TEK

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Product Group 38

2

2445B/2455B
OSCILLOSCOPES
SERVICE

CALIBRATION LAB
INSTRUCTION MANUAL FOR
Oscilloscopes
Model #: 2445/2455/2465 Date: 7 JUL 98
Cal Tech:

WARNING

THE FOLLOWING SERVICING INSTRUCTIONS ARE FOR USE BY QUALIFIED PERSONNEL ONLY. TO AVOID PERSONAL INJURY, DO NOT PERFORM ANY SERVICING OTHER THAN THAT CONTAINED IN OPERATING INSTRUCTIONS UNLESS YOU ARE QUALIFIED TO DO SO. REFER TO OPERATORS SAFETY SUMMARY AND SERVICE SAFETY SUMMARY PRIOR TO PERFORMING ANY SERVICE.

Please Check for CHANGE INFORMATION at the Rear of This Manual

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INSTRUMENT SERIAL NUMBERS

Each instrument has a serial number on a panel insert, tag, or stamped on the chassis. The first number or letter designates the country of manufacture. The last five digits of the serial number are assigned sequentially and are unique to each instrument. Those manufactured in the United States have six unique digits. The country of manufacture is identified as follows:

B000000	Tektronix, Inc., Beaverton, Oregon, USA
100000	Tektronix Guernsey, Ltd., Channel Islands
200000	Tektronix United Kingdom, Ltd., London
300000	Sony/Tektronix, Japan
700000	Tektronix Holland, NV, Heerenveen, The Netherlands

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OPERATORS SAFETY SUMMARY

The general safety information in this part of the summary is for both operating and servicing personnel. Specific warnings and cautions will be found throughout the manual where they apply and do not appear in this summary.

Terms in This Manual

CAUTION statements identify conditions or practices that could result in damage to the equipment or other property.

WARNING statements identify conditions or practices that could result in personal injury or loss of life.

Terms as Marked on Equipment

CAUTION indicates either a personal injury hazard that is not immediately accessible as one reads the markings, or a hazard to property (including the equipment itself).

DANGER indicates a personal injury hazard immediately accessible as one reads the marking.

Symbols in This Manual



This symbol indicates where applicable cautionary or other information is to be found. For maximum input voltage see Table 1-1.

Symbols as Marked on Equipment



DANGER — High voltage.



Protective ground (earth) terminal.



ATTENTION — Refer to manual.

Power Source

This product is intended to operate from a power source that will not apply more than 250 volts rms between the supply conductors or between either supply conductor and ground. A protective ground connection by way of the grounding conductor in the power cord is essential for safe operation.

Grounding the Product

This product is grounded through the grounding conductor of the power cord. To avoid electrical shock, plug the power cord into a properly wired receptacle before making any connections to the product input or output terminals. A protective ground connection by way of the grounding conductor in the power cord is essential for safe operation.

Danger Arising from Loss of Ground

Upon loss of the protective-ground connection, all accessible conductive parts (including knobs and controls that may appear to be insulated) can render an electric shock.

Use the Proper Power Cord

Use only the power cord and connector specified for your product.

Use only a power cord that is in good condition.

For detailed information on power cords and connectors, see Table 2-1.

Use the Proper Fuse

To avoid fire hazard, use only a fuse of the correct type, voltage rating and current rating as specified in the parts list for your product.

Do Not Operate in Explosive Atmospheres

To avoid explosion, do not operate this instrument in an explosive atmosphere unless it has been specifically certified for such operation.

Do Not Remove Covers or Panels

To avoid personal injury, do not remove the product covers or panels. Do not operate the instrument unless the covers and panels are properly installed.

SERVICING SAFETY SUMMARY

FOR QUALIFIED SERVICE PERSONNEL ONLY

Refer also to the preceding Operators Safety Summary.

Do Not Service Alone

Do not perform internal service or adjustment of this product unless another person capable of rendering first aid and resuscitation is present.

Use Care When Servicing With Power On

Dangerous voltages exist at several points in this product. To avoid personal injury, do not touch exposed connections or components while power is on.

Disconnect power before removing protective panels, soldering, or replacing components.

Power Source

This product is intended to operate from a power source that does not apply more than 250 volts rms between the supply conductors or between either supply conductor and ground. A protective ground connection by way of the grounding conductor in the power cord is essential for safe operation.

SPECIFICATION

INTRODUCTION

The TEKTRONIX 2445B and 2455B Oscilloscopes are portable 150-MHz and 250-MHz bandwidth instruments having four-channel vertical deflection systems. Channel 1 and Channel 2 provide calibrated deflection factors from 2 mV per division to 5 V per division. For each of these channels, input impedance is selectable between two values: either 1 MΩ in parallel with 15 pF, or 50-Ω internal termination. Input-signal coupling with 1-MΩ impedance can be selected as either AC or DC. Channel 3 and Channel 4 have deflection factors of either 0.1 V or 0.5 V per division. Each of these channels has an input impedance of 1 MΩ in parallel with 15 pF, with DC input-signal coupling.

The trigger system works automatically for most signals. They operate in various modes, from any channel, with couplings for a wide range of signals. The 2445B trigger system gives stable displays from dc to 250 MHz. The 2455B trigger system gives stable displays from dc to 500 MHz.

The horizontal deflection system provides calibrated sweep speeds from 1.5 s per division to 1 ns per division, including the effects of the X10 magnifier and the calibrated variable between the 1-2-5 steps. Horizontal displays include A Sweep, B Sweep (delayed), A alternated with B, and CH 1 (for X/Y displays).

The AUTO, SAVE, and RECALL features save time and prevent errors. Pressing the AUTO Setup button gives a workable setup for almost any signal. For repetitive measurements, the Save and Recall functions record and immediately or sequentially restore as many as 30 instrument setups. The SETUP buttons operate all instrument functions, including the extended function options.

Direct, on-screen readouts of time measurements, voltage measurements, scale factors, trigger levels, and auxiliary information also save time and improve operator confidence.

The instruments are shipped with the following standard accessories:

2 Probe packages
1 Snap-lock accessories pouch
1 Zip-lock accessories pouch
1 Operators manual
1 Power cord (installed)
1 2-A, 250-V fuse
1 Clear plastic CRT filter
1 Blue plastic CRT filter (installed)
1 Front-panel cover
1 Operators pocket reference card

For part numbers and further information about both standard and optional accessories, refer to "Options and Accessories" (Section 7) of the instrument's Operators manual or the Accessories information at the rear of this manual. Your Tektronix representative or local Tektronix Field Office can also provide accessories information and ordering assistance.

PERFORMANCE CONDITIONS

The following electrical characteristics (Tables 1-1 through 1-5) are valid for the instrument when it has been adjusted at an ambient temperature between +20°C and +30°C, has had a warm-up period of at least 20 minutes, and is operating at an ambient temperature between −15°C and +55°C (unless otherwise noted).

Items listed in the "Performance Requirements" column define the measurement capabilities of the instruments. Supplementary measurement conditions may also be listed in the "Performance Requirements" column.

Mechanical characteristics are listed in Table 1-6.

Environmental characteristics are given in Table 1-7. The oscilloscope meets the environmental requirements of MIL-T-28800C for Type III, Class 3, Style C equipment, with the humidity and temperature requirements defined in paragraphs 3.9.2.2, 3.9.2.3, and 3.9.2.4.

Table 1-1
2445B-2455B Electrical Characteristics

Characteristics	Performance Requirements
VERTICAL DEFLECTION SYSTEM—CHANNEL 1 AND CHANNEL 2	
Deflection Factor	
Range	2 mV/division to 5 V/division in a 1-2-5 sequence of 11 steps.
Accuracy	1 MΩ input, noninverted.
+15°C to +35°C	
On-Graticule Accuracy	Within ±2% at any VOLTS/DIV setting for a four or five-division signal centered on the screen.
ΔV Accuracy (using cursors over entire graticule area)	±(1.25% of reading +0.03 div + signal aberrations).
−15°C to +15°C and +35°C to +55°C	Add ±2% of reading.[a]
50 Ω Coupling	Add ±1% of reading.
CH 2 Inverted	Add ±1% of reading.
ΔV Range	±8 X VOLTS/DIV setting.[a]
V/DIV VARiable, noninverted	Continuously variable between VOLTS/DIV settings. Extends deflection factor to >12.5 V/division.
Frequency Response	Bandwidth is measured with a leveled, low distortion, 50-Ω source, sine-wave generator, terminated in 50 Ω. The reference signal amplitude is set at the lesser of 6 divisions or the maximum leveled amplitude. External termination bandwidth is checked with a 4 division reference signal. Bandwidth with probe is checked using a BNC-to-probe-tip (013-0227-00) adapter. Bandwidth with external termination is checked using a BNC 50 Ω feed-through terminator (011-0049-01).
−3 dB Bandwidth	Using standard accessory probe or internal 50-Ω termination.
2455B	
+15°C to +35°C	Dc to 250 MHz.[b]
−15°C to +15°C and +35°C to +55°C	Dc to 200 MHz.[a]
2445B	Dc to 150 MHz.
−4.7 dB Bandwidth	Using 50-Ω external termination on 1-MΩ input.
2455B	
−15°C to +35°C	Dc to 250 MHz.[b]
+35°C to +55°C	Dc to 200 MHz.
2445B	Dc to 150 MHz.

[a]**Performance requirements not checked in manual.**

[b]**If the instrument is subjected to "greater than" 85% relative humidity, bandwidth is reduced by 50 MHz. The instrument then requires more than 50 hours of operation at "less than" 60% relative humidity before full bandwidth is restored.**

Table 1-1 (cont)

Characteristics	Performance Requirements
AC Coupled, Lower −3 dB Frequency	10 Hz or less.
With Standard Accessory Probe	1 Hz or less.[a]
Step Response Rise Time	Calculated from $T_r = 0.35/BW$.[a]
2455B	≤1.4 ns.
2445B	≤2.33 ns.
Channel Isolation	≥100:1 attenuation of deselected channel at 100 MHz; ≥50:1 at 350 MHz, for an eight-division input signal from 5 mV per division to 500 mV per division, with equal VOLTS/DIV settings on both channels.
Displayed Channel 2 Signal Delay with Respect to Channel 1 Signal	Adjustable through a range of at least −500 ps to +500 ps.[a]
Input R and C (1 MΩ)	
Resistance	1 MΩ ±0.5%.[a]
Capacitance	15 pF ±2 pF.[a]
Maximum Input Voltage ⚠	
DC, AC, or GND Coupled	400 V (dc + peak ac). 800 V p-p ac at 10 kHz or less.[a]
Input R (50 Ω)	
Resistance	50 Ω ±1%.[a]
VSWR	≤1:3:1 for dc to Nominal Bandwidth.
Maximum Input Voltage ⚠	5 V rms, averaged for 1 second; ±50 V peak.
Cascaded Operation	Channel 2 Vertical Signal Output into Channel 1 input; DC coupled using a 50 Ω RG-58C/U coaxial cable, with 1 MΩ DC or 1 MΩ AC Channel 1 input coupling; with Channel 1 and Channel 2 VOLTS/DIV set at 2 mV and 20 MHz Bandwidth Limit On.
Deflection Factor	200 μV per division ±10%.
CMRR (ADD Mode with Channel 2 inverted)	At least 20:1 at 50 MHz for common-mode signals of eight divisions or less, with VAR VOLTS/DIV control adjusted for best CMRR at 50 kHz, at any VOLTS/DIV setting.

[a]**Performance requirement not checked in manual.**

Table 1-1 (cont)

Characteristics	Performance Requirements
VERTICAL DEFLECTION SYSTEM—CHANNEL 3 AND CHANNEL 4	
Deflection Factors	
Values	100 mV and 500 mV per division.
Accuracy	Within ±10%.
Frequency Response	Bandwidth is measured with a leveled, low distortion, 50-Ω source, sine-wave generator, terminated in 50 Ω. The reference signal amplitude is set at the lesser of 6 divisions or the maximum leveled amplitude. External termination bandwidth is checked with a 4 division reference signal. Bandwidth with probe is checked using a BNC-to-probe-tip (013-0227-00) adapter. Bandwidth with external termination is checked using a BNC 50-Ω feed-through terminator (011-0049-01).
−3 dB Bandwidth	Using standard accessory probe.
2455B	
+15°C to +35°C	Dc to 250 MHz.[b]
−15°C to +15°C and +35°C to +55°C	Dc to 200 MHz.[a]
2445B	Dc to 150 MHz.
−4.7 dB Bandwidth	Using 50-Ω external termination.
2455B	
+15°C to +35°C	Dc to 250 MHz.[a]
−15°C to +15°C and +35°C to +55°C	Dc to 200 MHz
2445B	Dc to 150 MHz.[a]
Step Response Rise Time	Calculated from $T_r = 0.35/BW$.
2455B	≤1.4 ns.
2445B	≤2.33 ns.
Channel Isolation	≥50:1 attenuation of deselected channel at 100 MHz with an 8-division input signal.
Signal Delay Between Channel 1 and Either Channel 3 or Channel 4	Within ±1.0 ns, measured at the 50% points.[a]
Input Resistance	1 MΩ ±1%.[a]
Input Capacitance	15 pF ±3 pF.[a]
Maximum Input Voltage ⚠	400 V (dc + peak ac). 800 V p-p ac at 10 kHz or less.[a]

[a]Performance requirements not checked in manual.

[b]If the instrument is subjected to "greater than" 85% relative humidity, bandwidth is reduced by 50 MHz. After the instrument is subjected to "greater than" 85% relative humidity, it requires more than 50 hours of operation at "less than " 60% relative humidity before full bandwidth is restored.

Table 1-1 (cont)

Characteristics	Performance Requirements
VERTICAL DEFLECTION SYSTEM—ALL CHANNELS	
Low-frequency Linearity	0.1 division or less compression or expansion of a two-division, center-screen signal when positioned anywhere within graticule area.
Bandwidth Limiter	Reduces upper 3 dB bandpass to a limit of 13 MHz to 24 MHz.
Vertical Signal Delay	At least 30 ns of the sweep is displayed before the triggering event is displayed at any SEC/DIV ≥10 ns/div. At 5 ns/div, at least 10 ns of the sweep is displayed before the triggering event.[a]
Chopped Mode Switching Rate	With displayed SEC/DIV in the 20 μs to 2 μs/div range, the switching rate is 2.5 MHz ±0.2%. Otherwise, the switching rate is 1 MHz ±0.2%. The display cycle rate equals the chop switching rate divided by the number of channels displayed. The chop switching rate is modulated slightly to minimize waveform breaks with repetitive signals.[a]
TRIGGERING	
Minimum P-P Signal Amplitude for Stable Triggering from Channel 1 or Channel 2 Source	
2455B	
DC Coupled	0.35 division from dc to 50 MHz; increasing to 1.0 division at 300 MHz and 1.5 divisions at 500 MHz.
NOISE REJ Coupled	≤1.2 divisions from dc to 50 MHz; increasing to 3 divisions at 300 MHz and 4.5 divisions at 500 MHz.
AC Coupled	0.35 division from 60 Hz to 50 MHz; increasing to 1.0 division at 300 MHz and 1.5 divisions at 500 MHz. Attenuates signals below 60 Hz.
HF REJ Coupled	0.5 division from dc to 30 kHz.
LF REJ Coupled	0.5 division from 80 kHz to 50 MHz; increasing to 1.0 division at 300 MHz and 1.5 divisions at 500 MHz.
2445B	
DC Coupled	0.35 division from dc to 50 MHz; increasing to 1.5 divisions at 250 MHz.
NOISE REJ Coupled	≤1.2 divisions from dc to 50 MHz; increasing to 4.5 divisions at 250 MHz.
AC Coupled	0.35 division from 60 Hz to 50 MHz; increasing to 1.5 divisions at 250 MHz. Attenuates signals below 60 Hz.
HF REJ Coupled	0.5 division from dc to 30 kHz.
LF REJ Coupled	0.50 division from 80 kHz to 50 MHz; increasing to 1.5 divisions at 250 MHz.

[a]Performance requirement not checked in manual.

Table 1-1 (cont)

Characteristics	Performance Requirements
Minimum P-P Signal Amplitude for Stable Triggering from ADD Source	Add 0.5 division to CH 1 or CH 2 requirement at 300 MHz and 500 MHz for 2455B.
Minimum P-P Signal Amplitude for Stable Triggering from CH 3 or CH 4 Source	0.5 X CH 1 or CH 2 requirement.
Minimum P-P Signal Amplitude for Stable Triggering from Composite, Multiple Channel Source, ALT Vertical Mode	Add 1 division to the single-channel source specification. Checked at 50 mV per division.
Maximum P-P Signal Rejected by NOISE REJ COUPLING Signals Within the Vertical Bandwidth	
CH 1 or CH 2 SOURCE	≥0.4 division for VOLTS/DIV settings of 10 mV/div and higher. Maximum noise amplitude rejected is reduced at 2 mV/div and 5 mV/div.
CH 3 or CH 4 SOURCE	≥0.2 division.[a]
Jitter	
2455B	≤50 ps with 5 divisions of 250 MHz at 1 ns/division.
2445B	≤100 ps with 5 divisions of 150 MHz at 1 ns/division.
LEVEL Control Range	
CH 1 or CH 2 SOURCE	±18 × VOLTS/DIV setting.[a]
CH 3 or CH 4 SOURCE	±9 × VOLTS/DIV setting.[a]
LEVEL Readout Accuracy	For triggering signals with transition times greater than 20 ns.
CH 1 or CH 2 SOURCE	
+15°C to +35°C	Within ±[3% of reading + 3% of p-p signal + 0.2 division + 0.5 mV + (0.5 mV × probe attenuation factor)] with Vertical Input at 1 MΩ DC, CH 2 Source Not Inverted, and Trigger DC Coupled.
−15°C to +35°C and +35°C to +55°C	Add 1.5 mV × probe attenuation to +15°C to +35°C specification.[a]
50 Ω Input	Add ±1% to 1 MΩ input specification.[a]
CH 2 Inverted	Add ±1% of reading to non-inverted specification.[a]
NOISE REJ Coupled	Add ±0.6 division to DC Coupled specifications.[a]
CH 3 or CH 4 SOURCE	Within ±[3% of reading + 4% of p-p signal + 0.1 division + (0.5 mV× probe attenuation factor)] and Trigger DC Coupled.
NOISE REJ Coupled	Add ±0.3 division to the DC Coupled specification.

[a]Performance requirement not checked in manual.

Table 1-1 (cont)

Characteristics	Performance Requirements
AUTO LVL Mode Maximum Triggering Signal Period	
A SEC/DIV Setting	
<10 ms	At least 20 ms.[a]
10 ms to 50 ms	At least four times the A-SEC/DIV setting.[a]
>50 ms	At least 200 ms.[a]
AUTO Mode Maximum Triggering Signal Period	
A-SEC/DIV Setting	
<10 ms	At least 80 ms.[a]
10 ms to 50 ms	At least 16 times the A-SEC/DIV setting.
>50 ms	At least 800 ms.
AUTO LVL Mode Trigger Acquisition Time	Eight to 100 times the AUTO LVL Mode maximum triggering signal period, depending on the triggering signal period and waveform.
Trigger Holdoff	
Minimum	The greater of the A-SEC/DIV setting value or 2 μs, within +33% to −10%, except 1 μs at 5 ns/div.[a]
Variable	Increases trigger holdoff time to 10 to 25 times the minimum holdoff.
SLOPE Selection	Conforms to trigger-source waveform or ac power-source waveform.
HORIZONTAL DEFLECTION SYSTEM	
A Sweep Time Base Range	500 ms/div to 10 ns/div in a 1-2-5 sequence of 24 steps. X10 MAG extends maximum sweep rate to 1 ns/div.
B Sweep Time Base Range	50 ms/div to 10 ns/div in a 1-2-5 sequence of 21 steps. X10 MAG extends maximum sweep rate to 1 ns/div.
Timing Accuracy	+15°C to +35°C, A Sweep, with SEC/DIV at 100 ms/div or faster.
Sweep Accuracy Unmagnified	±(0.7% of time interval + 0.6% of full scale).
Δt Accuracy With Cursors, Unmagnified	±(0.5% of time interval + 0.3% of full scale).
Δt Accuracy with Sweep Delay	±(0.3% of time interval + 0.1% of full scale + 200 ps).
Delay Accuracy, A-Sweep Trigger to Start of B Sweep	±(0.3% of delay setting + 0.6% of full scale) +0 to −25 ns.
B-Sweep Accuracy and Δt Accuracy with Cursors on B Sweep	Add ±0.3% of time interval to A-Sweep specifications.

[a]Performance requirement not checked in manual.

Table 1-1 (cont)

Characteristics	Performance Requirements
X10 MAG Accuracy	Add ±0.5% of time interval to unmagnified Sweep and Δt Cursors specifications. Exclude the first 0.5 division after the sweep starts (the first 0.5% of the full 100 division sweep).
500 ms or 200 ms/div Timing Accuracy (A Sweep only)	Add ±0.5% of interval to specifications for A SEC/DIV at 100 ms or faster.
SEC/DIV VAR Timing Accuracy	Add 2% of time interval to sweep accuracy specifications when VAR is out of detent.
Timing Accuracy (−15°C to +15°C and +35°C to +55°C)	Add ±0.2% of time interval to all Δt and delay specifications. Add ±0.5% of interval to sweep accuracy specification.[a]
Δt Readout Resolution	Greater of either 20 ps or 0.25% of full scale.[a]
Δt Range	±10 times A-SEC/DIV setting with Cursors, ±9.95 times A-SEC/DIV setting with Sweep Delay.[a]
Sweep Delay Range	0 to 9.95 times the A SEC/DIV setting, from 500 ms to 20 ns. A-Sweep triggering event is observable on B Sweep with zero delay setting for A SEC/DIV settings 10 μs or faster.[a]
Delay Jitter	Within 0.004% (one part or less in 25,000) of the maximum available delay, plus 50 ps.[a]
Horizontal POSITION Range	Start of 1 ms per division sweep can be positioned from right of graticule center to at least 10 divisions left of graticule center. Some portion of 1 ms per division sweep is always visible with X10 MAG off.[a]
X-Y Operation	
X-Axis Deflection Factor Range, Variable, and Input Characteristics	Same as Channel 1.[a]
Deflection Factor Accuracy	Same as Channel 1.
X-Axis Bandwidth	Dc to 3 MHz.
Phase Difference Between X and Y with BW Limit Off	≤1° from dc to 1 MHz; ≤3° from 1 MHz to 2 MHz.
X-Axis Low-frequency Linearity	0.1 division or less compression or expansion of a two-division, center-screen signal when positioned within the graticule area.

[a]Performance requirement not checked in manual.

Table 1-1 (cont)

Characteristics	Performance Requirements
DISPLAY	
Cursor Position Range	
Delta Volts (ΔV)	At least the center 7.6 vertical divisions.
Delta Time (Δt)	At least the center 9.6 horizontal divisions.
Graticule	
Size	80 mm X 100 mm.[a]
Markings	8 major divisions vertically and 10 major divisions horizontally, with auxiliary markings.[a]
Trace Rotation Range	Adequate to align trace with the center horizontal graticule line.
Z-AXIS INPUT	
Sensitivity	
Dc to 2 MHz	Positive voltage decreases intensity; +2 V blanks a maximum intensity trace.
2 MHz to 20 MHz	+2 V modulates a normal intensity trace.[a]
Input Resistance	10 kΩ ±10%.[a]
Maximum Input Voltage ⚠	±25 V peak; 25 V p-p ac at 10 kHz or less.[a]
SIGNAL OUTPUTS	
CALIBRATOR	With A SEC/DIV set to 1 ms.
Output Voltage and Current	0.4 V ±1% into a 1-MΩ load, 0.2 V ±1.5% into a 50-Ω load, or 8 mA ±1.5% into a short circuit.[a]
Repetition Period	Two times the A SEC/DIV setting for SEC/DIV from 100 ns to 100 ms.
Accuracy	±0.1% during sweep time.
CH 2 SIGNAL OUT	
Output Voltage	20 mV/division ±10% into 1 MΩ; 10 mV/division ±10% into 50 Ω.
Offset	±20 mV into 1 MΩ when dc balance has been performed within ±5°C of the operating temperature.
A GATE OUT and B GATE OUT	
Output Voltage	2.4 V to 5 V positive-going pulse, starting at 0 V to 400 mV.
Output Drive	Will supply 400 μA during HI state; will sink 2 mA during LO state.[a]

[a]Performance requirement not checked in manual.

Table 1-1 (cont)

Characteristics	Performance Requirements
AC POWER SOURCE	
Source Voltage Nominal Ranges 115 V	90 V to 132 V.
230 V	180 V to 250 V.
Source Frequency	48 Hz to 440 Hz.[a]
Fuse Rating	2 A, 250 V, AGC/3AG, Fast blow; or 1.6 A, 250 V, 5 X 20 mm Quick-acting.[a]
Maximum Power Consumption (fully optioned instrument)	120 watts (180 VA).[a]
Primary Circuit Dielectric Voltage Withstand Test	1500 V rms, 60 Hz for 10 seconds without breakdown.[a]
Primary Grounding	Type test to 0.1 Ω maximum. Routine test to check grounding continuity between chassis ground and protective earth ground.

[a]Performance requirement not checked in manual.

Table 1-1 (cont)

Characteristics	Performance Requirements
PARAMETRIC MEASUREMENTS	
Period	
Accuracy	
+15°C to +35°C	0.9% + 0.5 ns + Jitter Error. (500ps (<50ps)
-15 to +15°C and +35°C to +55°C	Add 0.3%.
Minimum Period	⩽ 2 ns.
Maximum Period	⩾100 ms (MINFREQ=10Hz).
Minimum Signal Amplitude	⩽ (60 mV + probe attenuation factor p-p). If DC coupling is used, the DC offset voltage must meet the following criteria: at a VOLTS/DIV setting which gives a p-p signal ⩾ 4 divisions, the peak signal + offset must be ,= 12 divisions.
Frequency	Calculated as 1/period.
Volts	
+Peak, −Peak, Peak-to-Peak, and Average	
Accuracy	
+15°C to +35°C	5% of reading + 5 mV + (0.5 mV * probe attenuation) + signal aberrations + 1 Least Significant Digit (LSD).
−15°C to +15°C and +35°C to +55°C	Add (1.5 mV * probe attenuation).
Minimum Width at Peak Amplitude	⩽ 10 ns.
Maximum Sine Wave Frequency	
+15°C to +35°C	⩾ 1 MHz.
−15°C to +15°C and +35°C to +55°C	Add 2%. Volts measurements depend on peak signal measurements. Noise on the input signal, even if at a low repetition rate that makes it difficult to see, will be detected and will affect the measurements.
Pulse Width (High or Low)	
Accuracy	
+15°C to +35°C	0.9% of reading + 1.0 ns + jitter error + 2 * offset error.
−15°C to +15°C and 35°C to +55°C	Add 0.3%.
Minimum Pulse Width	⩽ 5 ns.
Minimum Repetition Rate	⩽ 10 Hz (with MINFREQ = 10 Hz).

