dots eq. 8-15$$

Cs is derived from equation 8-15 as follows:

$$Cs = \frac{10}{\omega Rr} \cdot \frac{T2}{T1} \cdot \dots \cdot eq. 8-16$$

Substituting T1 in equation 8-6 produces:

$$C_{S} = \frac{10}{2\pi k_{2}R_{r}}T_{2}...$$
 eq. 8-17

Since T2 is counted by a 31.83kHz (its period is  $10\pi \times 10^{-6}$  sec) clock, equation 8-17 is:

$$C_S = n \frac{100}{2k 2Rr} \times 10^{-6} \dots eq. 8-18$$

where, n is number of clock counts.

If 4262A measurement frequency is 1kHz, Rr is  $1k\Omega$ , and k2 is 5, equation 8-18 becomes:

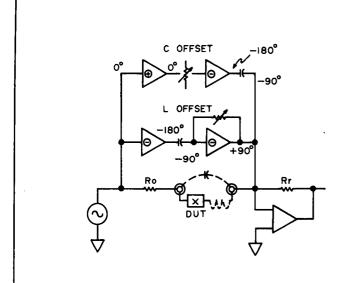
$$C_S = n \frac{100}{2 \times 5 \times 10^3} \times 10^{-6} = 10n \times 10^{-9} (F)$$

When the capacitance of the unknown is  $10\mu F$ , the 4262A displays 10.00 counts and the  $\mu F$  unit lamp lights.

(Sheet 2 of 2)

8-19. Display Control converts the measurement data signals from the nanoprocessor to display component signals which are so coded that corresponding numeric figures are displayed on the 7 segment LED displays. The measurement data is momentarily stored in a memory in this section and sent, in turn, to the matrix drive of each digit of the displays. The alphabetic PASS FAIL, U-CL, and O-F annunciations are illuminated directly on the display by annunciation signals coded by the nanoprocessor. This section also includes a clock generator which employs a crystal resonator to provide the digital section with accurate timing.

8-20. The nanoprocessor centered control and other digital sections are connected to a data bus line (8 bit) on which the measurement data and nanoprocessor I/O signals are transferred. This data bus line serves the overall digital section including the optional sections when the instrument is equipped with HP-IB Compatible (Option 101), BCD Data Output (Option 001), or Comparator (Option 004) option. The timing of the handshakes with system controller (such as a calculator), data transfer, and comparative data are also managed via the data bus line by the nanoprocessor. The operating principles of the option sections are discussed in the paragraphs entitled Options.



The influence of stray capacitance and residual inductance of the test jig can be offset from the current flowing through the range resistor Rr by establishing an opposition current flow through the junction of the unknown device and Rr. The C and L offset circuits develop, respectively, currents which are phase shifted by -90 and +90 degrees as referenced to the oscillator output. The changes in phase are reverse those of the effects of the capacitance and inductance of the test jig. When the offset currents are properly adjusted, the offset currents and the undesired component of the test jig measurement current cancel each other.

Figure 8-4. Offset Control Principle.

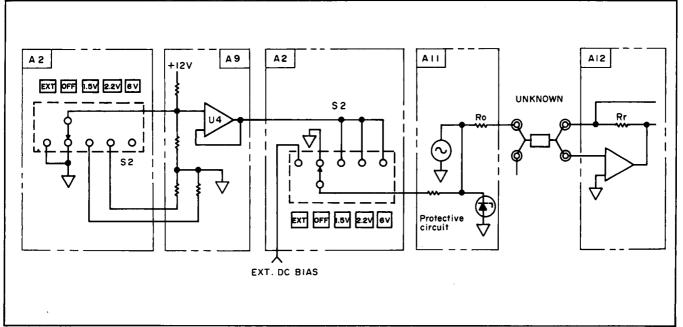


Figure 8-5. DC Bias Circuit.

							Note 2  • When distipation Extor range is 20, the plasse desertor output voltage is sitemated by 1/10 and D value is multiplied by 10.		
_	12.	Derivation of Dissipation Factor	$D = \frac{Gp}{\omega Cp} = \frac{T^4}{T^3}$ $= ^4n \times 10^3$	$D = \omega C_5 R_5 = \frac{T_4}{T_3}$ $= 4_B \times 10^3$	$D = \frac{Rs}{\omega Ls} = \frac{T^4}{T^3}$ $= *n \times 10^3$	$D = \frac{\omega L p}{R p} = \frac{T_4}{T_3}$ $= ^*n \times 10^3$	Note 2  When distipated by the sector out, and D value is 100.001	1000 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 10000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 1000 1 100	Cates
Vector voltage ratio measurement (D)	1000000   Photophysis   Photop	n:number of clock counts Detection phase of phase detector	$\begin{array}{c c} & cop_{R,S^1} \\ & eref = ex \\ & & & & & \\ & & & & & \\ & & & & & \\ & & & & & \\ & & & & & \\ & & & & & \\ & & & & & \\ & & & & & \\ & & & & & \\ & & & & & \\ & & & & & \\ & & & & & \\ & & & & & \\ & & & & & \\ & & & & & \\ & & & & \\ & & & & \\ & & & & \\ & & & & \\ & & & & \\ & & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & \\ & & \\ & & \\ & & \\ & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & \\ & & \\ & \\ & & \\ & \\ & & \\ & \\ & \\ & \\ & \\ & \\ & \\ & \\ & \\ & \\ & \\ & \\ & \\ & \\ & \\$	$\frac{k_{\text{PLE}}}{k_{\text{PLE}}}   \text{Ref.}$ $\frac{k_{\text{PLE}}}{k_{\text{PLE}}}   \text{Ref.}$ $\frac{k_{\text{PLE}}}{k_{\text{PLE}}}   \text{Ref.}$ $0 \qquad \text{a.12}$	eref = 6y	eref = 6x	1	R1, R0   2558   1050   1050   145   1050   1500   150   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   1050   10	100Ha SER 100 1000 1kg 1000 PARA 100
6,1		Integrator output waveform	T T	T1 T2	T.	T R	T. a.	T:	
	Phase Det	Value of k 120Hz 1kHz 10kHz	5 50	0.5 5	5 50 500	0.5 5 50			
	-6y 10 -6x 10 -0-0-0-0-0-0-0-0-0-0-0-0-0-0-0-0-0-0-	Derivation of unknown value	$C_{P} = \frac{T_{2}}{10\omega RrT^{1}}$ $= n \cdot \frac{1}{2kRr} \times 10^{-6}$	$C_8 = \frac{10T_2}{\omega R T T_1}$ = $n \cdot \frac{100}{2 R R r} \times 10^6$	$L_{S} = \frac{R_{L} T_{2}}{10\omega T_{1}}$ $= n \cdot \frac{R_{Z}}{2k} \times 10^{6}$	$L_{\rm p} = \frac{10 R \cdot T^2}{\omega \cdot T_1}$ $= n \cdot \frac{100 R r}{2k} \times 10^6$	Rs = <u>RrTr</u> = n · 100Rr x 10 <sup>-6</sup>	$Rp = \frac{10RrT2}{T1}$ $= n \cdot 100Rr \times 10^{-3}$	
Vector voltage ratio measurement (LCR)	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	k: constant n:number of clock counts.  Detection phase of phase detector	ους (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (1/2) (	61t0 & 62/659k.	ετεί = 6y π/2 0	eref = 6x	(T1 = 1 x 10.2 sec)   Ref = Cy	(T1 = # x 10.3 sec) eref = 6x	
1	**************************************	्रे Relationship between B1 and B2	.E2 = (Gp + j.o.Cp)RtE1	.E1 = ( Rs + 1 / 10/GRr )E2	$\cdot E_1 = (\frac{R_8}{R_T} + \frac{j\omega L_8}{R_T})E_2$	$.\mathbb{E}_2 = (\frac{R_x}{Rp} + \frac{R_t}{j\omega L_p})$	.51 = ( <del>11 + 11 )</del> .52	.E2 = (	
	ia () →		8 8	2 Z	<sup>14</sup> <sup>18</sup> <sup>18</sup> مربر	3 4 4 91 81	Rs jx	Rp Rp	

Figure 8-3. Measurement Principles.

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. Model 4262A

#### 8-21. BLOCK DIAGRAM DISCUSSION.

## 8-22. Analog Section Discussion.

These paragraphs describe how each individual circuit section operates to establish L, C, R and D measurement values as controlled by the digital section. Figure 8-6 is a schematic block diagram of the 4262A analog section. The table in Figure 8-6 shows the range and source resistor values selected by range and function controls.

#### 8-23. A11 Oscillator and Source Resistor.

The test signal is generated by an amplitude stabilized Wien Bridge type oscillator. Oscillator output is fed through an attenuator (A11R18 and R19) to a power amplifier. Attenuator switch A3Q3 turns on only when a Cp measurement is being made and the TEST SIGNAL LOW LEVEL button is pushed. The oscillator signal from the secondary of transformer T2 is designed to have a low output impedance via source resistor Ro to the unknown device (Cx in diagram). Transformer T2 isolates the power amplifier from dc bias voltages which can be applied to unknown device. The A11 Board includes an L Offset Control circuit which provides a compensation circuit to compensate for residual inductance of test leads or fixture. The operating principle of the L Offset Control is diagrammed in Figure 8-4.

8-24. The unknown connection is basically a four terminal (five terminals including GUARD terminal) configuration method. The GUARD terminal is connected directly to the instrument chassis. Circuit common for all PC boards is also eventually connected to the chassis. DC bias voltages up to +40 volts (+6V internally) can be applied to unknown device. The DC bias circuit is illustrated in Figure 8-5.

## 8-25. A12 Range Resistor.

The current that flows through Cx also flows through range resistor Rr. The range resistor amplifier causes the voltage across Rr to represent (exactly) the current flow through Cx. Ro and Rr are selected by a range control signal from the digital section. The table in Figure 8-6 describes how the resistors are controlled. C Offset Control circuit is capable of compensating for stray capacitance up to 10pF (see Figure 8-4 for operating principle).

#### 8-26. A13 Process Amplifier.

The very precise voltage across Cx and Rr are fed to differential amplifiers (A13U1 through U4). C2 and C4 are dc blocking capacitors. This assembly processes these signals to feed the Cref signal (reference phase signal used for phase detection) and the Cm signal (signal measured by the integrator) to the A6 board. The two input signals are selected according to specific measurement rules and are used as Cref and Cm signals. The Cref signal is chosen at the same time that the measurement cir-

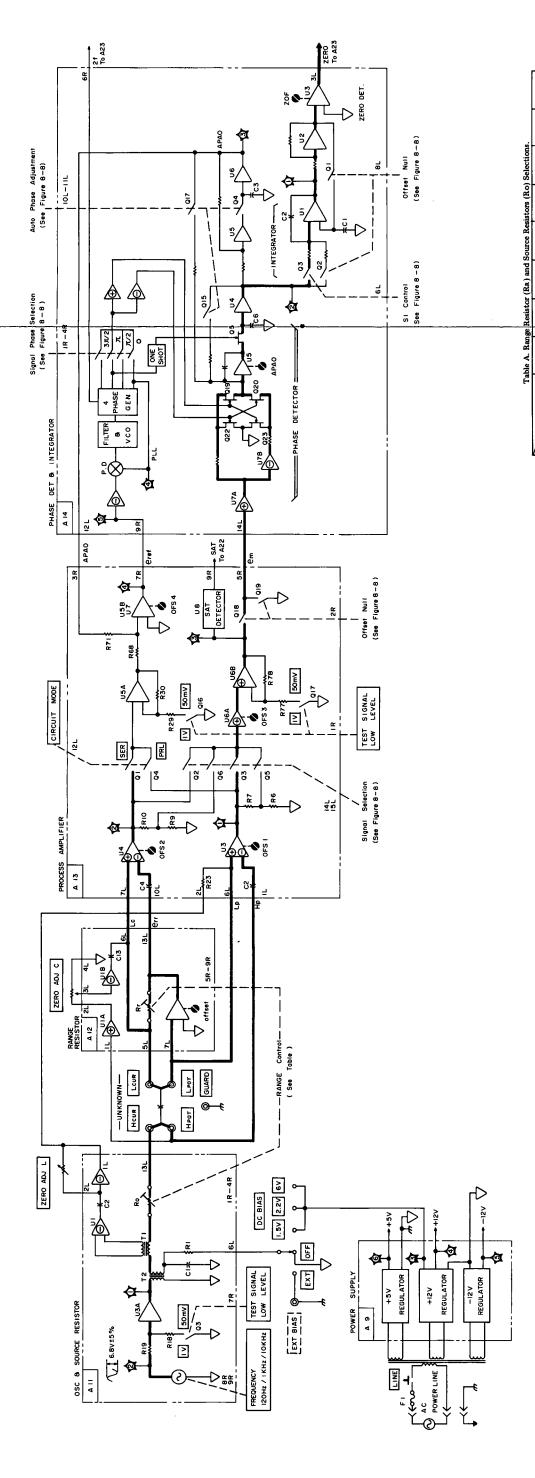
cuit mode is selected. Setting the CIRCUIT MODE to PRL selects the voltage across Cx as the eref signal. When the CIRCUIT MODE is set to SER, the voltage across Rr is selected as the eref signal. In the AUTO measurement mode, the eref signal selection is done automatically and applied in a manner similar to the above. The selected eref signal is amplified by A13U5A and is wave-shaped by A13U5B and U7 which also adjusts the phase angle of eref by a control input (APAO signal) from A14 Board.

The em signal is selected by FET switches A13Q2, Q3, Q5, and Q6 which are, in turn, controlled by signal selection signals from the digital section. The method of selecting the em signal is graphically shown in Figure 8-8 Timing Diagrams. The selected **e**m signal is amplified by A13U6A, U6B and becomes an input signal for the phase detector on A14 Board. The switches A13Q19 and Q18 turn on and off respectively to interrupt the em signal flow during integrator offset control period. When TEST SIGNAL LOW button is pushed and lights (this pushbutton functions in Cp measurement mode only), the gain of amplifiers A13U5A and U6B is increased. Thus, the voltage levels of eref and em signals remain the same as when making a measurement at the nominal (high) test signal level. An SAT detector detects any em signal level that exceeds approximately ±5 volts and transfers such SAT signals to digital section.

## 8-27. A14 Phase Detector and Integrator.

The A14 Board consists of three major circuit sections: PLL Reference Phase Generator, Phase Detector, and Integrator. The specific end functions of the two input signals, **e**ref and **e**m are to establish a ZERO signal whose time interval is equivalent to the desired measurement quantity. This ZERO signal is fed to A23 Board to be manipulated by the nanoprocessor.

The Reference Phase Generator produces four reference phase signals each being different by 90 degrees in phase one from the other (these four signals are phase shifted respectively 0,  $\pi/2$ ,  $\pi$  and  $3\pi/2$  in radius vector as referred to the input signal eref.). The reference phase signals are individually selected in a manner peculiar to the measurement modes (four types). The selected reference phase signal is fed to the Phase Detector to drive switches A14Q19, Q20, Q22 and A23 of the Phase Detector. The method of selecting the reference phase signal is illustrated in Figure 8-8 Timing Diagram. To establish the very accurate 90° phase difference, the Reference Phase Generator employs a Phase Locked Loop (PLL) circuit consisting of a local phase detector (PD), filter, and voltage controlled oscillator (VCO). Thus measurement error is minimized. An explanation of Reference Phase Generator operation is given on Service Sheet 14.



The input signal **e**m to the Phase Detector is a vector voltage representing the impedance of the unknown device. The voltage components of the **e**m signal are detected. These components correspond to the phase angles  $(0, \pi/2, \pi \text{ or } 3\pi/2)$  established by the reference phase signal. Consequently, the phase detector outputs are voltage components which represent the resistive, capacitive, or inductive characteristics of the unknown device. The phase detector output is converted to dc by a smoothing circuit which adopts the period averaging technique to accelerate transient response to the input signals. The special combination of this technique is to speed measurements at the 120Hz test frequency. An explanation of the period averaging technique is given on Service Sheet 14.

The Integrator is charged and discharged by input signals (dc) fed from the Phase Detector. The Zero Detector notes the time that the output of the Integrator crosses the zero level and sends a ZERO signal to the Digital Section (A23 Board). For accurate integrator operation, an integrator Offset Null sequence is excuted before integrator charging is began. Offset Null control details are described on Service Sheet 14.

The phase detector output is provided through A14U5 and U6 to A13 Board as an APAO (Auto Phase Adjustment Output) signal for the period of the auto phase adjustment. In this sequential period, the reference phase signals are adjusted to minimize any phase error which cause a measurement error. The operating principle of the auto phase adjustment is given on Service Sheet 14.

1000kg 10.00Mg 100kn 100.0kΩ 10.00kg 10001 1005 100.00 100.0dl 10.00വ 1000m100 100 1kHz 10kHz RR, RO SER 120HzPARA RR, RO RR, RO Full-scale Full-scale Full-scale

Figure 8-6. Analog Section Block Diagram.

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## 8-28. DIGITAL CONTROL SECTION.

8-29. Paragraphs 8-29 discusses how the 4262A digital section controls the analog section to measure LCR and D values of unknown device and how the built-in nanoprocessor creates unique performance in the 4262A. Figure 8-7 is the basic block diagram of 4262A digital section. All analog section control signals except for Test Signal and Circuit Mode Control Signals are sequentially outputted from A23 Processor & ROM in accord with nanoprocessor programming. The A21 Keyboard Control establishes the measurement function as selected when the front panel control keys are appropriately depressed. The A21 section also stores annunciation data and transfers it to A2 Display and Keyboard to display the annunciation information. A22 Display Control and RAM converts measured data transmitted from A23 into signals appropriate for display on the numeric displays (A2). The A21, A22, and A23 sections are connected to the bidirectional DATA BUS LINE (8 bit).

#### 8-30. A23 PROCESSOR AND ROM.

A23 board consists of Nanoprocessor (A23U1) located in the center of the digital section, Program Control ROM (U15andU16), Data Bus Driver/ Receiver (U5 and U6), Device Select Decoder (U3 and U4), and Analog Section Control Register (U7, U8 and U11). The Nanoprocessor governs the various sequences and timing of the digital section and also sends properly timed measurement control signals to the analog section. For control and data processing, the Nanoprocessor has four major input/output data bus lines: Program Address, Device Select Code, Direct Control Flag, and Data Bus lines. The nanoprocessor programs are filed in the Program Control ROM which has a 4 kilobyte total memory capacity. To extract measurement control instructions from the Program Control ROM, the Nanoprocessor sequentially addresses the ROM through the PROGRAM AD-DRESS BUS line (11 bit). The measurement control instructions outputted from the ROM are momentarily stored in the Analog Section Control Register when the Data Bus Driver/Receiver is set to receiver mode. The analog section control signals which are outputted from the Analog Section Control Register are shown on the block diagram. For accurate timing control of integrator operations, the integrator switch control, ZERO signal, and 2f (= double the test signal frequency) signals are transmitted directly from/to the Nanoprocessor through the Direct Control Flag bus line (bidirectional bus line).

The Nanoprocessor accesses its program data simultaneously by addressing the ROM while the ROM outputs the nanoprocessor program codes. When the ROM outputs an analog section control signal or while measured data is being transferred through the Data Bus line, the Nanoprocessor is not accessing. The Nanoprocessor sequentially excutes program steps in accord with the program data given by the ROM. Various timing in the digital section is controlled by Device Select Code signals (4 bit). These timing control signals are decoded to DSR (Device Select: Read) and DSW (Device Select: Write) signals and manipulate the individual devices, respectively, of the digital section as follows:

DSR: Causes Register or Memory to output data or sets Data Bus Driver/Receiver to driver mode. Nanoprocessor accesses (reads) the data sent from Memory or Data Bus Driver/receiver.

DSW: Enables Register or Memory to store data or sets Data Bus Driver/Receiver to receiver mode. Nanoprocessor sends (writes out) data to Register, Memory or Data Bus Driver/Receiver.

The Device Select Decoder (U3 and U4) each have 15 DSR and DSW output ports.

When 4262A function is selected or changed, the INT. REQ (INTerrupt REQuest) control line goes to high level. This INT. REQ signal requests the Nanoprocessor to pause before proceeding with the nanoprocessor program and to manage the function control prior to program processes. The INT. REQ control line is always active so as to allow for servicing of interrupt requests. The INT. ACK (INTerrupt ACKnowledge) line momentarily goes high to make the vector address line valid. The Nanoprocessor accesses the vector address code (VAO and VA1) to discriminate which control (or controller) originated the interrupt request. When the INT ACK line is at high level. interrupt control data is inputted to the nanoprocessor via A21 Keyboard Control. Successively, the INT ENA (INTerrupt ENAble) output line is set to "disable" status so as not to allow a second interruption before the present interrupt is processed and ends. The INT ENA line is also controlled in the program execute phase (specifically, this output line performs a "handshake" function when the 4262A is used as a component in an HP-IB system).