Table 1-1 (cont)

Characteristics	Performance Requirements
Duty Cycle	Calculated from Pulse Width and Period.
Rise Time, Fall Time, and Time Interval	
Accuracy	
+15°C to +35°C	5% of reading + 3.0 ns + jitter error + offset error.
Rise/Fall Time	Add 0.5 ns if measurement is made between CH1 and CH2.
Time Interval	0.5 % of reading + 5% of start event transition time + 5% of stop event transition time + 3.0 ns + jitter error + offset error.
	Rise and Fall time measurement is made at 20% and 80% points of transition and linearly extrapolated to the 10% and 90% points.
	Accuracy is relative to time interval as measured on screen using cursors. Measurement is made using peak-to-peak transition for measurement points in percent.
−15 to +15°C and +35°C to +55°C	Add 2%.
Minimum Time	⩽ 5 ns.
Minimum Repetition Rate	⩽ 10 Hz (with MINFREQ = 10 Hz).
Jitter Error	Noise on the input signal causes jitter which introduces errors in the measurements. The amount of jitter depends on the noise amplitude and the slew rate of the input signals.
	The amount of jitter can be calculated as:
	$\text{jitter} = \frac{\text{input noise amplitude (peak)}}{\text{input slew rate in div/sec}}$
	Input slew rate should be measured at 2 Volts/div settings more sensitive than the setting at the end of the measurements or at 5 mV/div, whichever is less sensitive.
	The slew rate must be measured at the same points at which the measurement will be taken. The points for the various measurements are:

	Measurement Points	
Measurement	**First Measurement point**	**Second Measurement point**
Frequency	50% amplitude	50% amplitude
Width	50% amplitude	50% amplitude
Rise, Fall Time	10% amplitude	90% amplitude
Time interval	Specified by Time Interval Configuration	Specified by Time Interval Configuration

Table 1-1 (cont)

Characteristics	Performance Requirements
	The algorithms used for the measurements result in the following equation for the total jitter error that must be applied to the accuracy specifications. Jitter Error = 2 * first point jitter + 2 * second point jitter.
Offset Error	Offset error is introduced when the trigger level is not set exactly at the expected points. This misplacement of the trigger level applied to any non-infinite slew rate produces a timing error. The magnitude of the error is given by: $\text{Offset Error} = \frac{\text{offset}}{\text{input slew rate}}$ Frequency measurements do not suffer from offset errors since measurements are made with the same trigger level and slope, so no offset is introduced. All other timing measurements suffer from offset errors. The slew rates used to calculate offset errors must be measured at the first and second measurement points given in the Measurement Points table. Offset error is calculated as: $\text{Offset Error} = \frac{0.2\ \text{div}}{\text{First Point slew rate}} + \frac{0.2\ \text{div}}{\text{Second Point slew rate}}$ If a time interval measurement is made using Volts mode, the offset at each measurement point is: 0.2 div + 5% of measurement point voltage converted to divisions.

Table 1-2
Option 06 (C/T/T) Electrical Characteristics

Characteristics	Performance Requirements
SIGNAL INPUT	
	With DC Coupling of A Trigger and B Trigger.
Maximum Input Frequency for Count and Delay by Events	⩾150 MHz.
Minimum Width of High or Low State of Input Signal for Count and Delay by Events	⩽3.3 ns.
Sensitivity	For Count, Delay by Events, and Logic Trigger Functions Excluding Word Recognizer.
Dc to 50 MHz (0.5 Hz to 50 MHz for Frequency and Period)	
CH 1 and CH 2	1.5 divisions.
CH 3 and CH 4	0.75 division.
50 MHz to 150 MHz	
CH 1 and CH 2	4.0 divisions.
CH 3 and CH 4	2.0 divisions.
FREQUENCY	
Ranges	**RANGE** — **LSD[a]** 1 Hz — 100 nHz 10 Hz — 1 μHz 100 Hz — 10 μHz 1 kHz — 100 μHz 10 kHz — 1 mHz 100 kHz — 10 mHz 1 MHz — 100 mHz 10 MHz — 1 Hz 100 MHz — 10 Hz 150 MHz — 100 Hz
Automatic Ranging	Upranges at 100% of full scale; downranges at 9% of full scale. Downrange occurs at 90 MHz on 150 MHz range.
	Full scale corresponds to the value given in the Range column. The maximum displayed value for any range is the Range value minus the LSD value.
Accuracy	±[Resolution + (Frequency × TBE)] Hz.
Time Base Error (TBE)	10 ppm with less than 5 ppm per year drift.
Resolution	$\frac{1.4 \times \text{Frequency}^2 \times \text{TJE}}{N}$ + LSD.
Display Update Rate	Twice per second or twice the period of the input signal, whichever is slower.

[a]**Performance requirement not checked in manual.**

Table 1-2 (cont)

Characteristics	Performance Requirements	
PERIOD		
Ranges	**RANGE**	**LSD[a]**
	10 ns	1 fs
	100 ns	10 fs
	1 μs	100 fs
	10 μs	1 ps
	100 μs	10 ps
	1 ms	100 ps
	10 ms	1 ns
	100 ms	10 ns
	1 s	100 ns
	2 s	1 μs
Minimum Period	≤6.7 ns.	
Automatic Ranging	Upranges at 100% of full scale; downranges at 9% of full scale. Full scale corresponds to the value given in the Range column. The maximum displayed value for any range is the Range value minus the LSD value.	
Accuracy	±[Resolution + (TBE × Period)].	
Resolution	±[LSD + (1.4 × TJE)/N].	
Display Update Rate	Twice per second or twice the period of the input signal, whichever is slower.	
TOTALIZE		
Maximum Count	9999999.	
Display Update Rate	Twice per second or once per event, whichever is slower.	
DELAY BY EVENTS		
Maximum Event Count	4194303.	
Minimum Time from Start Signal to Any Delay Event	4 ns.	
LOGIC TRIGGER		
Minimum Function-True Time	4 ns.	
Minimum Function-False Time	4 ns.	

[a]Performance requirement not checked in manual.

Table 1-2 (cont)

Characteristics	Performance Requirements
ADDED DELAY TIME CHARACTERISTICS WITH C/T/T	
Run After Delay	
Accuracy	LSD[b] + [0.0012 × (A SEC/DIV)] + [0.03 × (B Time/Div)[c] + A Trigger Level Error + 50 ns.[a] When the A Sweep is triggered by the Word Recognizer in synchronous mode, add 100 ns for probe delay; in asynchronous mode, add 200 ns for probe delay.
Triggerable After Delay	
Accuracy	For intervals within 70 ns to 10 times the A-SEC/DIV Setting. LSD[b] + [10 ppm × (measured interval)] + TJE + A-Trigger Level Error + B-Trigger Level Error + 0.5 ns.[a] If the A and B Sweeps are triggered from different channels, add 0.5 ns for channel-to-channel mismatch. When the A Sweep is triggered by the Word Recognizer in synchronous mode, add 100 ns for probe delay; in asynchronous mode, add 200 ns for probe delay.
Minimum Measurable Delay Time	≤70 ns.[a]
Display Update Rate	In Auto Resolution, twice per second or once for every sweep, whichever is slower.[a] In 1 ns, 100 ps, and 10 ps resolution modes, the update rate depends on the A SEC/DIV setting and the trigger repetition rate.

[a]Performance requirement not checked in manual.

[b]See Tables 1-3 and 1-4.

[c]B Time/Div includes SEC/DIV, X10 MAG, and VAR.

Table 1-2 (cont)

Characteristics	Performance Requirements
ADDED DELTA-DELAY-TIME CHARACTERISTICS WITH C/T/T	
Run After Delay	
Accuracy	$LSD^b + [0.0008 \times (A\ SEC/DIV)] + [0.01 \times (B\ Time/Div)^c] +$ 83 ps.[a] When the A Sweep is triggered by the Word Recognizer in synchronous mode, add 1 ns for probe jitter; in asynchronous mode, add 20 ns for probe jitter.
Triggerable After Delay	
Accuracy	Both delays are within 70 ns to 10 times the A-SEC/DIV setting.
Superimposed Delta Time	$LSD^b + [0.01 \times (B\ Time/Div)^c] + [10\ ppm \times (A\ SEC/DIV)] + [10\ ppm \times (measured\ interval)] + 50\ ps + TJE.$[a] If CH 3 or CH 4 is one channel of a two-channel measurement, add 0.5 ns for channel-to-channel delay mismatch.
Nonsuperimposed Delta Time	$LSD^b + \lvert t_{r_{REF}} - t_{r_{DELT}} \rvert^d + TJE +$ $[(0.0005\ div) \times (1/SR_{REF} + 1/SR_{DELT})] +$ $[10\ ppm \times (A\ SEC/DIV)] + [10\ ppm \times (measured\ interval)] + 50\ ps.$ If A and B sweeps are triggered from different channels, add 0.5 ns for channel-to-channel mismatch + $[0.5\ div \times (1/SR_{REF} + 1/SR_{DELT})]$ for trigger offset.
Display Update Rate	In Auto Resolution, twice per second or once for every four sweeps, whichever is slower.[a] In 1 ns, 100 ps, and 10 ps resolution modes, the update rate depends on the A SEC/DIV setting and the trigger repetition rate.

[a]Performance requirement not checked in manual.

[b]See Tables 1-3 and 1-4.

[c]B Time/Div includes SEC/DIV, X10 MAG, and VAR.

[d]This term assumes the trigger points are between the 10% and 90% points of the waveforms. Fall time is expressed as a negative risetime.

Table 1-2 (cont)

Characteristics	Performance Requirements

DEFINITIONS

A Trigger Level Error = (A Trigger Level Readout Error)/SR_A.

B Trigger Level Error = (B Trigger Level Readout Error)/SR_B.

$t_{r_{REF}}$ = rise time, reference trigger signal.

$t_{r_{DELT}}$ = rise time, delta trigger signal.

SR_A = slew rate at trigger point, A Sweep trigger signal in div/sec.

SR_B = slew rate at trigger point, B Sweep trigger signal in div/sec.

SR_{REF} = slew rate at trigger point, reference trigger signal in div/sec.

SR_{DELT} = slew rate at trigger point, delta trigger signal in div/sec.

TJE = trigger jitter error.

For delay or delta time, disregarding noise in the signal, this term contributes <1 LSD if the slew rate is greater than 0.03 vertical div/ns or if the slew rate is greater than 30000 vertical div/horizontal div.

Trigger Jitter = [(Reference Trigger Signal Jitter)^2 + (Delta Trigger Signal Jitter)^2 + (A Sweep Trigger Signal Jitter)^2]$^{1/2}$.

Reference Trigger Signal Jitter = $(e_{n_S} + e_{n_{REF}})/SR_{REF}$.

= 0 for Frequency mode.

e_{n_S} = scope noise in div.

= 0.05 div for HF REJ trigger coupling.

= 0.1 div for DC trigger coupling, 5 mV to 5 V sensitivity.

= 0.15 div for DC trigger coupling, 2 mV sensitivity.

$e_{n_{REF}}$ = reference signal rms noise in div.

Delta Trigger Signal Jitter = $(e_{n_S} + e_{n_{DELT}})/SR_{DELT}$.

= 0 for Frequency or Delay mode.

$e_{n_{DELT}}$ = delta signal rms noise in div.

A Trigger Signal Sweep Jitter = $(e_{n_S} + e_{n_A})/SR_A$.

e_{n_A} = A sweep trigger signal rms noise in div.

When the Word Recognizer supplies a trigger in synchronous mode, the trigger jitter of the associated trigger signal is <1 ns; in asynchronous mode, the associated trigger signal jitter is <20 ns.

N = number of averages during measurement interval.

= see Table 1-3 for Delay or Delta Time.

= (measured frequency) × (measurement interval) for Frequency or Period.

Measurement Interval = 0.5 s or two periods of measured signal, whichever is greater.

Table 1-3
Resolution Selections

A SEC/DIV	Selection	Least Digit	N for Average
10 ns to 500 ms	AUTO	See Table 1-4	See Table 1-4
10 ns to 5 μs	10 ps	10 ps	$> 10^6$
	100 ps	100 ps	$> 10^4$
	1 ns	1 ns	> 100
10 μs to 50 μs	10 ps or 100 ps	100 ps	$> 10^4$
	1 ns	1 ns	> 100
100 μs to 500 μs	10 ps to 1 ns	1 ns	> 100
1 ms to 5 ms	Any	10 ns	> 1
10 ms to 50 ms	Any	100 ns	> 1
100 ms to 500 ms	Any	1 μs	> 1

Table 1-4
Resolution Selections

A SEC/DIV	Trigger Rate	Least Digit	N for Average
10 ns to 2 μs	> 20 kHz	100 ps	$> 10^4$
10 ns to 2 μs	200 Hz to 20 kHz	1 ns	> 100
5 μs to 200 μs	> 200 Hz	1 ns	> 100
10 ns to 200 μs	< 200 Hz	10 ns	> 1
500 μs to 5 ms	Any	10 ns	> 1
10 ms to 50 ms	Any	100 ns	> 1
100 ms to 500 ms	Any	1 μs	> 1

Table 1-5
Option 09 (WR) Electrical Characteristics

Characteristics	Performance Requirements
SYNCHRONOUS MODE	
Data Setup Time D_0—D_{15} and Q	25 ns.
Data Hold Time D_0—D_{15} and Q	0 ns.
Minimum Clock Pulse Width	
High	20 ns.
Low	20 ns.
Minimum Clock Period	50 ns.[a]
Delay from Selected Clock Edge to Word Out from C/T/T	≤55 ns.
ASYNCHRONOUS MODE	
Maximum Trigger Frequency	10 MHz.[a]
Minimum Coincidence Between Data Inputs (D_0—D_{15} and Q) Resulting in a Trigger	<85 ns.
Maximum Coincidence Between Data Inputs (D_0—D_{15} and Q) Without Producing a Trigger	>20 ns.
Delay from Input Word Coincidence to Word Out	≤140 ns.
INPUTS AND OUTPUTS	
Input Voltages	
Minimum Input Voltage ⚠	−0.5 V.[a]
Maximum Input Voltage ⚠	5.5 V.[a]
Maximum Input Low Voltage	0.6 V.[a]
Minimum Input High Voltage	2.0 V.[a]
WORD RECOG OUT	
High	> 2.5 V LSTTL output.[a]
Low	< 0.5 V LSTTL output.[a]
Input High Current	≤20 μA.[a]
Input Low Current	≥−0.6 mA source.[a]

[a]**Performance requirement not checked in manual.**

Table 1-6
Mechanical Characteristics

Characteristics	Performance Requirements
Weight	
With Accessories and Pouch	10.2 kg (22.4 lb).
With Option 06 and 09	12.0 kg (26.44 lb).
Without Accessories and Pouch	9.3 kg (20.5 lb).
Domestic Shipping Weight	12.8 kg (28.2 lb).
With Option 06 and 09	17.6 kg (38.8 lb).
Height	
Without Accessories Pouch, with or without Options 06 and 09	160 mm (6.29 in).
With Feet and Accessories Pouch, with or without Options 06 and 09	243 mm ±25.4 mm (9.56 in ±1.0 in).
Width (with handle)	338 mm (13.31 in).
Depth	
With Front Panel Cover	434 mm (17.1 in).
With Handle Extended	508 mm (20.0 in).
Cooling	Forced-air circulation.
Finish	Tek Blue vinyl clad material on aluminum cabinet.
Construction	Aluminum-alloy chassis (sheet metal). Plastic-laminate front panel. Glass-laminate circuit boards.

Table 1-7
Environmental Requirements

Characteristics	Performance Requirements
	Environmmental requirements qualify the electrical and mechanical specifications. When not rack mounted, the instruments meet the environmental requirements of MIL-T-28800C for Type III, Class 3, Style C equipment, with the humidity and temperature requirements defined in paragraphs 3.9.2.2, 3.9.2.3, and 3.9.2.4. Rack mounting changes the temperature, vibration, and shock capabilities. The rack mounted instruments meet or exceed the requirements of MIL-T-28800C with respect to Type III, Class 5, Style C equipment with the rack-mounting rear-support kit installed. Rack mounted instruments will be capable of meeting or exceeding the requirements of Tektronix Standard 062-2853-00, class 5.
Temperature	
Operating	−15°C to +55°C.
	For a rack mounted instrument, ambient temperature should be measured at the instrument's air inlet. Fan exhaust temperature should not exceed +65°C.
Nonoperating (Storage)	−62°C to +85°C.
Altitude	
Operating	To 15,000 feet. Maximum operating temperature decreases 1°C for each 1000 feet above 5000 feet.
Nonoperating (Storage)	To 50,000 feet.
Humidity	
Operating and Storage	Stored at 95% relative humidity for five cycles (120 hours) from 30°C to 60°C, with operational performance checks at 30°C and 55°C.
Vibration (operating)	
Not Rack Mounted	15 minutes along each of three axes at a total displacement of 0.025 inch p-p (4 g at 55 Hz), with frequency varied from 10 Hz to 55 Hz in one minute sweeps. Hold 10 minutes at each major resonance or, if none exists, hold 10 minutes at 55 Hz (75 minutes total test time).
Rack Mounted	Change displacement to 0.015 inch p-p (2.3 g at 55 Hz).
Shock (operating and nonoperating)	
Not Rack Mounted	50 g, half sine, 11 ms duration, three shocks on each face, for a total of 18 shocks.
Rack Mounted	30 g.
Transit Drop (not in shipping package)	8-inch drop on each corner and each face (MIL-T-28800C, para. 4.5.5.4.3).

Table 1-7 (cont)

Characteristics	Performance Requirements
Bench Handling (cabinet on and cabinet off)	MIL-STD-810C, Method 516.2, Procedure V (MIL-T-28800C, para. 4.5.5.4.3).
Topple (operating with cabinet installed)	Set on rear feet and allow to topple over onto each of four adjacent faces (Tektronix Standard 062-2858-00).
Packaged Transportation Drop	Meets the limits of the National Safe Transit Assn., test procedure 1A-B-2; 10 drops of 36 inches (Tektronix Standard 062-2858-00).
Packaged Transportation (Vibration)	Meets the limits of the National Safe Transit Assn., test procedure 1A-B-1; excursion of 1 inch p-p at 4.63 Hz (1.1 g) for 30 minutes (Tektronix Standard 062-2858-00).
EMI (Electro-magnetic Interference)	Meets MIL-T-28800C; MIL-STD-461B, part 4 (CE-03 and CS-02), part 5 (CS-06 and RS-02), and part 7 (CS-01, RE-02, and RS-03)—limited to 1 GHz; VDE 0871, Category B; Part 15 of FCC Rules and Regulations, Subpart J, Class A; and Tektronix Standard 062-2866-00.
Electrostatic Discharge Susceptibility	Meets Tektronix Standard 062-2862-00. The instrument will not change control states with discharges of less than 10 kV.
X-Ray Radiation	Meets requirements of Tektronix Standard 062-1860-00.

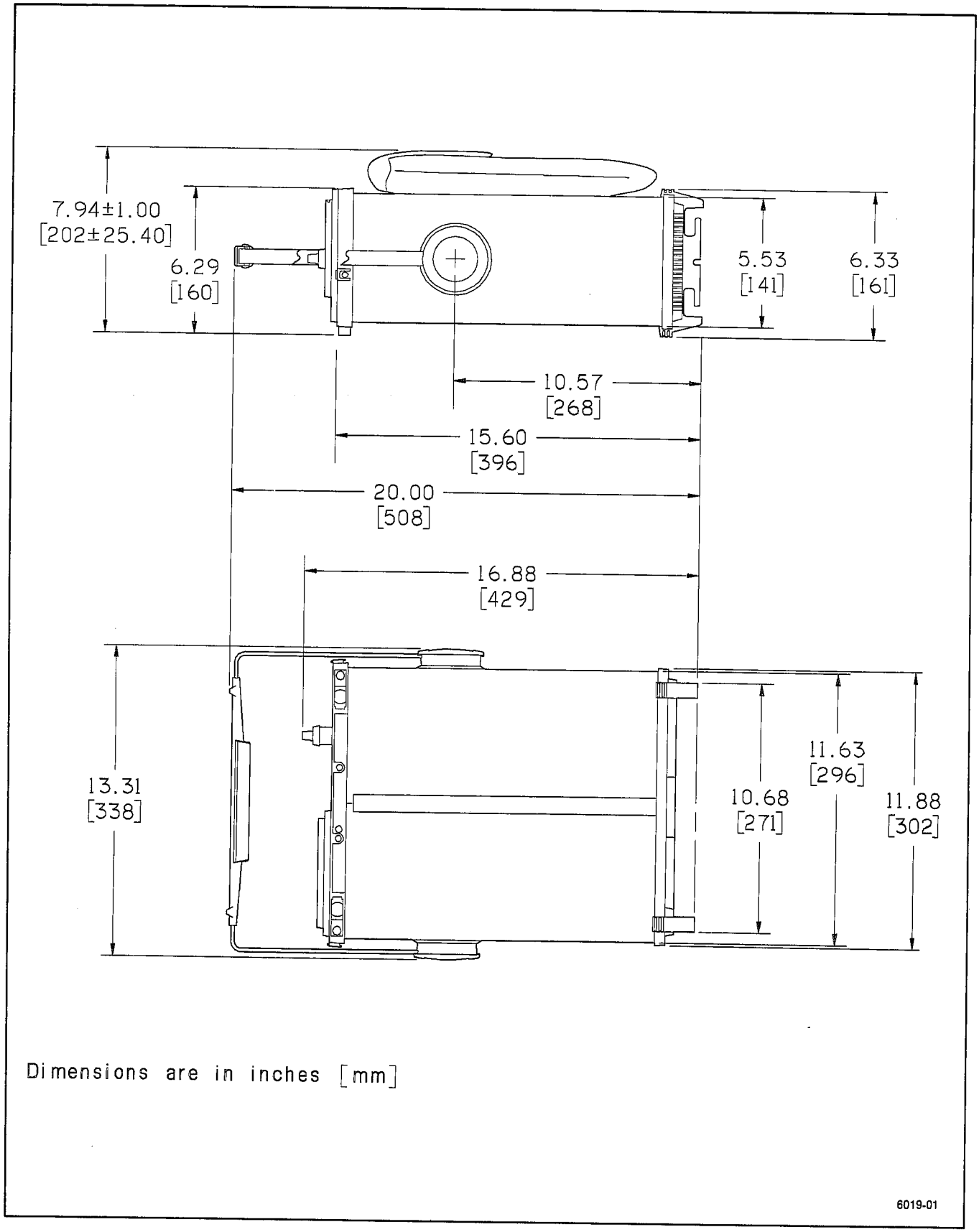


Figure 1-1. Dimensional drawing.

OPERATING INFORMATION

SAFETY

Before connecting the oscilloscope to a power source, read entirely both this section and the Safety Summary at the front of this manual. Be sure you have the training required to safely connect the instrument inputs to the signals you will be measuring. Refer to the Safety Summary for power source, grounding, and other safety considerations pertaining to the use of the instrument.

CAUTION

This instrument may be damaged if operated with the LINE VOLTAGE SELECTOR switch set for the wrong applied ac input-source voltage or if the wrong line fuse is installed.

LINE VOLTAGE SELECTON

The oscilloscope operates from either a 115-V or a 230-V nominal ac power-line with any frequency from 48 Hz to 440 Hz. Before connecting the power cord to a power source, verify that the LINE VOLTAGE SELECTOR switch, located on the rear panel (see Figure 2-1), is set correctly (see Table 2-1) and that the line fuse is correct. To convert the instrument for operation on the other line-voltage range, move the LINE VOLTAGE SELECTOR switch to the correct nominal ac source-voltage setting. The detachable power cord may have to be replaced to match the particular power source.

LINE FUSE

To verify the instrument power-input fuse rating, do the following steps:

1. Press in the fuse-holder cap and release it with a slight counterclockwise rotation. Pull the cap (with the attached fuse inside) out of the fuse holder.
2. Verify that the fuse is of the type listed on the back of the instrument. Then install the proper fuse and reinstall the proper fuse-holder cap. The two types of fuses listed are not directly interchangeable; they require different types of fuse caps. Included in the accessory pouch is a 5x20 mm fuse holder cap for use with 1.6 A, 250 V, 5x20 mm (IEC 127) fuses.

POWER CORD

This instrument has a detachable, three-wire power cord with a three-contact plug for connection to both the power source and protective ground. The power cord is secured to the rear panel by a cord-set-securing clamp. The protective-ground contact on the plug connects through the power-cord to the external metal parts of the instrument. For electrical-shock protection, insert this plug into a power-source outlet that has a properly grounded protective-ground contact.

Instruments are shipped with the required power cord as ordered by the customer. Available power-cord information is presented in Table 2-1, and part numbers are listed in "Options and Accessories" (Section 7). Contact your Tektronix representative or local Tektronix Field Office for additional power-cord information.

INSTRUMENT COOLING

To prevent instrument damage from internally generated heat, adequate air flow must be maintained. Before turning on the power, verify that the spaces around the air-intake holes on the bottom of the cabinet and the fan-exhaust holes in the rear panel are free of any obstruction to airflow.

OPERATING INFORMATION

All operating information pertaining to the use of these instruments is found in the respective instrument Operators Manual.

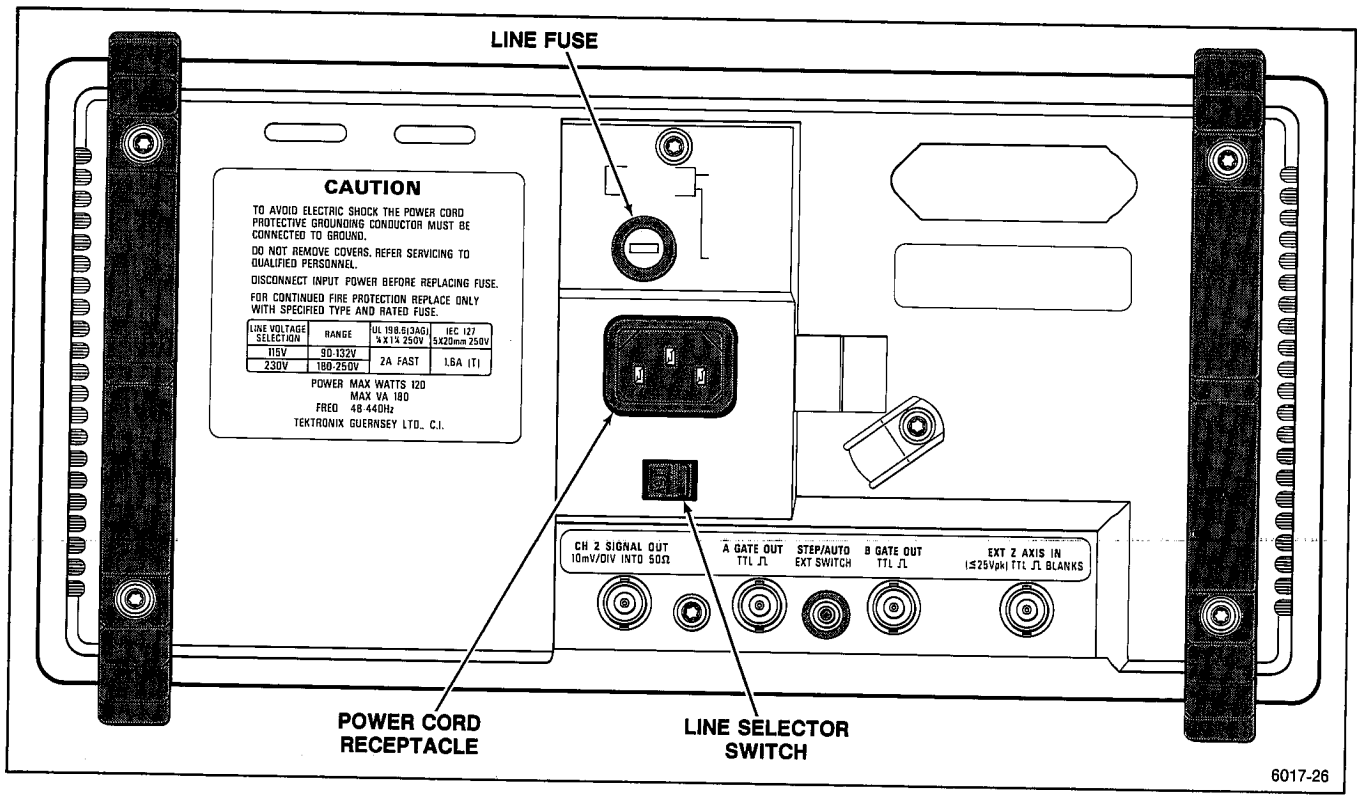


Figure 2-1. Line selector switch, line fuse, and detachable power cord.

START-UP

The oscilloscope automatically performs a set of diagnostic tests each time it is turned on. These tests warn the user of any available indication that the instrument may not be fully functional. The tests run for several seconds after power is applied. If no faults are encountered, the instrument operates normally. A failure of any of the power-up tests will be indicated by either a flashing TRIG'D indicator on the instrument front panel or a bottom-line readout on the CRT in the form: **TEST XX FAIL YY** (where XX is the test number and YY is the failure code of the failed test).

If a failure of any power-up test occurs, the instrument may still be usable for some applications. To operate the instrument after a power-up test failure, press the A/B TRIG button. Even if the instrument then functions for your particular measurement requirement, it should be repaired by a qualified service technician at the earliest convenience. Additional information on the power-up tests and troubleshooting may be found in the "Maintenance" section of this manual. Consult your service department, your local Tektronix Service Center, or nearest Tektronix representative if additional assistance is needed.

REPACKAGING FOR SHIPMENT

If this instrument is to be shipped by commercial transportation, it should be packaged in the original manner. The carton and packaging material in which your instrument was shipped to you should be retained for this purpose.