The Nanoprocessor is synchronized with the 1.27MHz Clock and calculates the measured quantity as a number counted toward the 31.83kHz ( $100k/\pi$ Hz) secondary clock pulse. To identify which, if any, option is installed and being used in the instrument, the Nanoprocessor accesses the option code from the option selection switch setting when the Data Bus Driver/Receiver is set to driver mode by a DSR signal. The Nanoprocessor controls the option section in accord with the nanoprocessor programs as appropriate to the selected option.

## 8-31. A21 KEYBOARD CONTROL.

The A21 Keyboard Control is composed of two major sections: one is the interrupt control consisting of the Interrupt Priority Encoder (U24), Multiplexer (U12 & U23), Row Scan Counter (U2), Gate (U1) and Flip-Flops (U3 & U14); the other is the Annunciator Register (U7, U8, U15 through U21) which stores and transfers manifold annunciation data (keyboard pushbutton indication, range indication, circuit mode indication, etc.).

The Row Scan Counter outputs periodic ROW signals (3 bit) to A2 board as driven by 31.83kHz secondary clock. These ROW signals are decoded to the keyboard scan signals which cause, in turn, specific groups of keys to become valid. Each group of control keys is enabled, in sequence, to perform its function. When a keyboard pushbutton is pressed, the output logic of U1 goes high and subsequently the Row Scan Counter stops. The contents of the ROW Scan Counter and the column number given by CLM Ø through CLM 3 signals are coordinated with the address of the key depressed. Simultaneously, U1 activates Flip-Flops U3 and U14 causing the INT & signal to be outputted. The Interrupt Priority Encoder converts its INT Ø through INT 3 input signals into the vector address signals (4 bit octal code) as appropriate for nanoprocessor input. INT 1, 2, and 3 signals are present only when the 4262A is equipped with option(s). The INT REQ signal is sent to A23 and the INT ACK signal actuates the Multiplexer so that the vector address and keyboard address signals pass through the Multiplexer toward the DATA BUS line.

The Annunciator Register stores manifold annunciation data which are serially transferred from the Nanoprocessor to each register file of IC's U7, U8 and U15 through U21. Specifically, U15 stores test signal annunciation data and, additionally, originates the test signal control signals which direct the Low Level, 120kHz, 1kHz and 10kHz measurement functions. U8 also originates the CMS (Circuit Mode Selection) signal. When the nano-

processor is transferring the annunciation data, the Data Bus Driver/Receiver is set to receiver mode.

#### 8-32. A22 DISPLAY CONTROL & RAM.

A22 section consists of three major circuits: Display control, Extender RAM and Clock generator. The Display control does conversion and storage of measured data to be displayed on the seven segment numeric display. When the Nanoprocessor begins to transfer measured counts (8 bit BCD signal), the Data Bus Driver/Receiver (U19 & U20) is set to receiver mode. L, C or R count data passes through the Data Bus Driver/Receiver and D or Q count data follows. These signals are simultaneously routed to both the Multiplexer (U10 & U18) and the BCD to Seven Segment Decoder (U5). When the measured data is being transferred, the Multiplexer continues selecting BCD to seven segment decoder output signals from its two channel input signals. Other signals, fed directly from the Data Bus Driver/Receiver, are disregarded. Thus, the measured data is translated into segment data which is coded as appropriate for driving the seven segment numeric displays and, is successively stored in the Display Register File (U9 & U17) to accomplish matrix drive of display. The Display Register File outputs the display segment signals which alternately illuminate the numeric figure of each measured count digit of the displays. These display segment signals are amplified to supply sufficient current to the LED displays (cathode driver output signals CAT1 - CAT8). The Scan Decoder U1 outputs periodic anode scan signals which activate, in sequence, the display for each digit. Both the Display Register File and the Scan Decoder are simultaneously driven by Scan Counter U2.

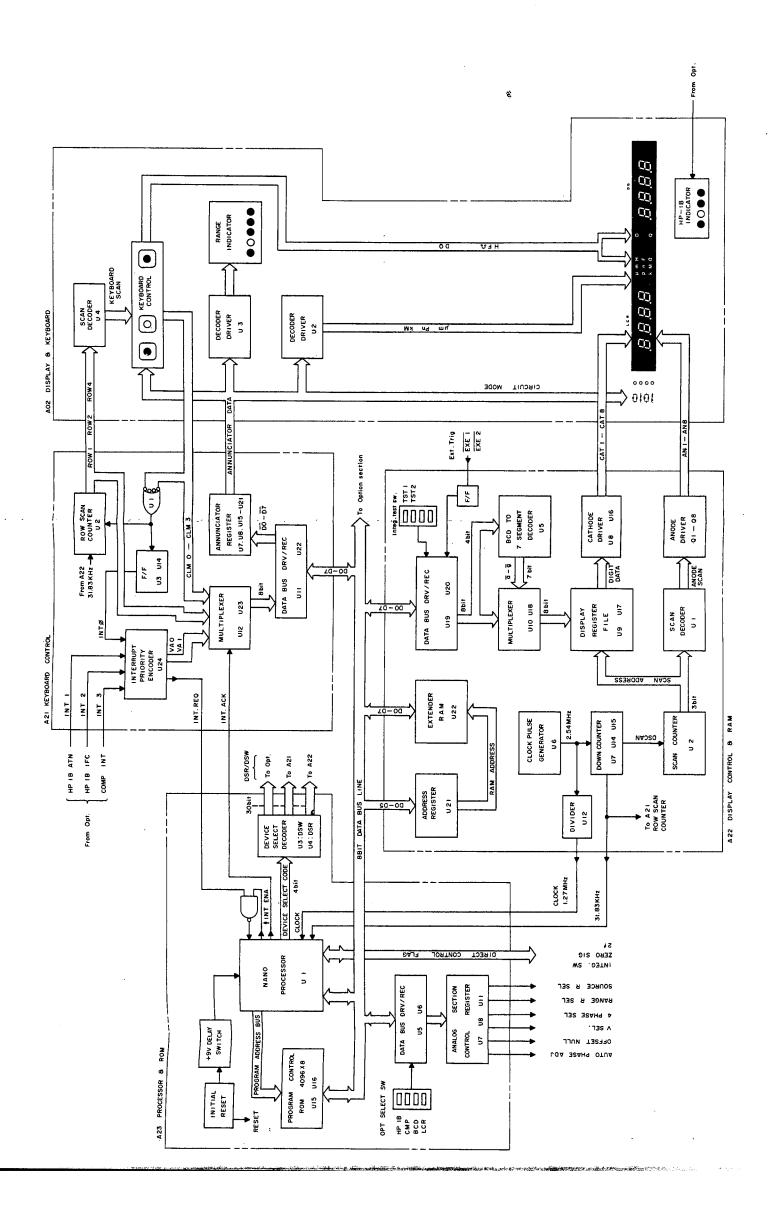
Alphabetic annunciations— PASS, FAIL, O-F and U-CL— are displayed in the following manner: the nanoprocessor encodes annunciation contents so that the annunciation data comprises the display segment signals appropriate for displaying annunciation figures. The annunciation data passes through the Data Bus Driver/Receiver and is inputted to the Multiplexer. In the annunciation execute phase, the Multiplexer selects the annunciation data and disregards the (unnecessary) signals from the BCD to Seven Segment Decoder. The Display Register File stores the annunciation data which coincides directly with the display segment signals. The Data Bus Driver/Receiver can be set to driver mode when the Integrator test switch is set to TST position or the instrument is triggered externally. The Extender RAM (U22) performs supplementary storage of data which is inputted or outputted to/from the Nanoprocessor. The Nanoprocessor sends address signals to the Address Register (U21) before storing data in the Extender RAM. When data is transferred to the RAM, the DSW signal actuates the RAM and the Address Register addresses the RAM to assign individual memories for storing the data. When a DSR signal actuates the RAM, the Nanoprocessor causes the RAM to output stored data. The RAM writes out data as addressed by signals inputted at the RAM ADDRESS signal port.

The clock pulse generator oscillates at 2.54MHz and is frequency stabilized by a crystal resonator. Divider U12 counts down the 2.54MHz basic clock by one half (to 1.27MHz) and provides the nanoprocessor with a stable time base for synchronizing various circuit timing. The Down Counter (U7, U14 and U15) produces the 31.83kHz frequency whose value coincides with the reciprocal number of pi ( $\pi$  = 3.14159...). This particular frequency is significant in derivation of the measured value. The secondary clock signal is fed to the Nanoprocessor for calculating the value of the DUT. Additionally, the Down Counter drives the Scan Counter (U2) which produces display timing signals.

### 8-33. A2 DISPLAY AND KEYBOARD.

A2 section includes the Keyboard Control, Displays, and certain decoders. The Keyboard Control manipulates the Keyboard Scan signals sent from the Scan Decoder (U4) and outputs the resulting CLM (CoLuMn) signals. All annunciator data except for alphabetic annunciations are transmitted from the A21 section. Because the range and multiplier annunciator data has been coded to minimum bit size, the Decoder Drivers U3 and U4 translate them so as to illuminate proper indicators. The Unit and DQ annunciator signals are fed, respectively, via the Function and Loss indicators assembled in the keyboard pushbutton. The numeric displays are independently driven by the A22 section.

8-11



#### 8-34. TIMING DIAGRAM DISCUSSION.

8-35. Figure 8-8 presents a timing diagram for the 4262A. The upper part of the diagram shows output waveforms of the integrator, execute time for each measurement sequence, and main control signals which direct the vector voltage ratio measurement. As may be seen from the diagram, the instrument first measures the L/C or R value and then the dissipation (D) and Q (calculated from D) factors. Approximately three seconds after the LINE switch is depressed to turn the instrument on, power voltage (VGG) is applied to the nanoprocessor through a delay switch (A23 board). The nanoprocessor is simultaneously set to its initial conditions ready for beginning the display test which precedes measurement. When the display test ends, the processor sets the 4262A to a predetermined measurement mode (automatic initial settings) and a capacitance measurement is initiated. When LCR and DQ ranges are set to AUTO, the autoranging recycle repeats until an LCR range suitable for the sample is selected. A front panel range indicator lamp lights and step-shifts to left or right. The displays show blanking signs (- - -) during autoranging period. If the sample is too (in PRL mode) or too small value (in SER mode) compared to the range, the Saturation Detector (A13) send a SAT signal to the nanoprocessor. Range is shifted just after Offset Null operations are completed (instrument does not cycle through steps in remaining measurement sequence). This permits faster ranging. Setting LCR RANGE to MANUAL bypasses autoranging cycle.

When a range is selected in which integrator discharge time interval is within 162 and 1820 clock periods (limits), the measurement sequence proceeds with an L/C/R measurement cycle. To minimize vector voltage ratio measurement error, Offset Null and Auto Phase Adjustment sequences precede integrator charge/discharge (by phase detected DUT signal). During Offset Null period, A13Q19 turns on and Q18 turns off to interrupt the em signal transfer. At this time, any output of the integrator caused by residual phase detector output voltage and integrator output offset voltage is fed back to the input of the integrator to reduce the output of the integrator to zero. And this feedback voltage is stored in a memory capacitor during the measurement to eliminate any measurement error

due to residual phase detector and integrator voltages. Refer to service sheet 14 for offset null control details.

At each integrator operating sequence change, a HOLD TIME is provided to prevent a switching transient waveform from entering the integrator and/or to permit full discharge of the integrator capacitor (from previous integrator operation). Now, an Auto Phase Adjustment consisting of two periods begins. During these periods, to minimize measurement error, the phase detector phase reference is precisely set. APA1 (Auto Phase Adjustment 1) and APA2 control signals administer switches A14Q13, Q14 and Q15 timing to accomplish phase adjustment of **e**ref signal (A14TP1) for establishing exact detection phases of Phase Detector. The Integrator disregards this phase adjustment sequence. Refer to service sheet 14 for auto phase adjustment details.

When an integrator charge period is initiated, the DUT signal (synchronously phase detected) is applied to the integrator input. The Integrator is charged with the incoming signal (dc) for a constant time interval (see table in timing diagram). Two kinds of integrator waveforms are developed depending on measurement function and circuit mode. In the Cp measurement mode, integrator output voltage is increased as its charge is proportional to the DUT current (voltage across Rr) and is decreased as its discharge is proportional to the (constant) voltage across the DUT (constant decay rate). On the other hand, in the Cs measurement mode, the integrator rapidly charges in a short time - the constant voltage across Rr representing the current flowing through the DUT. The integrator discharge depends on the voltage across the DUT (and is proportional to DUT). Detailed integrator operation peculiar to each measurement mode group is described in "Principles of Operation" on Page 8-4. The nanoprocessor counts the time of a 31.83kHz  $(10000/\pi \text{ kHz})$ clock for the time required to discharge the integrator until integrator output voltage reaches the zero level. When integrator output voltage crosses the zero level, a zero detector transfers the ZERO signal to the nanoprocessor. The Nanoprocessor stops counting and stores a number corresponding to the L, C or R value of DUT in its internal registers.

D/Q MEASUREMENT

MEASUREMENT CYCLE

L/C/R MEASUREMENT

the real to the imaginary part of the DUT current (voltage across DUT when circuit mode is SER), the integrator is charged when the detection degrees. In R measurements, the D measurement cycle is omitted. Since the electrical response time this frequency, the sequence execute times are different for measurement frequencies of 120Hz, 1kHz and 10kHz. Note that the execute time for set instrument to appropriate D range. After an Successively, the D measurement cycle begins. The integrator begins to charge - its incoming voltage being proportional to the conductance or resistance of the DUT. Discharge time is proportional to the reactance of the DUT. To calculate the ratio of at "90" for each measurement frequency is different and the charge cycle time is sometimes a function of D autoranging recycle is done or repeats once to offset null sequence for D measurement, the is. detected output the discharge sequence is variable. the oţ phase

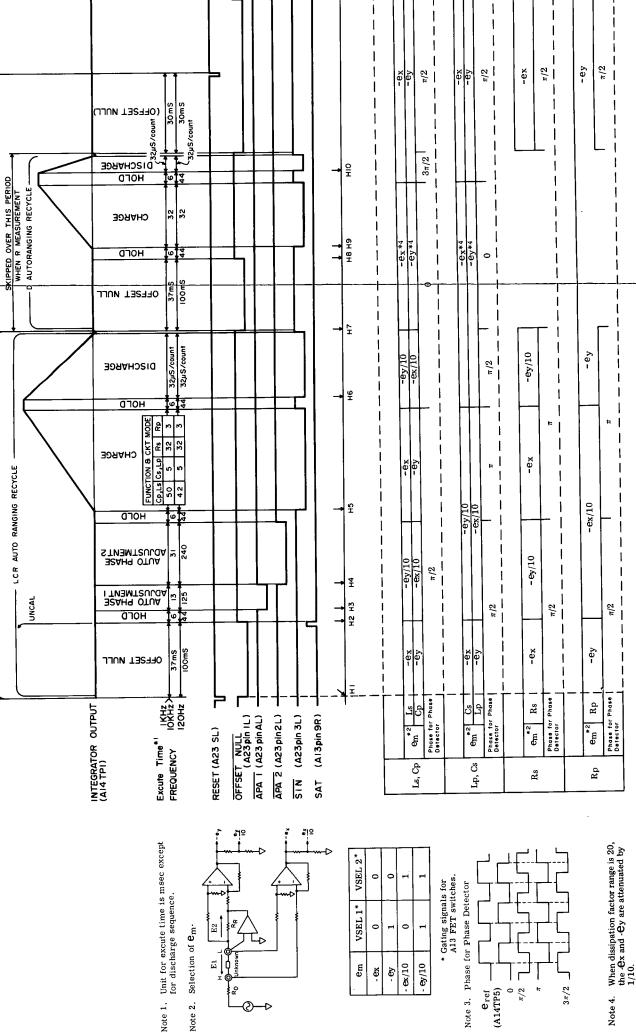
8-36. The table shown in the lower part of the diagram explains how voltage Em is selected by the instrument (either from voltage across Rr or the voltage across the UNKNOWN) and how the detection phase for the phase detection, employed in either PRL and SER circuit modes, is selected. Both upper and lower sections of the waveform timing diagram have the same time scale.  $-\mathbf{e}\mathbf{x}$ ,  $-\mathbf{e}\mathbf{x}/10$ ,  $-\mathbf{e}\mathbf{y}$  and  $-\mathbf{e}\mathbf{y}/10$  in the  $\mathbf{e}\mathbf{m}$  column are names for voltages shown in diagram Note 2. Diagram Note 3 shows the phase relationships of the voltages applied to phase detector FET switches A14Q19, Q20, Q22 and Q23 (detection phase) along with the phase of  $\mathbf{e}\mathbf{r}\mathbf{e}$  signal at A14 TP5. The detection phase is sequentially selected by PHASE control signals  $(\phi \sim 3\pi/2)$  which are transmitted to 4 Phase Selector on A14 board (from A23 Nanoprocessor & ROM board).

# Note

Labels H1 through H10 in the timing diagram denote the timing for trigger used when troubleshooting instrument using A23 service board (service kit 04262-87001). The 4262A measurement sequence can be stopped at or resumed from the desired point from among these triggering points by pushing specific 4262A front panel buttons.

8-13

Figure 8-8. Timing Diagram.



#### **8-37. OPTIONS.**

8-38. The theory of operation for the 4262A optional circuits is outlined in the following paragraphs. The currently available options (001, 004 and 101) with a summary of their functions and the material furnished are listed in Table 8-1.

Figure 8-9 is a block diagram showing the option section when all available optional equipment is installed. The basic instrument and the individual option sections are interconnected by 8 bit data bus lines through which both measured and control data are transferred.

8-39. OPTION 001 BCD DATA OUTPUT (A35). Option 001 BCD OUTPUT CONTROL (A35) consists of a Data Bus Driver/Receiver and two shift Register Files which momentarily store the measured data for simultaneous transfer of the complete data to BCD DATA OUTPUT connectors. Timing control of the A35 circuitry is done by nanoprocessor Device Select signals DSR11, DSW13, DSW14 and DSW15. When 4262A TRIG-GER function is set to EXT, the instrument can be triggerred by an EXE (external encode) signal inputted from either BCD DATA OUTPUT connector J7 or J8 (pin 46). After a measurement cycle ends, a DSR11 pulse signal sets Data Bus Driver/ Receiver U10 to driver mode. As long as the DSR11 signal is valid, the switch setting of the SER/PRL switch (A35S1) has access to the nanoprocessor for assigning the output data format in parallel (simultaneous) or serial (alternate) sequences. The data output timing for both simultaneous and alternate sequencing is diagrammed on Page 8-70. To simplify the explanation, only the parallel output sequence is discussed here. The measured data is stored in the shift registers in synchronism with DSW13 and DSW14 pulses (each outputted 8 times during the data transfer cycle). The Data Bus Driver/Receiver is set to receiver mode to allow the measured data to pass through the device. First, a DSW13 pulse train causes the shift registers U9, U11, U12 and U13 to store the LCR data which is simultaneously transferred with the pulse train. Successively, a DSW14 pulse train actuates shift registers U4, U5, U6 and U7 to store the sequentially transferred DQ data. One shot multivibrators U1A/B generate an output pulse train consisting of pulses that are somewhat shorter than the input DSW pulses. This eliminates the possibility of the shift register not storing the input data because of a DSW signal timing error. One transfer data group is stored in the first 1/8 stack of each shift register when triggered by the rising edge of the one shot multivibrator output pulse. Thus, a total of 16DSW pulses complete storage of all data in the shift register file during the data transfer phase. Next, a DSW15 pulse activates the "two times" Flip Flop U2 - one delayed for 1.2 msec after the other. Thus, the Flip Flop generates FLAG pulse which commands the external recorder to print the measured data concurrently presented at the LCR and DQ BCD output connectors. After the FLAG signal is transferred, a periodic DSR11 pulse actuates the Data Bus Driver/ Receiver and frequently sets it to driver mode to monitor the status of the INHIBIT signal outputted by the external recorder. The DSR11 pulse train continues until the nanoprocessor senses a change in the logic of the INHIBIT signal (meaning that printing is complete). In alternate data output format, the data storage and output cycle for LCR precedes that for DQ. Hence, Device Select signals are alternately provided for both an LCR and a DQ output cycle [as shown in Timing Diagram(Page 8-70)].

Table 8-1. Currently Available Options.