If the original packaging is unfit for use or is not available, repackage the instrument as follows:

1. Obtain a corrugated cardboard shipping carton having inside dimensions at least six inches greater than the instrument dimensions and having a carton test strength of at least 275 pounds.

2. If the instrument is to be shipped to a Tektronix Service Center for service or repair, attach a tag to the instrument showing the following: owner of the instrument (with address), the name of a person at your firm who can be contacted, complete instrument type and serial number, and a description of the service required.

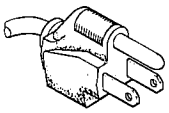
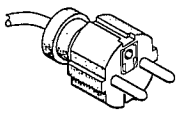
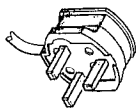
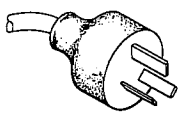
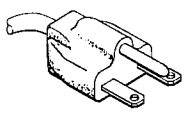
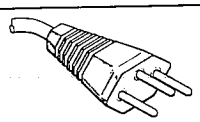
3. Wrap the instrument with polyethylene sheeting or equivalent to protect the outside finish and prevent entry of packing materials into the instrument.

4. Cushion the instrument on all sides by tightly packing dunnage or urethane foam between the carton and the instrument, allowing three inches on each side.

5. Seal the carton with shipping tape or with an industrial stapler.

6. Mark the address of the Tektronix Service Center and your return address on the carton in one or more prominent locations.

Table 2-1
Power Cord and Voltage Data

Plug Configuration	Option	Power Cord/ Plug Type	Line Voltage Selector	Reference Standards[b]
	U.S. Std.	U.S. 120V	115V	ANSI C73.11 NEMA 5-15-P IEC 83 UL 198.6
	A1	EURO 220V	230V	CEE(7), II, IV, VII IEC 83 IEC 127
	A2	UK[a] 240V	230V	BS 1363 IEC 83 IEC 127
	A3	Australian 240V	230V	AS C112 IEC 127
	A4	North American 240V	230V	ANSI C73.20 NEMA 6-15-P IEC 83 UL 198.6
	A5	Switzerland 220V	230V	SEV IEC 127

[a]A 6A, type C fuse is also installed inside the plug of the Option A2 power cord.

[b]Reference Standards Abbreviations:

ANSI—American National Standards Institute
AS—Standards Association of Australia
BS—British Standards Institution
CEE—International Commission on Rules for the Approval of Electrical Equipment
IEC—International Electrotechnical Commission
NEMA—National Electrical Manufacturer's Association
SEV—Schweizervischer Elektrotechnischer Verein
UL—Underwriters Laboratories Inc.

THEORY OF OPERATION (SN B050000 & ABOVE)

INTRODUCTION

SECTION ORGANIZATION

This section contains a functional description of the instrument circuitry. The discussion begins with an overview of the instrument functions and continues with detailed explanations of each major circuit. Reference is made to supporting schematic and block diagrams which will facilitate understanding of the text. These diagrams show interconnections between parts of the circuitry, identify circuit components, list specific component values, and indicate interrelationships with front-panel controls.

The detailed block diagram and the schematic diagrams are located in the tabbed "Diagrams" section at the rear of this manual, while smaller functional diagrams are contained within this section near their respective text. The particular schematic diagram associated with each circuit description is identified in the text, and the diagram number is shown (enclosed within a diamond symbol) on the tab of the appropriate foldout page. For optimum understanding of the circuit being described, refer to both the applicable schematic diagram and the functional block diagram.

HYBRID AND INTEGRATED CIRCUIT DESCRIPTIONS

Digital Logic Conventions

Digital logic circuits perform many functions within this instrument. The operation of these circuits is represented by specific logic symbology and terminology. Most logic-function descriptions contained in this manual use the positive-logic convention. Positive logic is a system of notation whereby the more positive of two levels is the TRUE (or 1) state; the more negative level is the FALSE (or 0) state. In the logic descriptions, the TRUE state is referred to as HI, and the FALSE state is referred to as LO. The specific voltages which constitute a HI or a LO state vary between individual devices. For specific device characteristics, refer to the manufacturer's data book.

Hybrids

Some of the circuits in this instrument are implemented in hybrid devices. The hybrids are specialized electronic devices combining thick-film and semiconductor technologies. Passive, thick-film components and active, semiconductor components are interconnected to form the circuit on a ceramic carrier. The end result is a relatively small "building block" with enhanced performance characteristics, all in one package. Hybrid circuits are shown on schematics simply as blocks with inputs and outputs. Information about hybrid functioning is contained in the related portion of the Detailed Circuit Description.

Linear Devices

The operation of individual linear integrated circuit devices is described in this section using waveforms or other graphic techniques to illustrate their operation.

BLOCK DIAGRAM

The following discussion is provided to aid in understanding the overall operation of the instrument circuitry before the individual circuits are discussed in detail. A simplified block diagram of the instrument, showing basic interconnections, is shown in Figure 3-1. The diamond-enclosed numbers in each block refer to the schematic diagram(s) at the rear of this manual in which the related circuitry is located.

BLOCK DESCRIPTION

The Low Voltage Power Supply is a high-efficiency, switching supply with active output regulation that transforms the ac source voltage to the various dc voltages required by the instrument. The High Voltage Power Supply circuit develops the high accelerating potentials required by the crt, using voltage multiplication techniques, and the DC Restorer provides interfacing for the low-potential intensity signals from the Z-Axis Amplifier to the crt control grid.

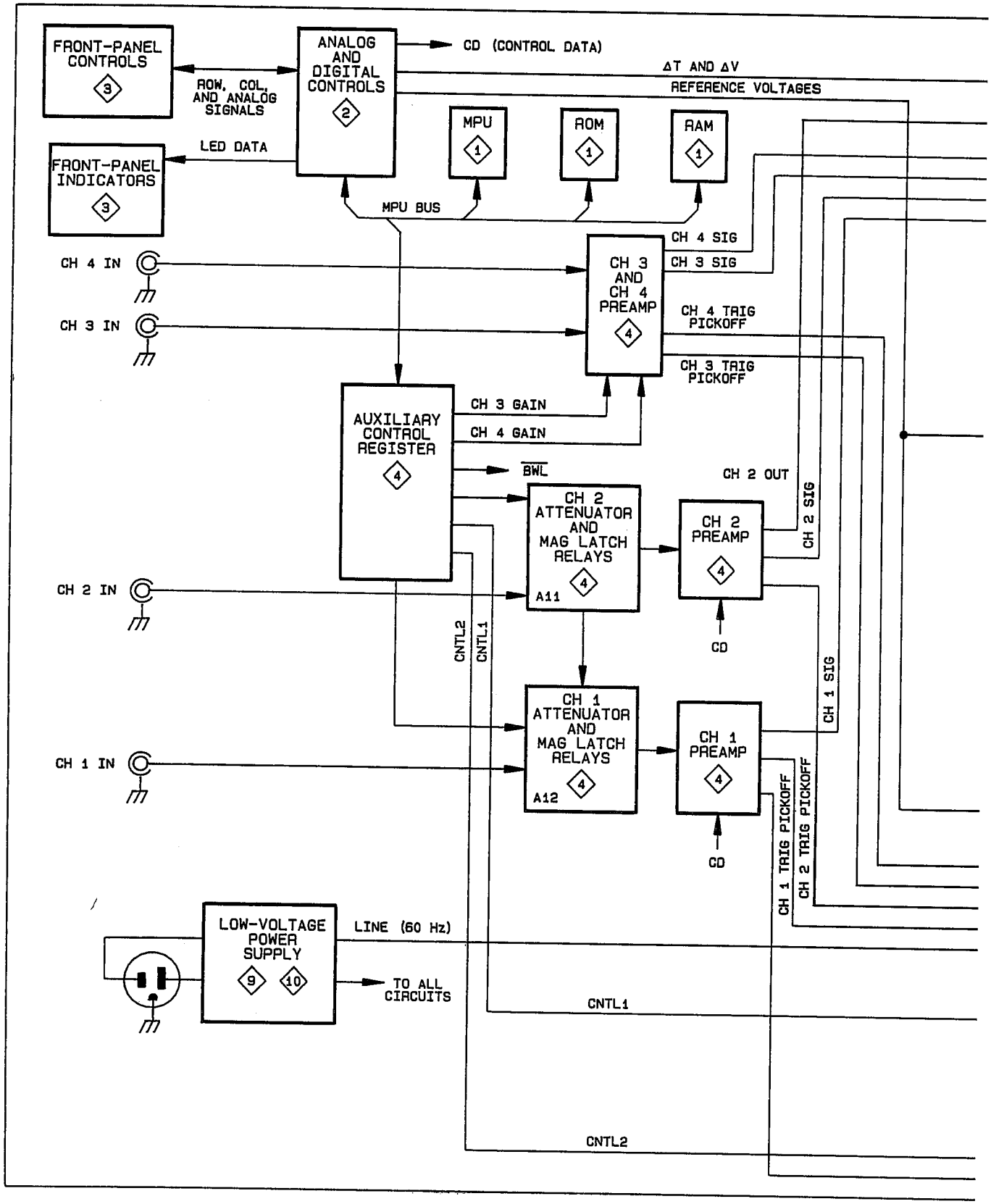


Figure 3-1. Instrument block diagram.

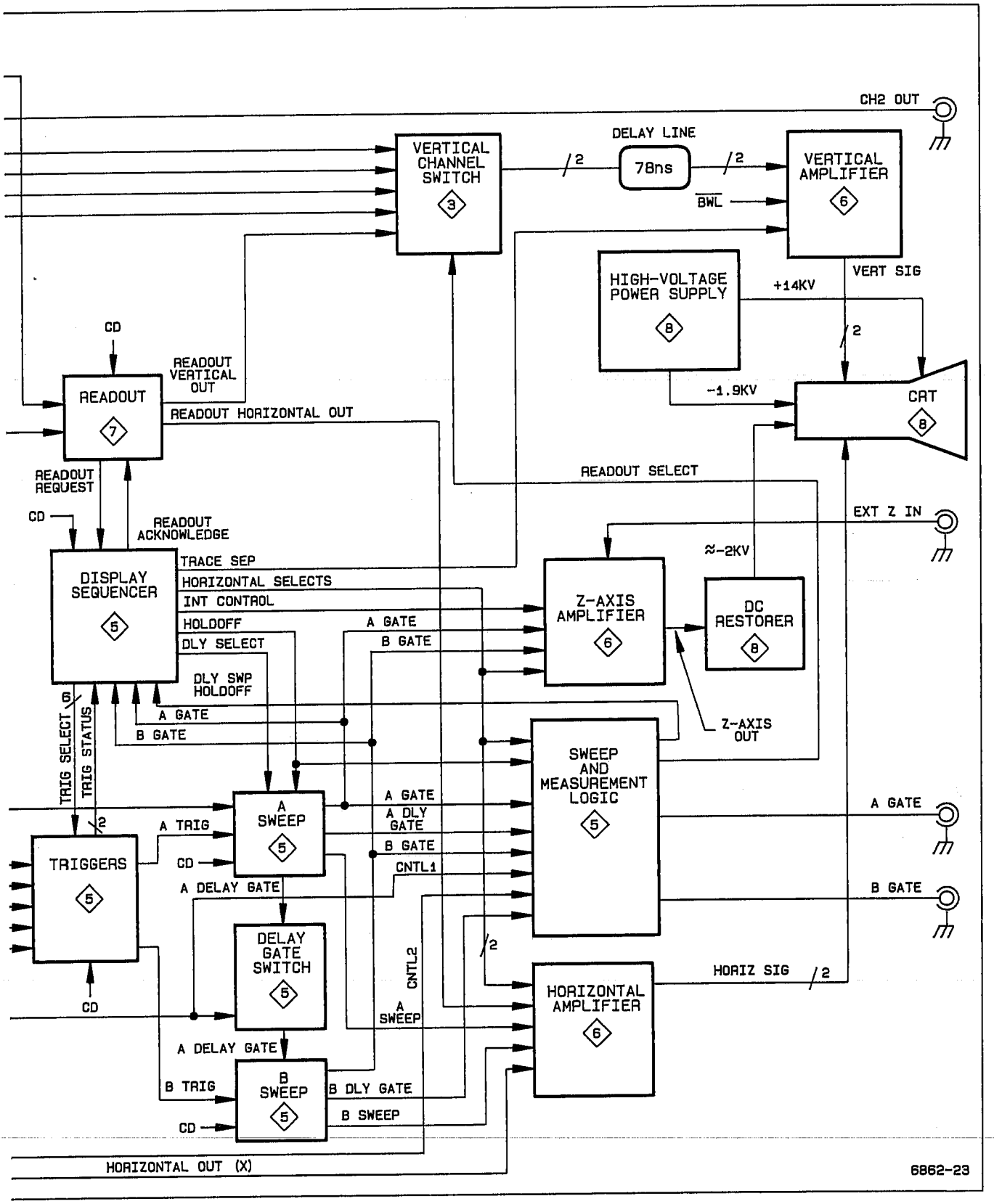


Figure 3-1. Instrument block diagram (cont).

Most of the activities of the instrument are directed by a microprocessor. The microprocessor, under firmware control (firmware is the programmed instructions contained in read-only memory that tells the processor how to operate), monitors instrument functions and sets up the operating modes according to the instructions received.

Various types of data read to and from the Microprocessor (program instructions, constants, control data, etc.) are all transferred over a group of eight bidirectional signal lines called the Data Bus. The Data Bus is dedicated solely to microprocessor-related data transfer.

Another group of signal lines, called the Address Bus, are responsible for selecting or "addressing" the memory location or device that the Microprocessor wants to communicate with. Typically, depending on the instruction being executed, the processor places an address on the Address Bus to identify the location the Microprocessor must communicate with. This address, along with some enabling logic, opens up an appropriate data path between the processor and the device or memory location via the Data Bus; and data is either read from or written to that location by the processor.

While executing the control program, the Microprocessor retrieves previously stored calibration constants and front-panel settings and, as necessary places program-generated data in temporary storage for later use. The battery backed up RAM provides these storage functions.

When power is applied to the instrument, a brief initialization sequence is performed, and then the processor begins scanning the front-panel controls. The switch settings detected and the retrieved front-panel data from the battery backed up RAM causes the processor to set various control registers and control voltages within the instrument that define the operating mode of the instrument. These register settings and voltage levels control the vertical channel selection and deflection factors, the sweep rate, the triggering parameters, the readout activity, and sequencing of the display. Loading the control data into the various registers throughout the instrument is done using a common serial data line (CD). Individual control clock signals (CC) determine which register is loaded from the common data line.

Coordination of the vertical, horizontal, and Z-Axis (intensity) components of the display must be done in real time. Due to the speed of these display changes and the precise timing relationships that must be maintained between display events, direct sequencing of the display is beyond the capabilities of the processor control. Instead, control data from the processor is sent to the Display Sequencer (a specialized integrated circuit) which responds by setting up the various signals that control the stages handling real-time display signals. The controlled stages are stepped through a predefined sequence that is determined by the control data. Typically, as the sequence is being executed, the Display Sequencer will be changing vertical signal sources, Z-Axis intensity levels, triggering sources, and horizontal sweep signal sources. The specific activities being carried out by the Display Sequencer depend on the display mode called for by the control data.

Vertical deflection for crt displays comes from one or more of the four front-panel vertical inputs and, when displaying readout information, from the Readout circuitry. Signals applied to the front-panel Channel 1 and Channel 2 inputs are connected to their respective Preamplifiers via processor-controlled Attenuator networks. Control data from the Microprocessor defining the attenuation factor for each channel is serially loaded into the Auxiliary Control Register and then strobed into the Attenuator Mag-Latch Relays in parallel. The relay switches of each Attenuator network are either opened or closed, depending on the data supplied to the Mag-Latch Relay Drivers. The relays are magnetically latched and remain as set until new control data is strobed in. The Auxiliary Control Register is therefore available, and different mode data is clocked into the register to set up other portions of the instrument.

Attenuated Channel 1 and Channel 2 input signals are amplified by their respective Preamplifiers. The gain factor for the Channel 1 and Channel 2 Preamplifiers is settable by control data from the processor. The Channel 3 and Channel 4 input signals are amplified by their respective Preamplifiers by either of two gain factors set by control bits from the Auxiliary Control Register. All four of these preamplified signals are applied to the Vertical Channel Switch where they are selected by the Display Sequencer for display when required.

Each of the vertical signals is also applied to the A and B Trigger circuitry via trigger pickoff outputs from the Preamplifier stages. Any one of the signals may be selected as the trigger SOURCE for either the A or the B Trigger circuitry as directed by the Display Sequencer. The line trigger signal provides an added trigger source for A Sweeps only. Control data from the Microprocessor is written to the Trigger circuitry to define the triggering LEVEL, SLOPE, and COUPLING criteria. When the selected trigger signal meets these requirements, a sweep can be initiated. The Trigger circuit initiates both the A Sweep and the B Sweep as required by the display mode selected.

In the case of A Sweeps, the LO state of the THO (trigger holdoff) signal from the Display Sequencer enables

the A Sweep circuit and the next A trigger initiates the sweep. For B sweeps, and in the case of intensified sweeps, the A Sweep delay gate signal (DG) enables the B Sweep circuit. Depending on the B trigger mode selected, a B Sweep will be initiated either immediately (RUN AFT DLY) or on the next B trigger signal (TRIG AFT DLY). The slope of the sweep ramp is dependent on Microprocessor-generated control data loaded into the internal control register of the A and B Sweep circuit hybrids.

Sweep signals generated by each of the Sweep hybrids are applied to the Horizontal Amplifier. The Horizontal Amplifier is directed by the Display Sequencer to select one of the sweep ramps for amplification in sequence. In the case of Readout and X-Y displays, the X-Readout and CH 1 input signals are selected to be amplified, also under direction of the Display Sequencer.

To control the display intensity, the Display Sequencer directs the Z-Axis circuit to unblank the display at the appropriate time for the sweeps and readout displays. When the display is unblanked, the Display Sequencer selects the display intensity for either waveform displays or for readout displays by switching control of the Z-Axis beam current between the front-panel INTENSITY and READOUT INTENSITY potentiometers as appropriate.

During readout displays, the vertical dot-position signal from the Readout circuitry is applied to the Vertical Amplifier via the Vertical Channel Switch. Horizontal dot-position deflection for the readout display is selected by internal switching in the Horizontal Amplifier.

The vertical, horizontal, and Z-Axis signals are applied to their respective amplifiers where they are raised to crt-drive levels. The output signals from the Vertical and Horizontal Amplifiers are applied directly to the crt deflection plates. The Z-Axis Amplifier output signal requires interfacing to the high-potential crt environment before application to the crt control grid. The necessary Z-Axis interfacing is provided by the DC Restorer circuit located on the High-Voltage circuit board. The resulting display may be of waveforms, alphanumeric readout, or a combination of both.

DETAILED CIRCUIT DESCRIPTION

INTRODUCTION

The following discussion provides detailed information concerning the electrical operation and circuit relationships of the instrument. Circuitry unique to the instrument is described in detail, while circuits common in the electronics industry are not. The descriptions are accompanied by supporting illustrations and tables. Diagrams identified in the text, on which associated circuitry is shown, are located at the rear of this manual in the tabbed foldout pages.

PROCESSOR AND DIGITAL CONTROL

The Processor and Digital Control circuitry (diagram 1) directs the operation of most oscilloscope functions by following firmware control instructions stored in memory. These instructions direct the Microprocessor to monitor the front-panel controls and to send control signals that set up the various signal processing circuits accordingly.

Microprocessor

The Microprocessor (U2140) is the center of control activities. It has an eight-bit, bidirectional data bus for data display transfer (D0 through D7) and a 16-bit address bus (A0 through A15) for selecting the source or destination of the data. Precise timing of instruction execution, addressing, and data transfer is provided by an external, crystal-controlled clock signal.

The clock signal is developed by the Microprocessor Clock stage and applied to the Microprocessor at pin 39. Using the external clock as a reference, the Microprocessor generates synchronized control output signals, R/$\overline{W}$ (read-write), E (enable), and VMA (valid memory address) that maintain proper timing relationships throughout the instrument.

Microprocessor Clock

The Microprocessor Clock stage generates a 5-MHz square-wave clock signal to the Microprocessor and a 10-MHz clock signal to portions of the Readout circuitry. Flip-flop U2440A is a divide-by-two circuit that reduces the 10-MHz clock down to a 5-MHz square-wave signal used to clock the Microprocessor and the Display Sequencer. The 10-MHz clock is supplied to the Readout circuitry for dot timing and is also available for use with option circuitry.

Reset Control

The Reset Control circuitry ensures that, at power up, the Microprocessor begins program execution from a known point in memory and with all the processor registers in known states. It also allows the processor to reset itself when power is turned off so that the instrument powers down in a known state.

POWER UP SEQUENCE. Reset generator U2240 generates the power-up reset. As power is applied to the instrument U2240 tests the voltage at U2240 pin 7. The reset generator forces U2240 pin 5 LO, and the LO is applied to the processor $\overline{\text{RESET}}$ input (pin 40). After the SENSE input reaches its nominal voltage level, the reset condition continues to allow the microprocessor system time to reset. The reset continues for the time determined by C2350. The effect of power supply transients is reduced by C2241. After the suplies reach their nominal level and the delay period ends U2240 pin 5 goes HI. The RESET signal to the processor then goes HI to enable normal execution to begin, and the processor is directed to the starting address of the power-up routine, which it then performs.

POWER DOWN SEQUENCE. When the instrument power switch is turned off, the PWR UP signal from J251 pin 12 immediately goes LO. This LO generates the NMI (non-maskable interrupt) request to the processor on pin 6 which causes the processor to branch to the power-down routine. Under direction of that routine, the processor begins shutting down the instrument in an orderly fashion before the power supply outputs can drop below the operating thresholds. This routine disconnects the CH1 and CH2 50-Ω input terminations to protect them from accidental application of excessive voltage during storage or bench handling.

As the operating voltages are falling, the Reset circuitry must not generate a false RESET signal to the processor. Such a restart when the power supply voltages are outside their normal operating range would produce unpredictable processor operation that could alter the contents of the battery backed up RAM. When the processor has completed all the other power-down tasks, it finally sets the PWR DOWN signal HI via U2310 (diagram 2). This signal is applied to inverter U2540E at pin 11. Pin 10 of U2650C goes LO and immediately pulls pin 2 of Reset Generator U2240 LO to prevent a reset to the processor. Reset Generator U2240 immediately switches state to assert the $\overline{\text{RESET}}$ signal to the processor. The $\overline{\text{RESET}}$ signal is held LO until the power supplies have fully discharged.

For diagnostic purposes, the PWR DOWN reset signal can be disabled. Moving jumper P503 to the DIAG (diagnostic) position keeps U2240 pin 2 HI. The RESET signal is therefore held HI, and the processor can execute a free-running NOP (no operation) loop without interruption if the PWR DOWN bit is set HI while the Address Bus is incrementing.

Address Bus

Octal Latches, U2415 and U2425 are used to buffer the address signals to the circuitry on the Processor Control board as well as provide additional drive current for the options. The RC network composed of R2465 and C2465 and inverter U2540B provide an additional >30 ns of address hold time on the buffered address signals for the options.

U2415 and U2425, along with Octal Latch U2405, allow the buffered Address Bus and Microprocessor control signals to be disconnected from the microprocessor. This allows in-circuit testing of the Processor Control board without having to remove the Microprocessor.

Data Bus

Tri-state buffer U2350 is used to buffer the data signals to the Microprocessor from other devices on the bus. When not enabled, the device is switched to isolate the processor from the buffered Data Bus. Buffer U2350 is enabled via the Read-Write Latch U2440B when the processor reads data from another device on the bus.

When the processor writes data onto the bus, Octal Latch U2450 is enabled by the Read-Write Latch U2440B. When the E (enable) signal at pin 11 of U2450 is HI, processor data bits are passed asynchronously through the latch to the buffered data bus. When the E signal goes LO, data bits meeting setup times are latched into the device. The latched Q outputs provide the required drive current to the various devices on the bus and ensure that data hold times are met for correct data transfer. When the Read-Write Latch places a HI on pin 1 of U2450, latch U2450 is disabled, and the outputs are switched to their high-impedance state.

Data transfers to and from the processor may be interrupted by removing Diag/Norm Jumper P503. This forces a NOP (no operation) condition that is useful for verifying the functionality of the processor (when a data-bus device is suspected of causing a system failure) or for troubleshooting the Address Bus and Address Decode circuitry. Moving the jumper to the DIAG position disables both U2350 and U2450 and disconnects the microprocessor from the buffered Data Bus. With the Data Bus disconnected, a resistor network pulls the processor Data Bus lines (D0 through D7) to a NOP (no operation) instruction. A NOP causes the Microprocessor to continuously increment through its address field. The Address Decode circuitry may then be checked to determine if it is operating properly.

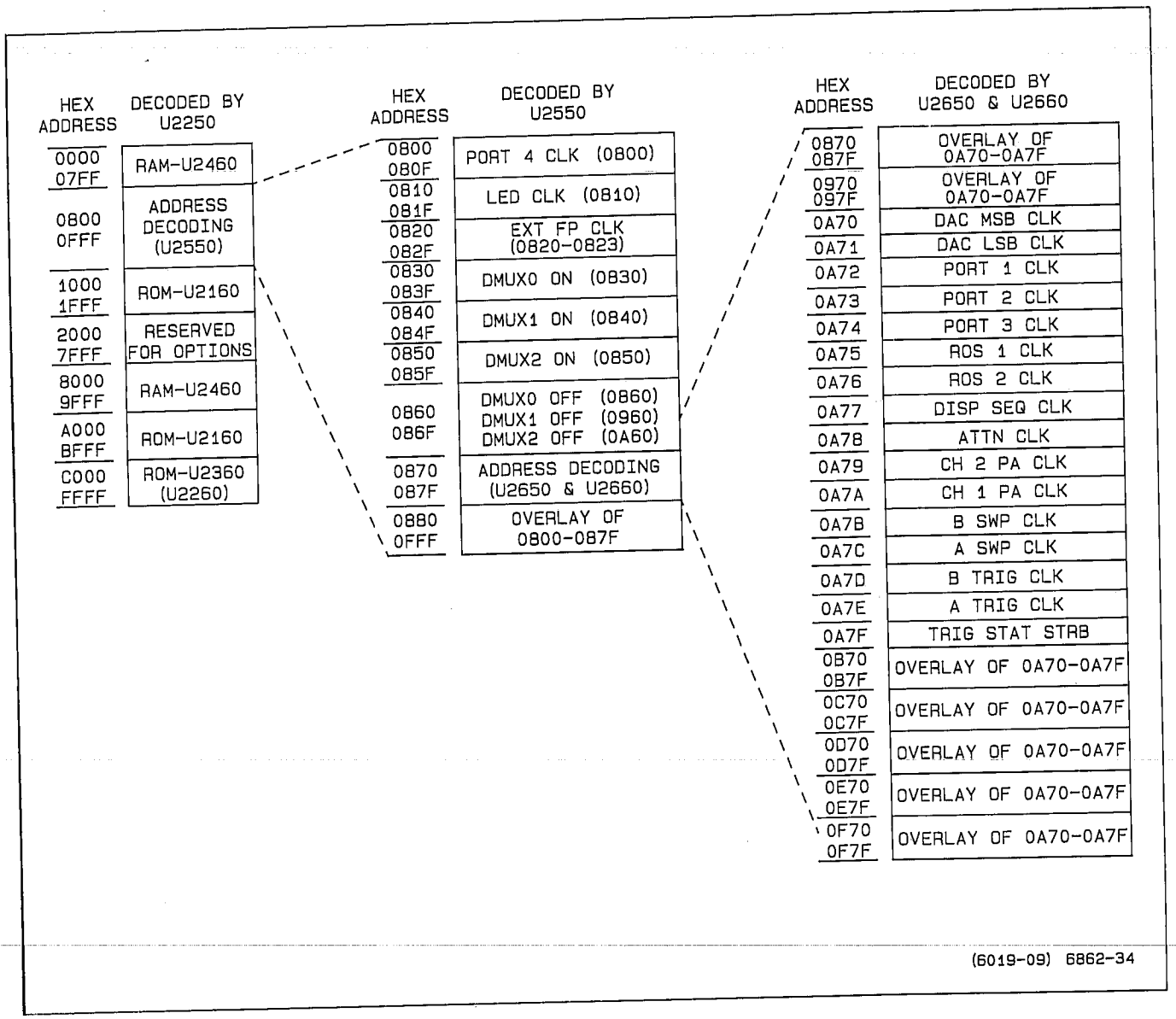


Figure 3-2. Address decoding.