OPTION	FUNCTION	MATERIAL
OPT. 001 BCD DATA OUTPUT	Provides measured LCR and DQ data with Polarity, Decimal Point, Unit, and measurement status in BCD code at rear panel connectors.	A35 BCD OUTPUT CONTROL (04262-66535)
OPT. 004 COMPARATOR	Built-in comparator compares measured value with LCR and DQ HIGH and LOW limits. Provides decision data in display and by Relay and TTL output.	A24 COMPARATOR CONTROL (04262-66524) A4 THUMBWHEEL SWITCH (04262-66504) A5 COMPARATOR KEYBOARD (04262-66505)
OPT. 101 HP-IB COMPATIBLE	Provides system interface capabilities in accordance with IEEE-STD-488-1975 recommendations.	A25 HP-IB INTERFACE (04262-66525) A3 HP-IB CONNECTOR (04262-66503)

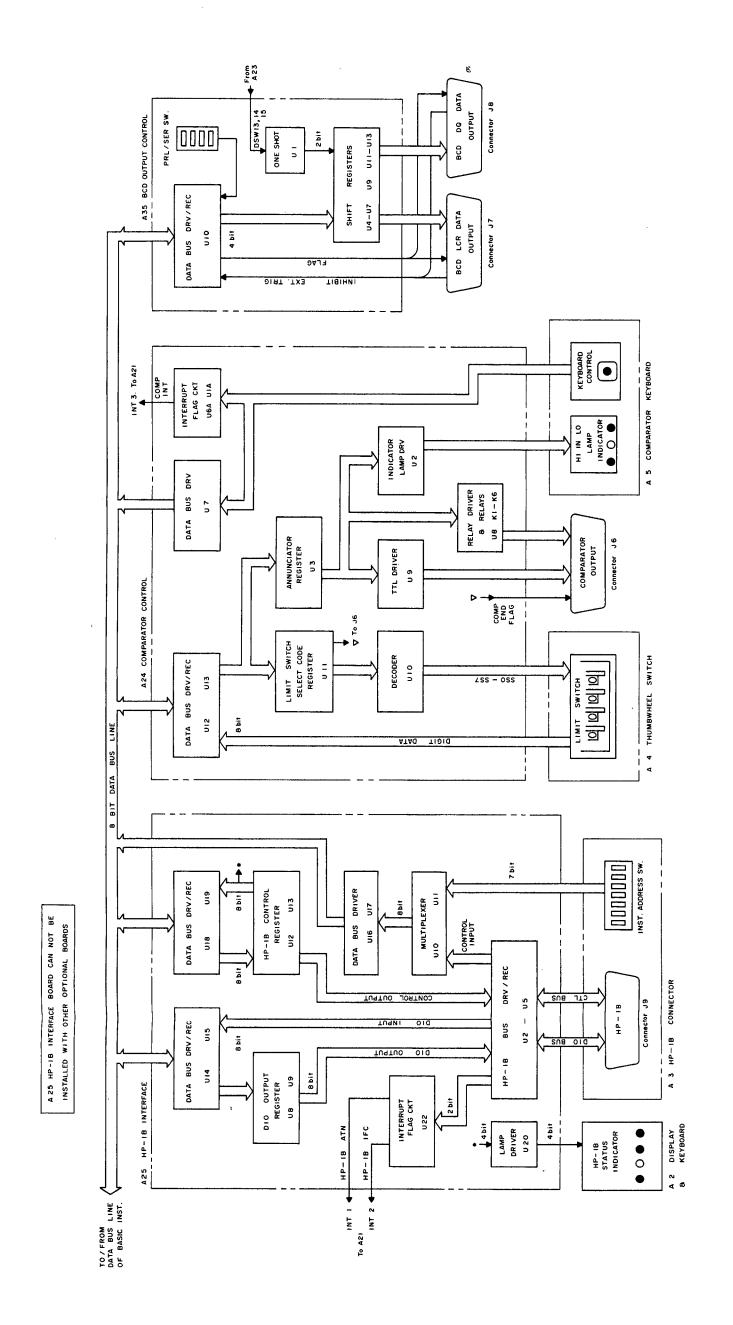
8-40. OPTION 004 COMPARATOR (A4, A5, & A24). Option 004 adds A24 COMPARATOR CON-TROL and the front panel control unit comprised of A4 Thumbwheel Switch and A5 Comparator Keyboard. The A24 Comparator Control manages the control data set into the panel controls as well as the decision data transferred from the nanoprocessor so that comparison results are provided (in three output configurations). The panel control functions are managed in the following manner: An instrument equipped with option 004 includes a front panel control assembly which includes four 4 digit thumbwheel switches used to assign the desired respective limits of L, C or R and D or Q. The thumbwheel switch assembly provides output data for each digit in a 4 bit code which correspondes to the set number indicated in the control panel window. To transmit the high and low limit data from the thumbwheel switch assembly through an 8 bit digit data transmission line, the thumbwheel switches are assigned 8 addresses (each set of four digits occupies two addresses). The nanoprocessor alternately accesses the thumbwheel switch output code in the order of their address numbers. First. Data Bus Driver/Receiver is set to receiver mode and a 4 bit address code is stored in the Limit Switch Select Code Register U11. The Decoder U10 sets its output logic (SS0) to low level in response to the 4 bit address code (output logic of the SS1 through SS7 outputs stay at high level). The SSO signal causes the 8 bit digit data to change depending on the setting of the most significant digit and third digit of the LCR HIGH LIMIT switch that is first addressed. The digit data is transferred to the nanoprocessor (passing through the Data Bus Driver/Receiver set to driver mode).

Successively, the other digit data is transferred in like manner. The commands of all pushbutton controls on the A5 Comparator Keyboard are processed during the interruption phase. The Interrupt Flag circuit directs the nanoprocessor to act on the interrupt request. When a comparator keyboard control pushbutton is pressed, Gate U6A sets its output logic to high level. This causes Flip Flop U1A to generate an INT 3 output pulse. The INT 3 signal is sent to A21 Keyboard Control circuit which forwards the interrupt requests to the nanoprocessor. At this point, the comparator keyboard signals access the nanoprocessor via the Data Bus Driver U7.

The nanoprocessor compares the measured values with the limit numbers (values) and stores the decision data — the results of the comparison in Annunciator Register U3. The decision data is inputted, in parallel, to the TTL, Relay and Indicator Lamp drivers.

8-41. OPTION 101 HP-IB COMPATIBLE (A25). An instrument equipped with Option 101 HP-IB Compatibility includes the A25 HP-IB INTER-FACE board which provides the circuitry to enable intercommunications with external devices in accord with IEEE-STD-488-1975 recommendations. The A25 circuitry is basically composed of data bus driver/receivers and data registers which provide the timely actions for handling the HP-IB data bus input/output and control bus input/output flow as directed by the nanoprocessor. Since the circuit configuration is of general HP-IB design and since general instructions on HP-IB interface is otherwise readily available, a detailed circuit description is not given in this manual.

8-15



#### 8-42. TROUBLESHOOTING.

#### CAUTION

THE OPENING OF COVERS OR THE REMOVAL OF PARTS, EXCEPT THOSE TO WHICH ACCESS CAN BE GAINED BY HAND, IS LIKELY TO EXPOSE LIVE PARTS. IN ADDITION, ACCESSIBLE TERMINALS MAY ALSO BE LIVE.

THE APPARATUS SHALL BE DISCONNECTED FROM ALL VOLTAGE SOURCES BEFORE ANY ADJUSTMENT. PARTS REPLACEMENT, OR MAIN-TENANCE AND REPAIR ARE PERFORMED FOR WHICH THE INSTRUMENT MUST BE OPEN-ED. IF, AFTERWARDS, ANY ADJUSTMENT, MAINTENANCE OR REPAIR OF THE OPENED INSTRUMENT UNDER VOLT-AGE IS REQUIRED, IT SHALL BE CARRIED OUT ONLY BY A PERSON WHO IS SKILLED AWARE OF THE HAZARD IN-VOLVED.

8-43. When 4262A is inoperative or readings for the sample connected to the UNKNOWN terminals are incorrect, you should first check power line voltage used and next the behavior of instrument with respect to the DUT when a measurement is attempted. The two may be incompatible. In addition, check for appropriate test leads or fixture. Determining whether the trouble is in an external device connected to the instrument or is in the actual instrument is primary and a fundamental procedure which must precede troubleshooting the LCR Meter. Occasionally, the unknown sample may have characteristics not measurable by the 4262A. Table 8-2 lists the examples of symptoms likely to mislead. You should also be concerned about the operating environmental conditions in which the instrument is operated. Surrounding magnetic fields or the presence of a strong radiowave will sometimes disturb the measurement. To isolate any instrument trouble from the above possibilities, perform the following examinations:

1) Measure a sample whose characteristics and value (L, C or R and D/Q value) are known to be measurable with the 4262A. Thus, if the problem is restricted to difficulty in measuring a particular sample, it might suggest that the sample is not measurable with the 4262A.

- 2) Next, connect sample directly to the UN-KNOWN terminals without using any test fixture or test leads. Any external equipment being used with 4262A should be disconnected from the connectors of the 4262A. These tests isolate troubles on the external equipment or test jig from those on the instrument.
- 3) Securely ground the instrument to earth. If environmental conditions are suspected, change the location of instrument.
- 4) Use a four terminal connection configuration and measure a sample. An improper connection to unknown will cause a measurement error.
- 5) Properly terminate UNKNOWN terminals (short or open circuit), and press SELF TEST button. Confirm that normal PASS annunciator readouts occur on the LCR DISPLAY.

8-44. Figure 8-10, "How to Use Troubleshooting Guides", is helpful when starting to troubleshoot the 4262A. This flow diagram shows the fundamental procedures which breakdown the trouble possibilities to the component level. The troubleshooting guides are divided into the following major procedures:

## Power Supply Section Isolation Procedure (Fig. 8-17).

Basically used for checking internal dc power supply voltages of the instrument. The guide for checking the power supply section is included in Figure 8-17.

#### Option Section Isolation Procedure (Fig. 8-17).

This procedure, which is used to isolate the option section from the overall unit, is included in Figure 8-17. If the instrument is a standard unit equipped with no option, omit this procedure.

## Analog and Digital Section Isolation Procedure (Fig. 8-17).

The troubleshooting guide in Figure 8-17 describes how to distinguish whether the faulty assembly is located in the analog or in the digital section. In conjunction with the troubleshooting flow diagram of Figure 8-17, the built-in self test function is used to assist in isolating the analog section from the digital section. To study the self test function, refer to Figure 8-11.

## Analog Section Troubleshooting Procedure to Assembly Level (Fig. 8-18).

The troubleshooting flow diagram in Figure 8-18 helps to isolate a faulty board assembly in the analog section. The built-in self test function is also helpful in troubleshooting to the assembly level.

### Component Level Troubleshooting Guides.

Component level troubleshooting guides are provided for each major assembly (other than for A21, A22 and A23 boards of the digital control section) in the service sheets. Procedures for narrowing down the trouble possibilities in A21, A22 and A23 boards to the component level are covered in "Digital Section Troubleshooting Guide". Refer to guideline below.

## Digital Section Troubleshooting Guide.

The search for and location of a faulty component in the digital control section is done in accord with the troubleshooting flow diagrams in Figure 8-19. To facilitate an "easy to make" failure diagnosis, a "signature analysis" method was adopted for troubleshooting both the digital and option sections. When diagnosing with this method, a Signature Analyzer (HP 5004A) is necessary to properly employ the procedures and associated signature maps (see service sheets). Refer to Figure 8-12 for signature analysis guidelines.

8-45. Table 8-3 describes typical front panel symptoms present when 4262A internal controls

(adjustable points) are not well-adjusted. A search for and interpretation of trouble symptoms by operating front panel controls is important and often gives hints as to trouble location. Table 8-4, Front Panel Isolation Procedure provides such an approach to troubleshooting. These primary troubleshooting procedures are supplemental to and should be used with the main procedures in the flow diagrams.

#### WARNING

WHENEVER IT IS LIKELY THAT THE PROTECTION PROVIDED BY THE FUSE HAS BEEN IMPAIRED, THE INSTRUMENT MUST BE SECURED AGAINST ANY UNINTENDED OPERATION.

#### **CAUTION**

CAPACITORS INSIDE THE INSTRUMENT MAY STILL BE CHARGED EVEN THOUGH THE INSTRUMENT HAS BEEN DIS-CONNECTED FROM ALL VOLTAGE SOURCES. BE SURE THAT ONLY FUSES OF THE REQUIRED RATED CURRENT AND THE SPECIFIED TYPE ARE USED FOR REPLACEMENT. THE USE OF REPAIRED FUSES AND THE SHORT-CIRCUITING OF HOLDERS MUST BE FUSE AVOIDED.

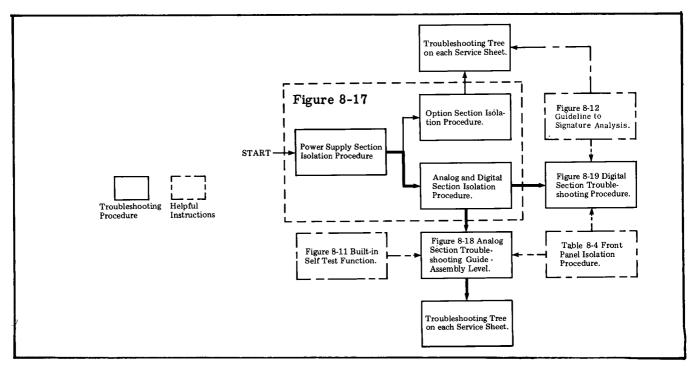


Figure 8-10. How To Use Troubleshooting Guides.

Table 8-2. Symptoms Likely to Mislead.

Category	Symptoms	Probable cause
	When LCR RANGE setting is in AUTO, the range is shifted alternately up and down between two ranges and does not settle on a specific range.	This symptom occurs when the inductance of an inductor with core changes because of the current flowing through the coil.
L MEASUREMENT	Measured values differ depending on the range selected.	Permeability of inductor core changes with measurement signal level (current), which differs for each range. (Measure in MANUAL ranging mode.) See Note below.
	Measured values differ depending on the selected test signal frequencies. Specifically, a large difference exists be- tween the measured value at 120Hz and that at another test signal frequency.	This symptom is because of a difference in the permeability of the inductor core developed by two different measurement frequencies.
C MEASUREMENT	When measuring a small capacitance at 120Hz test signal frequency, measured counts on the LCR DISPLAY fluctuates by several counts.	Interference of ac frequency hum noise. Check for any ac line cables close to the test leads. Check for grounding of the instrument chassis.
R MEASUREMENT	Both LCR and D/Q DISPLAYS are blank () with respect to the sample connected to the UNKNOWN terminals.	The DUT is a wirewound resistor having a large inductance. (Note that some standard resistors are used only with dc current and their calibrated values are so certified.)
Common to all LCR MEASUREMENTS	When measuring an inductance, capacitance or resistance of a large value, a measurement error over the specified limits occurs.	C OFFSET control (related to inductance and resistance measurements) or L OFFSET control (related to capacitance measurement) is misadjusted.

Note: For example, if value of sample is  $187.0\mu H$  on the  $100\mu H$  range, the auto ranging function moves to  $1000\mu H$  range. Then the sample may develop a lower inductance at the applied measurement signal on the  $1000\mu H$  range. It may, for example, develop an inductance of  $160.0\mu H$  that is suitable for measurement on  $100\mu H$  range. The range will again be reset to the  $100\mu H$  range and, as a result will repeat (auto range) up and down between the lower and the higher ranges.

Table 8-3. Front Panel Symptoms of Internal Control Misadjustment.

Adjustment	Symptom
A12R1	When TEST SIGNAL setting is LOW LEVEL, autoranging operation sometimes does not work well.
A12C3	Measurement accuracy of $10 \mathrm{kHz}$ measurements is lower on the highest L and R measurement ranges or the lowest C measurement range.
A12C11	C ZERO ADJ control range is improper.
A13C1	The 10kHz measurement error is excessive.
A13R1 (OFS-1)	When making a measurement in the series equivalent mode, the measurement accuracy is sometimes lower (due to improper dc level at A13TP3).
A13R2 (OFS-2)	When making a measurement in the parallel equivalent mode, the measurement error is sometimes excessive (due to improper dc level at A13TP3) — especially when TEST SIGNAL is set to LOW LEVEL.
A13R66 (OFS-3)	Measurement accuracy will become lower when offset voltage at A13U6 pin 7 is not zero volts. This is usually more noticeable when TEST SIGNAL is set to LOW LEVEL.
A13R67 (OFS-4)	D measurement error sometimes exceeds specifications (impossible to automatically adjust the detection phase of phase detector). This symptom is present when auto phase adjustment signal at A14TP3 exceeds 0 ±3 volts.
A14R1 (ZOF)	Measurement errors for both LCR and D/Q values has increased. The error is maximum at count displays of 1999 for all three measurement functions (Cs, Lp and Rp).
A14R15 (APAO)	D measurement has significant error (detection phase error).
A23R12 (VR1)	Instrument is inoperative or measurement sometimes stops.

Table 8-4. Front Panel Isolation Procedure.

Symptoms	Probable Faulty Board
ZERO ADJ L control malfunctions but measurement is made correctly.	A11
Measured value is incorrect at a particular range setting.	A11, A12
Measurement is not made correctly when TEST SIGNAL setting is at LOW LEVEL.	A11, A13 Note 1
Displayed count is unstable and fluctuates several counts at 120Hz measurement.	A11, A14
ZERO ADJ C control malfunctions but measurement is made correctly.	A12
Autoranging operation skips a particular range.	A12
U-CL is displayed on every range.	A13
Measurement is made only in either PRL or SER mode.	A13
Display count changes randomly.	A14
Figure(s) in numeric display is (are) defective.	A2
An indicator lamp does not light.	A2, A21
Pushbutton controls do not work (always invalid).	A2, A21, A23
An indicator lamp stays lit.	A21
All numeric display are blank.	A22
Trigger lamp does not light or stays lit.	A22, A23
Autorange control is inoperative.	A23

Note 1: If test signal voltage at H<sub>CUR</sub> terminal is correct (140mVp-p), A13 board is faulty. If not, A11 board is faulty.

#### **SELF TEST FUNCTION**

Pressing the SELF TEST button (located at left in line with the CIRCUIT MODE selection buttons) directs the instrument to begin a sequence of instrument operated self-test functions. This is an outline of how to use the self test function for failure diagnosis.

#### Automatic self test settings:

An appropriate equivalent circuit mode (either to SER or PRL) is automatically selected for the duration of the self test. Since self testing is done in a particular equivalent circuit mode for each of the measurement parameters (L, C and R), auto testing is limited to the ranges specified for these circuit modes. The table below shows measurement ranges tested by self-test function. However, since, during self test, all instrument measurement functions are brought into action (including all the range resistors), this test is broad check of overall instrument performance for all ranges.

Table 8-5. Self Test Ranges

Range	Cs —	Ls	Rs -11
1	100pF	100μΗ	$1000 \mathrm{m}\Omega$
2	1000pF	1000μΗ	10Ω
3	10nF	10mH	100Ω
4	100nF	100mH	1000Ω
5	1000nF	1000mH	10kΩ

Note

Multiply range by 10 at 120Hz and by 0.1 at 10kHz test signal frequencies.

#### How the self test function operates:

To perform the self test, the instrument simulates a measurement of either zero or infinite impedance. For these tests, the UKNOWN terminals are appropriately terminated (short or open). Under these test conditions, the integrator develops an output voltage corresponding to a 1000 count display (full scale) for the LCR measurement test cycle and a 000 count display for the DQ measurement test cycle. The nanoprocessor monitors the 1000 and 000 counts calculated from the integrator output. If either or both of the counted numbers differ by more than 5 counts from their respective nominal values, a FAIL annunciation is displayed on the LCR DISPLAY. The nanoprocessor also monitors a SAT signal from Saturation Detector (A13) to further categorize the failures into other subdivisions.

#### Self Test Diagnostic Guide

Table 8-6 "Self Test Displays and Trouble Possibilities" is helpful in troubleshooting the analog section. No pushbuttons except for the FUNCTION and TEST SIGNAL controls should be depressed while the self test is being performed (if a pushbutton is inadvertently pressed, the self test function will be reset and will require reactivating).

Table 8-6. Self Test Displays and Trouble Possibilities.

Display	Source of FAIL signal	Probable Cause of Trouble
FAIL 1	Process Amplifier has been saturated by a signal of excessive amplitude. Saturation Detector is generating SAT signal.	<ol> <li>One of the range resistor selection switches on the A12 board is defective.</li> <li>One of the signal selection switches on the A13 board is defective.</li> <li>Saturation Detector on A13 board is faulty.</li> <li>A13Q17 is always conducting (display will change to PASS when LOW LEVEL button is pressed).</li> </ol>
FAIL 2	Integrator has developed an incorrect output voltage in an LCR measurement cycle.	<ol> <li>Test signal is not present at HCUR terminal. A11 board is faulty.</li> <li>A12 range resistor amplifier is faulty.</li> <li>An amplifier or an active switch on A13 board is faulty.</li> <li>PLL circuit or Phase Selector in A14 board is faulty.</li> <li>Phase Detector or Integrator on A14 board is faulty.</li> <li>Auto Phase Adjustment malfunctioning.</li> <li>Integrator Offset Null control malfunctioning.</li> </ol>
FAIL 3	Integrator has developed an incorrect output voltage in the D/Q measurement cycle.	A23 Processor and ROM board assembly is faulty.

Note: The trouble possibilities outlined in the table above presupposes that the digital control section is operating correctly. A FAIL indication can also be generated by trouble in the digital section.

Figure 8-11. Self Test Function (sheet 2 of 2).

### Digital Section Troubleshooting Using Signature Analyzer.

The advantage of troubleshooting based on "Signature Analysis" is accuracy and ease in finding failures. It is generally difficult to search for an error by means of observing waveforms on an oscilloscope for the reason that bit trains in a digital circuit seem to be much the same whichever is observed. Specifically, to find the errors in stream of a large bit size (or word length) data takes much time and requires the use of an instrument such as a logic state analyzer. Hewlett-Packard has proposed a method called "Signature Analysis" which recognizes the bit pattern measured in a 4 digit hexa-decimal code (signature) for running an easy diagnostic test program. With the Signature Analyzer (HP 5004A), the signatures are displayed in a readable 4 digit-figure set of alphanumeric figures (0 1 2 3 4 5 6 789 ACFHPU). The signature analysis is based the usual signal tracing method followed in troubleshooting an analog circuit. According to signature analysis, devices in a digital circuit are checked with the signal analyzer by comparing signal input and output signatures to and from each device for the "correct" signature denoted in the service manual signature map. If a signature is not identical, the troubleshooter need only trace the bit train in opposite direction to the signal flow and, when a device is noted which generates an erratic signature despite a correct input, the component may be regarded as faulty. One additional important consideration, since the actual program ROM board (P/N: 04262-66523) in the 4262A does not include a self-test program for signature analysis (as part of the program ROM), a troubleshooting board is required when diagnosing with the Signature Analyzer.

When the troubleshooting board is installed in the instrument, a test program is written out from a special ROM which activates overall the digital control circuit, and, if included, any optional circuits. For convenience in troubleshooting the 4262A, this signature test board is supplied as Service Kit (04262-87002).

#### HOW TO USE THE SIGNATURE ANALYZER TEST BOARD.

#### Note

Use either procedure 1 or 2 depending upon instrument serial number.

#### 1. Serial numbers 1710J00340 and below.

- a. Remove A11, A12, A13 and A14 boards from instrument.
- b. Take out A23 Board.
- c. Disconnect A23U16 (ROM) from socket J2 and put aside.
- d. Disconnect signature program ROM from socket J3 (labeled TEST ROM) on test board and install the ROM in place of A23U16.
- e. Reinstall A23 Board in its normal position.

#### Note

When testing ROM's with A23 board assembly, install the ROM in socket J1 (labeled 2708A) on the test board. Install the test board in place of A13 board assembly. Observe signatures at test points D0 through D7 on the board and follow troubleshooting procedures. Test board flat cable need not be connected anywhere.

- f. Turn instrument off and on (press LINE button) to reset digital control circuit and to return test program to its initial address line.
- 2. Serial numbers 1739J00341 and above.
  - a. Remove A11, A12, A13 and A14 boards from instrument.
  - b. Install Signature test board in place of A13 board.
  - c. Take out the A23 board.
  - d. Disconnect A23U15 (ROM) from socket J2 and put aside.
  - e. Connect 24 pin plug of the test board flat cable assembly to socket J2 on A23 board.
  - f. Reinstall A23 board in its normal position.
  - g. Turn instrument off and on (press LINE button) to reset digital control circuit and to return test program to its initial address line.

#### Note

When testing ROM's on A23 board assembly, install the ROM in socket J2 (labeled 2316A) on test board. Observe signatures at test points D0 through D7 on the board and follow troubleshooting procedures. Test board flat cable may be left connected to A23 board.

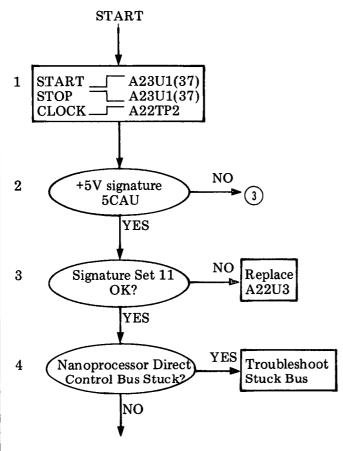
### SIGNATURE ANALYZER TECHNIQUE.

An active digital hand-held logic tracer coupled with an active pod (with four miniature clip connection leads) is sufficient for detecting the test signal and for development of the signature on the Signature Analyzer display. The active probe has access to the desired node in the circuit being tested and transfers this input data to the analyzer. The four input leads of the test cable active pod, connect the gate signals — START, STOP, and CLOCK — from the instrument being tested to the analyzer. The remaining lead is connected to instrument GND. The START signal is an open "window" (measurement gate) signal which causes the signature analyzer to prepare for receiving data via the active probe. The STOP signal causes the window to close. The CLOCK is taken from the time base of the instrument and permits receiving input data and gate signals in synchronization. Polarity of the gate signal active (enable) edges (positive or negative) can be selected by the front panel controls of the signature analyzer. Probing points and connection locations of START, STOP and CLOCK leads are designated on the troubleshooting flow diagrams.

Use an -hp- Model 547A Current Tracer to trace a "stuck" node current.

Figure 8-12. Signature Analysis Guide (sheet 2 of 3).

### Signature Analysis Diagnostic Flow Diagram Notes.



1. Both START and STOP signals are taken from A23U1 pin 37. CLOCK signal is taken from A22TP2. Front panel control settings for Signature Analyzer are:

START button: released (1)
STOP button: depressed (1)
CLOCK button: released (1)

- 2. Checks that signature of +5V supply is 5CAU. If incorrect, go to Flow Diagram number 3.
- 3. Compares actual signatures with signature set (1) on the signature map (see Figure B). If not identical, replace A22U3.
- 4. Check signatures with respect to nanoprocessor direct control bus line. If incorrect, check every component on faulty bus line.

Figure A. Diagnostic Flow Diagram Notes.

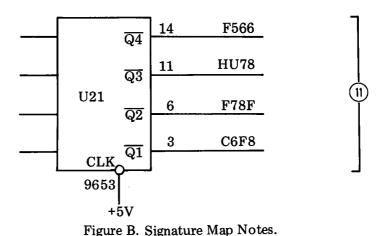


Figure 8-12. Signature Analysis Guide (sheet 3 of 3).

#### 8-46. REPAIR.

### **WARNING**

BEFORE PROCEEDING WITH REPAIR, BE SURE THAT INSTRUMENT IS DISCONNECTED FROM POWER LINE!

## 8-47. REMOVAL OF 02 or 03.

- a. Fully loosen top cover retaining screw located at rear of instrument and lift off top cover.
- b. Remove left handle mounting screws (2). Slide left side panel toward the rear of instrument and take off.
- c. Remove the two transistor retaining screws.
- d. Lift out transistor.
- e. Install new transistor. To maintain good thermal diffusion, use fresh silicone paste on transistor and insulator sheet.

## 8-48. LINE SWITCH (S1) REMOVAL.

- a. Perform steps a and b of paragraph 8-47, removal of Q2 and Q3.
- b. Remove the two screws which fasten LINE switch S1 to plate on side frame.
- c. Remove the cable clamp screw (located at center near top of side frame).
- d. Pull LINE switch toward the rear of instrument and take out switch with extender shaft from instrument.
- e. Pull extender shaft out of switch shaft. Unsolder cable from switch.
- f. Install new switch. Envelop the switch with heat contractible tubing.

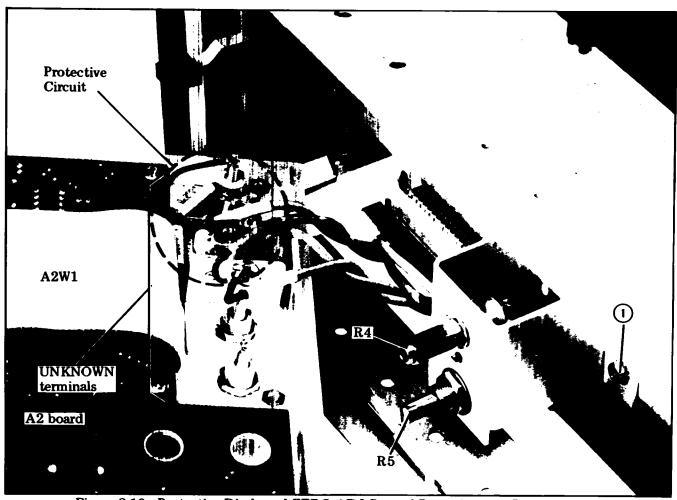


Figure 8-13. Protective Diode and ZERO ADJ Control Potentiometer Replacement.

## 8-49. PROTECTIVE DIODE REPLACEMENT (CR4, CR5, CR6 and CR7).

To replace protective circuit diodes connected to UNKNOWN terminals (Low side), perform the following procedure:

- a. Remove top trim strip from front frame (use a screwdriver to lift out the trim).
- b. Remove the two left hand screws from among the four screws located at the top side of the front frame.
- c. Turn instrument upside down.
- d. Remove the two right-hand screws from among the four screws located at bottom side of the front frame.
- e. Carefully pull unknown terminal binding posts forward and front panel assembly out.

#### **CAUTION**

DO NOT USE EXCESSIVE FORCE OR WIRE CONNECTIONS TO UNKNOWN TERMINALS MAY BREAK.

f. Disconnect flat cable 40 pin connector A2W2 from the plug mated with A21 board assembly. See Figure 8-14.

- g. Disconnect flat cable 40 pin connector A2W1 from the plug mated with mother board. See Figure 8-14.
- h. Unsolder wire leads to diode and disconnect diode from the binding post soldering lugs of UNKNOWN terminals.
- Install new diode. Solder wire leads to new diode.

## 8-50. ZERO ADJ CONTROL POTENTIOMETER (R4 and R5) REPLACEMENT.

- a. Perform steps a through g of paragraph 8-49 Protective Diode Replacement.
- b. Remove retaining screw 1 shown in Figure 8-13.
- c. Remove the potentiometer retaining nut and unsolder wiring leads to the potentiometer.
- d. Install new potentiometer.

## 8-51. A2 KEYBOARD AND DISPLAY BOARD DISASSEMBLY.

- a. Perform steps a through g of paragraph 8-49 Protective Diode Replacement.
- b. Remove the 8 screws (1) through (8) in Figure 8-14) fastening A2 board to front panel.

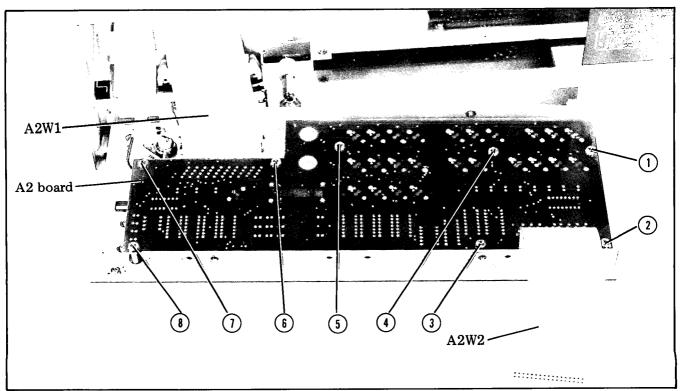


Figure 8-14. A2 Keyboard and Display Board Disassembly.

## 8-52. KEYBOARD SWITCH LED REPLACEMENT.

- a. Perform steps a through g of paragraph 8-49, Protective Diode Replacement.
- b. Remove 8 screws (1) through (1) in Figure 8-14) fastening A2 board to front panel.
- c. Take out A2 board from instrument.
- d. Remove pushbutton switch by melting plastic legs of the switch. Use tool HP P/N 5951-8516.
- e. Unsolder defective LED.
- f. To assure that the newly installed LED will not rub against the switch plunger (when pushbutton is pressed), a soldering guide is required. Fabricate a soldering guide from a piece of 3.18mm (0.125 inch) internal diameter, thin walled plastic tubing 4.76mm (3/16 inch) in length. If tubing is not available, use a 4.76mm strip of paper rolled to make up an approximate I. D. of 3.18mm.
- g. Insert tubing (or rolled paper) into bottom of plunger of new switch (see Figure 8-15).
- h. Insert the new LED into bottom of switch plunger containing tubing.
- i. Rotate LED (in bottom of switch plunger) so that the shortest lead passes through the P. C. board mounting hole (identified with dot marking). See Figure 8-16.

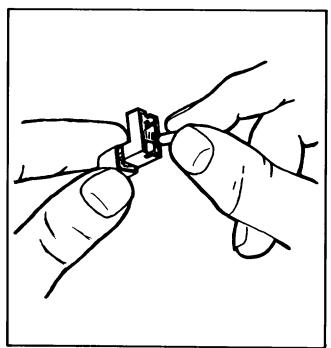


Figure 8-15. Inserting Tubing Into Switch Plunger.

- Install switch and LED combination onto A2 board assembly.
- k. Grasp LED leads (back side of A2 board) and pull LED flush against front side of A2 board.
- l. Solder LED to A2 board assembly.

#### **CAUTION**

WHILE SOLDERING LED, PRESS SWITCH AGAINST FRONT SURFACE OF A2 BOARD ASSEMBLY. BE CAREFUL NOT TO MELT PLASTIC LEGS OF SWITCH OR TO CONTAMINATE IT WITH SOLDERING FLUX.

- m. Take off switch and remove tubing (or rolled paper) from switch plunger. Clean any reresidual flux from A2 board assembly.
- n. Mount switch over LED and operate switch several times to assure that switch plunger does not rub against LED, and that the lightpipe in key-cap does not contact LED before switch plunger bottoms.

#### Note

If the results of step n are not satisfactory, repeat the LED installation procedure.

 Install switch (over new LED) onto A2 board assembly.

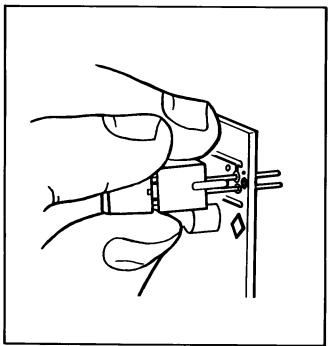


Figure 8-16. LED Installation in Switch.

#### 8-53. PRODUCT SAFETY CHECKS.

### WARNING

WHENEVER IT APPEARS LIKELY THAT SAFETY PROTECTIVE PROVISIONS HAVE BEEN IMPAIRED, THE APPARATUS SHALL BE MADE INOPERATIVE AND BE SECURED AGAINST ANY UNINTENDED OPERATION. THE PROTECTION IS LIKELY TO BE COMPROMISED IF, FOR EXAMPLE;

- -- THE APPARATUS SHOWS VISI-BLE DAMAGE.
- -- THE INSTRUMENT FAILS TO PERFORM THE INTENDED MEAS-UREMENT.
- -- THE UNIT HAS UNDERGONE PRO-LONGED STORAGE UNDER UN-FAVORABLE CONDITIONS.
- -- THE INSTRUMENT HAS SUFFERED SEVERE TRANSPORT STRESS.

8-54. The following five checks are recommended to verify the product safety of the 4262A LCR Meter (these checks may also be done to check for product safety after troubleshooting and repair). When such checks are needed, perform the following:

- Visually inspect interior of instrument for any signs of abnormal, internally generated heat, such as discolored printed circuit boards or components, damaged insulation, or evidence of arcing. Determine and remedy cause of any such condition.
- 2. Using a suitable ohmmeter, check resistance from instrument enclosure to ground pin on power cord plug. The reading must be less than 0.5 ohm. Flex the power cord while making this measurement to determine whether intermittent discontinuities exist.
- 3. Check GUARD terminal on front panel using procedure (2).
- 4. Disconnect instrument from power source. Turn power switch to on. Check resistance from instrument enclosure to line and neutral (tied together). The minimum acceptable resistance is two megohms. Replace any component which fails or causes a failure.
- 5. Check line fuse to verify that a correctly rated fuse is installed.

Section VIII Model 4262A

## TROUBLESHOOTING FLOW DIAGRAMS

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Flow Diagram G.	A21 Board Diagnostic Flow Diagram	8-41
Flow Diagram H.	A21 Board Diagnostic Flow Diagram	8-42

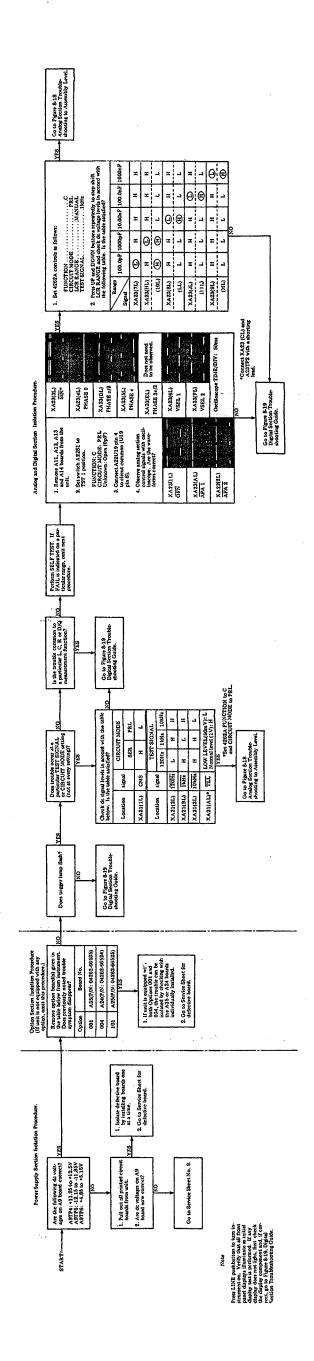


Figure 8-17. Analog and Digital Sections Isolation Procedure.

Model 4262A

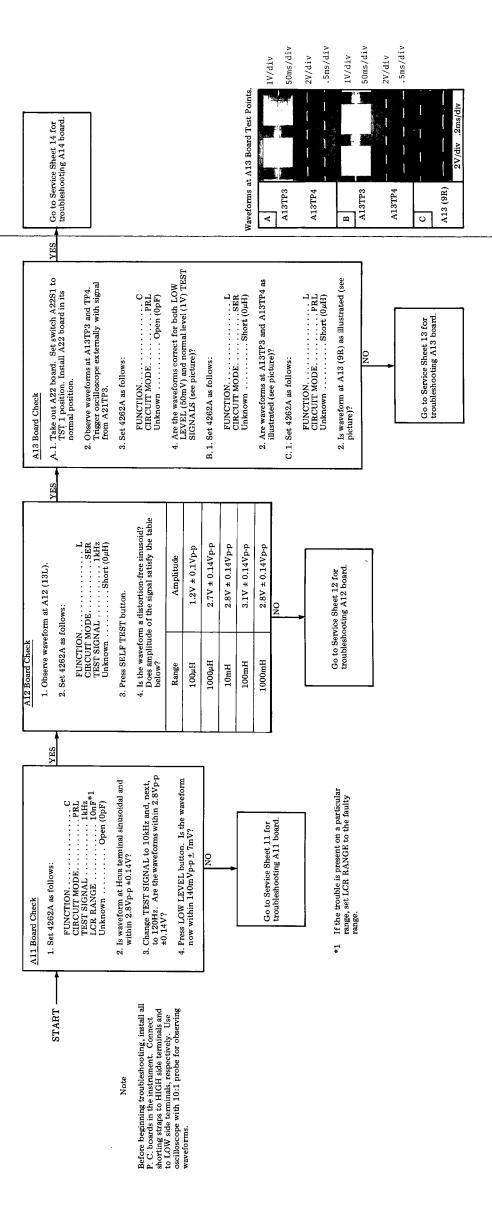


Figure 8-18. Analog Section Troubleshooting Procedure to Assembly Level.

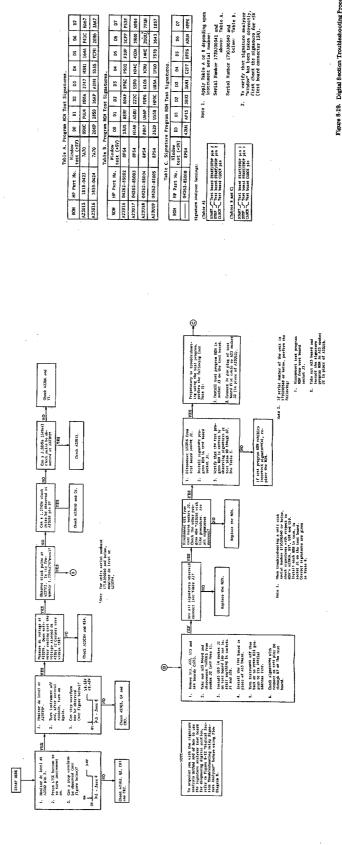


Figure 8-18. Digital Section Troubleshooting Proceduses. Flow Diagram A. Primary Diagnostic Row Diagram. Flow Diagnam B. Program ROM Diagnostic Flow Diagram.