Address Decode

The Address Decode circuitry generates enabling signals and strobes that allow the Microprocessor to control the various devices and circuit functions. The controlling signals are generated as a result of the Microprocessor placing specific addresses on the Address Bus. Figure 3-2 illustrates the enables and strobes generated by the Address Decode circuitry.

Address decoding is performed by a programmable logic device and 3 three-to-eight line decoders attached to the Address Bus. The five most significant address bits are decoded by U2250. This device initially separates the total addressable-memory space (64K-bytes) into thirty-two 2K-byte blocks. Addresses in the top 24K-byte memory space (address bit BA15 HI and either BA14 or BA13 HI) select one of two read-only memories (ROM); U2160 or U2360 (or U2260). When the VMA (Valid Memory Address) and E (Enable) outputs from the Microprocessor go HI, the selected ROM is enabled, and the data from the selected address location is read from the ROM. The remaining 8K-byte memory space (address bit BA15 HI and both BA14 and BA13 LO) select random-access memory (RAM); U2460. Both outputs of flip-flop U2440B are used to generate the $\overline{OE}$ and $\overline{WE}$ signals to the RAM.

Of the addresses in the bottom 32K-byte memory space, only the lowest 8K-bytes are further decoded. Addresses in the lowest 2K-byte block of addresses will cause U2250 to generate an enable signal to the RAM, U2460. Addresses in the next 2K-byte block of addresses will enable U2550 to do the next state of address decoding. The next 4K-byte block of addresses will enable the Buffer Board ROM section of U2160.

The level of decoding performed by U2550 uses address bits BA4, BA5, and BA6 to separate the addresses within the 2K-byte block of addresses 0800 thru 0FFF into 128 groups of 16 addresses. Address bits BA7 thru BA10 are not used in the decoding scheme, so each of these 128 blocks is not uniquely identified. This results in sixteen duplicate sections within the address block, each consisting of eight groups of 16 addresses. The upper fifteen sections in the address space are never used; therefore, decoding by U2550 may be more simply thought of as eight groups of 16 address locations. Addresses within these eight groups generate control signals to other portions of the instrument.

The final level of address decoding is done by a pair of three-to-eight-line decoders, U2650 and U2660. When enabled by the Y7 output of U2550, these decoders separate the highest 16-address group decoded by U2550 into 16 individual control signals.

Each of the control signals generated by the Address Decode circuitry are present only as long as the specific address defining that signal is present on the Address Bus. However, four of the addressable control signals decoded by U2550 are used to either set or reset flip-flops U2560A and B, and U2570A. The control signals are, in effect, latched and remain present to enable multiplexers U2521, U2530, (diagram 2), and U170 (diagram 4). When enabled, these multiplexers route analog control signals from the DAC (digital-to-analog converter) U2101 (diagram 2) to the various analog control circuits.

Read-only Memory (ROM)

The Read-only Memory consists of one 128K-byte ROM and one 64K-byte ROM that contain operating instructions (firmware) used to control processor (and thus oscilloscope) operation. Addresses from the Microprocessor that fall within the top 24K-bytes of addressable space cause one of the two read-only memory integrated circuits to be enabled. (See Address Decode description.) Instructions are read out of the enabled ROM (or PROM) IC from the address location present on its address input pins. The eight-bit data byte from the addressed locations is placed onto the Buffered Data bus (BD0 through BD7) to be read by the Microprocessor.

Random-Access Memory (RAM)

The RAM consists of integrated circuit U2460 and provides the Microprocessor with 8K-bytes of battery backed up temporary storage space for data that is developed during the execution of a routine. The RAM is enabled whenever an address in the lowest 2K-byte of addresses is placed on the Address Bus or whenever an address of 8000 thru 9FFF is placed on the Address bus. When writing into the RAM, the write-enable signal (WE) on pin 27 of U2460 is set LO along with the chip enable (CE) signal on pin 20. At the same time, the output-enable (OE) on pin 22 is HI to disable the RAM output drivers. Data is then written to the location addressed by the Microprocessor. If data is to be read from the RAM, the WE signal is set HI to place the RAM in the read mode, and the OE signal is set LO to enable the output drivers. This places the data from the addressed location on the buffered Data Bus where it can be read by the Microprocessor.

The RAM also provides non-volatile storage for the calibration constants and the power-down front-panel settings. When power is applied to the instrument, the Microprocessor reads the calibration constants and generates control voltages to set up the analog circuitry. The front-panel settings that were present at power-off are recalled and the instrument is set to the operating mode previous to power-off.

Timing Logic

The Timing Logic circuit composed of U2440B, and U2540D generates time- and mode-dependent signals from control signals output from the Microprocessor. The enable (E) signal output from the Microprocessor is a 1.25 MHz square wave used to synchronize oscilloscope functions to processor timing.

Data applied to the Address Bus, Data Bus, and various control signals are allowed to settle (become valid) before any of the addressed devices are enabled. This is accomplished by switching the E signal HI a short time after each processor cycle begins. Inverter U2540D inverts the polarity of the delayed enable signal and enables the Address Decode stage only after the address bus has settled.

Read-Write Latch U2440B is used to delay the processor's read/write signal (R/$\overline{W}$) from the Microprocessor to meet hold-time requirements of the RAM. At the same time, it generates delayed read and write enabling signals of both polarities to meet the requirements of Buffer U2350 and Latch U2450 (in the Microprocessor Data Bus) and various other devices in the Readout circuitry (diagram 7).

When R/$\overline{W}$ goes LO for a write cycle and E goes HI, Read-Write Latch U2440B is reset, and Q output (pin 9) is held LO, Latch U2450 is in its transparent state at this time, and data from the Microprocessor is applied asynchronously to the buffered Data Bus. At the end of the write cycle, the R/$\overline{W}$ signal goes HI. The E signal also goes through a negative transition, and data on the Microprocessor data bus lines is latched into U2450. The next positive transition of the 1.25-MHz E signal (1/2 E cycle after the R/$\overline{W}$ signal goes HI) clocks the HI level at U2440B pin 12 (the D input) to the Q output, and the $\overline{Q}$ output (pin 8) goes LO. The 1/2 E cycle delay between the time R/$\overline{W}$ goes HI and the time that the Q output of U2440B goes HI keeps Latch U2450 outputs on long enough to meet the data hold time for the RAM. At the end of that delay time, pin 1 of U2450 goes HI, and the Latch outputs are switched to the high-impedance state to isolate it from the buffered Data Bus.

READOUT FRAMING AND INTERRUPT TIMING. Binary counter U2640 is used to generate a readout-framing clock to the Readout circuitry and a real-time interrupt request to the Microprocessor via inverter U2540C. The readout-framing clock is a regular square-wave signal obtained from U2640 pin 12 by dividing the 1.25-MHz E signal by 512 (2^9). This clock tells the readout circuitry to load the next block (subframe) of readout information to be displayed. (See "Readout" description for further information concerning alphanumeric display.) The real-time interrupt request, which occurs every 3.3 ms, is obtained from pin 2 by dividing the E signal by 8192 (2^{13}).

When the real-time request occurs, $\overline{IRQ}$ (pin 4 of U2140) goes LO, and the processor breaks from execution of its mainline program. The Microprocessor first resets Binary Counter U2640 by setting pin 19 of U2301 (diagram 2) HI (to generate the reset), then it resets pin 19 LO to allow the counter to start again. At this time, the Microprocessor sets analog control voltages and reads trigger status from the Display Sequencer (diagram 5). When this is completed, it reverts back to the mainline program.

In addition to the analog control and trigger status update that occurs with each interrupt, on every fifth interrupt cycle, the Microprocessor also scans the front-panel potentiometers. Every tenth interrupt cycle, scanning the front-panel switches and checking the 50-Ω DC inputs for overloads is added to the previously mentioned tasks. If all the tasks are not completed at the end of one interrupt cycle, the real-time interrupt request restarts the analog updates, but as soon as those are accomplished, the Microprocessor will pick up with its additional tasks where it was before the interrupt occurred. This continues until all tasks are completed. If any pot or switch changes are detected, the Microprocessor updates the analog control voltages and the control register data to reflect those changes prior to reverting back to the mainline program instructions.

FRONT-PANEL SCANNING and ANALOG CONTROLS

The Analog Control circuitry (diagram 2), under Microprocessor control, reads the front-panel controls and sets various analog control voltages to reflect these front-panel settings. The calibration constants determined during instrument calibration and the last "stable" front-panel setup conditions are stored in battery backed up RAM. At power-on the stored front panel information is used to return the instrument to its previous state.

Hardware I/O

Data transfer from the Analog Control circuitry to the Microprocessor is via Status Buffer U2220. Data bits applied to the input pins are buffered onto the Data Bus when enabled by the Address Decode circuitry. Via the Status Buffer, the processor is able to (1) determine the settings of front- and rear-panel pots and switches, (2) determine instrument type (2445B or 2455B), (3) determine if a triggered sweep is in progress, and (4) read the contents of the Readout RAM. When disabled, the buffer outputs are switched to high impedance states to isolate them from the buffered Data Bus.

Data transfer from the Microprocessor to the Analog Control circuitry is via registers U2210 and U2310. Via register U2210, the Microprocessor is able to select the pot-scanning multiplexers, turn the trigger LED on and off, and control other hardware via serial control data and the attenuator strobe. Via register U2310, the processor controls pot selection, ROM addressing, and power down timing.

Front-Panel Switch Scanning

The Front-Panel Switches are arranged in a matrix of ten rows and five columns. Most of the row-column intersections contain a switch. When a switch is closed, one of the row lines is connected to one of the column lines through a diode. Reading of the switches is accomplished by setting a single row line LO and then checking each of the five column lines sequentially to determine if a LO is present (signifying that a switch is closed). After each of the five columns have been checked, the current row line is reset HI and the next row line is set LO for the next column scan cycle. A complete Front-Panel scan consists of all ten row lines LO in sequence and performing a five-column scan for each of the rows.

Row lines are set LO when the microprocessor writes a LO to one of the flip-flops in octal registers U2301 or U2201. The row data placed on the buffered Data Bus by the Microprocessor is clocked into the registers as two, eight-bit words by clocks from the Address Decode circuitry (DAC LSB CLK for the lower eight bits and DAC MSB CLK for the upper eight bits). All eight outputs of register U2201 and two outputs of U2301 drive the ten rows of the front-panel switch matrix (the fifth line of the matrix is not used). Series resistors in the lines limit current flow and eliminate noise problems associated with excessive current flow.

While each row is selected, the processor will scan each of the five column lines. To scan the columns, the microprocessor enables U2410 by the address decode circuitry. Data bits applied to the input pins are buffered onto the Data Bus.

In addition to the front-panel switches, the CAL/NO CAL jumper (P501) is checked to determine whether the instrument should be allowed to execute the calibration routines. The levels on U2410 pin 11 and 12 are read by scanning two additional columns at power-up. If the jumper is pulling the CAL bit LO, the operator will be allowed to use the calibration routines stored in firmware. If the NO CAL bit is pulled LO, the calibration routines may not be performed. If the jumper is removed, and neither bit is pulled LO, the Microprocessor is forced into a special diagnostic mode (CYCLE) used to record certain operating failures during long-term testing of the instrument. (See the "Maintenance" section for an explanation of the diagnostic modes.) Removing P501 or switching it between the CAL and NO CAL positions will not be recognized by the Microprocessor until the instrument is powered down and then turned back on.

The resistors in series with the input lines to U2410 are current-limiting resistors that protect the CMOS data buffer from static discharges. The resistors connected from the input lines to the +5 V supply are pull-up resistors for the front-panel column lines.

Digital-to-Analog Converter (DAC)

DAC U2101 is used to set the various analog references in the instrument and is used to determine the settings of the front panel potentiometer. The 12-bit digital values to be converted are written to octal registers U2301 and U2201 for application to the DAC input pins. The DAC then outputs two complementary analog currents that are proportional to the digital input data. (Complementary, in this case, means that the sum of the two output currents is always equal to a fixed value.)

The maximum range of the output currents is established by a voltage-divider network composed of R2010, R2012, R2013, R2014 and R2011 conected to the positive and negative reference current inputs of the DAC (pins 14 and 15 respectively). A +10-V reference voltage applied to the DAC through R2013 sets the basic reference current. Resistor R2011 and R2014 and potentiometer R2010 provide a means to adjust this current over a small range for calibration purposes. The nominal reference current is 1 mA, the DAC full-scale output current is 4 mA. The output currents flow through series resistors R2520 and R2521, connected to the +1.36-V reference, and proportional voltages result.

Pot Scanning

The Pot Scanning circuitry, in conjunction with the DAC, derives digital values for each of the various front-panel potentiometers. Scanning of the pots is accomplished by data selectors U2401, U2501, and U2601. Three bits are written to register U2310 and select the pot to be read. The bits are latched in the register and keep the pot selected until the register is reset. The Microprocessor writes a LO to the inhibit input pin (pin 6) of either U2401, U2501 or U2601 via register U2210 to enable the device. The enabled data selector connects the analog voltage at the wiper of the selected pot to comparator U2510.

Comparator U2510 compares the analog voltage of each pot to the output voltage from the DAC (pin 18). To determine the potentiometer output voltage, the processor performs a binary search routine that changes the output voltage from the DAC in an orderly fashion until it most closely approximates the voltage from the pot.

The conversion algorithm is similar to successive approximation and generates an eight-bit representation of the analog level. When the pot's value is determined, the Microprocessor stores that value in memory. Once all of the pots have been read and the initial value of each has

been stored, the processor uses a shorter routine to determine if any pot setting changes. To do this the DAC output is set to the last known value of the pot (plus and minus a small drift value), and the status bit is read to see that a HI and LO occurs. If within the limits, the processor assumes that the pot setting has not changed and scans the next pot. When the processor detects that a pot setting has changed, it does another binary search routine to find the new value of that pot.

Analog Control

The operating mode and status of the instrument requires that various analog voltages (for controlling instrument functions) be set and updated. The digital values of the controlling voltages are generated by the Microprocessor and converted by the DAC. Analog multiplexers U2521 and U2530 (on diagram 2) and U170 (on diagram 4) route the DAC voltages to sample-and-hold circuits that maintain the control voltages between updates.

The Microprocessor writes three selection bits to register U2301 that directs the DAC output to the appropriate sample-and-hold circuit and charges a capacitor (or capacitors) to the level of the DAC. When the processor disconnects the DAC voltage from the sample-and-hold circuit (by disabling the multiplexer) the capacitor(s) remains charged and holds the control voltage near the level set by the DAC. Due to the extremely high input impedance of the associated operational amplifiers, the charge on the capacitor(s) remains nearly constant between updates.

FRONT-PANEL CONTROLS

The Front Panel is the operator's interface for controlling the user-selectable oscilloscope functions. Along with the crt, it provides visual feedback to the user about the present operating state of the instrument.

Most of the Front-Panel controls (diagram 3) are "cold" controls; i.e., they are not connected directly into the signal path. Therefore, associated circuits are not influenced by the physical parameters (such as capacitance, resistance, and inductance) of the controls. In addition, translating the analog output levels of most of the potentiometers to digital equivalents allows the processor to handle the data in ways that result in a variety of enhanced control features.

To maintain the front-panel operating setup between uses of the instrument, the digitized values of the potentiometers and front-panel switch settings are stored in battery backed up RAM so that when the instrument power is turned off, these control settings are not lost. Then, when power is next applied, the instrument will power up to the same configuration as when the power was last removed (assuming the settings of the non-digitized pots and switches remain the same).

The Front-Panel Controls also allow the user to initiate and direct the diagnostic routines (and when enabled, the calibration routines) programmed into the read-only memory (ROM). These routines are explained in the Maintenance section of this manual.

Front-Panel Switches

The Front Panel Switches are arranged in a ten-row-by-five-column matrix, with each switch assigned a unique location within the matrix (see Figure 3-3). A closed switch connects a row and a column together through an isolating diode. To detect a switch closure, the switch matrix is scanned once every 32 ms (every tenth Microprocessor interrupt cycle). When scanning, the Microprocessor sequentially sets each individual row line LO. A closed switch enables the LO to be passed through the associated diode to a column line. When the processor checks each of the five column lines associated with the selected row, the LO column is detected. The intersection of the selected row and the detected column uniquely identifies the switch that is closed. Further information about switch scanning is found in the "Front-Panel Scanning" description located in the "Analog Control" discussion.

As each switch is read, the processor compares the present state of the switch to its last-known state (stored in memory) and, if the same, advances to check the next switch. When a switch is detected as having changed, the processor immediately reconfigures the setup conditions to reflect the mode change and stores the new state of the switch in memory. The detected status of the switch on each of the following scan cycles is then compared against the new stored data to determine if the switch changes again. The 32-ms delay between the time a switch is detected as having changed and the next time it is read effectively eliminates the effects of switching noise (switch bounce) that may occur after the switch is actuated.

Front-Panel Pots

The thirteen Front-Panel Potentiometers, READOUT INTENSITY, and INTENSITY are "cold" controls that control the linear functions of the instrument. (SCALE ILLUM and FOCUS are not considered part of the Front-Panel Control circuitry for the purposes of this description.) All are digitized and control their functions indirectly. Data Selectors U2401, U2501, and U2601 in the Analog Control circuitry (diagram 2) route the wiper arm voltage of the pot being read to comparator U2510 where it is compared

with the output of DAC U2101. The processor changes the DAC output until it most closely matches the output voltage of the pot, then stores the digital value of the "match". See the "Pot Scanning" description in the "Analog Control" discussion for further information on the reading of pot values.

Like the switch matrix scanning, the Front-Panel pot scanning routine is performed every 16 ms. When entered, the routine reads the settings of the "last-moved" pot and one "unmoved" pot. Each succeeding scan continues to read the last-moved pot in addition to a new unmoved pot. In this way, each pot is monitored, but most of the scan time is devoted to the pot that is still moving (needing continuous updating).

As the initial pot settings are determined, a digital representation of each value is stored in memory. The processor then checks each pot against its last-known value to determine if a pot has moved. If a pot is detected as moving, the processor executes a routine that converts the movement (displacement from last-set value) into a corresponding control voltage.

When producing the actual analog control levels, the processor can manipulate the digital values read for the various pots before sending the output data to the DAC. This allows many of the oscilloscope parameters to vary in an enhanced fashion. The pot data is manipulated by the processor in a manner that produces such features as variable resolution, continuous rotation, fine-resolution backlash, and electrically detented controls.

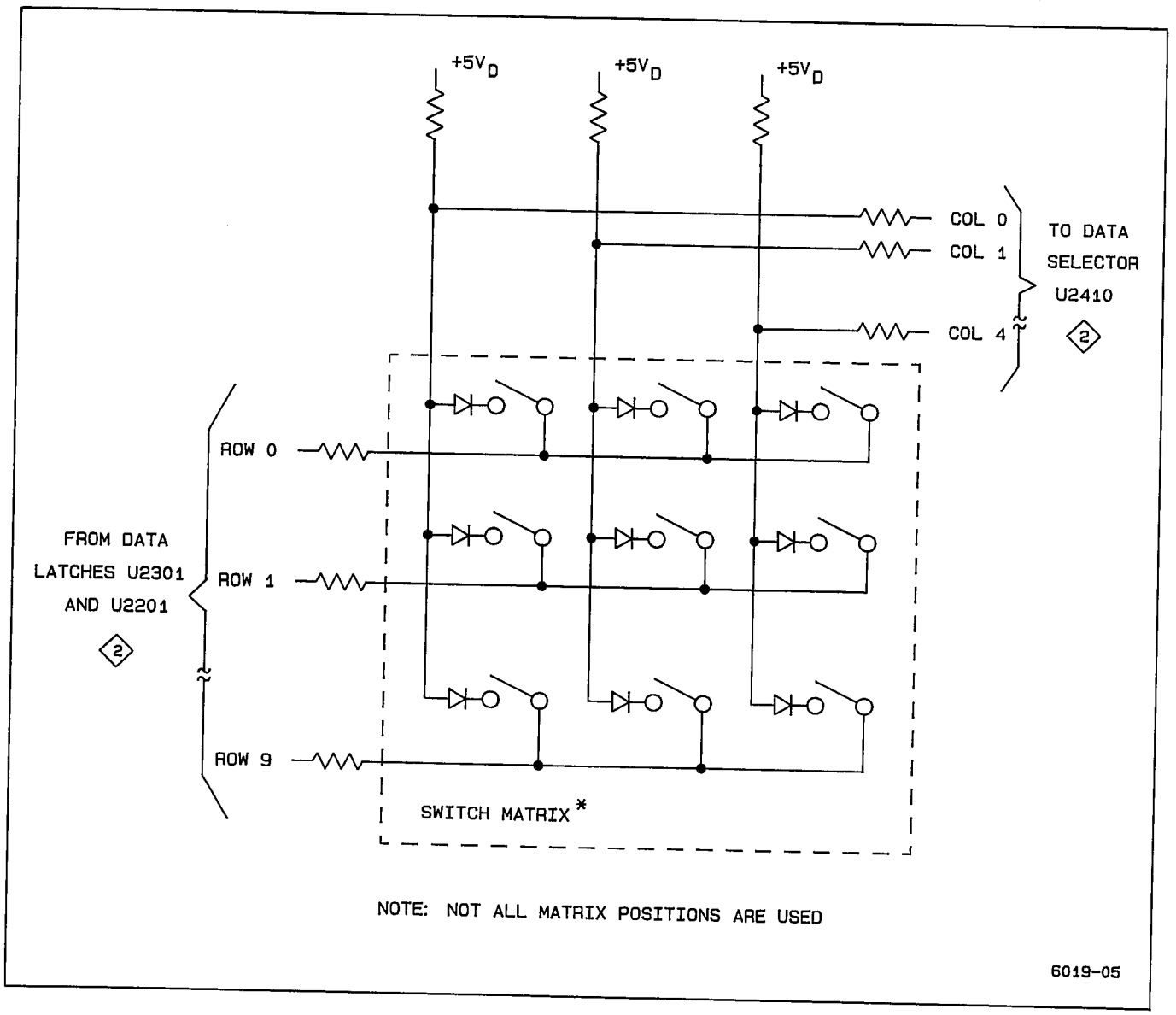


Figure 3-3. Front-panel switch matrix.

With all thirteen Front-Panel Potentiometers, READOUT INTENSITY, and INTENSITY controls, the processor reads the magnitude and direction of pot rotation and produces variable-resolution control voltages. If a pot's direction of rotation changes, the magnitude of the change from the last-set position remains small, or if it was not the last pot moved, a fine-resolution control voltage results. In the fine-resolution range, a given rotation displacement will cause a small control voltage change. The same displacement farther away from the last-set reference will cause a proportionally larger control voltage change, producing a coarse-resolution effect. If the changing pot is the last one moved and the direction of rotation remains the same, the algorithm continues from where it left off during the preceding scan; producing control voltage changes with the same increment as it was last using.

The delta reference controls (Δ REF OR DLY POS and Δ) are continuous-rotation potentiometers. They each consist of two pots ganged together with their wiper arms electrically oriented at 180° apart. As the wiper of one pot is leaving its resistive element, the wiper of the other pot comes onto its element. The Microprocessor has the ability to watch the output voltage from each wiper and when it detects that the controlling wiper is nearing the end of its range, it will switch control over to the other wiper. The routine the processor uses to watch these pots sets the associated control voltage on the basis of relative voltage changes (ΔV) that occur. Switching between the pots to change control to the opposite wiper arm is based on specific voltage levels being sensed.

Sensing specific voltage levels is also used when reading the VOLTS/DIV VAR, SEC/DIV VAR, and HOLDOFF controls. These pots have both a mechanical detent and a processor-generated electrical detent. As one of these controls is moved out of the mechanical detent position, the processor watches the analog voltage changes that occur; but the associated control voltage will not change until a specific voltage level (the electrical detent level) is reached. Once the electrical detent value is exceeded, the processor begins to vary the associated control voltage in response to further pot rotation. When returning to the mechanical position, the electrical detent level is reached first, and the variable voltage action is stopped before the mechanical detent is entered.

Front-Panel Status LEDs

Light-emitting diodes (LEDs) are used to provide visual feedback to the operator about the oscilloscope status and operating mode by backlighting front-panel nomenclature. A 48-bit status word, defining the diodes to be illuminated, is generated by the processor and then serially clocked into the six LED-Status Registers (U3001, U3002, U3003, U3004, U3005, and U3006). The registers hold the selected diodes on until the next update. Whenever the processor detects that a front-panel control has changed (and a new status display is required), a new status word is generated and applied to pin 1 of U3002. As each of the bits is clocked into the Q_A position of U3002, the preceding bit is shifted to the next register position. After 48 bits have been clocked into (and 40 bits through) U3002, all six LED-Status registers are full and contain the LED illumination pattern to be displayed to the user. A LO at any Q output of the registers illuminates the corresponding front-panel LED.

The TRIG'D LED is not driven by the LED-Status Register. It is driven by the Analog Control circuitry and illuminated whenever a triggered sweep is in progress.

ATTENUATORS AND PREAMPS

The Attenuators and Preamps circuitry (diagram 4) allows the operator to select the vertical deflection factors. The Microprocessor reads the Channel VOLTS/DIV switches and VOLTS/DIV VAR controls and then digitally switches the attenuator and sets the preamplifier gains accordingly.

CHANNEL 1 AND CHANNEL 2 ATTENUATORS

The Channel 1 and Channel 2 Attenuators are identical in operation, with corresponding circuitry in each channel performing the same function. Therefore, only the Channel 1 circuitry is described.

Input signals from the Channel 1 input connector are routed through an attenuator network by four pairs of magnetic-latch relay contacts. The position of the relays is set by Microprocessor data placed into Auxiliary Control Register U140. Relay buffer U110 provides the necessary drive current to the relays.

Four input coupling modes (1MΩ AC, GND, 1MΩ DC, and 50 Ω DC) and three attenuation factors (1X, ÷10, and ÷100) may be selected by closing different combinations or relay contacts. The three attenuation factors, along with the variable gain factors of the Vertical Preamplifier, are used together to obtain the crt deflection factors. The relays are magnetically latched and once set, remain in position until new attenuator-relay-setting data and strobes are generated. (See the "Auxiliary Control Register" description for a discussion of the relay-latching procedure.)

The 50 Ω termination resistor has a thermal sensor associated with it that produces a dc voltage (CH 1 OVL) proportional to the input power. Should the input power exceed the normal safe-operating level for the 50 Ω DC input, the termination resistor temperature will exceed the

normal operating limit and change the output voltage of the thermal sensor. The amplitude of this dc level is periodically checked via comparator U2510 and DAC U2101 (on diagram 2) and allows the Microprocessor to detect when an overload condition is present. When an overload occurs, the processor switches the input coupling to the 1 MΩ position to prevent damage to the attenuator and displays 50 Ω OVERLOAD on the crt.

Compensating capacitor C105 is adjusted at the time of calibration to normalize input capacitance of the preamplifier to the attenuator.

A probe-coding ring around the BNC input connector passes probe coding information (a resistance to ground) to the Analog Control circuitry for detection of probe attenuation factors. The readout scale factors are set to reflect the detected attenuation factor of the attached probe.

Auxiliary Control Register

The Auxiliary Control Register allows the Microprocessor to control various mode and range dependent functions of the instrument. Included in these functions are: attenuation factors, input coupling, Channel 3 and Channel 4 gains, vertical-bandwidth limiting, the X-Y display mode, and the state of the measurement PAL.

When the Microprocessor sets the input coupling mode and attenuation factors for Channel 1 and Channel 2, a series of eight, 16-bit control words are serially clocked into shift registers U140 and U150 (eight bits in each register). Each control word is used to set the position of one of the eight attenuator and coupling relays (four relays are in each attenuator assembly). Each control word will have one HI bit. This bit will correspond to the specific relay contact to be closed. Relay buffers U110 and U130A (for Channel 1) and U120 and U130B (for Channel 2) are Darlington configurations that invert the polarities of all bits. This results in a LO being applied to only the coil lead associated with the contact to be closed; all other coil leads are held HI.

To set a relay once the control word is loaded, the Microprocessor generates a ATTN STRB (attenuator strobe) to U130G pin 7 via R129 and C130. The strobe pulses the output of U130G LO for a short time. This output pulse attempts to turn on both Q130 and Q131 (relay drivers) via their identical base-bias networks. Due to the lower level from the turned on Darlington relay buffer (coupled through the associated coil diode and either CR130 or CR131 to one of the bias networks), one transistor will turn on harder as the ATTN STRB pulse begins to forward bias the transistors. The more positive collector voltage of the transistor turning on harder is fed through the bias diode (again either CR130 or CR131) to further turn off the opposite transistor. This action results in one transistor being fully on and the other one being fully off. The saturated transistor sources current through the two stacked relay coils to the LO output of either U140 or U150 (current sink) to close the selected contacts. Once set, the magnetic-latch feature will hold the relay set to this position until opposing data is clocked into the Auxiliary Control Register and strobed into the relay. All coil leads for the remaining relays are set HI, and only the selected relay will be set.

To set the seven remaining Attenuator and coupling relays, the sequence just described is repeated seven more times. Whenever the Microprocessor determines that the attenuation factor or input coupling has changed, the entire relay-setting procedure is repeated for all eight relays.

After the coupling and attenuator relays have been latched into position, the Auxiliary Control Register is free to be used for further circuit-controlling tasks. Eight more bits of control data are then clocked into U140 either to enable or disable the following functions: vertical bandwidth limiting (BWL), triggered X-Y mode (TXY), the A and B Sweep Delay Comparators (BDCA and BDCA), and slow-speed intensity limit (SIL); or to alter the Channel 3 and Channel 4 gain factors (GA3 and GA4). Four other bits are clocked into register U150: one to produce the CTC signal, one to control the scale illumination circuit during SGL SEQ display mode, and two (CNTL1 and CNTL2) to control the state of the measurement PAL, U975. The CTC control bit is used to enable a sweep-start linearity circuit in the A Sweep circuitry (diagram 5) on the 2 ns and 20 ns per division sweeps.

Analog Control Demultiplexer

When enabled by the Address Decode circuitry, Analog Control Demultiplexer U170 directs the analog levels applied to pin 3 from DAC U2101 (diagram 2) to one of six sample-and-hold circuits. In the Preamplifier circuitry, the sample-and-hold circuits maintain the VAR gain and DC Bal control-voltage levels applied to both the Channel 1 and Channel 2 Preamplifiers U100 and U200 between updates. Two of the Demultiplexers outputs direct analog levels to the Holdoff and Channel 2 Delay offset sample-and-hold circuits (diagram 5). Routing is determined by the three-bit address from register U2301 (diagram 2) applied to Demultiplexer U170 on pins 9, 10, and 11.

Channel 1 Preamplifier

Channel 1 Preamplifier U100 converts the single-ended input signal from the Channel 1 Attenuator to a differential output signal used to drive the Vertical Channel Switch. The device produces either amplification or attenuation in predefined increments, depending on the control data writ-

ten to it from the Microprocessor. The preamp also has provisions for VAR gain, vertical positioning, and a trigger signal pickoff.

The Channel 1 vertical input signal is applied to pin A of Channel 1 Preamplifier U100. Control data from the processor is clocked into the internal control register via pin 22 (CD) by the clock signal applied to pin 23 ($\overline{CC}$). The data sets the device to have an input-to-output gain ratio of 2, 4, or 10, depending on the VOLTS/DIV control setting.

Two analog control voltages set by DACs modify the differential output signal at pins 9 and 10. The front-panel Channel 1 POSITION control supplies a position signal to U100 pin 17 (via MUX U2530 and sample-and-hold U2430 and C2432) that vertically positions the Channel 1 display on the CRT. A DC Bal signal is applied to pin 2 of U100 from MUX U170 via the sample-and-hold circuit composed of U160A and C177. This DC BAL signal is a dc offset-null level that is determined during the automatic DC Bal procedure. The offset value is stored as a calibration constant in RAM and is recalled at regular intervals to set the DC Bal level, holding the Preamplifier in a dc balanced condition.

The Channel 1 VOLTS/DIV VAR control is monitored by the Microprocessor during the front-panel scanning routine. When the processor has determined where the VOLTS/DIV VAR control is positioned, it causes DAC U2101 (diagram 2) to produce a corresponding control level and routes it to the VAR gain sample-and-hold circuit composed of U160D, C179, and associated components. The control voltage at the output of U160D (pin 14) sets the variable gain of the Preamplifier.

A pickoff amplifier internal to U100 conditions the trigger signal and provides the proper signal level at pin 15 to drive the A/B Trigger Generator (U500, diagram 5). The pickoff point for the trigger signal is prior to the addition of the vertical position offset, so the position of the signal on the crt has no effect on the trigger operation. However, the pickoff point is after the DC Bal and Variable gain signals have been added to the signal so both of these functions will affect trigger operation.

Common-mode signals are rejected from the trigger signal by the circuitry composed of operation amplifier U450B and associated components. The inverting input of U450B (pin 6) is connected to the common-mode point between APO+ (pin 12) and TPO− (pin 15) of U100. Any common-mode signals present are inverted and applied to a common-mode point between R451 and R453 to cancel the signals from the differential output. A filter network composed of LR 180 and the built-in circuit board capacitor (5.6 pF) reduces trigger noise susceptibility. Trigger signals for options are obtained from J100.

The Channel 1 input signal used to provide the horizontal deflection for the X-Y displays is obtained from U100 pin 11. The components between pin 11 and the Horizontal Output Amplifier provide phase compensation of the signal. During instrument calibration, the delay produced by C115, C116, L115, R115, and variable capacitor C118 is matched to the 78-ns delay of the vertical delay line (DL100, diagram 6).

Channel 2 Preamplifier

Operation of Channel 2 Preamplifier U200 is nearly identical to that of the Channel 1 Preamplifier just described. The exceptions are that the output polarity of the Channel 2 signal may be either normal or inverted and that the signal obtained from the BPO+ output (pin 11) is conditioned differently for a different purpose than in the Channel 1 Preamplifier circuitry.

Inverting the Channel 2 signal for the CH 2 INVERT feature is accomplished by biasing on different amplifiers. The control data clocked into the internal control register from pin 22 sets up the necessary switching.

The Channel 2 BPO+ signal at U200 pin 11 provides an accurate representation of the Channel 2 signal at the rear-panel CH 2 OUT connector.

Channel 3 and Channel 4 Preamplifier

The functions provided by the Channel 3 and Channel 4 Preamplifier are similar to those provided by the Channel 1 and Channel 2 Preamplifiers. The single-ended CH 3 and CH 4 input signals are converted to differential signals, and vertical gain and vertical positioning are added to the output signals. Trigger pickoff signals are generated for both channels and are routed to the Trigger hybrid.

Channel 3 and Channel 4 gains may be either 0.1 volt per division or 0.5 volt per division. The logic levels of control bits applied to U300 pin 30 (GA3) and pin 31 (GA4) from Auxiliary Control Register U140 sets the gain of the Channel 3 and Channel 4 preamplifiers respectively. Vertical positioning of the Channel 3 and Channel 4 signals on the crt is controlled by the voltage levels applied to pin 29 (POS3) and pin 32 (POS4) from the front-panel CH 3 and CH 4 POSITION potentiometers (via MUX U2530 and sample-and-hold amplifiers U2430C and C2333 and U2430D and C2332).

Dc offsets in the output signal due to any tracking differences between the +5-V and the −5-V supply to

U300 are reduced by the tracking regulator circuit composed of U165A, Q190, and associated components. Operational amplifier U165A and Q190 is configured so that the output of voltage at the emitter of Q190 follows the −5-V supply applied to R198. This tracking arrangement ensures that the supply voltages are of equal magnitudes to minimize dc offsets in the output signals.

Scale Illumination

The Scale Illumination circuit consists of U130C, U130D, U130E, U130F, and associated components. The circuit enables the operator to adjust the illumination level of the graticule marks on the crt face plate using the SCALE ILLUM control.

Components U130C through U130F, depicted on diagram 4 as inverters, are actually Darlington transistor pairs. Figure 3-4 is a simplified illustration of the Scale Illumination circuitry, redrawn to show U130C through F as Darlington transistor pairs for the purpose of the following description.

Darlington transistors U130D and U130E control the current flow to scale-illumination lamps DS100, DS101, and DS102. Base drive current for U130D and U130E via R133 is set by the front-panel SCALE ILLUM pot R134. Voltage at the more negative end of the pot is set by the self-biasing configuration of U130F and R135. The voltage level established by these two components is two diode drops above ground (≈ 1.2 V) so that, at full counterclockwise rotation, the wiper voltage of the SCALE ILLUM pot will just match the turn-off point of U130D and U130E. The voltage at the other end of the pot is set by the collectors of U130D and U130E. As the SCALE ILLUM pot is advanced, the base drive to U130D and U130E increases, and the voltage on their collectors moves closer to ground potential. This increases the current through the scale-illumination lamps to make them brighter and produces some negative feedback to the base circuit through the SCALE ILLUM pot. Negative feedback stabilizes the base drive to U130D and U130E to hold the illumination level constant at the selected setting of the SCALE ILLUM control.

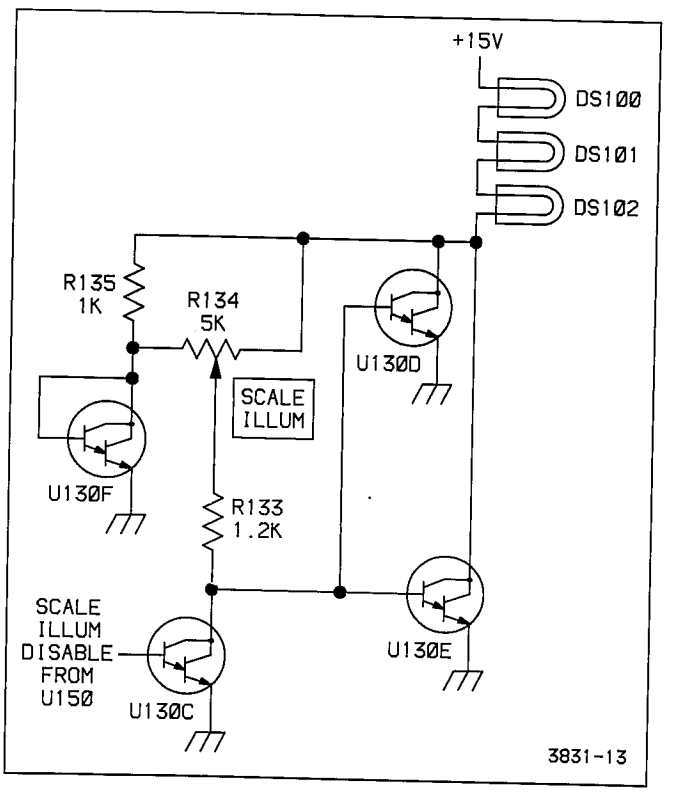


Figure 3-4. Scale illumination circuit.

During SGL SEQ display mode, the graticule is illuminated only once during the sequence for photographic purposes. In this mode, a HI is initially written to Auxiliary Control Register U150 (bit Q_H). This turns on U130C and shunts the base drive current of U130D and U130E to ground. At the point in the sequence when the graticule should be illuminated, the processor writes a LO to bit Q_H, and Q130C is turned off. This enables U130D and U130E to turn on the lamps to the illumination level set by the SCALE ILLUM pot.

DISPLAY SEQUENCER, TRIGGERS, AND SWEEPS

The Display Sequencer circuitry (diagram 5) controls and sequences the "analog-type" oscilloscope functions in real time, dependent on control data it receives from the Microprocessor. The A/B Trigger circuitry, under control of the Display Sequencer, detects when triggering requirements are met and initiates the appropriate sweep. The A Sweep and B Sweep circuits generate sweep ramps under control of the Display Sequencer when triggered by the A/B Trigger circuitry.

Display Sequencer

The Display Sequencer consists primarily of integrated circuit U650. This IC accepts analog and digital control signals from various parts of the instrument and, depending on the control data string clocked into its internal control register from the Microprocessor, will change control signals that it sends to other, signal-handling circuits.

In the course of developing waveform displays, the Display Sequencer selects one or more vertical channels, sets the trigger source, and selects the horizontal display

mode. In most cases, the trigger selection does not change after it has been set unless a front-panel trigger control is changed. An exception is that in VERT TRIGGER MODE, the trigger source tracks the sequencing of the vertical channels (unless AUTO LVL MODE, or CHOP VERTICAL MODE is also selected). Trigger source selection lines are changed only during trigger holdoff time between sweeps.

Fifty-five bits of serial data from the processor defining the instrument's operating sequence are applied to the Display Sequencer data input, pin 25. The data string is clocked into U650 to the internal control register by the processor-generated control clock applied to pin 24. The data string is organized in several fields, with each field defining the operating mode of one specific instrument function.

Display Sequencer U650 controls the various functions defined by the data fields by setting the levels of the associated control lines. The functions and controlling signal lines for each function are as follows:

VERTICAL DISPLAY SELECTION. CH 1, CH 2, CH 3, CH 4, ADD, and Readout Y signals are selected by the $\overline{VS1}$, $\overline{VS2}$, $\overline{VS3}$, and $\overline{VS4}$ control signals. See the Vertical Channel Switch description for further information.

HORIZONTAL DISPLAY SELECTION. A Sweep, B Sweep, CH 1 (for X-Y displays) and Readout X are selected by the $\overline{HSA}$ and $\overline{HSB}$ control signals. See the Horizontal Output Amplifier description for further information.

TRIGGER SOURCE SELECTION. CH 1, CH 2, CH 3, CH 4, ADD, Line, and a sample of the vertical output signal (for calibration purposes only) are selectable as the Trigger SOURCE by the $\overline{SR0A}$, $\overline{SR1A}$, $\overline{SR2A}$, $\overline{SR0B}$, $\overline{SR1B}$, and $\overline{SR2B}$ control lines (pins 28, 27, 29, 32, 31, and 30 respectively). See the A/B Trigger description for further information.

TRIGGER HOLDOFF. Sweep recovery time and the circuit initialization time required when front-panel controls are changed are controlled by the THO (trigger holdoff) signal.

DELTA TIME (Δt) DELAY SELECTION. DLY REF 0 or DLY REF 1 is selected by the $\overline{DS}$ (delay select) signal.

TRIGGER and SWEEP ACTIVITY (STATUS). The activity of the Trigger and Sweep circuits, as indicated by the $\overline{SGA}$, $\overline{SGB}$, $\overline{TSA}$, and $\overline{TSB}$ lines, is reported to the Microprocessor via the TSO (trigger status output) line when clocked by the $\overline{TSS}$ (trigger status strobe) signal.

INTENSITY CONTROL. The readout intensity, display intensity, and display intensity compensation are controlled by the BRIGHT output level.

DISPLAY BLANKING. Display blanking for CHOP VERTICAL MODE, Readout transitions, and front-panel control changes is controlled by the BLANK output.

READOUT CONTROL. The vertical selection, horizontal selection, and intensity controls are all set to their readout modes either at the end of an A Sweep ($\overline{SGA}$ goes HI) or in response to a readout request ($\overline{ROR}$) from the Readout circuitry (diagram 7). While in the readout mode, the BLANK control signal is driven by the readout blank ($\overline{ROB}$) input signal on pin 5 (also from the Readout circuitry). The readout active line ($\overline{ROA}$, pin 6), when set LO, tells the Readout circuitry that readout dots may be displayed if necessary. The $\overline{ROA}$ signal is always set LO at the start of the trigger holdoff time following sweeps, and it is held there until the holdoff time is almost over. This allows the majority of holdoff time to be used for displaying readout dots. The Display Sequencer will switch the $\overline{ROA}$ signal back to HI before the end of holdoff so that the readout display does not interfere with display of the vertical signal at the triggering event.

TRACE SEPARATION. Vertical separation between the A Sweep trace and the B Sweep traces (for alternate horizontal sweep displays), and between the reference B Sweep trace and the delta B Sweep trace (when delta time is selected in B Sweep only mode), is enabled by the TS1+TS2 output.

X10 HORIZONTAL MAGNIFICATION. Horizontal X10 magnification is controlled by the $\overline{MAG}$ output.

CALIBRATOR TIMING. The 5-Hz to 5-MHz drive signal to the Calibrator circuitry is provided by the CT output.

DELAY GATE OPERATION. Analog Switches U850B and U850C select the delay references for each sweep. Depending on the display mode and point in the display sequence, the DS control signal (U650 pin 40) routes one of the two analog delay references through U850B and U850C to the two sweep hybrids. The selected reference level is compared against the changing sweep ramp voltages to generate the delay gates that control each sweep's functions.

After an A Sweep has been initiated by a trigger, a delay gate circuit within U700 compares the A Sweep

ramp voltage to the selected delay reference. When the sweep ramp reaches the delay reference level, the DG (delay gate) output goes LO, enabling the B trigger portion of U500 and B Sweep hybrid U900. Then, when B triggering occurs (for TRIG AFT DLY mode), the A/B Trigger hybrid sets the $\overline{TGB}$ (trigger gate B) signal LO, initiating the B Sweep. In RUN AFT DLY mode, however, the $\overline{TGB}$ signal to U900 is held LO, and the B Sweep is initiated at the end of the A Sweep delay time when the A Sweep delay gate goes LO.

STATUS MONITORING. As the Display Sequencer controls the display system in real time, it continually monitors the trigger and sweep operations and updates the internal trigger status register accordingly. The Microprocessor checks the contents of this register every 3.3 ms to determine the current status of the trigger and sweep circuitry. The Microprocessor reads the trigger status register by generating a series of trigger status strobe ($\overline{TSS}$) pulses (U650 pin 19) to serially clock the contents of the register out to the TSO (trigger status output) line and onto the Data Bus (via Status Buffer U2220 on diagram 2). The system status information obtained by this check is used for AUTO LVL triggering, AUTO free-run triggering, detecting the completion of all sweeps in a SGL SEQ display, automatic measurement functions, and during instrument calibration.

INTENSITY CONTROL. The Display Sequencer controls the intensity for both sweep and readout displays. The analog levels at pins 22 and 23 determine the basic intensity level of the displays. Two internally generated DAC currents (developed by multiplying the IREF current at pin 20 by two processor-generated numbers stored internally) are added to the basic intensity level currents to produce the display intensity seen on the crt (see Table 3-1). The two DAC currents added to the INTENSITY current are dependent on sweep speed, number of channels being displayed, and whether or not the X10 MAG feature is in use. These added currents increase crt beam current and hold the display intensity somewhat constant under the varying display conditions. The resulting current is applied to Z-Axis Amplifier U950 (diagram 6) from the BRIGHT output of the Display Sequencer (pin 21).

To produce the intensified zone on the A Sweep trace for A intensified by B Sweep displays, an additional current is added to the crt drive signal by the Z-Axis Amplifier during the concurrence of the $\overline{SGAZ}$ and $\overline{SGBZ}$ (sweep gate A and B z-axis) signals.

The readout intensity (ROI) level, controlled from the front-panel READOUT INTENSITY pot (via MUX U2530 and sample-and-hold U2630A and C2732). The Microprocessor increases readout intensity when the pot is rotated

Table 3-1
Intensity Control

Type of Display	Horizontal Selects		Resulting Current at BRIGHT Output
	$\overline{HSA}$	$\overline{HSB}$	
X/Y	LO	LO	DI (display intensity) only
A Sweep	LO	HI	DI + A SWP DAC current
B Sweep	HI	LO	DI + B SWP DAC current
Readout	HI	HI	ROI (readout intensity) only

either direction from center. Minimum readout intensity current occurs at the midpoint of the READOUT INTENSITY pot rotation. The Microprocessor also detects to which side of center the READOUT INTENSITY control is set. Depending on the status received, the processor sets up the Readout circuitry (diagram 7) to display either all of the readout information or just the "delta type" readouts.

Blanking of the crt display during CHOP VERTICAL MODE displays or when switching between dot positions in the readout displays is controlled by the Display Sequencer's BLANK output (pin 3). When the signal is LO, the crt z-axis is turned on to the selected intensity level; when HI, the crt display is blanked.

READOUT CONTROL. The readout request signal ($\overline{ROR}$), the readout active signal ($\overline{ROA}$), and the readout blank signal ($\overline{ROB}$) control readout displays. During the first part of the holdoff time, up until one or two holdoff ramps before holdoff time ends (dependent on the sweep rate), the Display Sequencer sets the $\overline{ROA}$ signal line LO. While the $\overline{ROA}$ line is LO, the Readout circuitry may display readout character dots if necessary. During readout displays, the horizontal and vertical select signals ($\overline{HSA}$, $\overline{HSB}$, $\overline{VS1}$, $\overline{VS2}$, $\overline{VS3}$, and $\overline{VS4}$) are all set HI. This deselects the waveform-related sweep and deflection signals and gives display control to the Readout circuitry. While readout information or cursors are being displayed, the BLANK output signal (pin 3) is controlled by the readout blank ($\overline{ROB}$) signal from the Readout circuitry, and the readout intensity (ROI) signal pin (pin 23) controls the BRIGHT output level.

During holdoff, the Display Sequencer always sets the readout active ($\overline{ROA}$) line LO. As previously described, setting the $\overline{ROA}$ signal LO allows the Readout circuitry to display readout dots. In some settings of the SEC/DIV switch, with adequate trigger rates, holdoff time is provided for the Readout circuitry to display all the readout information without causing noticeable display flicker.

In those cases where the holdoff time is insufficient to prevent flicker, a portion of the Readout circuitry will request display control by setting the readout request ($\overline{\text{ROR}}$) signal LO. The Display Sequencer recognizes all readout requests immediately and switches the horizontal and vertical select lines to the readout display mode. The Readout circuitry displays one readout dot and then resets the readout request HI to switch back to the display of waveforms. Readout requests occur as required during sweep times to keep the readout display up to date. (See "Readout" description for further information).

TRACE SEPARATION. The TRACE SEP feature is used to position the alternate B Delayed Sweep trace downward from the A Sweep when Alternate Horizontal Display Mode (TURN-ALT) is active. It is also used when either the Δt or $1/\Delta t$ measurement function is used with B Sweep only displays. In the latter case, the TRACE SEP control vertically positions the trace(s) associated with the Δ control.

When the Display Sequencer determines that trace separation should be active, the LO TSIN level at pin 7 is routed to pins 9 and 8, the TS1 and TS2 outputs (connected together). This LO output turns off transistor Q600 (diagram 6), thereby enabling the trace separation voltage from the front-panel TRACE SEP pot (via MUX U2530 and sample-and-hold U2630C and C2631) to be applied to pin 42 of Vertical Output Amplifier U600. To disable the trace separation function, the Display Sequencer sets the TS1 + TS2 control line HI, turning on Q600 and shunting the trace separation signal to ground.

X10 MAG SELECT. The $\overline{\text{MAG}}$ (sweep magnifier) output (pin 39) drives the magnifier control input (pin 14) of Horizontal Output hybrid U800 and the select input (pin 9) of analog switch U860C (diagram 6). Analog switch U860C routes a magnifier gain-control voltage to the Horizontal Amplifier to set the horizontal gain for the X10 magnified displays.

CH 2 DELAY OFFSET. The $\overline{\text{VS2}}$ (vertical select, channel 2) output applied to analog switch U860B at pin 10 routes a calibrated offset voltage from sample-and-hold buffer U165D to both sweep hybrids when the Channel 2 vertical signal is being displayed. The offset voltage is used to eliminate the apparent propagation delay between the Channel 2 and the Channel 1 (or CH 2 and either one of the other channels). A step in the calibration procedure allows use of the front-panel Channel 2 Delay Offset feature to be either enabled or disabled. When enabled, the Channel 2 offset may be adjusted up to $\pm$500 ps (with respect to Channel 1) using the Δ control.

CALIBRATOR TIMING. The Calibrator timing signal (CT) from the Display Sequencer is generated by an internal counter. The counter divides the 5-MHz clock input at pin TC (timing clock) by a value that is a function of sweep speed. The resulting square-wave output signal drives the Calibrator circuit. For ease of sweep rate verification, the Calibrator signal provides a display of five complete cycles on the crt at sweep speeds from 100 ms per division to 0.1 μs per division. Below 100 ms per division, the Calibrator output frequency remains at 5 Hz; and above 0.1 μs per division, the Calibrator frequency remains at 5 MHz.

When chopping between vertical channels, the Display Sequencer adds a 200-ns skew at the end of some sweeps to desynchronize the chop frequency from the sweep speed (to prevent the sweep from locking onto the chop frequency). Due to this, the Calibrator signal has an irregular pulse repetition characteristic between sweeps. This will not be apparent when observing the Calibrator signal on the instrument crt since the skew is synchronized to the sweep, but may be observed when the Calibrator output signal is used with other instrumentation. The skew can be eliminated by setting the instrument to SGL SEQ Mode (to shut off the sweeps).

Holdoff Circuitry

The holdoff circuit, used to delay the start of a sweep until all circuits have recovered from the previous sweep, is made up of U165C, Q154, Q155, and associated components. Operational Amplifier U165C and capacitor C180 form a sample-and-hold buffer used to set the charging current for holdoff-ramp integrating capacitor C171. A control voltage from digital-to-analog converter (DAC) U2201 (diagram 2) via multiplexer U170 (diagram 4) is stored on C180. The stored voltage level sets the base voltage for both Q154 and Q155 via amplifier U165C. Transistors Q154 and Q155 form a current-mirror with nearly equal collector currents. Transistor Q154 is a current-to-voltage converter that provides negative feedback to U165C, setting loop gain. Transistor Q155 acts as a constant-current source that charges integrating capacitor C171, producing a linear holdoff ramp.

A comparator circuit in U650 detects when the ramp crosses a predefined threshold voltage (approximately +3 V). When the threshold is reached, pin 10 of U650 (HRR) goes LO and the integrating capacitor is discharged. At that same time, an internal counter that keeps track of the holdoff ramp cycles is incremented. The ramps continue to be generated and reset until the holdoff ramp counter has counted the number of ramp cycles defined by the sweep-rate-dependent holdoff data field stored in the Display Sequencer control register. At all sweep speeds except 5 ns per division, the count is at least two holdoff ramp cycles. The front-panel variable HOLDOFF control affects holdoff time by varying the HOLDOFF control voltage to

U165C (from the DAC), changing the charging rate of integrating capacitor C171.

When holdoff time requirements are met (determined by the number of ramps counted), the Display Sequencer sets the THO (trigger holdoff) signal LO. This enables both the A Sweep hybrid (U700) and the A Trigger circuitry in U500. The Trigger circuit begins monitoring the selected trigger source line and, when a triggering event is detected that meets the triggering requirements defined by the stored control data, initiates the A Sweep and sets the $\overline{TSA}$ (trigger status, A Sweep) line to Display Sequencer U650 LO (indicating that the A Sweep has been triggered).

As the A Sweep circuit (U700) responds to the trigger, it sets the $\overline{SGA}$ (sweep gate A) line LO (via U980A) indicating that an A Sweep is in progress. After the sweep has run to completion, U700 sets the $\overline{SGA}$ line HI signaling the end of sweep. The Display Sequencer then sets the THO line HI, resetting A/B Trigger hybrid U500 and A Sweep hybrid U700 in preparation for the next sweep.

Triggers

The A/B Trigger hybrid (U500) and associated circuitry select the triggering signal source for each horizontal sweep as directed by the Display Sequencer. When the proper triggering criteria to initiate a sweep are detected, a triggering gate signal is produced to start the selected sweep.

Control data from the processor defining trigger mode, coupling, and slope parameters for each trigger is clocked into two storage registers internal to U500 by the A TRIG CLK signal on pin 23 ($\overline{CCA}$) and the B TRIG CLK signal on pin 47 ($\overline{CCB}$). The Display Sequencer selects the A trigger source with the $\overline{SR0A}$, $\overline{SR1A}$, and $\overline{SR2A}$ signal lines; the B trigger source is selected using the $\overline{SR0B}$, $\overline{SR1B}$, and $\overline{SR2B}$ signal lines. Table 3-2 illustrates trigger source selection.

To initiate the A Sweep, the trigger hybrid compares the selected signal to the analog trigger level input at pin 13, the TLA (trigger level A). B trigger signals are compared to the TLB (trigger level B) signal at pin 37 when trigger B Sweeps are required. When the proper trigger signal is detected, U500 outputs a trigger gate ($\overline{TGA}$ or $\overline{TGB}$) to the appropriate sweep circuit to initiate that sweep.