Noto 1. Apply Table A or 8 depending upon instrument serial number: Serial Number 172500591 and Serial Number 173500591 and Serial Number 17300550

| STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STATE | STAT

| Applied | Applied A | Program RGN Text Signatures. | Applied A |

To verify that signature analyzer "window" has been taken correctly first check the signature for +5V (test board connector 13%).

(Tables 8 and C)

\$1MCT\_Test board \$1M(7)\$10p pin 1

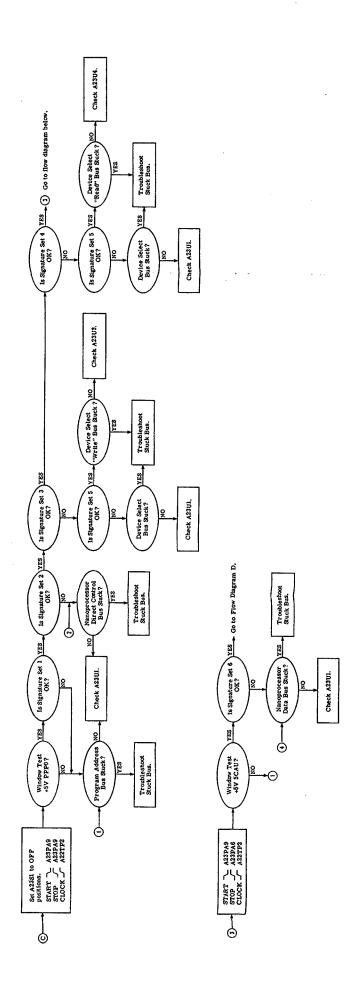
\$10CT\_Test board \$1M(7)\$0p pin 1

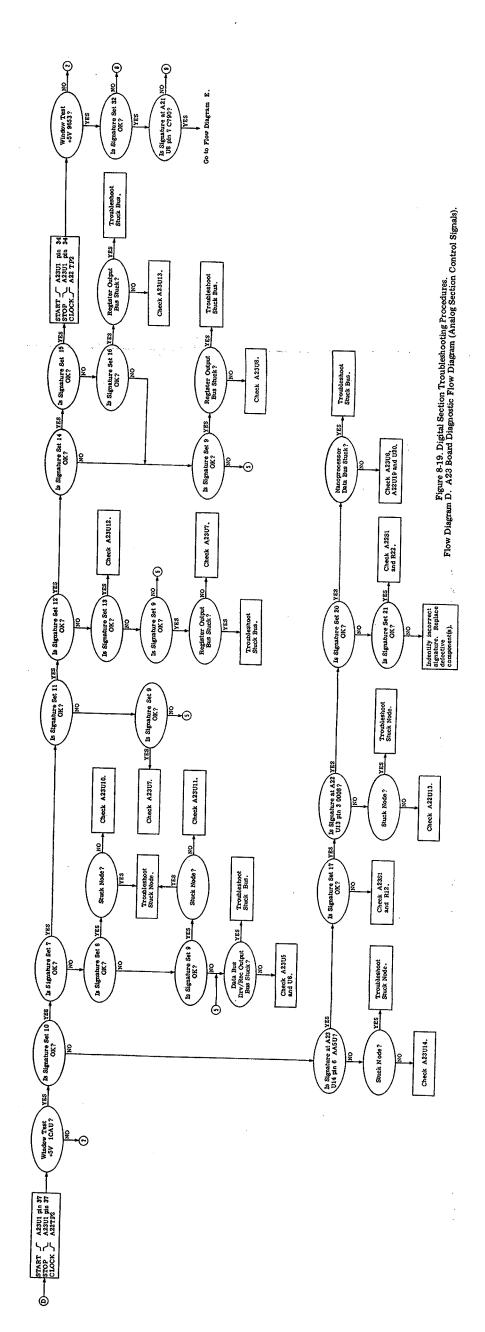
CLOCKT\_Test board \$1M(7)\$0p pin 1

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8-37





Pigital Section Troubleshooting Procedures
Flow Diagram C

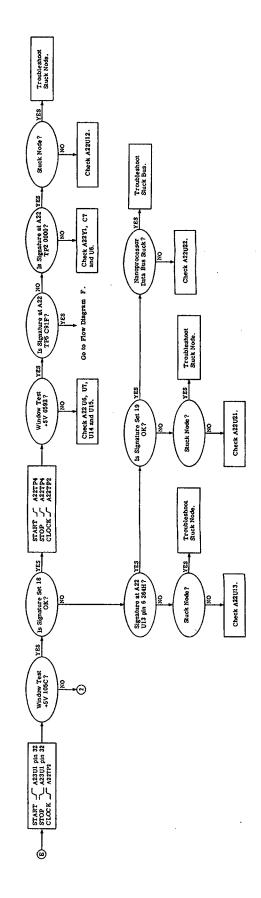


Figure 8-19. Digital Section Troubleshooting Procedure Iow Discram E. A 22 Board Dismostic Flow Discram (Clock and RAM)

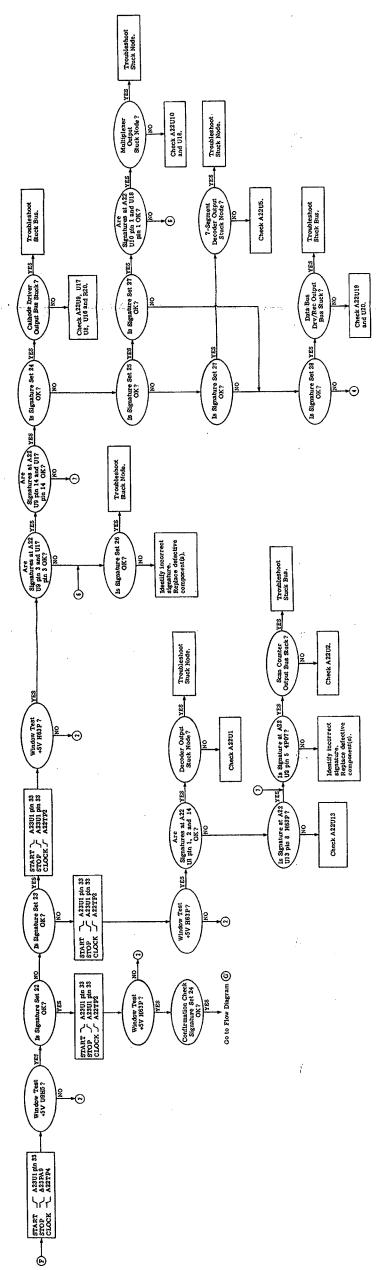


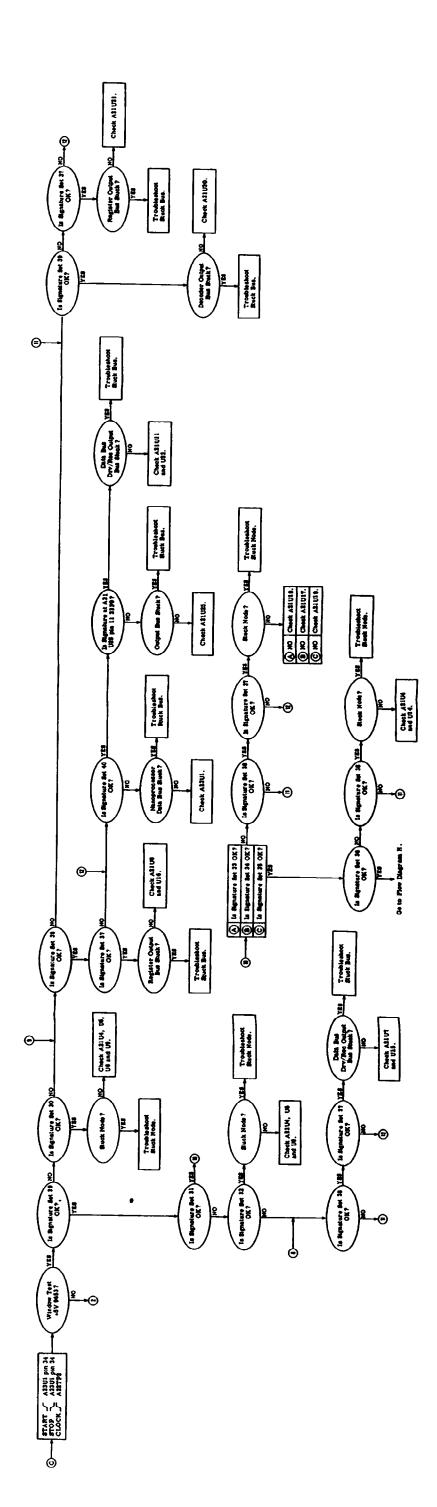
Figure 8-19. Digital Section Troubleshooting Procedures. Flow Diagram F. A22 Board Diagnostic Flow Diagram (Dispaly Control).

Figure 8-19

Pigital Section Troubleshooting Procedures
Flow Diagram E
see INSIDE
8-39

8-40

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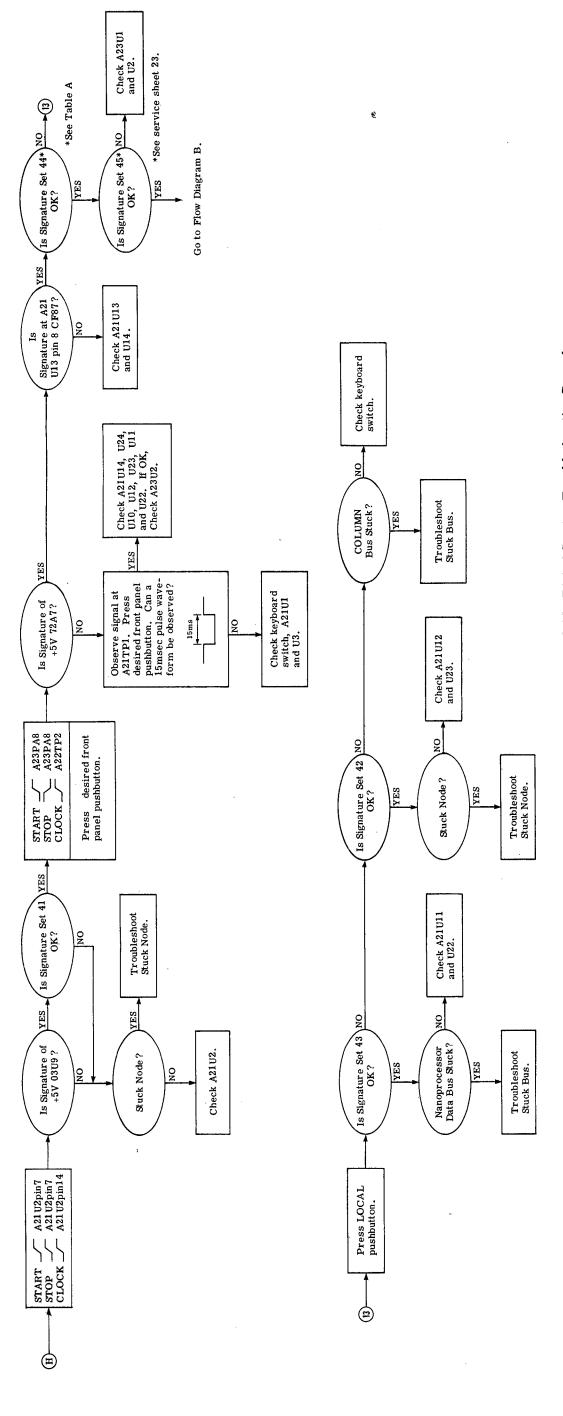


Figure 8-19. Digital Section Troubleshooting Procedures. Flow Diagram H. A21 Board Diagnostic Flow Diagram.

Figure 8-19
Digital Section Troubleshooting Procedures
Flow Diagram G
see inside

Table A. Keyboard Switch Test Signature.

Key*	U22(3)D0	U22(6)D1	U22(10)D2	U22(13)D3	U11(3)D4	U11(6)D5	U11(10)D6	U11(13)D7
LOCAL	บ0บ7	35U8	H64U	4548	5754	209 F	H4H9	2 FH7
SELF TEST	U0U7	35 U8	H64U	4548	9974	PPCF	H4H9	2FH7
CMD AUTO	Մ0Մ7	35U8	H64U	4548	9974	209 F	1AU9	2FH7
CMD PRL	U0U <b>7</b>	UCH8	H64U	4548	5754	209F	H4H9	2FH7
CMD SER	U0U7	UCH8	H64U	4548	9974	PPCF	Н4Н9	2FH7
FUNC L	U0U7	UCH8	H64U	4548	9974	209F	1AU9	2FH7
FUNC C	U0U7	35 U8	186U	4548	5754	209F	H4H9	2FH7
FUNC R	U0U7	35 U8	186U	4548	9974	PPCF	H4H9	2FH7
FUNC A LCR	U0U7	35 U8	186U	4548	9974	209 F	1AU9	2FH7
LCR RNG AUTO	U0U7	UCH8	186U	4548	5754	209F	H4H9	2FH7
LCR RNG MANUAL	U0U7	UCH8	186U	4548	9974	PPCF	H4H9	2 FH7
LCR RNG DOWN	U0U7	UCH8	186U	4548	9974	209 F	1AU9	2FH7
LCR RNG UP	UOU7	35 U8	H64U	8C68	5754	209F	H4H9	2FH7
LOSS D	U0U7	35 U8	H64U	8C68	9974	PPCF	H4H9	2FH7
LOSS Q	U0U7	35U8	H64U	8C68	9974	209F	1 AU9	2FH7
DQ RNG AUTO	U0U7	UCH8	H64U	8C68	5754	209F	H4H9	2FH7
DQ RNG MANUAL	U0U7	UCH8	H64U	8C68	9974	PPCF	H4H9	2FH7
DQ RNG STEP	U0U7	UCH8	H64U	8C68	9974	209 F	1AU9	2FH7
TEST SIG LOW LEVEL	U0U7	35U8	186U	8C68	5754	209 F	Н4Н9	2FH7
TEST SIG 120Hz	U0U7	35U8	186U	8C68	9974	PPCF	H4H9	2 FH7
TEST SIG 1kHz	U0U7	35U8	186U	8C68	9974	209 F	ÎAU9	2FH7
TEST SIG 10kHz	U0U7	UCH8	186U	8C68	5754	209F	Н4Н9	2FH7
TRIG INT	U0U7	UCH8	186U	8C68	9974	PPCF	Н4Н9	2FH7
TRIG EXT	U0U7	UCH8	186U	8C68	9974	209F	1AU9	2FH7
TRIG HOLD/MANUAL	U0U7	35U8	H64U	4548	9974	209 F	Н4Н9	P2U7

## Signature Analyzer Settings:

START A23PA8 \_\_\_\_ STOP A23PA8 \_\_\_\_ CLOCK A22TP2 \_\_\_ Window Test (+5V): 72A7 \* Depressing the keys listed will result in the signatures defined in Table A.

Model 4262A

Section VIII Figures 8-20, 8-21 and 8-22

P/0	Part of.	Encloses front panel designations.	
0	Knob control.	Encloses rear panel	
0	Screwdriver adjustment.		
	Circuit assembly boarderline.		
*	Asterisk denotes a factory selected value, part may be omitted.	e. Value shown is typical	
	Heavy line indicates main signal path.		
	Heavy dashed line indicates main feedback path.	ck path.	
₩ ••••••••••••••••••••••••••••••••••••	Wiper moves towards CW with clockwise rotation of control (as viewed from shaft or knob).	e rotation of control (as viewed	
	Numbered test point. Measurement aid provided.	provided.	
	Denotes wire color code. Code used is the same as the resistor color code (e.g., 9.4.7 denotes white/yellow/violet),	the same as the resistor color //violet),	
<b>-</b>  ı	Indicates direct conducting connection to earth.	earth.	
4	Indicates conducting connection to chassis or frame.	is or frame.	
<b>→</b>	Indicates circuit common connection.		

Figure 8-20. Schematic Diagram Notes.

8-43

Figure 8-22. Adjustment Locations.

A22 Display Control & RAM A35 BCD Output Control A24 Comparator Control A21 Keyboard Control A25 HP-IB Interface Koleman and Antonomical Control of the Control of t Option board slots Figure 8-21. Assembly Locations. ALON WIND LIES, John HIDDEN A9R6 (+12V) To reference of A13R67 (OFS-4) A9 Power Supply A13R2 (OFS-2) A14R15 (APAO) A13C1. A12C11~ A12R1 A14 Phase Detector & Integrator A13 Process Amplifier A11 OSC & Source Resistor A12C3 A12 Range Resistor A14R1 (ZOF) A13R1 (OFS-1)

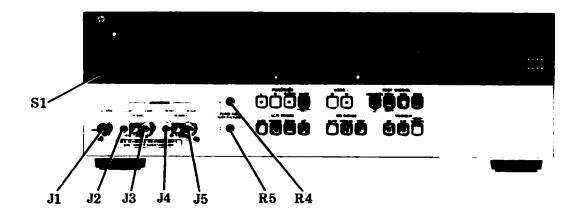


Figure 8-23. Front Panel Component Locations.

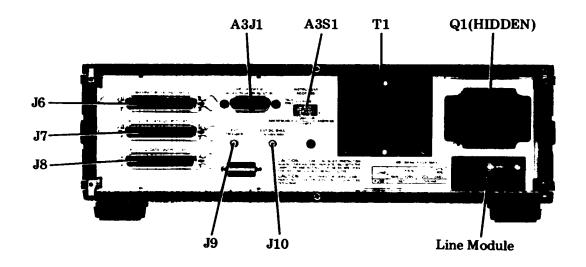
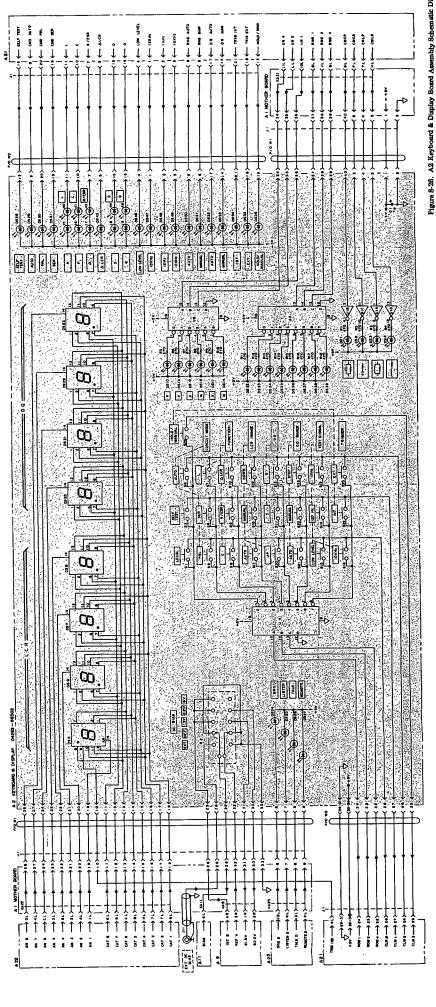
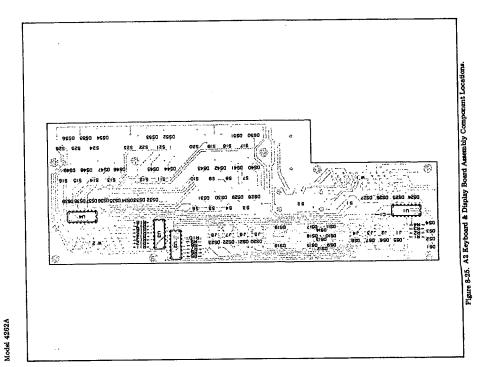


Figure 8-24. Rear Panel Component Locations.

# A9 Board Troubleshooting Tree Under Fold





.

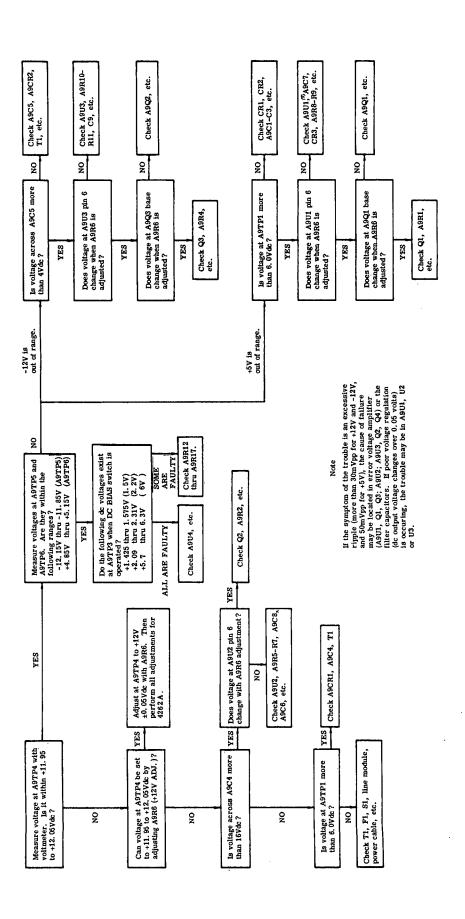
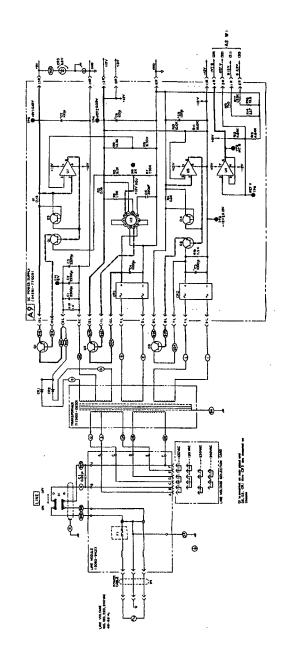
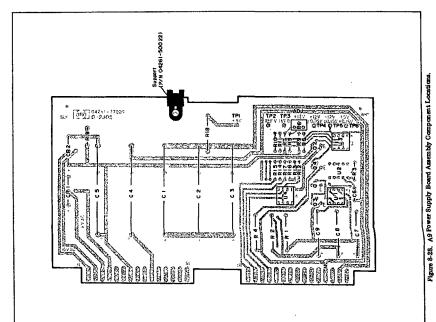


Figure 8-27. A9 Power Supply Board Troubleshooting Tree.

# A11 Board Troubleshooting Tree Under Fold







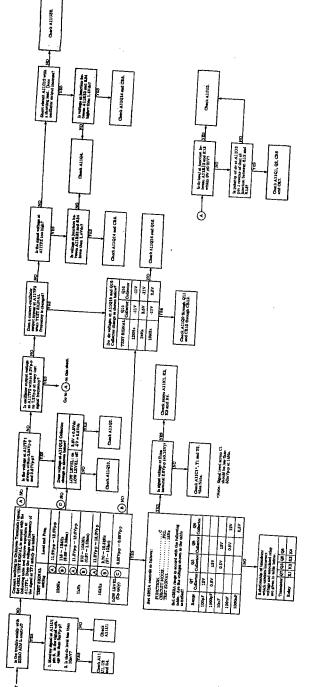


figure 8-30. A11 OSC & Source Resistor Board Troubleshooting Trees

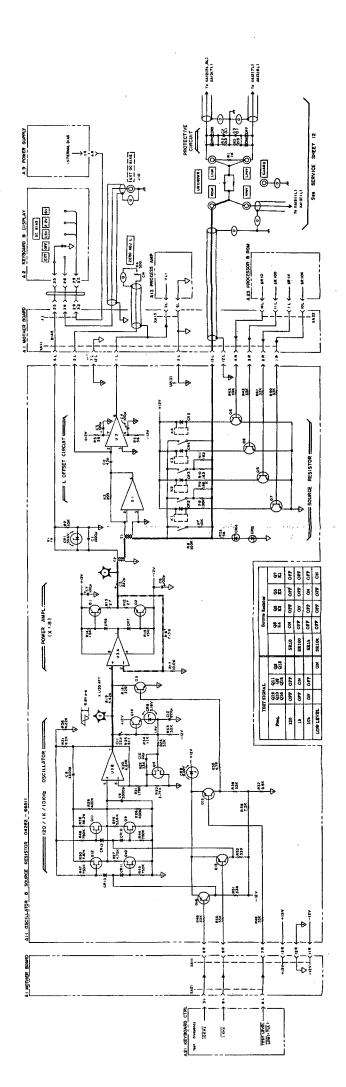
### A11 BOARD CIRCUIT DESCRIPTION.

The Wien bridge oscillator frequency is derived from the equation:  $1/(2\pi\sqrt{RaRbC8C9})$ . Associated resistances Ra and Rb are selected from resistors R25 through R30 by active switches Q9 through Q12 to set desired test signal frequency (both Ra and Rb have the same value). The relationships of the switches to the oscillation frequency are shown in the table with the circuit diagram. Automatic level control circuit Q4 and Q14 operates to maintain a constant oscillator output level against changes in oscillator circuit parameters and supply voltage as follows: if the oscillator output level rises above 6.8Vp-p, Q14 is turned on for a longer period, the voltage across C12 increases, and Q3 is moved nearer to an OFF condition. Therefore, the feedback to U3B increases, and the gain of U3B is decreased to lower its output level to the proper amplitude. This provides stable amplitude characteristics to the oscillator. The table below shows the relationship of selected source and range resistors to 4262A FUNCTION, CIRCUIT MODE, and RANGE settings. At any setting, both the range resistor Rr and source resistor Ro have the same value. Note that the  $100\Omega$  and  $10\Omega$  source resistances include the total series resistance of the range resistors and the output resistance of transformers T1 and T2.

I able A.	kange Resistor	(Kr	and Source I	Resistor	(Ro)	Selections.

Tuble 11. Italige Resistor (Ita) and Source Resistor (Ita) Selections.										
Range Function		1	2	3	4	5	6	7	8	
L	Full- scale	120Hz	1000µH	10.00mH	100.0mH	1000mH	10.00H	100.0H	1000H	
		1kHz	100.0μH	1000µH	10.00mH	100.0mH	1000mH	10.00H	100.0H	
		10kHz	10.00μH	100.0μΗ	1000µH	10.00mH	100.0mH	1000mH	10.00H	
	RR, Ro	SER	$10\Omega$	100Ω	1kΩ	10kΩ	100kΩ			
		PARA				10Ω	100Ω	1kΩ	10kΩ	
C	Full- scale	120Hz	1000pF	10.00nF	100.0nF	1000nF	$10.00 \mu \mathrm{F}$	100.0μF	1000μF	10.00mF
		1kHz	100.0pF	1000pF	10.00nF	100.0nF	1000nF	10.00μF	100.0µF	1000µF
		10kHz	10.00pF	100.0pF	1000pF	10.00nF	100.0nF	1000nF	10.00µF	100.0μF
	RR, Ro	PARA	100kΩ	10kΩ	1kΩ	$100\Omega$	$10\Omega$			
		SER				100kΩ	10kΩ	1kΩ	100Ω	10Ω
R s	Full- scale	120/ 1kHz/ 10kHz	1000mΩ	10.00Ω	100.0Ω	1000Ω	10.00kΩ	100.0kΩ	1000kΩ	10.00MΩ
	RR, Ro	SER	10Ω	100Ω	1kΩ	<b>10k</b> Ω	100kΩ			
		PARA				$10\Omega$	100Ω	1kΩ	<b>10k</b> Ω	100kΩ

A12 Board Troubleshooting Tree Under Fold



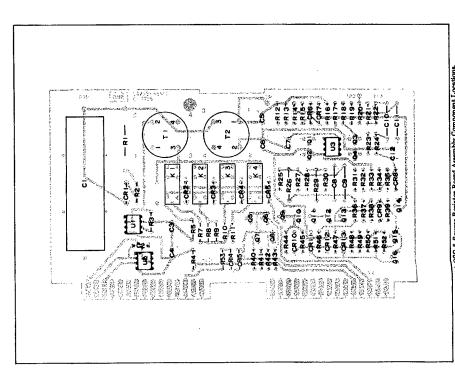


Figure 8-51. A11 08C & Source Resistor Board Assembly Component Locali

MAG.

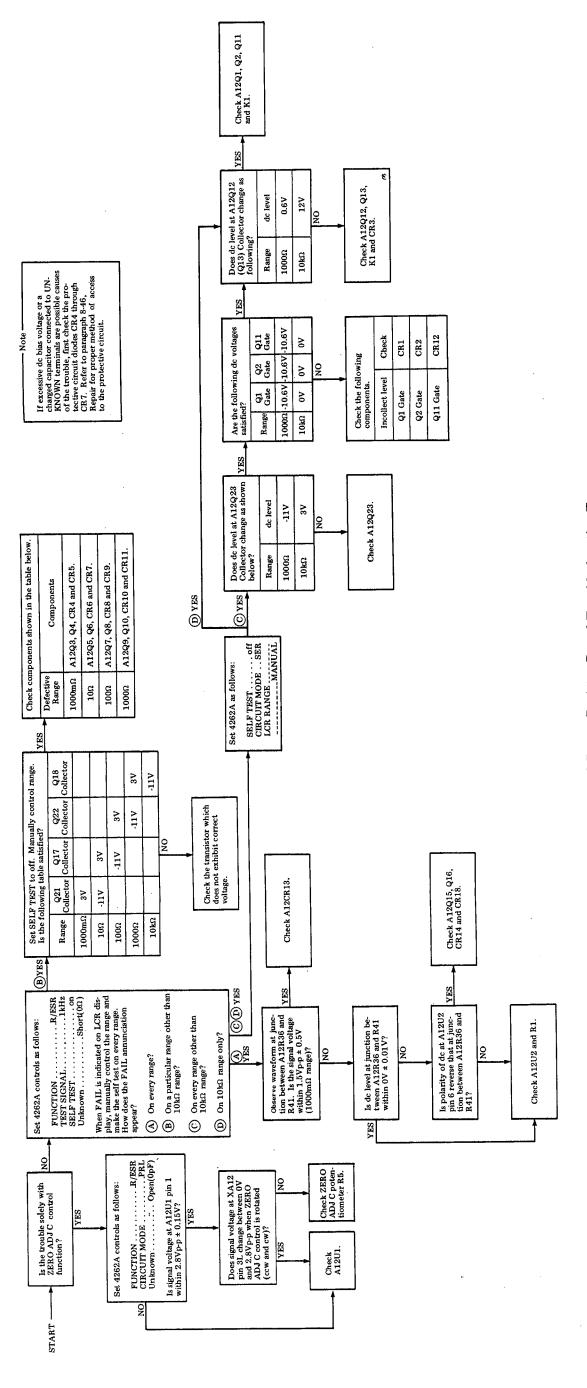


Figure 8-33. A12 Range Resistor Board Troubleshooting Tree.

### A12 BOARD CIRCUIT DESCRIPTION.

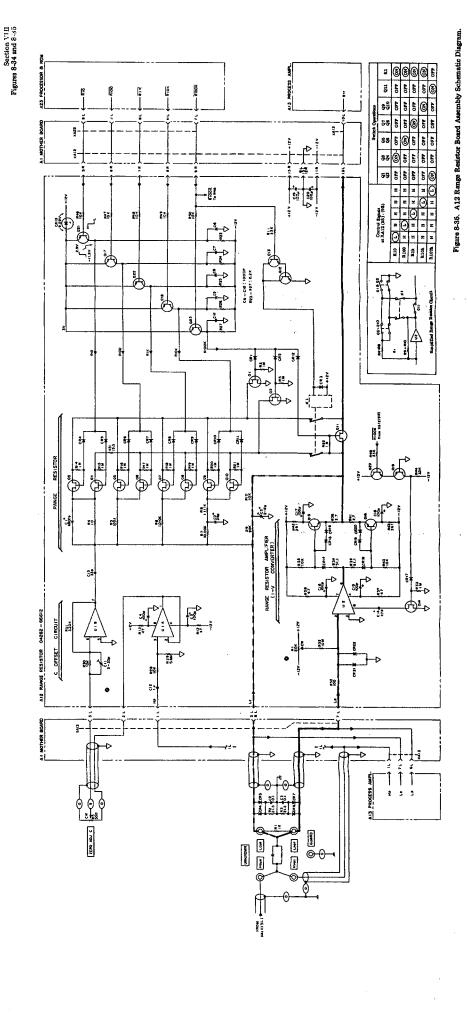
Table A below shows the relationship of selected range resistors to 4262A FUNCTION, CIRCUIT MODE and RANGE settings. Range selector switches (active switches) Q3 through Q10 and associated switches Q1, Q2, Q11 and K1 (relay) are controlled to select the range resistance which will provide an appropriate full scale range (see table with circuit schematic). Two switches concurrently act to enable detection of an exact voltage drop across the range resistor regardless of the resistance of the range selection switch through which the range resistor current flows. For example, both Q3 and Q4 turn on to sense the voltage drop and to simultaneously route the DUT current flow through range resistor R4 (10 $\Omega$ ). R4 and Q4 compose a feedback loop in the Range Resistor amplifier on the selected range. The exact voltage drop across R4 is routed through Switch Q3 (K1: ON, Q11: OFF). The selectable  $10\Omega$ ,  $100\Omega$ ,  $1k\Omega$  and  $10k\Omega$ range resistances are always placed in parallel with the permanent 100kΩ range resistance (R9 plus R10). The  $100k\Omega$  range resistance alone is selected by causing Q11 to turn on, K1 to deenergize, and its contacts to open. The open contacts of K1 also interrupt the error current flowing through the stray capacitance in the range resistor circuit. This eliminates any error current effect on the circuit being used (R9, R10 and C3) on this range. In addition, to further reduce the error current, Q1 and Q2 conduct the current flowing through the stray capacitance of the relay contacts to ground.

Table A. Range Resistor (RR) and Source Resistor (Ro) Selections.

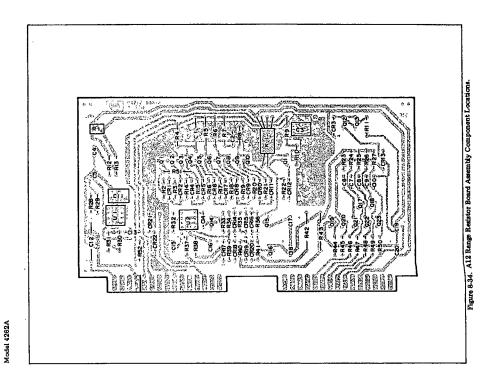
Fu	nction	Range	1	2	3	4	5	6	7	8
L	Full- scale	120Hz	1000µH	10.00mH	100.0mH	1000mH	10.00H	100.0H	1000H	
		1kHz	100.0μΗ	1000µH	10.00mH	100.0mH	1000mH	10.00H	100.0H	
		10kHz	10.00μH	100.0μH	1000μH	10.00mH	100.0mH	1000mH	10.00H	
	Rr, Ro	SER	10Ω	100Ω	1kΩ	$10 \mathrm{k}\Omega$	100kΩ			
		PARA				10Ω	$100\Omega$	1kΩ	$10 \mathrm{k}\Omega$	
C	Full- scale	120Hz	1000pF	10.00nF	100.0nF	1000nF	10.00μF	100.0μF	1000μF	10.00mF
		1kHz	100.0pF	1000pF	10.00nF	100.0nF	1000nF	$10.00 \mu F$	100.0µF	$1000 \mu F$
		10kHz	10.00pF	100.0pF	1000pF	10.00nF	100.0nF	1000nF	10.00µF	$100.0 \mu \mathrm{F}$
	Rr, Ro	PARA	<b>100</b> kΩ	10kΩ	1kΩ	100Ω	10Ω			
		SER				100kΩ	10kΩ	1kΩ	$100\Omega$	$-10\Omega$
R	Full- scale	120/ 1kHz/ 10kHz	1000mΩ	10.00Ω	100.0Ω	1000Ω	10.00kΩ	100.0kΩ	1000kΩ	10.00MΩ
	RR, Ro	SER	$10\Omega$	100Ω	1kΩ	10kΩ	100kΩ			
		PARA				$10\Omega$	100Ω	1kΩ	$10 \mathrm{k}\Omega$	100kΩ

## A13 Board Troubleshooting Tree Under Fold





8-51



. . .

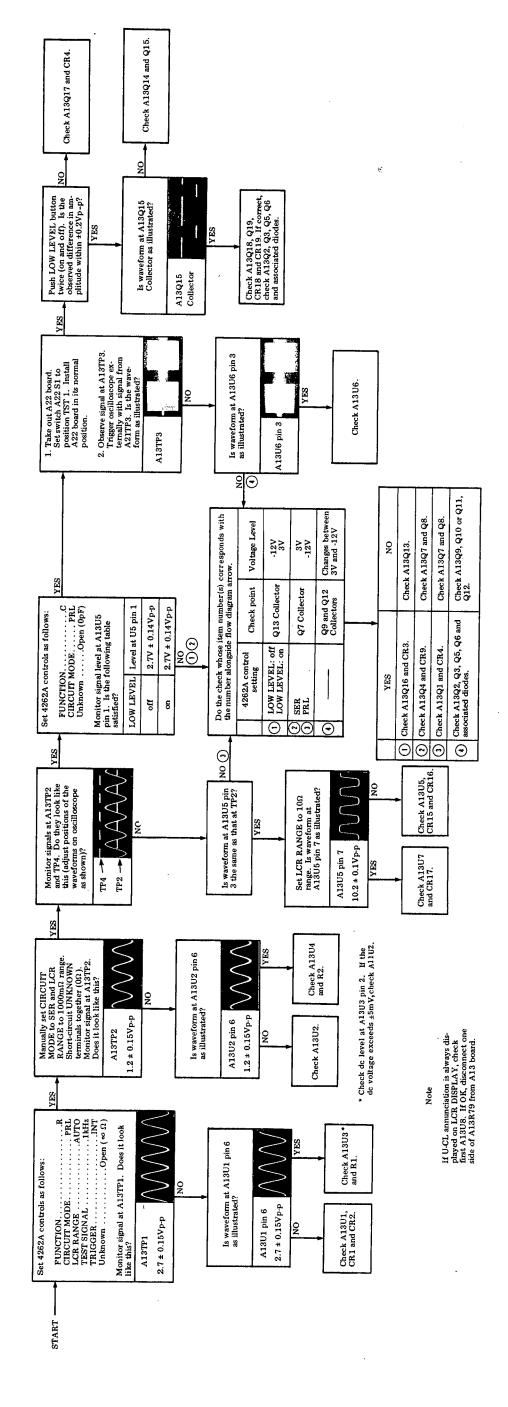


Figure 8-36. A13 Process Amplifier Board Troubleshooting Tree.

#### A13 BOARD CIRCUIT DESCRIPTION.

The input circuitry of the A13 board is composed of impedance converters and differential amplifiers which sense the exact voltage drops across the DUT (E1) and across range resistor (E2). The choice of the eref and Em signals by Q1 through Q6 depends upon the FUNCTION and CIRCUIT MODE settings. Switches Q1 and Q4 select the phase detector phase references (eref) from either ex or ey (representing E1 and E2, respectively) differential amplifier outputs as directed by the CMS (Circuit Mode Selection) signal. Switches Q2, Q3, Q5 and Q6 sequentially select the em signal (as components of the measured quantity) from among the  $e_x$ ,  $e_x/10$ ,  $e_y$  and  $e_y/10$  signals. The method of the selection, relative to the measurement mode, is graphically illustrated in Figure 8-8 Timing Diagram. When the TEST SIGNAL function is set to LOW LEVEL, both Q16 and Q17 turn on. To maintain the amplitudes of eref and em signals the same as in taking a measurement with a standard test signal level, the amplification factors of amplifiers U5A and U6B are now increased by 20 times. If the amplitude of U6B output (em) exceeds ±5.2V peak, the window comparator U8 outputs a SAT (saturation) pulse which signals that an improper FUNCTION or RANGE setting is being attempted for measuring the unknown device. Switches Q18 and Q19 operate during the integrator null offset sequence (refer to Page 8-56 for the null offset control details). An APAO (Auto Phase Adjustment Output) signal, added to the eref signal at the input stage of the Phase Shifter U5B causes a change in the phase of the eref signal. This phase change on the APAO voltage is determined by a comparison of the phase shifter output to the zero level. Circuit operating theory of the Phase Shifter is given in the following paragraph.

#### AUTO PHASE ADJUSTMENT (Phase Control).

This paragraph should be read along with the general description of the auto phase adjustment (on service sheet 14). A DC input (APAO) to the Phase Shifter is added to the ac input signal (eref) for the purpose of shifting the ac waveform upwards or downwards depending on the dc input level (as illustrated in Figure A). Additionally, the phase shifter reverses polarity of the signal. The phase shifter output is wave-shaped to a square wave which changes its polarity every time that the phase shifter output waveform crosses the zero level. The waveforms drawn in solid lines in Figure A are those that exist when 0V dc input (APAO) is applied. Waveforms in dotted lines are those that are present when a plus dc input (APAO) is applied. When an ac signal with a certain dc (APAO) level is inputted, the duty factor of the eref signal is shifted (narrowed or widened) as the phase shifter output is wave-shaped with respect to a fixed (0V) reference. Therefore, the phase of the PLL output used for phase detection will vary since the PLL circuit detects only the trailing edge of an eref signal.