When an A Sweep is initiated, the trigger-status line ($\overline{TSA}$) (trigger status A, U500 pin 20) goes LO to signal the Display sequencer that a trigger has occurred. Until the sweep is completed, the $\overline{TGA}$ signal on pin 18 (or $\overline{TGB}$ signal on pin 42 for B Sweeps) remains LO. After the A Sweep is completed, the A Sweep Gate ($\overline{SGA}$) from A Sweep hybrid U700 (via U980A) will go HI, causing the Display Sequencer to set its THO (trigger holdoff) line (pin 13) HI. This resets the sweep hybrid and the trigger hybrid in preparation for the next trigger event.

Table 3-2
Trigger Source Selection

Select Inputs			Trigger Source
$\overline{SR2A(B)}$	$\overline{SR1A(B)}$	$\overline{SR0A(B)}$	
H	H	L	CH 1
H	L	H	CH 2
H	L	L	ADD
L	H	L	CH 3
L	L	H	CH 4
H	H	H	LINE (or BWLB)[a]

[a]During calibration routines from the Diagnostic Monitor.

The B Trigger Holdoff input (THOB, U500 pin 39) is held HI (keeping the B Trigger reset) until the A Sweep Delay Gate (DG, U700 pin 41) goes LO (see the following A Sweep description). When DG goes LO, the B Trigger portion of U500 is enabled. The B Sweep Trigger functions in a manner similar to that of the A Sweep Trigger just described. During a parametric measurement, the THOB line may be driven by either A Sweep Delay Gate, or BHO from the measurement PAL, U975. If CNTL1 is LO, THOB is driven by A Sweep Delay Gate through the buffer transistor Q741. If CNTL1 is HI, Q741 is held off by Q742 and THOB is driven by BHO.

A Sweep

When properly triggered, the A Sweep circuit generates linear sweep ramps of selectable slopes. When amplified, these ramp signals horizontally sweep the crt beam across the face of the crt. The A Sweep circuitry consists of U700, Q709, Q710, Q741, U910B, U980A, and associated components.

The A Sweep ramp signal is derived by charging one of several selectable capacitors from a programmable constant-current source. Capacitor selection depends on the sweep-rate-dependent control data (CD) on pin 29 that is clocked into A Sweep hybrid U700 by the A SWP CLK on pin 28 ($\overline{CC}$). This sweep-rate data causes some internal logic to select either hybrid-mounted capacitors CT0 or CT1 or capacitor C708 at the CT2 (timing capacitor two) pin. An additional capacitor, C709, may be selected (via Q709 and Q710) if the control data asserts the TCS (timing capacitor select) signal on pin 9. TCS will be HI for A

Sweep speeds slower than 1 ms per division. Capacitor C707 and associated circuitry form a linearity compensation circuit.

The constant current to charge the selected capacitor is derived from the DAC-controlled voltage, A TIM REF (A timing reference), generated on the Control Board. The ITREF input (U700 pin 24) is held at zero volts by an internal programmable current-mirror circuit at that input (see Figure 3-5). The A TIM REF voltage is applied to the current mirror via series resistors R723 and R724 to establish the input reference current (ITREF). The output of this current mirror is related to the input reference current by a multiple "M" that is set by a control data field stored in the internal control register of U700. The derived output current (M x ITREF) is connected to another programmable current-mirror circuit, U910B, external to the hybrid. The output of U910B provides the actual charging current and is a control-data-selected multiple of the M x ITREF current.

At the time of calibration, the processor will vary the ITREF input current until the slope of the output ramp for specific current-mirror/timing capacitor combinations is precisely set. The values of A TIM REF at these settings allow the processor to precisely calculate the characteristics of the current-mirror circuits at their various multiplication factors and the charging characteristics of the timing capacitors. These values are stored as calibration constants in nonvolatile memory (RAM U2460, diagram 1).

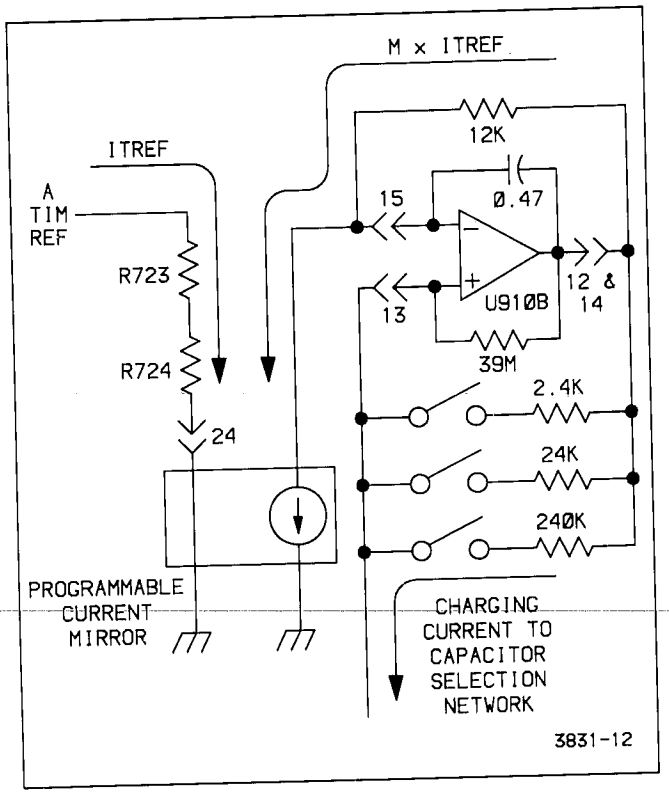


Figure 3-5. Sweep generator.

Once the calibration constants are set, any setting of the SEC/DIV switch causes the Microprocessor to recall the associated calibration constants from RAM. The processor then calculates the proper value of A TIM REF based on the selected timing capacitor and the current-mirror multiplication factors.

If the SEC/DIV VAR control is out of the calibrated detent position, the processor will decrease the A TIM REF voltage from the maximum, in-detent value by an amount proportional to the position setting of the VAR control. At the maximum, fully counterclockwise setting of the VAR control, the ITREF current is one-third that of the normal, in-detent current.

For A Sweep hybrid U700 to initiate a sweep at the selected rate, the $\overline{\text{AUXTRIG}}$ (auxiliary trigger) input (pin 3), the THO (trigger holdoff) line from the Display Sequencer (on pin 1), and the $\overline{\text{TRIG}}$ (trigger) line from the trigger hybrid (on pin 2) must all be LO. With these three inputs LO, the A SWEEP ramp begins, and the sweep gate ($\overline{\text{SG}}$) output (pin 45) goes LO. The buffered sweep gate signal ($\overline{\text{SGA}}$) at the output of U975 returns to the Display Sequencer through R981 to indicate that the A Sweep is active. The sweep gate signal is used by various other circuits for their timing activities and is held LO until the A SWEEP ramp ends. The buffered (negative) sweep gate is inverted and routed to the rear-panel A GATE output connector via U975.

Diodes CR752 and CR753 and associated components form a charging network that permits delaying the timing of the end-of-A-Sweep gate signal ($\overline{\text{SGAZ}}$) for B Sweep displays. For normal A Sweep operation with the $\overline{\text{SGBZ}}$ signal HI, the SGAZ signal will end quickly, since the capacitance associated with Z-Axis hybrid U950 input (diagram 6) will be charged positively through both R753 and R754. For B Sweep operation ($\overline{\text{SGBZ}}$ is LO), the end of the SGAZ gate signal will be delayed slightly (with respect to the normal sweep gate) since charging of the Z-Axis input capacitance will be at a slower rate through R754 only. This allows more of the B Sweep to be displayed than would otherwise be possible.

The A Sweep Delay Gate (DG) signal acts as the trigger holdoff (THO) signal for the B Sweep and the B Trigger circuitry. It is generated by comparing the A SWEEP ramp voltage to the selected delay reference (DR) level from analog switch U850C. As the ramp voltage crosses the delay reference level, the delay gate (DG) output signal goes LO, removing the HI THO level to the B

Sweep. This enables the B Sweep to run immediately in RUN AFT DLY B Trigger Mode or, when in TRIG AFT DLY B Trigger Mode, enables the B Sweep to run when a B triggering event occurs.

The BDCA (A Sweep bypass-delay comparator) input (U700 pin 39) is a data bit from Auxiliary Control Register U140 (diagram 4) that, when HI, sets the A Sweep DG output LO at the beginning of the A Sweep. This enables the B Sweep to run immediately at the start of the A Sweep and is used for calibration purposes and for options.

The capacitive load (part of the etched-circuit board) at the RDA (retrace delay adjust) input (U700 pin 4) is used to delay the retrace of the sweep until the Z-Axis drive is fully turned off in response to the SGAZ gate going HI. This delay prevents any part of the retrace from being seen.

B Sweep

Operation of B Sweep hybrid U900 is similar to that just described for the A Sweep with the following exceptions: the THO input (and thus sweep enabling) is controlled by the A Sweep hybrid or the Measurement PAL, U975, and not the Display Sequencer (see the preceding A Sweep description). The timing capacitor select output, TCS, is not used, and only three timing capacitors are selectable (two on the B Sweep hybrid at CT0 and CT1 and one externally at CT2).

Calibrator

The Calibrator circuit, composed of Q550, U165B, U550A, B, C, and D, and associated components, generates a square wave output of precise amplitude and frequency characteristics. The CALIBRATOR signal provided at the front-panel output connector is useful for adjusting probe compensation and verifying VOLTS/DIV, SEC/DIV, and Δt (delta time) calibration. Output frequency is controlled by the Display Sequencer and is set to display five cycles across the ten crt graticule divisions at sweep speed settings from 100 ns per division to 100 ms per division. This feature allows quick and easy verification of the sweep rates. The Calibrator circuitry is essentially a voltage regulator that is alternately switched on and off, producing the square-wave output signal.

When the timing signal (CT) from the Display Sequencer to the base of U550D is LO, U550C (configured as a diode) is forward biased, shunting bias current away from Q550, keeping it turned off. When transistor Q550 is off, the front-panel CAL OUT connector is pulled to ground potential through R558, setting the lower limit of the CALIBRATOR output signal.

As the CAL signal goes from LO to HI, the emitter of U550D is pulled HI to reverse bias U550C. Bias current for Q550 is established, and the transistor is turned on. The voltage at the emitter of Q550 rises to a level of +2.4 volts, determined by the voltage regulator composed of U165B, U550A, U550B, and associated components. This regulated level is applied to the front-panel CALIBRATOR connector through a voltage-divider network composed of R557 and R558. This produces an output voltage of 400 mV with an effective output impedance of 50 Ω.

Since the frequency of the CALIBRATOR signal is controlled by the same divider chain that controls operation of the vertical chopping rate, the intentional 200-ns shift added to the chop signal at the end of some sweeps (to desynchronize the chopping rate from the sweep rate) shows up on the CALIBRATOR signal as an irregular-width pulse. This shift is not apparent when viewing the CALIBRATOR signal on the instrument providing the signal (since the skew occurs during sweep-retrace time), but it should be taken into account when using the CALIBRATOR signal with other instrumentation. The skew can be eliminated from the signal by setting the instrument TRIGGER MODE to SGL SEQ (to shut off the sweeps).

PARAMETRIC MEASUREMENTS

The VOLTS Parametric Measurement is made using the same methods and circuitry that is used in the Auto Level trigger mode to find the peak voltages. The accuracy of the VOLTS measurement is based on the accuracy of the trigger level and the DC balance of the instrument.

All of the time-based Parametric Measurements use the A and B Sweep gates and delay gates as the basis for the measurements.The measurement PAL, U975, controls the signal flow while in the Parametric mode. The measurement flip-flop, U980B, reports the state of a variety of conditions to the SLIC through the $\overline{\text{SGB}}$ line. The SLIC data is read by the processor system and used to compute the desired measurement.

VERTICAL CHANNEL SWITCH AND OUTPUT AMPLIFIERS

The Vertical Channel Switch (diagram 6) selects the signal source for vertical deflection of the crt beam. The Vertical, Horizontal, and Z-Axis output amplifiers provide the signal amplification necessary to drive the crt.

Vertical Channel Switch

The Vertical Channel Switch consists of hybrid Channel Switch U400, that selects one of the vertical signals for application to the Vertical Output Amplifier, and a combined switch/amplifier circuit that converts the single-ended readout vertical signal into a differential signal for application to the Channel Switch.

Table 3-3
Vertical Display Selection

Select Inputs				Vertical Display
$\overline{VS1}$	$\overline{VS2}$	$\overline{VS3}$	$\overline{VS4}$	
L	H	H	H	CH 1
H	L	H	H	CH 2
L	L	H	H	ADD
H	H	L	L	CH 3
H	H	H	L	CH 4
H	H	H	H	Readout (Y)

Channel selection is controlled by the Display Sequencer $\overline{VS1}$ through $\overline{VS4}$ signals applied to the vertical channel selection pins (pin 24, pin 25, pin 13, and pin 14 respectively). (See Table 3-3 for the Vertical Display Selection.) When a vertical select line is LO, the associated input signal pins are connected to the differential output (+OUT, pin 11 and −OUT, pin 3). The CH 5 input signal (Readout Vertical) is added to the output whenever both the $\overline{VS3}$ and $\overline{VS4}$ select signals are HI but will only contain readout information when the readout select logic (U975A and U975C) detects that the Display Sequencer has set both the Horizontal Select signals ($\overline{HSA}$ and $\overline{HSB}$) HI (readout selected).

READOUT SWITCH/AMPLIFIER. Transistors U485A, U485B, U485C, U485D, and U475C, along with their associated components, make up an analog switch circuit that routes either the readout vertical signal at the base of U485A or the ground reference at the base of U485C to the output at the emitter of U475C. The signal selected depends on the complementary voltages applied to the emitter junctions of the two emitter-coupled transistor pairs, U485A and B and U485C and D. The selection voltages are developed by voltage-divider networks on the complementary logic outputs of U975A and U975C.

When readout information is to be displayed, the horizontal select inputs to U980B and U980C go HI and the output of NAND-gate U975C goes LO. The LO applied to the divider network of R498, R484, and R471 pulls the anode of CR484 low enough to reverse bias it. This forward biases the emitter-coupled pair U485A and B via R483. NAND-gate U975A inverts the LO and applies a HI to the junction of R497 and R485. The HI forward biases CR485, and the emitters of U485C and D are pulled to a level in excess of +2 V, reverse biasing the transistor pair. With U485C and D reverse biased, the ground reference level at the base of U485C is isolated from the output, while the readout vertical information is allowed to pass through the forward-biased transistor pair.

When readout information is not being displayed, a HI is present at the output of NAND-gate U975C. The HI forward biases CR484 and, when inverted by U975A, reverse biases CR485. With the biasing conditions reversed, the transistor pair of U485C and D becomes forward biased and U485A and B becomes reversed biased. The ground reference level present at the base of U485C is coupled to the output, while the readout vertical signal is isolated.

The output signal (either the readout vertical signal or the ground reference level) is applied to the CH5+ input of Channel Switch U400 via R495 and R412. The inverting amplifier circuit composed of U475A, U475B, U475D, and associated components inverts the readout vertical signal for application to the CH5− input. The amplifier is an inverting unity-gain configuration with transistors U475A and U475B connected as an emitter-coupled pair. The base of U475A is referenced to ground through R482. The base of U475B is pulled to the same level by the negative feedback from emitter-follower U475D through R478. The noninverted signal is applied to the base of U475B through R492 and will attempt to increase or decrease the current to the base of U475B, depending on the amplitude and polarity of the signal. However, the negative feedback from the collector of U475B (via U475D and R478) will hold the base of U475B at the ground reference level. The feedback current through R478 develops a voltage drop across R478 that is equal in amplitude but opposite in polarity to the noninverted vertical readout signal. The inverted readout signal is applied to the Channel Switch on pin 2 (CH5−) via R476 and R402.

The HF ADJ (high-frequency adjust) potentiometer R417 (2455B only) and resistor R416 (connected to pin 16) adjust the high-frequency response of the Channel Switch hybrid.

Vertical Output Amplifier

Vertical Output Amplifier U600 is a hybrid device that provides the final amplification of the selected vertical signal, raising it to the level required to drive the crt deflection plates. Vertical deflection signals from the Vertical Channel Switch are delayed approximately 78 ns by Delay Line DL100. This delay allows the Sweep and Z-Axis circuits to turn on before the triggering event begins vertical deflection of the crt beam, thereby permitting the operator to view the triggering event. The bridged-T network, composed of inductors and capacitors built into the circuit board, corrects phase-distortion introduced by the delay line. The RLC networks connected between the output pins of U400 are adjusted during calibration to obtain the correct overall high-frequency response of the vertical deflection system. The vertical signal from the Delay Line is applied to pins 10 and 3 of U600. The RL network connected between pins 8 and 5 (COMPA and COMPB) of U600 compensates the signal for the skin-effect losses associated with the delay line.

Amplifier gain and vertical centering are adjusted by R638 and R639 respectively, primarily to match the

amplifier hybrid to the crt installed in the instrument. The Dynamic Centering circuit sinks an intensity-dependent correction current away from the vertical centering input at pin 39. The correction signal holds the vertical centering stable over a wide range of varying display intensities. Readout jitter adjustment pot R618 is used to minimize thermal distortion in the output amplifier to reduce jitter in the display readout.

The vertical output signal at pins 28 and 33 of U600 (OUT A and OUT B) is applied to the vertical deflection plates of the crt (diagram 8) via L628 and L633. The deflection plates form a distributed-deflection structure that is terminated by a hybrid resistor network. One element of the terminating network is an adjustment potentiometer used to match the network impedance to that of the crt.

BANDWIDTH LIMITING. Bandwidth limiting coils L644 and L619, along with capacitors built into U600, form a three-pole filter used to roll off high-frequency response of the Vertical Output amplifier above 20 MHz. To limit the vertical bandwidth, the $\overline{\text{BWL}}$ (bandwidth limit) input to U600 (pin 16) is pulled LO. It may be set LO either by the BWL control data bit from Auxiliary Control Register U140 (diagram 4) when the operator selects the Bandwidth Limit feature or automatically by the output of NAND-gate U975A in the Vertical Channel Switch circuitry (via CR616) when the readout is being displayed.

TRACE SEPARATION. The voltage applied to the TS (trace separation) input of U600 (pin 42) is used to offset the output levels to vertically shift the position of the trace on the crt. During normal sweep displays, TS1 + TS2 signal applied to the base of Q600 by the Display Sequencer (diagram 5) is HI, and the transistor is turned on. The TRACE SEP level at the junction of R642 and CR600 is shunted to ground, and no offsetting at the output signal will occur. For those displays in which trace separation should occur, the Display Sequencer switches the base of Q600 to ground level to turn off the transistor. The trace separation level set by front-panel TRACE SEP control R3190 (via MUX U2530 and sample-and-hold circuit U2630C and C2631) is applied to the TS input of U600, and a corresponding offset of the displayed trace will occur.

BEAM FIND. As an aid in locating off-screen or overscanned displays, the instrument is provided with a beam-finding feature. When the front-panel BEAM FIND button is pushed, the beam-find input pin (BF, pin 15) of U600 will be pulled HI. While BF is HI, the dynamic range of Vertical Output Amplifier U600 is reduced, and all deflected traces will be held to within the vertical limits of the crt graticule.

Also, the activation of the BEAM FIND switch is detected by the microprocessor during its normal Front-Panel Switch Scanning. When detected, this generates a User Request SRQ if option 10 is installed.

OUTPUT PROTECTION CIRCUIT. A current-limit circuit composed of transistors Q623 and Q624 protects the Vertical Output Amplifier from a short-circuited output or a bias-loss condition. Either of these fault conditions will cause excessive current to flow into pins 30 and 31 of U600. Current in FET Q624 is limited to the IDSS current, so the voltage at pins 24, 30 and 31 will drop. This decreases the forward bias on pass-transistor Q623 and lowers the voltage at pin 23 of U600 enough to provide some degree of protection for the device.

Horizontal Amplifier

The Horizontal Amplifier circuitry consists of a Horizontal Output Amplifier U800, a unity-gain buffer amplifier made up of the five transistors in U735, and associated components.

UNITY-GAIN BUFFER AMPLIFIER. The amplifier circuit composed of U735A, B, C, D, and E along with their associated components, form a unity-gain amplifier that buffers the ramp signal from A Sweep Generator U700 to the Horizontal Output Amplifier. Transistors U735C and D form a differential pair with the negative excursion of their emitters limited to −5 V (clamped by U735E). Negative feedback from the collector of U735C to its base is via emitter-followers U735A and B (in parallel) which drive the A Sweep input (pin 18, A+) of Horizontal Output Amplifier U800.

HORIZONTAL OUTPUT AMPLIFIER. Integrated circuit U800 provides the final amplification of the selected horizontal-deflection signal required to drive the crt. One of the single-ended input signals applied to the four input pins is converted to a differential-output signal at the output pins of the amplifier. The four deflection signals to U800 are: the A sweep (pin 18, A+), the B Sweep (pin 16, B+), the Readout Horizontal signal (pin 17, RO) and the Channel 1 signal (used for horizontal deflection of the X-Y displays) at pin 20, the X+ input pin. Signal selection is done by an internal channel switch and is controlled by the $\overline{\text{HSA}}$ (horizontal select A) and $\overline{\text{HSB}}$ (horizontal select B) signals from the Display Sequencer (see Table 3-4).

Switching between unmagnified (X1) gain and magnified (X10 gain) is also controlled by signals from the Display Sequencer. For normal horizontal deflection, the $\overline{\text{MAG}}$ signal on pin 14 of U800 is HI, and the gain of the output amplifier produces normal sweep deflection. Precise X1 deflection gain is set by adjusting X1 Gain pot R860.

Table 3-4
Horizontal Display Selection

Control Level		Selected Signal
$\overline{\text{HSA}}$	$\overline{\text{HSB}}$	
H	H	Readout (X)
H	L	B Sweep Ramp
L	H	A Sweep Ramp
L	L	X Input (from CH 1)

When the X10 MAG feature is selected, amplifier gain for the magnified sweeps is increased by a factor of 10. The $\overline{\text{MAG}}$ signal from the Display Sequencer goes LO when magnified sweep is to be displayed. This switches the amplifier gain and switches analog switch U860C from the X1 position to the X10 position. Amplifier gain in the magnified mode is adjusted by adding or subtracting a small bias current using X10 Gain control R850. Dc offsets in the amplifier and crt are compensated for, using Horiz Centering pot R801 to precisely center the display. An intensity-dependent position correction signal, used to hold the horizontal centering stable over a wide range of varying display intensities, is also added at this point by the Dynamic Centering circuitry.

Timing and linearity of the sweep is affected by the amplifier transient response; and Trans Resp pot R802, connected to pin 2, is adjusted during calibration for optimum accuracy of the high-speed sweeps.

As with the Vertical Output Amplifier, the Beam Find feature reduces the dynamic range of the Horizontal Output Amplifier. While the front-panel BEAM FIND button is pressed in, a HI is placed on U800 pin 15 via pull-up resistor R615, and the horizontal deflection is reduced, moving horizontally off-screen displays to within the graticule viewing area.

Z-Axis Amplifier

Z-Axis Amplifier U950 turns the crt beam off and on at the desired intensity levels as the oscilloscope goes through its display sequence. The BRIGHT (brightness) signal applied to U950 pin 44 from the Display Sequencer U650 (diagram 5) is amplified to the level required to drive the crt control grid (via the DC Restorer circuitry) and sets the crt beam intensity. The BLANK input signal applied to U950 pin 5, also from the Display Sequencer, blanks the trace during sweep retrace, chop switching, and readout blanking by reducing the VZ OUT signal to a blanked level. Sweep gate z-axis signals ($\overline{\text{SGAZ}}$ and $\overline{\text{SGBZ}}$) from the A Sweep and B Sweep hybrids (U700 and U900) respectively, (diagram 5) are applied to the Z-Axis Amplifier on pins 4 and 3. These signals turn the beam current on and off for the related displays and, when used in conjunction with the BLANK signal on pin 5, enable the sweeps to be blanked while still allowing the Readout circuitry to blank and unblank the crt for the readout displays.

Control signals applied to U950 pin 48, pin 2, and pin 1 ($\overline{\text{HSA}}$, $\overline{\text{HSB}}$, and TXY respectively) switch some internal logic circuitry to enable or disable different input signals for the various types of displays. Table 3-5 illustrates the effects of the various input signals on the output signal for different combinations of $\overline{\text{HSA}}$, $\overline{\text{HSB}}$, and $\overline{\text{TXY}}$.

The Z-Axis hybrid has an internal limiter circuit that prevents the crt from being damaged during high-intensity, high-repetition-rate displays. A signal representative of the intensity setting and the sweep repetition rate is integrated on C957 and results in a control level at pin 7 of U950 used to limit intensity of the crt beam. Maximum Grid drive is controlled by R949 on U950 pin 9.

Focus tracking for intensity (VZ OUT) level changes is provided by the VQ OUT (quadrapole output voltage) signal at pin 22 of U950. The VQ OUT signal varies the focusing voltages (and thus the focusing strength) of two quadrapole lenses in the crt (diagram 8). The VQ OUT signal is related to the VQ OUT level exponentially and provides the greatest auto-focus control at high intensity levels. Gain of the VQ OUT signal is set by the High-Drive Focus adjustment, R1842. The VQ OUT signal also drives the Dynamic Centering circuit and holds the display position stable during wide-range intensity level changes.

Dynamic Centering

The circuit composed of U3401, U3402, and associated components generates compensating signals to offset positioning effects that occur in the crt when the intensity is varied over a wide range. The VQ OUT signal from Z-Axis Amplifier U950 is exponentially proportional to the display intensity and dynamically controls the intensity-dependent offsets.

Dynamic Centering adjustment pots R3401 and R3407 set the gain and polarity of the signals at their related outputs by varying the current in the emitter circuit of one of two emitter-coupled pairs of transistors. Adjusting the bias level, at either pin 4, above $\simeq -10.6$ volts (determined by R3410 and R3411 at the complementary inputs, pins 1) will generate an inverted signal, while adjusting the

Table 3-5
Blanking and Intensity Control Selection

Control Inputs			Intensity Affected By	Blanking Affected By	Typical Display
TXY	$\overline{HSA}$	$\overline{HSB}$			
X[a]	H	H	BRIGHT (RO level)	BLANK	Readout
X	H	L	BRIGHT, Z EXT	BLANK, $\overline{SGAZ}$, $\overline{SGBZ}$	Delayed Sweep
X	L	H	BRIGHT, $\overline{SGBZ}$,Z EXT	BLANK, $\overline{SGAZ}$	Main Sweep
L	L	L	BRIGHT, $\overline{SGBZ}$, Z EXT	BLANK	X-Y
H	L	L	BRIGHT, $\overline{SGBZ}$, Z EXT	BLANK, $\overline{SGAZ}$	X-Y

[a]X = State doesn't matter.

bias levels below −10.6 volts will cause a noninverted signal. Amplitude of the resulting signal is dependent on how far from the −10.6-volt reference the bias is set. The output signal is added or subtracted from the position voltage applied to the Vertical and Horizontal Output Amplifiers. Both pots are adjusted so that position shifts due to display intensity variations are minimized.

READOUT

The Readout circuitry (diagram 7) is responsible for displaying the alphanumeric readout characters in the crt. An eight-bit character code specifying each character (or cursor segment) to be displayed is written from the Microprocessor to a corresponding location in the Character RAM U2920 (a 8K-x-8-bit, random access memory integrated circuit). Each of the following 128 locations in the RAM, address locations 0 through 63 for the first and fourth readout lines and 128 through 191 for the second and third readout lines, corresponds to one of the 128 possible character locations in the crt readout display (see Figure 3-6). The next 128 RAM locations, address locations 64 through 127 for the first and fourth readout lines and 192 through 255 for the second and third readout lines, are used to store cursor segment information for the display of the ΔV and Δt measurement cursors. The eight-bit character code written to each location in RAM points to a block of addresses in Character ROM U2930. This block in the ROM contains the dot-position information for the specific character to be displayed at the associated crt position.

Each character is made up of zero (for a space character) or more dots displayed in an eight-wide by sixteen-high dot matrix. Specific blocks of ROM addresses contain all the X-Y offset coordinates for the dots in a particular character in the readout. The coordinates are referenced to the lower-left corner of the character dot matrix. Each individual data byte in the block of ROM addresses contains both the X and the Y coordinates for one dot of the associated character.