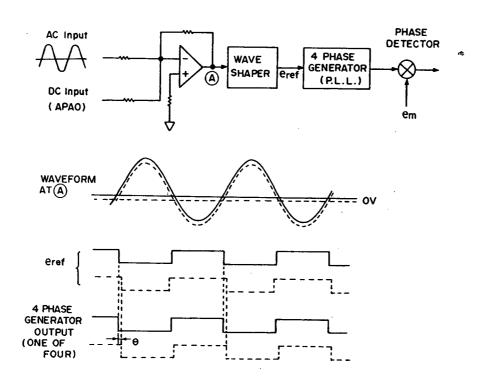


Figure A. Phase Control.

# A14 Board Troubleshooting Tree Under Fold



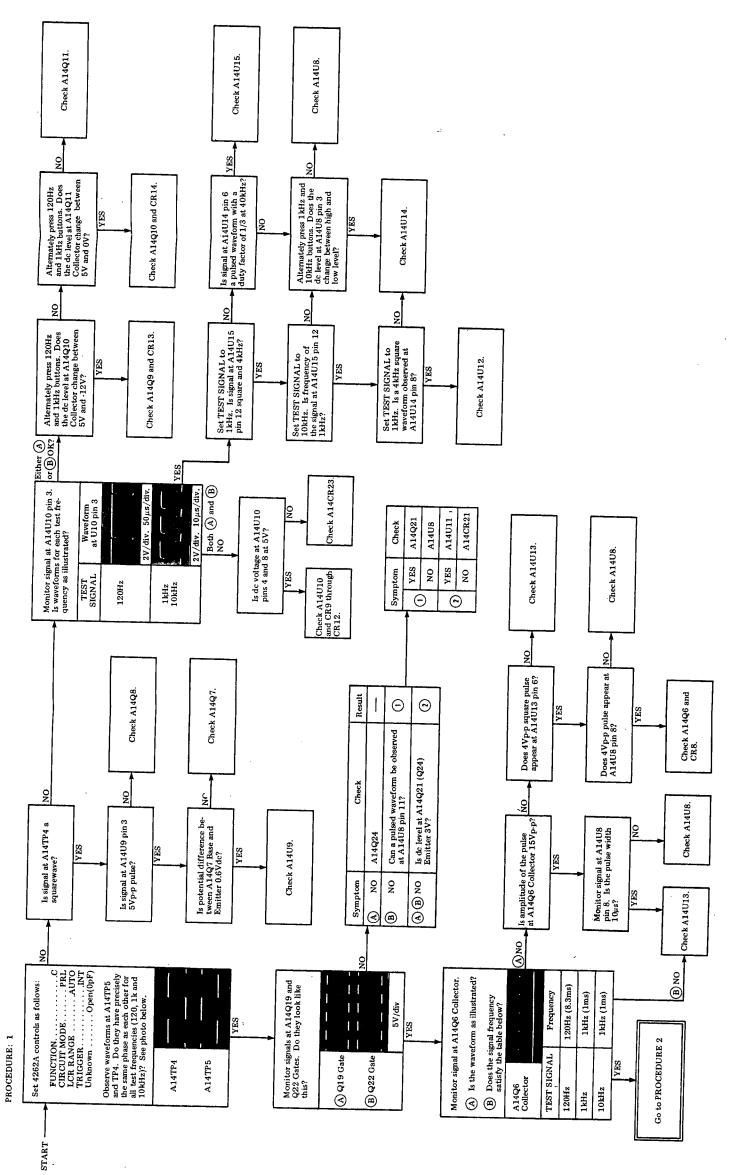
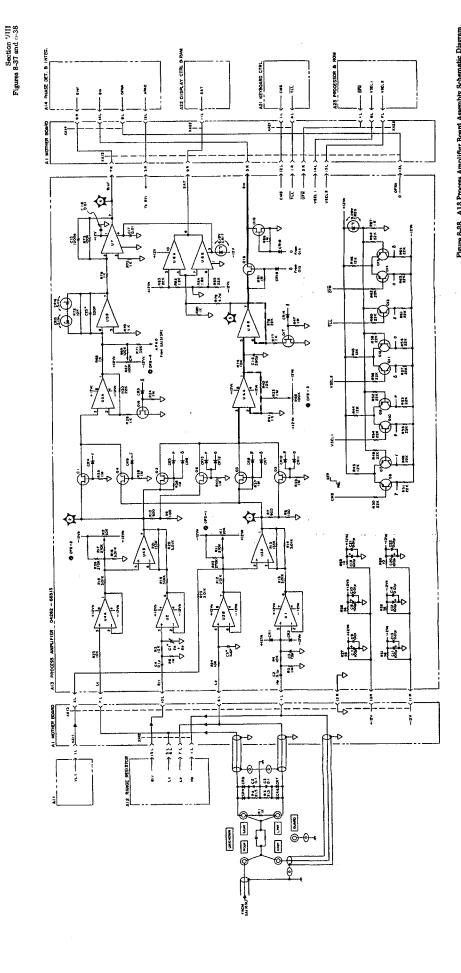
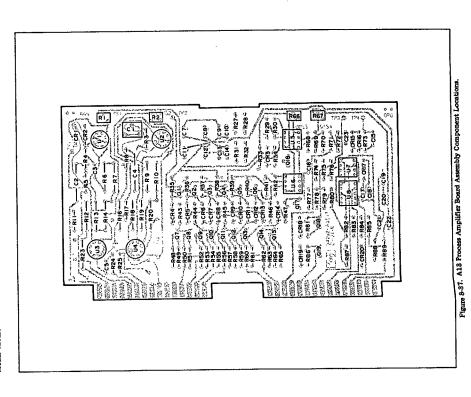


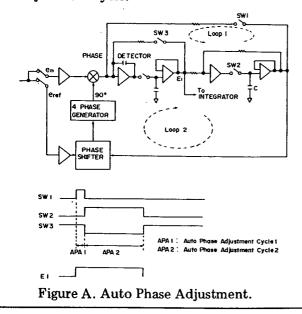
Figure 8-39. A14 Phase Detector & Integrator Board Troubleshooting Tree (A).





## AUTO PHASE ADJUSTMENT.

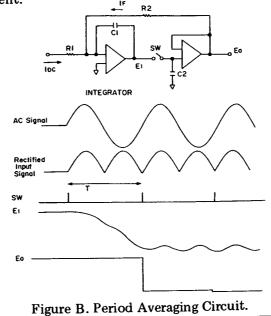
By using a feedback control technique in the Auto Phase Adjustment period, the phase of the Phase Locked Loop circuit is automatically adjusted to minimize detection phase error. This paragraph describes how the phase error is eliminated during APA (Auto Phase Adjustment) cycle periods 1 and 2. The basics of the auto phase adjustment circuit are diagrammed in Figure A. In the APA 2 period, the same signal is applied to both the phase detector and the phase shifter. The four phase generator outputs a 90 degree phase shifted pulse. Assuming that the detection phase is accurate, the average level of the phase detector output should necessarily be zero. If any phase error exists between the em and eref signal channels, the phase detector outputs an E1 signal which is the integrator output for such error signal. Because SW3 is open, the period averaging circuit functions as an ordinary integrator. APA amplifier (A14U5A and U6) which follow develop an APAO signal (dc) proportional to E1 and supply it to the phase shifter. In response to the APAO voltage, the phase shifter output tends to lower the E1 level. The phase error of detection phase is thus minimized (refer to phase shifter circuit description on service sheet 13). The APA2 cycle is performed by LOOP 2 (SW1: OFF, SW2: ON, SW3: OFF) as denoted in the diagram. After the APA2 period, SW3 is closed and SW2 is turned off to memorize the dc voltage stored in capacitor C. The memory capacitor C maintains the dc voltage to continuously provide an effective APAO signal during the measurement cycle. In the APA1 (Auto Phase Adjustment cycle 1) period which is done prior to APA2, SW1 is closed and the current flow through the LOOP 1 charges the capacitors in the period averaging circuit. APA1 control is provided to accelerate development of an appropriate APAO signal during the APA2 period and helps to reduce the time of the auto phase adjustment cycle.



### A14 BOARD CIRCUIT DESCRIPTION.

## PERIOD AVERAGING CIRCUIT.

A period averaging technique was adopted to get pure dc voltage at high speed from a rectified ac signal having a large ripple component. Generally, a filtering circuit has a long transient response time in converting a low frequency burst input signal to a pure dc voltage. The period averaging technique enables a dc output voltage to be produced which is almost equal, (in a precise fashion) to the final value in only several periods of the input ac signal. The 4262A employs the period averaging circuit for smoothing the phase detector output to a dc and for combining specified measurement accuracies and provides an improved measurement speed at the 120Hz test frequency. Figure B shows the full-wave rectified current input signal of this circuit. During the first T (time) period, the input current charges the integrator capacitor C1 (A14C8 in the actual circuit. In a 120Hz measurement, Q18 conducts to add C7 in parallel with C8). At the end of this period, the integrator output E1 is proportional to the dc current of the input signal (since T is equal to one period of the input ac signal). After the first T period, voltage E1 is memorized as a charge on C2 (A14C6) when switch SW (A14Q5) is momentarily closed, and E0 (period averaging circuit output) becomes a step function. As the feedback current (IF) from E0 to the integrator input is designed to be almost equal to Inc (input current to the period averaging circuit) in magnitude, the difference between Inc and IF is integrated during the next (T) integrating period so that output voltage E0 becomes exactly proportional to Inc. After four or five periods, E0 will be a pure dc signal having no ac component and be precisely proportional to (Ix) the input current.



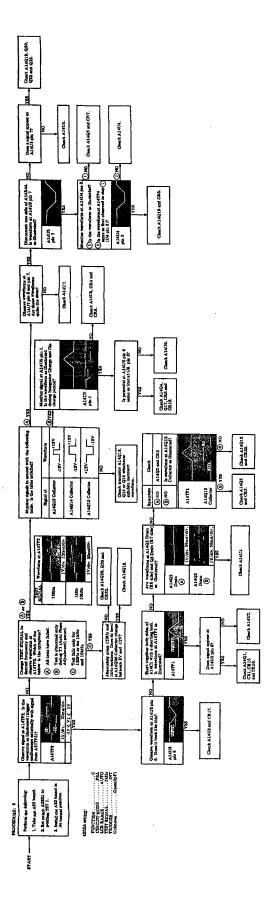


Figure 8-40. A14 Phase Detector & Integrator Board Troubleshooting Tree (B).

### A14 BOARD CIRCUIT DESCRIPTION.

PHASE LOCKED LOOP (PLL) CIRCUIT AND 4 PHASE GENERATOR.

Figure C shows the block diagram of the phase locked loop circuit used to establish an accurate detection phase in the phase detector. The PLL technique was incorporated to develop an input to the Four Phase Generator which satisfies the requirements of phase and frequency accuracies for establishing the exact relationships between the four phase generator output and the measurement signal. When the PLL control is off, the VCO oscillates at a frequency close to 40 times the frequency of the input signal (eref) to the Phase Shifter. In the 120Hz measurement setting, the frequency of VCO output becomes 4.8kHz. A 1/10 down counter U15 and the Four Phase Generator U12 (a 1/4 down counter) count down the VCO output frequency to 120Hz. This becomes the frequency of the feedback signal ef to the local phase detector (LPD) U9. The output voltage of the LPD (converted to a dc by Low Pass Filter Q7 and Q8) directs the oscillation of VCO so that the difference in both frequency and phase between the two input signals (eref and ef) to the LPD tends to become minimum. Eventually, both the phase and frequency of the four phase generator output (one of four) is precisely the same as that of the eref signal (120Hz). In a 1kHz measurement frequency setting, switch Q9 is turned off to change the oscillation frequency of the VCO to 40kHz. In a manner similar to that for the 120Hz measurement, the four phase generator output is fixed to the exact frequency of eref signal (1kHz). When measurement frequency is switched to 10kHz, the 40kHz VCO output passes through the gate circuitry (U14) and bypasses the 1/10 down counter. Thus, the frequency of the feedback signal ef

becomes 10kHz. The frequency of the four phase generator input is always four times the eref signal frequency. The 4f pulse train is converted to four square wave signals, each having an exact phase difference of 0°, 90°, 180° and 270° with respect to the negative edge of the Cref signal. The U13 Gate circuitry periodically creates a short pulse which drives sampling switch (Q5) of the period averaging circuit in synchronism with the measurement signal. In a 10kHz measurement, the four phase generator output is fed to the 1/10 down counter whose output is inputted to gate circuitry U13. The U13 output is a 1msec (1kHz) pulse train which drives the sampling switch Q5 at a rate of once in 20 periods of the period averaging circuit input (phase detector output) signal. The periodic rate is sufficient for period averaging of the high frequency input signal.

## INTEGRATOR NULL OFFSET CONTROL.

During the offset null sequence period, the Amplifier output offset voltages present in the phase detector and the integrator stages are reduced to zero at the integrator output. While the offset null is being performed, switches A13Q18 and Q19 interrupt em signal transfer to the Phase Detector. Simultaneously, A14Q1 and Q2 turn on. Q2 provides the integrator with a lower input resistance and advances charging to achieve a shorter null offset control period. The Integrator produces a dc output which represents the accumulated charge of the offset voltages. The integrator output is stored in capacitor C1 to maintain its voltage during the measurement cycle. Any incoming voltage to the integrator is referenced to the voltage across the charged capacitor. Thus, any offset voltages present are eliminated and are not a factor in the integrator output.

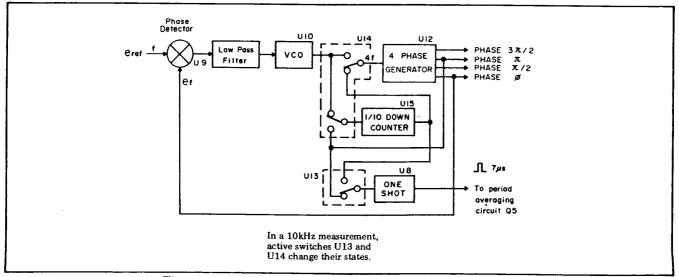


Figure C. Phase Locked Loop Circuit Block Diagram.

Model 4262A

Section VIII

KEYBOARD CONTROL.

Figure A below shows the simplified echematic of the Keyboard Control. Pressing a pushbutton key creates a connection between one of the 8 'vow" lines and one of the 4 'voolumn' lines. In the keyboard Coutroity, an individual switch is distinguished by its "address" which is related to a specific "vow" and "column" line. Identification of the pushbutton pressed and its associated function is coordinated by a time sharing operation of the keyboard courted system. A "keyboard court of the keyboard courtile standing concept contributes to the simplification of the circuit nereting a keyboard address code unique to each keyboard sawitch. The operation of the keyboard control may be explained as follows: The Scan Counter (A2102), whose time base is the 31.83kHz clock, divess the Decoder (A2404). The scan counter outputs are 3 bit ROW signals (binary ROWs 1, 2 and 4). These three signals are sufficient to achieve binary outputs of 1 through 8 from the Decoder. For example, a binary input of 101 will cause an output to occur on decoder couptut to will be super pulse width as that of the

011(6)D6 U11(10)D8

Table A. Keyboard Switch Test Signature,

U22(6)D1 U32(10)D2

input signals (ROW signals) on the 8 channel "row" drive lines and corresponding with the binary ROW signals. The 8 channel row lines, in turn, become low level as illustrated in Figure A. A row signal causes the three or four pushbutton keys on the row line to become valid (enabled). If a pushbutton is pressed while it is enabled (able to function), Gate (AZJUJ) switches its output logic and instantaneously stops the Scan Counter. Because the keyboard "scan" speed is extremely fast compared to the time it takes to depress the pushbutton, all the keyboard controls are seemingly always valid (enabled). When a pushbutton is pressed, the counter input to the decoder is momentarily interrupted and the column line peculiar to the individual pushbutton key goes to low level. Thus, each key can be identified by observing the RoW and CLM signals, but before the ROW and the CLM signals, but before through the Data Bus Driver/Receiver towardt the data bus line, flip (flop (AZJUJ4) outputs an INT of signal, the Interrupt Priority Encoder outputs a VAC (Vector Address O) signal which informs the nanoprocessor in response to the INT O signal, the Interrupt was generated from the Reyboard Control. The interrupt is managed in accord with the interrupt process routing of the nanoprocessor.

DECODER C 3333

Figure A. Keyboard Control Simplified Schematic Diagram

8-58

8-57

EASP\*

LOCAL

SELF TEST

CMD ANTO

CMD STR

FUNC I

FUNC I

FUNC I

FUNC RNG ANTO

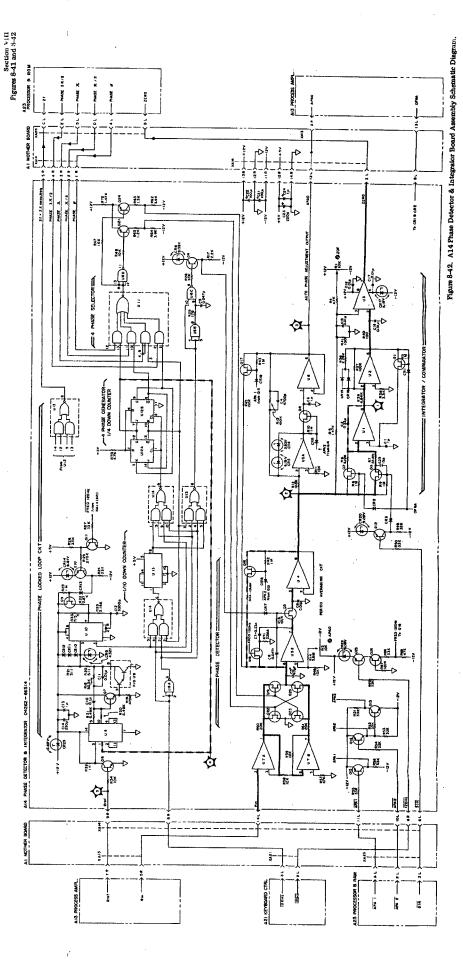
LCR RNG ANTO

LOSS D

LOSS Q

\* Depressing the keys listed will result in the signatures defined in Table A.

A14 Phase Detector & Integrator SERVICE SHEET 14



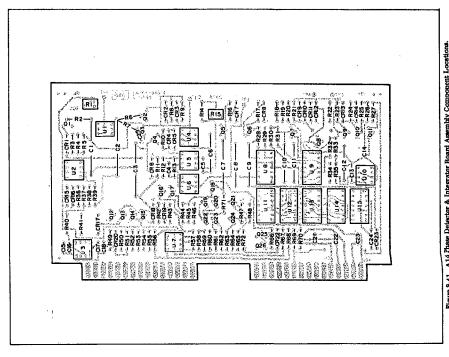


Figure 8-41. A14 Phase Detector & Integrator Board

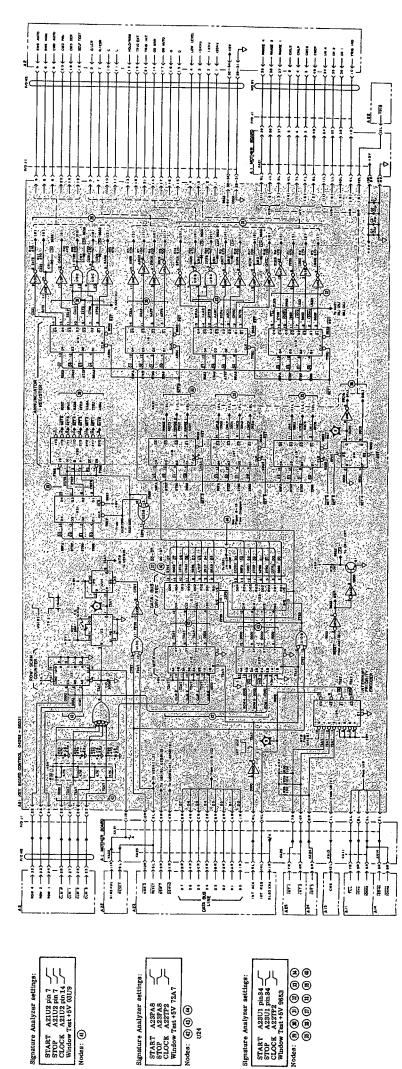


Figure 8-43. A21 Key board Control Board Assembly Com

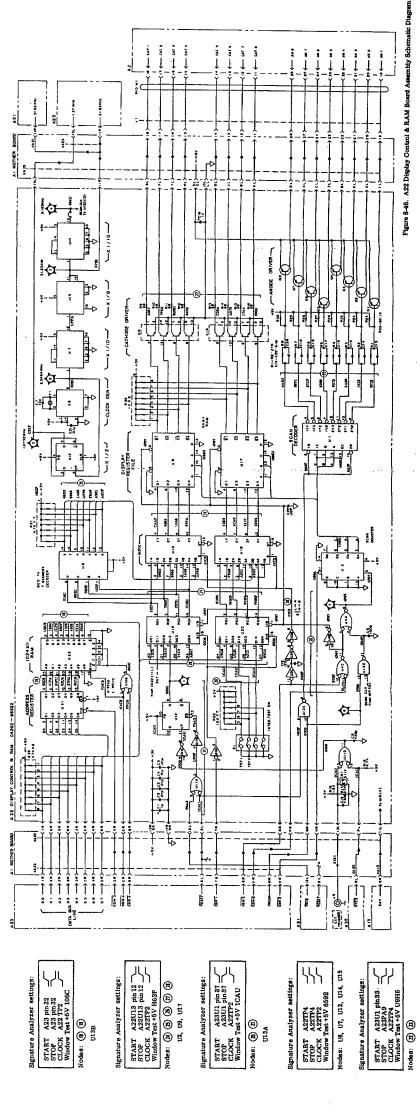


Figure 845. A22 Display Control & RAM Board Assembly Component Locations.