To display a character, a combination of the character position on the crt (the RAM address) and the byte of X-Y position data from Character ROM U2930 (relative to that character position) is applied to Horizontal and Vertical DAC (digital-to-analog converters) circuits, U2910 and U2905 respectively. In these circuits, the X-Y position data is converted to analog deflection signals used to position each dot in the crt readout display. Each of the position bytes are read from the block of ROM defining the character under control of the readout timing and sequencing circuitry. The resulting dots, when displayed in sequence, form the character at the proper location on the crt.

Readout I/O

The Readout I/O circuitry, composed of U2860, U2865, U2960, and associated components, provides the interface between the Microprocessor and the Readout board. Two types of data, Readout mode data and character data, are written to the Readout board serially via data bus line BD0.

STORING A CHARACTER. Displaying a character starts with serially clocking 16 character data bits into a 16-bit shift register formed by registers U2960 and U2860. The $\overline{ROS1}$ strobe (readout strobe one) from the Address Decode circuitry (diagram 1) is the clocking signal. The first eight bits of the loaded data indicate the character to be displayed, while the last eight select the location on the crt that the character is to be displayed.

On positive-going transitions of the $\overline{ROS1}$ strobe, the data bit present on the BD0 data line is shifted into the first latch of character address register U2960. The following negative-going edges of the $\overline{ROS1}$ strobe are inverted

by U2965A to produce a positive transition that shifts the data bit present at U2960 pin 9 (Q_{SH}) into U2860. After 15 $\overline{ROS1}$ strobes have occurred, seven bits of character data are latched into U2860, and the eighth character bit and seven of the character address bits are latched into character address register U2960 (though they have not been shifted into their correct positions for addressing the RAM).

At this point, the last character bit remains to be shifted into the registers, but the operating mode must be set up first to ensure correct operation upon shifting in the final bit. The eight bits of mode data are shifted into the mode control register U2865 by the $\overline{ROS2}$ strobe. Bit Q_4 ($\overline{WRITE}$), along with the $\overline{ROS2}$ and the R/ $\overline{W}$ DLYD signal are applied to the RAM enabling circuitry and determine when new character information will be written into the Character RAM. With U2865 loaded with the mode data, a final $\overline{ROS1}$ strobe clocks the eighth bit of character data from U2960 to U2860 on the negative edge, and the positive edge of the strobe clocks the eighth character address bit into U2960.

With control bit Q_4 from U2865 LO, the outputs of U2860 are enabled and the eight bits of character data (CD0 through CD7) are written in parallel into the Character RAM at the location selected by the eight-bit address from U2960. Register U2960 is enabled only when the Readout is not displaying characters (the REST signal at pin 15 of U2960 is HI).

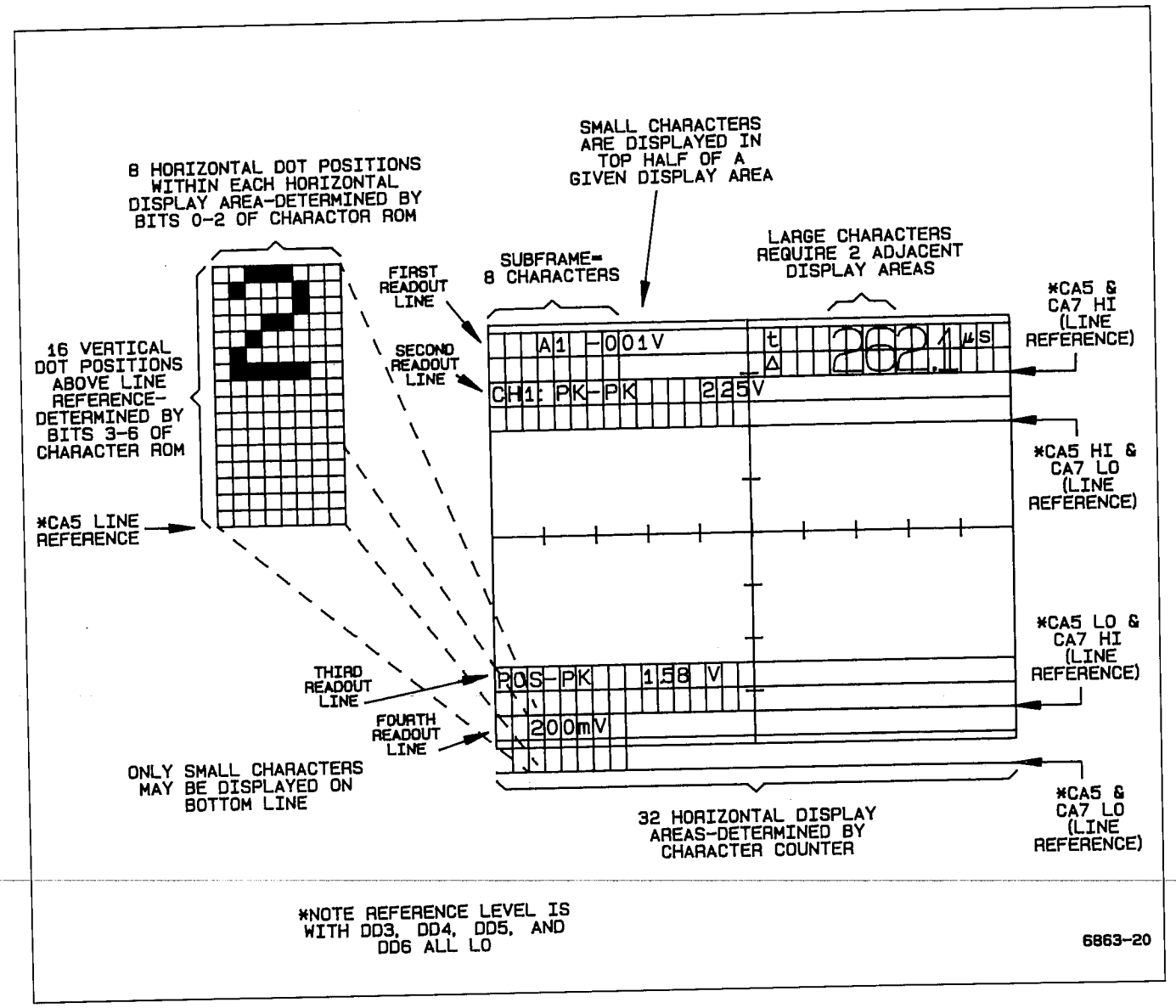


Figure 3-6. Developing the readout display.

The character data register U2860 also provides a means for the Microprocessor to read data from the Character RAM for partial verification of Readout circuit operation (during the power-up tests). The eight bits of parallel data from the Character RAM location selected by character address register U2960 are loaded into U2860 by setting bit Q_3 of mode control register U2865 LO. Inverter U2965C converts the LO to a HI and applies it to character-register U2860 at pin 1. The HI on pin 1, in combination with the fixed HI on pin 19 of U2860, switches the character register to the Parallel Load mode. The next positive transition of the $\overline{ROS1}$ strobe loads the eight data bits placed on the CD0 through CD7 bus lines into the register in parallel. Bit Q_3 is then returned HI, and the next positive transition of the $\overline{ROS1}$ strobe shifts the Q_A bit to pin 8 (Q_A'), the RO DO (readout data out) line. Seven more $\overline{ROS1}$ strobes shift the remaining seven bits of character data out onto the RO DO line to Status Buffer U2220 (diagram 2) to be read, one at a time, by the processor.

Character RAM

Character RAM U2920 provides temporary storage of the readout character selection data. This character data is organized as 256 eight-bit words that define the character that should be displayed at any given readout position on the crt. Cursor information is also stored in U2920 when cursors are to be displayed.

RAM locations may be addressed either from the Readout I/O stage by character address register U2960, as previously described, or by the Character Counter stage. Each of the following 128 address locations corresponds to a specific readout location on the crt. Address locations 0 through 63 correspond to the first and fourth readout lines and 128 through 191 to the second and third readout lines. The next 128 address locations store cursor information. Address locations 64 through 127 correspond to the first and fourth readout line storage and 192 through 255 to the second and third readout line storage. The eight bits of data written to one of these locations from the Readout I/O stage is a code that identifies the specific character (or cursor segment) that should be displayed at the associated crt location. After the display data is written into the RAM, the Character Counter is allowed to address the RAM, incrementing through the RAM address field. The eight-bit character codes for each display location are output to Character ROM U2930 in sequence.

Character Counter

The Character Counter stage consists of two four-bit counters (both within U2940) cascaded together to form an eight-bit counter and tristate buffer U2935 which drives the RAM address lines.

As the Character Counter addresses each RAM location (the counter also determines the character screen location), a sequence of "dot display cycles" is performed in which the individual dots that make up the character are positioned on the crt and turned on. The $\overline{EOCH}$ (end of character) signal applied to U2855A prevents the counter from incrementing until all dots of the character have been displayed. As the last dot of a character is addressed, the $\overline{EOCH}$ bit at pin 2 of U2855A goes LO. The next $\overline{GETDOT}$ pulse increments U2940 (via U2855A), and the next RAM location is addressed to start the display of the next character. Space characters have the $\overline{EOCH}$ bit set LO for the first "dot" of the character and merely advance the Counter to the next character address without displaying any dots. See the Character ROM description for further explanation of the $\overline{EOCH}$ bit.

Character ROM

Character ROM U2930 contains the horizontal and vertical dot-position information for all of the possible characters (or cursor segments) that may be displayed. The eight bits of character data from the Character RAM are applied to the eight most-significant address inputs (A4 through A11) of the Character ROM and select a block of dot-positioning data unique to the character to be displayed. The Dot Counter increments the four least-significant address lines (A0 through A3), causing the ROM to output a sequence of eight-bit words, each defining a dot position for the selected character.

The three least-significant bits of a ROM dot-data word (DD0 through DD2) select one of eight horizontal positions for the dot within an eight-by-sixteen character matrix (see Figure 3-6). The next four bits (DD3 through DD6) define the vertical position of the dot within the matrix. These dot-data bits are applied to the Horizontal and Vertical Character DACs, where they are converted to the analog voltages used to position the dot on the crt.

The last dot-data bit DD7 is the $\overline{EOCH}$ (end of character) bit and, when LO, indicates that the last dot of the character is addressed. It is used to reset the Dot Counter (via U2855B) and enables the Character Counter to be incremented (via U2855A) after the last dot of a character has been displayed.

Two servicing jumpers, J401 and J402, have been provided to disable the Character ROM and force the DD7 bit ($\overline{EOCH}$) LO. In certain instances, these two conditions may be useful when troubleshooting the Readout circuitry. To prevent damage to the ROM output circuitry, J402 should only be installed after J401 is installed (to disable the ROM).

Dot Counter

The Dot Counter consists of two four-bit counters (both within U2870), OR-gate U2835A, inverter U2980D, and inverting input AND-gate U2855B. It sequences through a

block of addresses containing dot-position data for a selected character. The Dot Counter is incremented when a dot is finished (via Inverter U2975A) by the $\overline{\text{GETDOT}}$ signal from the Dot Cycle Generator.

The counter increments through the block of dot-position data until the last byte of the block is encountered (last dot). This last data byte has the $\overline{\text{EOCH}}$ (end of character) bit (DD7) set LO. The dot is positioned and displayed in the normal manner, but when the $\overline{\text{GETDOT}}$ signal occurs for the next dot display cycle, the $\overline{\text{EOCH}}$ bit is latched into U2905 and generates the $\overline{\text{EOCH1}}$ (end of character, delayed one dot) signal at U2905 pin 15. With $\overline{\text{EOCH}}$ and $\overline{\text{EOCH1}}$ both LO, the HI reset pulse produced at pin 1 of NOR-gate U2855A resets the counter and, except for space characters, the $\overline{\text{EOCH}}$ bit returns HI. As the reset is removed from the Dot Counter, it is reenabled for display of the next character. For space characters, the $\overline{\text{EOCH}}$ bit will be detected as a LO when the first dot is read from the Character ROM, and the Character Counter will advance to the next character on the next rising edge of $\overline{\text{GETDOT}}$

Counter U2870 and OR-gate U2835D enable characters of more than 16 dots to be displayed. Since most of the readout characters are small, using 16 dots or less, efficient data storage is achieved by storing the dot-position data as 16 consecutive bytes. For displaying these smaller characters, the least significant four bits from U2870 are sufficient to address the 16 possible dot-position bytes.

When larger characters (up to 32 dots) are to be displayed, an additional bit of counter data must be used to address the ROM. This fifth bit comes from U2870 pin 11 and is ORed by U2835D with bit CD0 from the Character RAM. The block address for these larger characters always has bit CD0 set LO, so the counter bit from U2870 pin 11 is in control of the ROM address line at pin 7 of U2930. When displaying these larger characters, the dot count goes beyond 16 dots before the $\overline{\text{EOCH}}$ bit is set LO. On the seventeenth character, the fifth counter bit (pin 11 of U2870) will go HI to address the next 16-byte block of character data in ROM U2930. The lower four bits of the DOT Counter then sequence through this additional block in the normal manner until the $\overline{\text{EOCH}}$ bit is encountered, resetting the counter.

Horizontal DAC

The Horizontal DAC generates the voltages used to horizontally position dots of the readout display on the crt. Five data bits (CA0 through CA4) from the Character Counter stage position a character to the correct column in the display (32 possible columns across the crt), while three data bits from Character ROM U2930 (DD0 through DD2) horizontally position the dots within the eight-by-sixteen character matrix (see Figure 3-6).

The eight bits of position data are written to the permanently enabled DAC each time a new dot is requested by the Dot Cycle Generator. The $\overline{\text{GETDOT}}$ signal applied to pin 11 (Chip Select) enables the DAC to be written into, and the falling edge of the 5-MHz clock applied to pin 12 (Write) writes the data at the eight DAC input pins into an internal latch. The voltage at the DAC output pin changes to reflect the data present in the latch.

Vertical Character DAC

The function of Vertical Character DAC U2875A and U2905 is similar to that of the Horizontal DAC just described. It is responsible for vertically positioning each character dot on the crt. The Vertical DAC circuit is made up of seven, D-type flip-flops (contained within U2905 and U2875) and an accompanying resistor weighting network. The outputs of the flip-flops source different amounts of current to a summing node through a resistor weighting network.

The seven data bits are latched into U2875A and U2905 on the rising edge of the $\overline{\text{GETDOT}}$ signal. Two bits of character address data (CA5 and CA7) from the Character Counter switches the vertical display position between the four readout display lines. When the display is to be in the bottom line, bit CA5 is set LO. With CA5 LO, transistor Q2805 saturates pulling pin 3 of U2820 toward ground and a small current is sourced to the summing node via R2925. Vertical position above this reference is determined by dot data bits DD3 through DD6. When the top line is to be displayed, the CA5 bit is set HI, biasing Q2805 off and allowing pin 3 of U2820 to be pulled up to +5 V through the resistor divider composed of R2928 and R2929. A larger current is now sourced into the summing node via R2925 and enough voltage is developed across R2926 to move the display to the top row of the crt. The CA7 bit is used to offset the top and bottom readout display lines to form the center two readout display lines. As before, the individual dots are then positioned above this reference level by dot data bits DD3 through DD6.

Mode Select Logic and Analog Channel Switch

The Mode Select Logic circuitry is composed of analog switches U2800 and U2805, buffers U2820B and C, gates U2810A, B, C, and D, U2900B and C, and part of U2905. It controls the readout display mode by selecting which deflection signals should drive the Horizontal and Vertical Deflection Amplifiers during a readout display. Five display modes are decoded by the Mode Select Logic: character display, vertical cursor 0, vertical cursor 1, horizontal cursor 0, and horizontal cursor 1.

For normal character displays, cursor select bit CA6 on U2800 pin 1 is LO. This LO signal passes through analog switch U2800 and is latched into U2905 when the $\overline{\text{GETDOT}}$ request from the Dot Cycle Generator goes HI. This latched LO selects the character display mode by

Table 3-6
Readout Display Mode Selection

Control Bits				Mode Selected	Horizontal Signal	Vertical Signal
CA6 (Cursor Select)	DD5	DD4	DD3			
L	X[a]	X	X	Character Display	Horiz DAC	Vert DAC
H	L	H	L	Vert Cursor 1	Horiz DAC	DLY REF 1
H	L	H	H	Horiz Cursor 1	DLY REF 1	Horiz DAC
H	H	L	L	Vert Cursor 0	Horiz DAC	CURSOR 0
H	H	L	H	Horiz Cursor 0	CURSOR 0	Horiz DAC
H	L	L	X	Return to character display Mode		

[a]X = State doesn't matter.

forcing the outputs of U2900B and C and U2810A and B HI. The HI outputs of U2900B and C applied to the select input pins of analog switch U2805 cause the Horizontal DAC output signal applied to U2805 pin 11 to be routed to the Horizontal Amplifier (diagram 6) via buffer U2820B. The same HI logic levels cause NOR-gates U2810C and D to produce a LO at their outputs. This causes analog switch U2800 to route the Vertical DAC output signal applied to pin 12 to the Vertical Output Amplifier (also diagram 6) via buffer U2820A.

For cursor displays, cursor select bit CA6 goes HI. This HI is routed through analog switch U2800 and latched into U2905 when $\overline{\text{GETDOT}}$ next goes HI. This produces a HI at U2905 pin 16, enabling the Mode Select Logic to decode output bits DD3, DD4, and DD5 (from U2905) to determine which of the four possible cursor modes is selected (see Table 3-6). Once one of the cursor modes is entered, analog switch U2800 routes a fixed HI from pin 5, pin 2, or pin 4 to U2905 to keep the Mode Select Logic enabled. Character display mode is reentered only when return-to-character-mode data is decoded (DD4 and DD5 both LO). When that occurs, U2800 routes the CA6 bit to U2905 and, if the bit is LO, the cursor display mode is halted.

CURSOR DEVELOPMENT. Cursors are displayed in short sections, alternating between both vertical positions (for the delta voltage cursors) or both horizontal positions (for the delta time cursors). When displaying delta voltage cursors, the CURSOR 0 level is routed to the Vertical Amplifier by analog switch U2800. This level determines the vertical position of one of the voltage cursors. Horizontal-positioning voltages for one segment of the cursor are routed from Horizontal DAC through analog switch U2805 and buffer U2820B to horizontally position each of the dots making up the cursor segment. DLY REF 1 is then used to vertically position the second cursor, and the Horizontal DAC positions each of the dots for that cursor segment. The cycle is repeated until all segments of both cursors are displayed.

Delta time cursor displays are similar in that the CURSOR 0 and DLY REF 1 signals are used to position the cursors. In this case, however, analog switch U2805 selects the CURSOR 0 and DLY REF 1 signals alternately to position the cursors horizontally, and the Horizontal DAC output is routed via analog switch U2800 and buffer U2820C to vertically position the dots within each cursor segment.

Refresh Prioritizer

The Refresh Prioritizer circuitry consists of U2850A and B, U2950B, U2990A, and U2985. It keeps track of how well the Readout circuitry is doing in displaying all the required readout information and maintains the overall refresh rate. Since the readout display must remain flicker-free and at a constant intensity over the entire sweep rate range, various modes of displaying readout information are provided. The Refresh Prioritizer keeps track of the display status and enables the various readout-display modes as required to produce minimal interference with the displayed waveform trace(s).

Ideally, readout information should be displayed only when the oscilloscope is not trying to display waveform traces. These times occur before a trace commences, after a trace is completed, or between consecutive traces. Displaying in this mode corresponds to "priority one" in

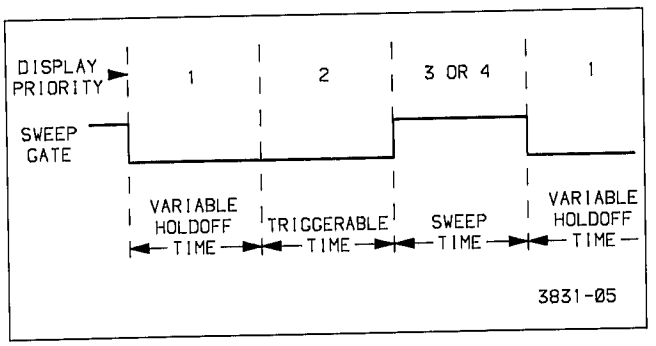


Figure 3-7. Readout display priorities.

Figure 3-7 and causes no interference with the displayed waveforms. If the Readout circuitry is able to display all the required readout dots during the holdoff time between sweeps, the prioritizer U2985 will turn off the Dot Start Governor until the next subframe of readout information is to be displayed. When the sweep times are either too fast to finish a readout display during holdoff (at 5 ns per division no identifiable holdoff time exists) or too slow to allow flicker-free readout, readout display modes other than priority one are initiated.

The next most desirable time for dots to be displayed is during "triggerable" time: that time between sweeps when the oscilloscope is waiting for a sweep trigger event to occur. This is designated priority two and may cause slight interference on the leading edge of the displayed trace if a dot is being displayed when the actual trigger occurs.

Finally, the least desirable dot display time is during a waveform trace display. This display time is designated either priority three or priority four. (Priority four indicates a higher demand of display time.) In priorities three and four, dot displays occur during the main portion of the waveform display. However, the waveform blanking associated with these displays is relatively random in nature and is usually not noticeable.

To start a readout display, the ROSFRAME (readout subframe) request from the Timing Logic (diagram 1) clocks the Q output of flip-flop U2850A HI. ROSFRAME is a periodic clocking signal used to hold the overall refresh rate constant and occurs at regular intervals, regardless of the state of the display.

As the Dot Cycle Generator runs, it resets half of U2830 in the Dot Timer at somewhat irregular intervals with the STARTDOT signal (via inverter U2890A). The Dot Timer then starts a timing sequence, and the rising edge of the REFRESH signal from U2830 pin 4 clocks the latched ROSFRAME request from U2850A pin 5 to the Q output (pin 9) of flip-flop U2850B. This HI, applied to the S1 input (pin 10) of prioritizer U2985, sets it up to increment with the next REFRESH clock applied to its clock input (pin 11). The LO $\overline{Q}$ output of U2850B (pin 8) applied to the reset input of U2850A resets the latched ROSFRAME request. See Figure 3-8 for an illustration of the timing sequence involved.

The next REFRESH clock increments the display priority to one by clocking a HI to the Q_D output (pin 12) of prioritizer shift register U2985. (Table 3-7 illustrates the operation of U2985.) The same clock latches the now LO ROSFRAME request at U2850B pin 12 to the Q output (pin 9), where it is applied to the S1 input (pin 10) of prioritizer U2985. The LO on the S1 input of the prioritizer will remain until another ROSFRAME request from the Timing Logic occurs, and the encoded priority at the output pins of U2985 will remain as it is presently set.

As each of the consecutive dots of the readout frame are displayed, the Dot and Character Counters increment until all dots of the subframe have been displayed (eight characters). As the Character Counter increments to address the next character of the display (first character of the next frame), the fourth bit of counter U2940 goes HI and sets the S0 input (pin 9) of prioritizer U2985 HI via exclusive-OR-gate U2990A. The Dot Timer then clocks the prioritizer with a REFRESH clock on pin 11 of U2985, and the priority is decremented back to zero (indicating that the subframe is completed). The next ROSFRAME request starts the process over again to display the next subframe of readout display. The sequence just described is the priority one display mode and is used when holdoff time between sweeps allows all dots of the subframe to be displayed before the next ROSFRAME request occurs.

If a second ROSFRAME request occurs before the Character Counter indicates the end of the subframe (to decrement the prioritizer back to zero), input S1 of U2985 will be set HI (while the S0 input pin remains LO) and the

Table 3-7
Operation of Prioritizer Shift Register

Select Inputs		Mode
S0	S1	
H	H	Parallel Load
H	L	L → Q_A (decrease priority)
L	H	H → Q_D (increase priority)
L	L	Hold Data

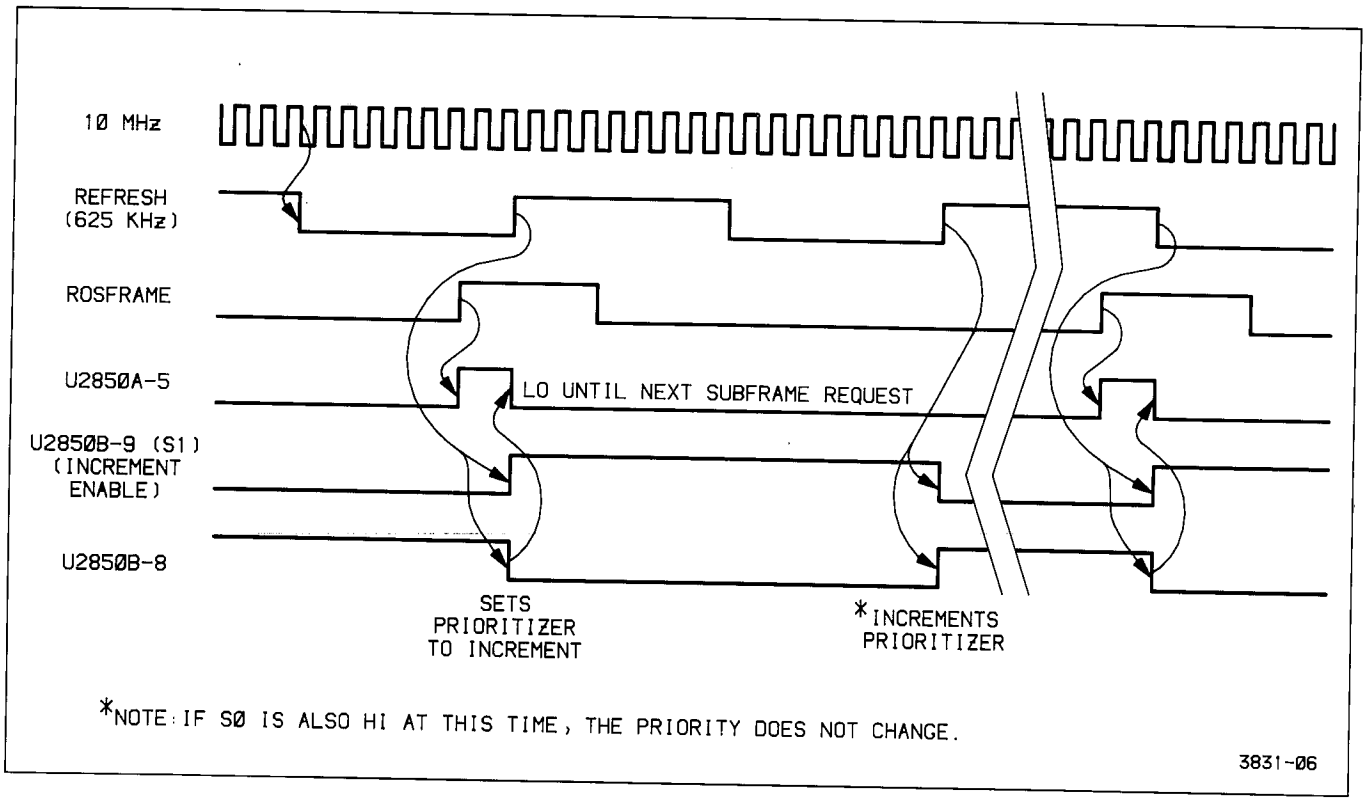


Figure 3-8. Timing of Refresh Prioritizer.

Prioritizer will increment to priority two (outputs Q_C and Q_D go HI) on the next STARTDOT cycle. If this display priority still is inadequate to complete the subframe display before the next ROSFRAME request occurs, priority two will be incremented up to priority three, or even to priority four should the condition persist. Priority four is operationally the same as priority three, but it is used to keep the readout circuitry continuously displaying readout data on through the next subframe, thus allowing the display to catch up. If priority four is in effect, the next decrement that occurs at the end of a subframe only returns the prioritizer to priority three, not to priority two.

The circuit composed of flip-flop U2950B and exclusive-OR-gate U2990A enables either edge of the CA3 bit to decrement the priority of the display when a subframe is completed. Either a negative or positive transition on pin 2 of U2990A will cause the output at pin 3 go HI since the Q output of U2950B is still at the opposite level. The HI from U2990A indicates that the end of the present subframe has occurred, and it sets up the prioritizer to decrement with the next REFRESH clock. At the same time that the prioritizer decrements, the changed level of the CA3 bit is clocked through U2950B and causes the output of exclusive-OR-gate U2990A to return LO until the next subframe is completed.