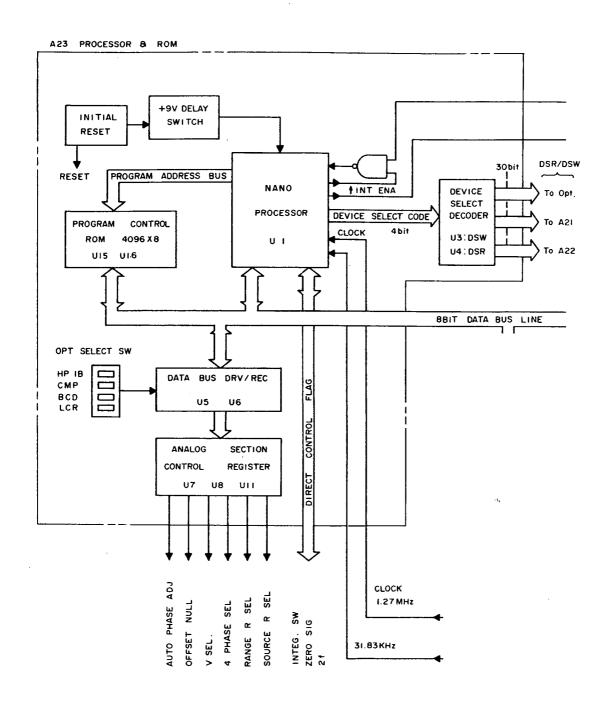
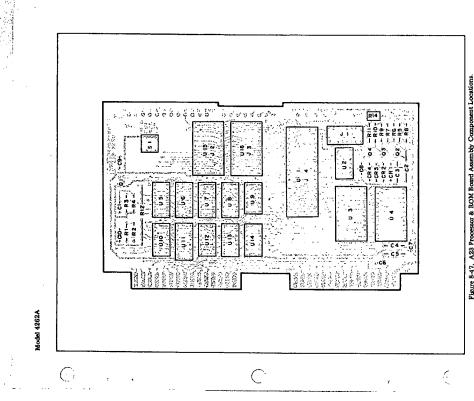


Figure A. A23 Processor & ROM Block Diagram.





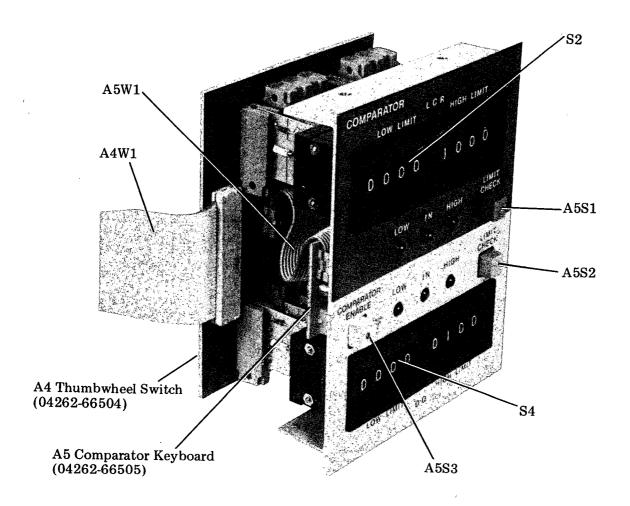
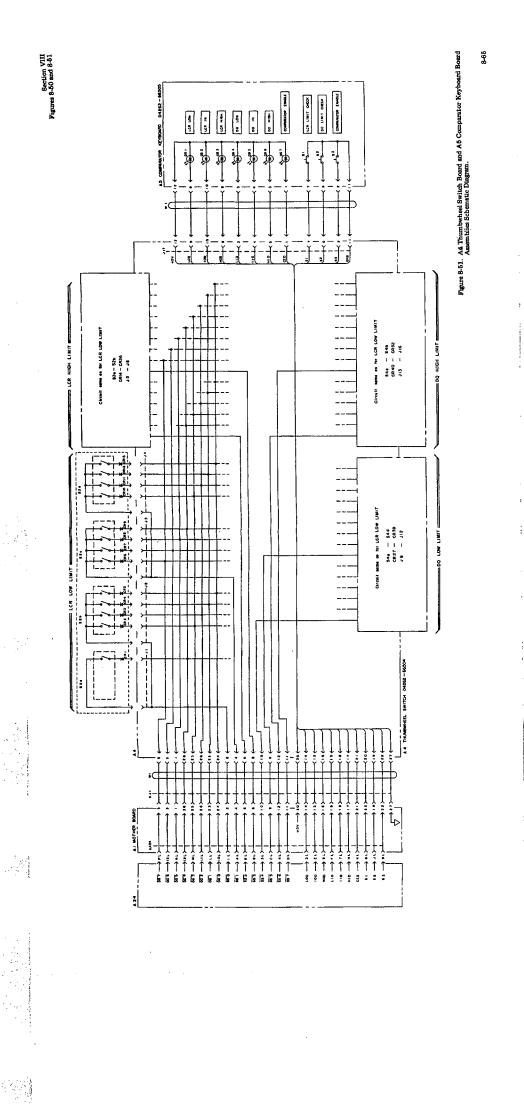


Figure A. Comparator Control Panel Assembly Component Locations.



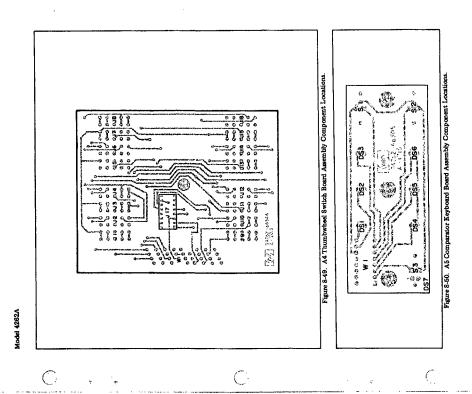


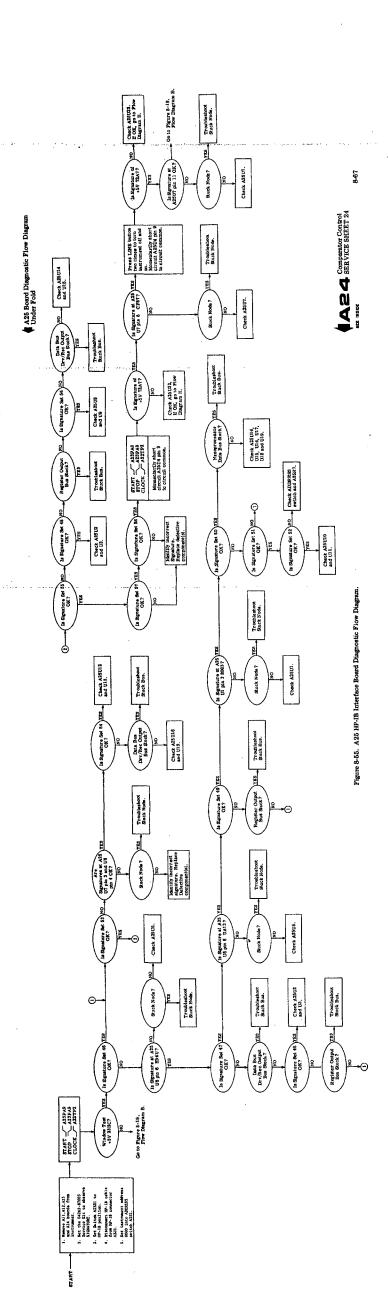
Table A. Comparator Keyboard Test Signatures.

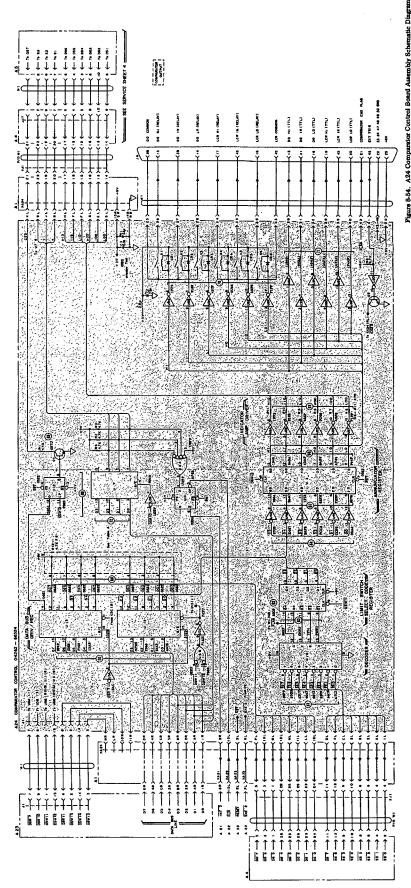
Keyboard Switch	U7(3) D0	U7(6) D1	U7(10) D2
LCR LIMIT CHK	AF4C	AF08	AF08
DQ LIMIT CHK	AF0C	AF48	AF08
CMP ENA	AF0C	AF08	AF48

Window Test (+5V): U9FF

## Note

To observe window test signature, continue pressing COMPARATOR ENABLE button for the duration of the initial window test. Then, press pushbuttons in accord with Table A above. Signatures for each individual circuit node can be observed while the appropriate button is being pushed.





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ei 4262A

## A35 Board Diagnositc Flow Diagram Under Fold



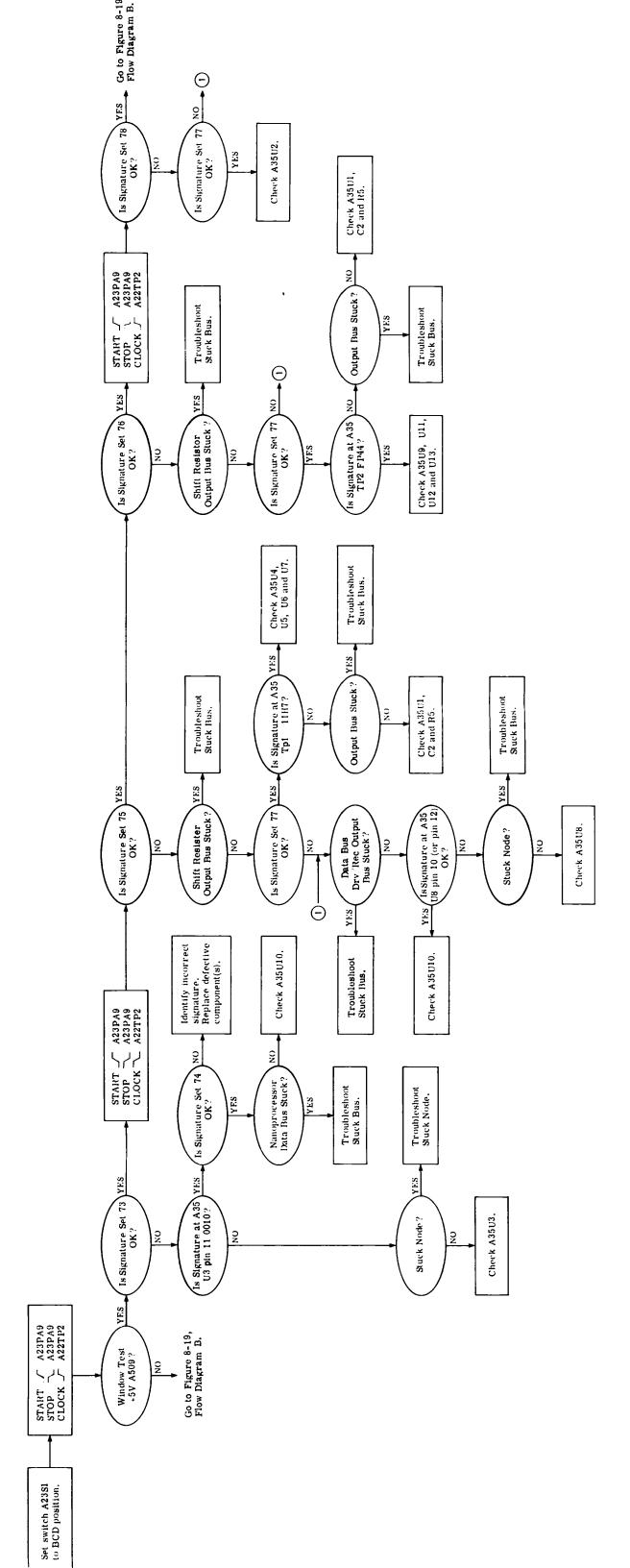


Figure 8-58. A35 BCD Output Control Board Diagnostic Flow Diagram.

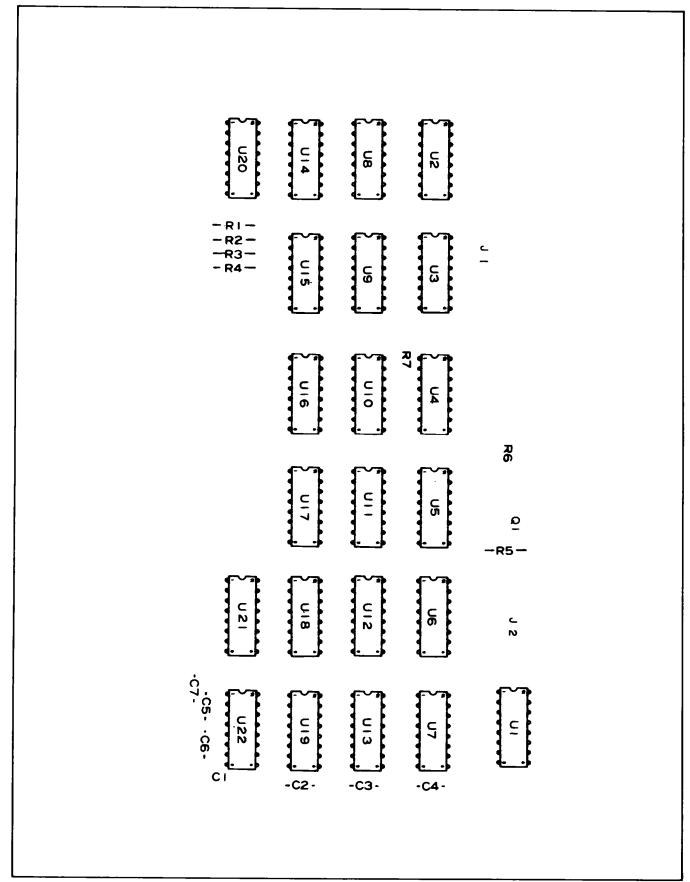


Figure 8-56. A25 HP-IB Interface Board Assembly Component Locations.

## Signature Analyzer settings:

START A23PA8 STOP **A23PA8** CLOCK A22TP2 Window Test +5V 72A7

Nodes: U22, U7B, U7D

## Signature Analyzer settings:

A23PA9 A23PA9 **START STOP** CLOCK A22TP2 Window Test +5V 23HC

Nodes: (46) (17) (48) (49)

(52) (53) (54) (55) (56) (57)

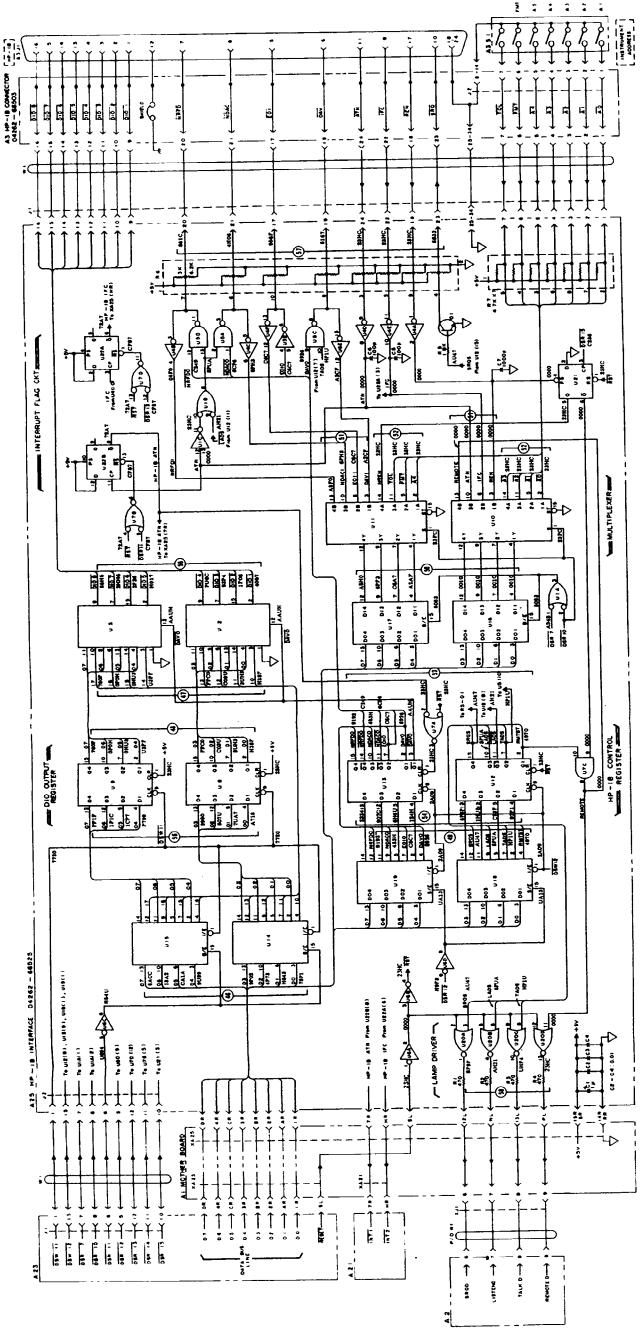


Figure 8-57. A25 HP-IB Interface Board Assembly Schematic Diagram.

8.69

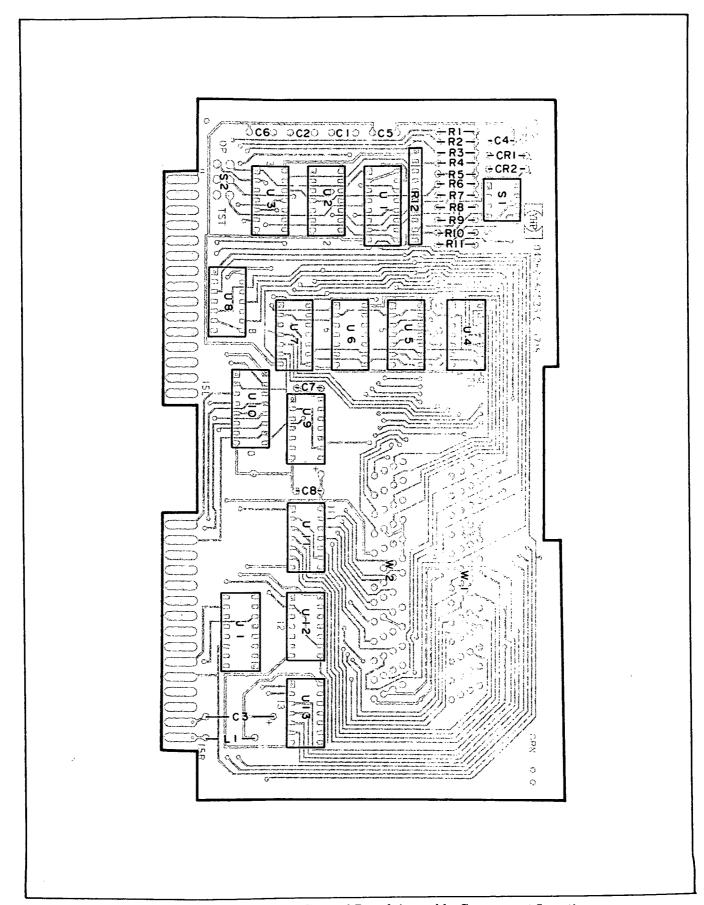


Figure 8-59. A35 BCD Output Control Board Assembly Component Locations.

Signature Analyzer settings:

START A23PA9
STOP A23PA9
CLOCK A22TP2
\*CLOCK
Window Test +5V A509

Nodes: (13) (14) (15) (16) (17) (18)