If the subframe is completed (S0 on U2985 goes HI) when a ROSFRAME request is also pending (S1 is also HI), U2985 does a parallel load, reloading the present priority back into the prioritizer. Since, in this case, the subframe display was completed at the same rate as the ROSFRAME request occurred, the readout display priority is not changed.

Dot Start Governor

The Dot Start Governor detects the display priority from the Refresh Prioritizer and initiates dot-display cycles as the appropriate conditions are met. The conditions tested include display priority, sweep gate completion, dot completion, readout control status, and the readout active enable from the Display Sequencer.

When the readout board status line (ACTIVE/$\overline{\text{ADDRESSABLE}}$) is HI (signifying display) and the REST line goes HI to indicate that the dot cycle is complete, NAND-gates U2890C and D generates a HI at pin 11 (DOTOK) to signal that a new dot display is allowed. The HI from U2890C and D enables most of the gating in the Dot Start Governor. If the Refresh Prioritizer has encoded a display priority of either one or two, the output of exclusive-OR-gate U2990B is HI. When DOTOK from U2890C and D goes HI to enable a dot display, the LO reset from pin 8 of U2970C and D to pin 1 of flip-flop U2880 is removed. Now, when the A Sweep gate ($\overline{\text{SGA}}$) goes HI (beginning of Holdoff), the HI at the D input of U2880B is clocked to the Q output and the $\overline{Q}$ output at

pin 8 will go LO, requesting display of a priority one or two dot. This LO dot request is propagated through U2885C, U2965C and D, and U2890B and sets the STARTDOT signal LO. STARTDOT going LO resets Dot Cycle Generator shift register U2995 and counter U2830B of the Dot Timer. Resetting the Dot Cycle Generator shift register causes the REST signal from U2995 pin 13 to go to a LO, removing the HI DOTOK signal at U2890 pin 11. As DOTOK goes LO, STARTDOT at pin 8 of U2890B goes HI to start the DOT Cycle Generator. At the same time the reset to U2880B is asserted via U2970C and D and the dot request is removed. Both the Dot Timer and the Dot Cycle Generator are now enabled and start the first dot-display cycle during holdoff time.

After the Display Sequencer U650 (diagram 5) has time to respond to the end of the sweep gate, it sets the readout active signal ($\overline{\text{ROA}}$) to pin 10 of U2880B LO. This sets pin 9 of U2885C LO, and the signal is propagated through U2885C, U2965C and D, and U2890B, as before, resetting the Dot Timer and the Dot Cycle Generator. REST then goes LO as before and starts the Dot Cycle Generator and Dot Timer. This cycle continues, displaying one dot per cycle (except for the first non-displayed dot of a character which is automatically initiated by $\overline{\text{EOCH2}}$, until the Display Sequencer determines that the readout time is over (sets $\overline{\text{ROA}}$ HI) or until the display priority is decremented to zero.

When a display priority of three or four exists, the output of U2990B will be LO, and U2970C and D, U2880B, and the associated logic gates following it will not be able to initiate a dot cycle. In either of these display priorities, U2970A and B, U2835C, U2965A and B, and flip-flop U2950A detect the higher priority and generate a readout request signal ($\overline{\text{ROR}}$) to the Display Sequencer. The LO from U2950A pin 6 propagates through U2965C and U2890B to initiate a STARTDOT cycle. When the Display Sequencer recognizes that the readout request signal is LO, it will perform the mode-dependent setup functions necessary to give display control to the Readout Board and will then set the $\overline{\text{ROA}}$ (readout active) line LO. The LO will be clocked into U2880B, and the Dot Cycle Generator will generate a $\overline{\text{GETDOT}}$ signal, resetting the readout request from flip-flop U2950B. Only one dot is displayed for each readout request.

A similar readout display request will be generated when priority-two-or-higher displays are required when sweep gates are not present (dot display during triggerable time after holdoff). This condition is detected by NAND-gate U2885A. NAND-gates U2970A and B allows a readout request to be generated when in the interfere mode. This mode is always invoked in 2467B instruments and invoked only during a single-sequence waveform display in 2465B instruments and ensures that all of the selected sweep combinations are displayed once, followed by a complete readout frame (for the purpose of crt photography).

Dot Cycle Generator

The Dot Cycle Generator, composed of shift register U2995, flip-flop U2880A, and associated gating circuitry, generates time-related signals for the following purposes: unblanking the crt to display a dot; requesting the next byte of dot data in preparation for displaying the next dot; and reenabling itself to repeat the tasks, via the Dot Start Governor (dependent on the display priority).

The timing relationships of the Dot Cycle Generator output signals are controlled by shift register U2995. When the Dot Start Governor initiates a STARTDOT cycle as previously described, the STARTDOT signal initially goes LO, resetting all the Q outputs of U2995 LO and setting the Q output of flip-flop U2880A to a HI. The STARTDOT signal is then returned HI, and the Dot Timer counter U2830 and shift register U2995 are enabled. The shift register begins to consecutively shift HI logic levels to its Q output pins with each 5-MHz clock from the Dot Timer. After approximately 400 ns, pin 5 (Q_C) of the shift register will go HI. The HI at Q_C propagates through exclusive-OR-gate U2990D and NAND-gates U2980A and D to unblank the crt by setting the readout blanking signal ($\overline{\text{ROB}}$) HI.

When the Q_F output of U2995 goes HI (1 μs after STARTDOT), the output of U2990D goes LO and the output of U2990C goes HI. The LO from U2990D propagates through U2980A and D to blank the crt ($\overline{\text{ROB}}$ goes LO) and to clock flip-flop U2880A via NAND-gate U2980B. The $\overline{\text{ROA}}$ (readout active) level from the Display Sequencer (diagram 5) is clocked from the D input (pin 2) of U2880A to the Q output; and, if LO (indicating that the readout circuitry had control of the crt when unblanking occurred; thus the dot was displayed), the output of U2980C is set HI. With three HI levels applied to NAND-gate U2885A, a $\overline{\text{GETDOT}}$ request is generated to get the next byte of dot-position data for display. The next 5-MHz clock sets the Q_G output of U2995 HI, and the output of U2990C goes LO, removing the LO $\overline{\text{GETDOT}}$ signal.

At 1.4 μs after STARTDOT goes HI, U2995 pin 13 (Q_H) goes HI to produce the REST signal, indicating that the current dot cycle is complete and the Dot Cycle Generator is at REST. If the readout ACTIVE/ $\overline{\text{ADDRESSABLE}}$ mode bit at U2980C pin 10 is still HI, the REST signal going HI produces a HI DOTOK signal (next dot is allowed) at pin 11 of U2890D. This HI applied to pin 4 of U2890B, along with any of the possible dot requests from the Dot Start Governor, will initiate another STARTDOT cycle for the next dot of the display. As long as the Display Sequencer holds the readout active line ($\overline{\text{ROA}}$) LO, U2885B, U2965C and D of the Dot Start Governor will automatically initiate dot cycles as soon as the previous one ends (REST goes HI), until the Refresh Prioritizer is decremented to zero.

When the last dot of the character is called from the Character ROM, the $\overline{\text{EOCH}}$ bit (DD7) applied to latch U2905 at pin 18 (in the Vertical Character DAC circuitry) is LO. At the end of that dot display cycle, the $\overline{\text{GETDOT}}$ signal (going HI) clocks the LO $\overline{\text{EOCH}}$ bit into latch U2905 and increments character counter U2940. The latched bit becomes the $\overline{\text{EOCH1}}$ signal (end of character, delayed one dot request) and is applied to U2855A, along with the already LO $\overline{\text{EOCH}}$ bit, to reset Dot Counter U2870. The least-significant bits to the Character ROM address pins (A0 through A4) are then zeros, and the first dot of the next character is addressed. The Horizontal and Vertical DACs don't write this first dot position data into their registers until the end of the next $\overline{\text{GETDOT}}$ signal. That same $\overline{\text{GETDOT}}$ signal also clocks $\overline{\text{EOCH1}}$ into U2905 which becomes $\overline{\text{EOCH2}}$ at pin 16 (end of character, delayed by two dot requests). $\overline{\text{EOCH2}}$ is applied to NAND-gate U2980D and disables the gate prior to the time the Dot Cycle Generator attempts to unblank the crt for the first dot display; thus the first dot of a character is never displayed.

Disabling the unblanking path for the first dot of each character in the manner just described allows the more radical voltage changes between characters to settle before the actual display of the next character begins. When the dot data for one of these undisplayed dots also has the $\overline{\text{EOCH}}$ bit set LO, it is a space character, and the display is advanced to the next character.

Dot Timer

The Dot Timer, composed of U2890A and U2830, generates three, time-related signals used to synchronize the display and maintain the proper sequencing of the individual character dots.

The two least-significant bits of the Dot Timer, from U2830 pins 11 and 10, are reset at the beginning of a dot cycle by a LO STARTDOT signal applied to the reset input of the counter via U2890A. As the dot-display cycle begins, the STARTDOT signal returns HI and the Dot Timer begins counting in a binary fashion. The 10-MHz clock applied to pin 13 is divided by two to produce the 5-MHz clocking signal at output pin 11. The 5-MHz clock sequences the Dot Cycle Generator through the various phases of the dot-display cycle. The REFRESH output signal from U2830 pin 4 updates the Refresh Prioritizer as each subframe is displayed.

A third clock, from U2830 pin 6, occurs at approximately 8-μs intervals and allows any pending dot requests to generate a $\overline{\text{ROR}}$ signal to the Display Sequencer via flip-flop U2950B. (Readout request generation is described in the Dot Start Governor discussion.)

HIGH VOLTAGE POWER SUPPLY AND CRT

The High-Voltage Supply and CRT circuit (diagram 8) provides the voltage levels and control circuitry for operation of the cathode-ray tube (crt). The circuitry consists of the High Voltage Oscillator, the High Voltage Regulator, the Cathode Supply, the Anode Multiplier, the DC Restorer, Focus Amplifiers, the CRT and the various CRT Control circuits.

High-Voltage Oscillator

The High-Voltage Oscillator transforms power obtained from the −15 volt unregulated supply to the various ac levels necessary for the operation of the crt circuitry. The circuit consists of transformer T1970, switching transistor Q1981, and associated circuitry. The low-voltage oscillations set up in the primary winding of T1970 are raised by transformer action to high-voltage levels in the secondary windings. These ac secondary voltages are applied to the DC Restorer, the Cathode Supply, and the anode multiplier circuits.

Oscillation occurs due to the positive feedback from the primary winding (pin 4 to pin 5) to the smaller base-drive winding (pin 3 to pin 6) for transistor Q1981. The frequency of oscillation is about 50 kHz, and is determined primarily by the resonant frequency of the transformer.

When power is first applied, the High-Voltage Regulator circuit detects that the negative crt cathode voltage is too positive and pulls pin 2 of transformer T1970 negative. The negative level forward biases transistor Q1981 via the base-drive winding of the transformer. Current begins to flow in the primary winding through transistor Q1981, inducing a magnetic field around the transformer primary winding. The increasing magnetic field induces a current in the base-drive winding that further increases the base drive to the transistor. This in-phase feedback causes current in Q1981 to increase until the primary winding current reaches its maximum value. As the rate of change of the primary current peaks and then reverses, the induced magnetic field begins to decay. This decreases the base-drive current and begins turning Q1981 off.

As Q1981 is beginning to turn off, the magnetic field around the primary winding continues to collapse at the resonant frequency rate of the transformer. This induces into the base-drive winding a voltage that completely turns off the transistor. The collapsing magnetic field goes to zero, then builds in the opposite direction to a maximum before collapsing again (resonant flywheel effect). This sequence of events occurs repetitively as the circuit continues to oscillate.

The oscillating magnetic field in the primary winding couples power into the secondary windings of the transformer. The amplitude of the voltages induced in the secondary windings is a function of the turns ratios of the transformer windings.

High-Voltage Regulator

The High-Voltage Regulator consists of U1956A and B and associated components. It monitors the crt Cathode Supply voltage and varies the bias point of the switching transistor in the High Voltage Oscillator to hold the Cathode Supply voltage at the nominal level. Since the output voltages at the other secondary winding taps are related by turns ratios to the Cathode Supply voltage, all voltages are held in regulation.

When the Cathode Supply voltage is at the proper level (−1900 V), the current through R1945 and the 19-MΩ resistor internal to High Voltage Module U1830 holds the voltage developed across C1932 at zero volts. This is the balanced condition and sets base drive in Q1981 via integrator U1956A and voltage-follower U1956B. Varying base drive to Q1981 holds the secondary voltages in regulation.

If the Cathode Supply voltage level tends too positive, a slightly positive voltage will develop across C1932. This voltage causes the outputs of integrator U1956A and voltage-follower U1956B to move negative. The negative shift charges capacitor C1951 to a different level, around which the induced feedback voltage at the base-drive winding will swing. The added negative bias causes Q1981 to turn on earlier in the oscillation cycle, and a stronger current pulse is induced in the secondary windings. The increased power in the secondary windings increases the secondary voltages until the Cathode Supply voltage returns to the balanced condition (zero volts across C1932). Opposite action occurs should the Cathode Supply voltage tend too negative.

Cathode Supply

The Cathode Supply circuit is composed of a voltage-doubler and an RC filter network contained within High-Voltage Module U1830. This supply produces the −1900 V accelerating potential applied to the CRT cathode and the −900 V slot lens voltage. The −1900 V supply is monitored by the High Voltage Regulator to maintain the regulation of all voltages from the High Voltage Oscillator.

The alternating voltage (950 V peak) from pin 10 of transformer T1970 is applied to a conventional voltage-doubler circuit at pin 7 of the High Voltage Module. On the positive half cycle, the input capacitor of the voltage doubler (0.006 μf) is charged to −950 V through the forward-biased diode connected to ground at pin 9 of the module (charging path is through the diode, so stored charge is negative). The following negative half cycle adds its ac component (−950 V peak) to this stored dc value and produces a total peak voltage of −1900 V across the capacitor. This charges the 0.006-μf storage capacitor (connected across the two doubler diodes) through the second diode (now the forward-biased diode) to −1900 V. Two RC filters follow the voltage doubler to smooth out the ac ripple. A resistive voltage divider across the output of the filter network provides the −900-V slot lens potential.

Anode Multiplier

The Anode Multiplier circuit (also contained in High Voltage Module U1830) uses voltage multiplication to produce the +14 kV CRT anode potential. Circuit operation is similar to that of the voltage-doubler circuit of the Cathode Supply.

The first negative half-cycle charges the 0.001-μf input capacitor (connected to pin 8 of the High Voltage Module) to a positive peak value of +2.33 kV. The following positive half cycle adds its positive peak amplitude to the voltage stored on the input capacitor and boosts the charge on the second capacitor of the multiplier (and those following) to +4.66 kV. Following cycles continue to boost up succeeding capacitors to values 2.33 kV higher than the preceding capacitor until all six capacitors are fully charged. This places the output of the last capacitor in the multiplier at +14 kV above ground potential. Once the multiplier reaches operating potential, succeeding cycles replenish current drawn from the Anode Multiplier by the crt beam. The 1-MΩ resistor in series with the output protects the multiplier by limiting the anode current to a safe value.

Focus Amplifier

The Focus Amplifier, in conjunction with the auto-focus circuitry of Z-Axis hybrid U950 (diagram 6), provides optimum focus of the crt beam for all settings of the front-panel INTENSITY control. The Focus Amplifier itself consists of two shunt-feedback amplifiers composed of Q1851, Q1852, and associated components. The outputs of the amplifiers set the operating points of a horizontally converging quadrapole lens and a vertically converging quadrapole lens within the crt. The convergence strength of each lens is dependent on the electric field set up between the lens elements.

Since the bases of Q1851 and Q1852 are held at constant voltages (set by their emitter potentials), changing the position of the wiper arms of the ASTIG and FOCUS pots changes the amount of current sourced to the base junctions through R1856 and R1857 respectively. This changes the base-drive currents and produces different

output levels from the Focus Amplifiers; that, in turn, changes the convergence characteristics of the quadrapole lenses.

Initially, at the time of adjustment, the FOCUS and ASTIG potentiometers are set for optimum focus of the crt beam at low intensity. After that initial adjustment, the ASTIG pot normally remains as set, and the FOCUS control is positioned by the user as required when viewing the displays. When using the FOCUS control, transistor Q1852 is controlled as described above; however, an additional current is also supplied to the base node of Q1851 from the FOCUS pot through R1855. This additional current varies the base-drive current to Q1851 and provides tracking between the two lenses as the FOCUS control is adjusted during use of the instrument.

The convergence strengths of the quadrapole lenses also dynamically track changes in the display intensity. The VQ OUT signal, applied to the crt at pins 5 and 6, is exponentially related to the VZ OUT (intensity) signal driving the crt control grid and increases the strength of the lenses more at higher crt beam currents. (A higher beam current requires a stronger lens to cause an equal convergence of the beam.)

DC Restorer

The DC Restorer provides crt control-grid bias and couples both the dc and the low-frequency components of the Z-Axis drive signal to the crt control grid. This circuit allows the Z-Axis Amplifier to control the display intensity by coupling the low-voltage Z-Axis drive signal (VZ OUT)

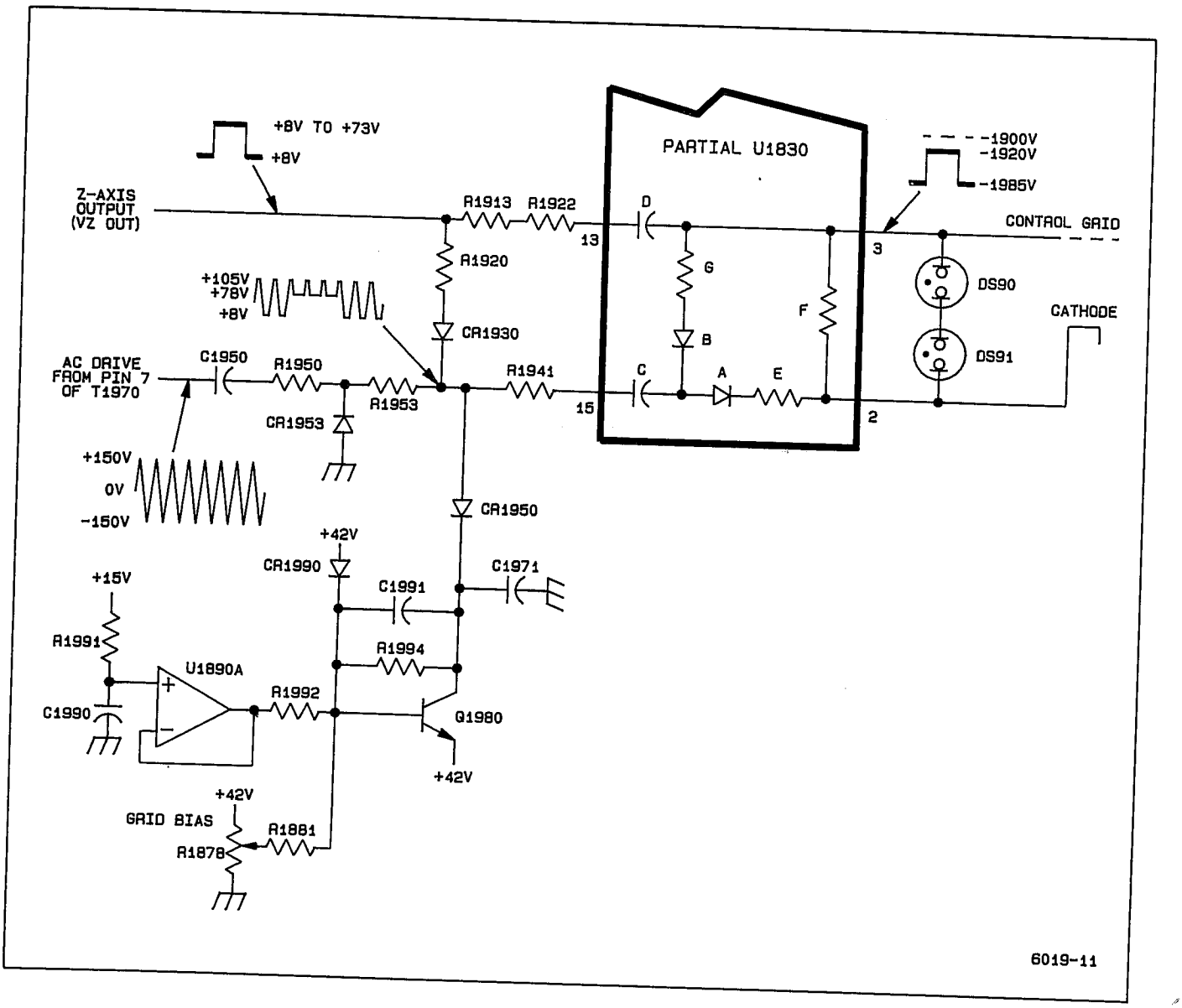


Figure 3-9. Dc restorer circuit.

to the elevated crt control-grid potential (about −1.9 kV).

The DC Restorer circuit (Figure 3-9) operates by impressing the crt grid bias setting and the Z-Axis drive signal on an ac voltage waveform. The shaped ac waveform is then coupled to the crt control grid through a coupling capacitor that restores the dc components of the signal.

GRID BIAS LEVEL. An ac drive voltage of approximately 300 V peak-to-peak is applied to the DC Restorer circuit from pin 7 of transformer T1970. The negative half cycle of the sinusoidal waveform is clipped by CR1953, and the positive half cycle (150 V peak) is applied to the junction of CR1930, CR1950, and R1941 via R1950 and R1953. Transistor Q1980, operational amplifier U1890A, and associated components form a voltage clamp circuit that limits the positive swing of the ac waveform at the junction.

Transistor Q1980 is configured as a shunt-feedback amplifier, with C1991 and R1994 as the feedback elements. The feedback current through R1994 develops a voltage across the resistor that is positive with respect to the +42.6 V on the base of the transistor. The value of this additive voltage plus the diode drop across CR1950 sets the upper clamping threshold. Grid Bias potentiometer R1878 sinks varying amounts of current away from the base node of the transistor and thus sets the feedback current through R1994. The adjustment range of the pot can set the nominal clamping level between +71 V and +133 V.

When the amplitude of the ac waveform is below the clamping threshold, series diode CR1950 will be reverse biased and the ac waveform is not clamped. During the time the diode is reverse biased, transistor Q1980 is kept biased in the active region by the charge retained on C1971 from the previous cycle. As the amplitude of the ac waveform at the junction of CR1930 and CR1950 exceeds the voltage at the collector of Q1980, diode CR1950 becomes forward biased, and the ac waveform is clamped at that level. Any current greater than that required to maintain the clamp voltage will be shunted to the +42 V supply by transistor Q1980.

Operational amplifier U1890A sinks a time-dependent variable current away from the base node of Q1980 that modifies the crt control-grid bias during the first few minutes of instrument operation. The circuit compensates for the changing drive characteristics of the crt as it warms up.

At power-up, capacitor C1990 begins charging through R1991 toward the +15 V supply. The output of U1890A follows the rising voltage on pin 3; and after about ten minutes (for all practical purposes), it reaches +15 V. As the output voltage slowly increases, the charging current through R1992 causes the Grid Bias voltage to gradually lower about ten volts from its power-on level. The charge on C1990 dissipates slowly; therefore, if instrument power is turned off and then immediately back on again, the output of U1890A will still be near the +15 V limit rather than starting at zero volts as when the crt was cold.

Z-AXIS DRIVE LEVEL. The variable-level Z-Axis signal (VZ OUT) establishes the lower clamping level of the ac waveform applied to the High Voltage Module. When the amplitude of the waveform drops below the Z-Axis signal, CR1930 becomes forward biased, and the ac waveform is clamped to the Z-Axis signal level. The VZ OUT level may vary between +8 V and +75 V, depending on the setting of the front-panel INTENSITY and READOUT INTENSITY controls.

The ac waveform, now carrying both the grid-bias information and the Z-Axis drive information, is applied to a DC Restorer circuit in the High Voltage Module where it is raised to the high-voltage levels of the crt control grid.

DC RESTORATION. The DC Restorer circuit in the High Voltage Module is referenced to the crt cathode voltage via a connection within U1830. Capacitor C (in Figure 3-9), connected to pin 15 of U1830, initially charges to a level determined by the difference between the Z-Axis signal level and the crt cathode potential. The Z-Axis signal sets the level on the positive plate of capacitor C through R1920, CR1930, and R1941; the level on the negative plate is set by the crt cathode voltage through resistor E and diode A. Capacitor D is charged to a similar dc level through resistors F, R1922, and R1913.

When the ac waveform applied to pin 15 begins its transition from the lower clamped level (set by the Z-Axis signal) towards the upper clamped level (set by the Grid Bias potentiometer), the charge on capacitor C increases. The additional charge is proportional to the voltage difference between the two clamped voltage levels.

When the ac waveform begins its transition from the upper clamped level back to the lower clamped level, diode A becomes reverse biased. Diode B becomes forward biased, and an additional charge proportional to the negative excursion of the ac waveform (difference between the upper clamped level and the lower clamped level) is added to capacitor D through diode B and resistor G. The amount of change added to capacitor D depends on the setting of the front-panel INTENSITY control, as it sets the lower clamping level of the ac waveform. This added charge determines the potential of the control grid with respect to the crt cathode.

The potential difference between the control grid and the cathode controls the beam current and thus the display intensity. With no Z-Axis signal applied (INTENSITY control off), capacitor D will be charged to its maximum negative value, since the difference between the two clamped voltage levels is at its maximum value. This is the minimum intensity condition and reflects the setting of the Grid Bias potentiometer. During calibration, the Grid Bias pot is adjusted so that the difference between the upper clamping level (set by the Grid Bias pot) and the "no signal" level of the Z-Axis drive signal (VZ OUT) produces a control grid bias that barely shuts off the crt electron beam.

As the INTENSITY control is advanced, the amplitude of the square-wave Z-Axis signal increases accordingly. This increased signal amplitude decreases the difference between the upper and lower clamped levels of the ac waveform, and less charge is added to capacitor D. The decreased voltage across capacitor D decreases the potential difference between the control grid and the cathode, and more crt beam current is allowed to flow. Increased beam current increases the crt display intensity.

During the periods that capacitor C is charging and discharging, the control-grid voltage is held stable by the long-time-constant discharge path of capacitor D through resistor F. Any charge removed from capacitor D during the positive transitions of the ac waveform will be replaced on the negative transitions.

The fast-rise and fast-fall transitions of the Z-Axis signal are coupled to the crt control grid through capacitor D. This ac-coupled fast-path signal quickly sends the crt electron beam to the new intensity level, then the slower DC Restorer path "catches up" to handle the dc and low-frequency components of the Z-Axis drive signal.

Neon lamps DS90 and DS91 prevent arcing inside the crt should the control grid potential or cathode potential be lost for any reason.

CRT Control Circuits

The CRT Control circuits provide the various potentials and signal attenuation factors that set up the electrical elements of the crt. The control circuitry is divided into two separate categories: (1) level setting and (2) signal handling. The level setting circuitry produces voltages and current level necessary for the crt to operate, while the signal-handling portion is associated with changing crt signal levels.

LEVEL-SETTING CIRCUITRY. Operational amplifier U1890B, transistor Q1980, and associated components form an edge-focus circuit that sets the voltages on the elements of the third quadrapole lens. The positive lens element is set to its operating potential by Edge Focus adjustment pot R1864 (via R1897). This voltage is also divided by R1893 and R1982 and applied to the non-inverting input of U1890B to control the voltage on the other element of the lens.

The operational amplifier and transistor are configured as a feedback amplifier, with R1891 and R1990 setting the stage gain. Gain of the amplifier is equal to the attenuation factor of divider network R1893 and R1892, so total overall gain of the stage from the wiper of R1864 to the collector of Q1890 is unity. The offset voltage between lens elements is set by the ratio of R1891 and R1990 and the +10 V reference applied to R1990. This configuration causes the two voltages applied to the third quadrapole lens to track each other over the entire range of Edge Focus adjustment pot R1864.

Other adjustable level-setting circuits include Y-Axis Alignment pot R1848, used to rotate the beam alignment after vertical deflection. This adjustment controls the amount of current through the Y-Axis Alignment coil around the neck of the crt and is set to produce precise perpendicular alignment between x- and y-axis deflections. The TRACE ROTATION adjustment R975 is a front-panel screwdriver-adjustable control. The effect of the adjustment is similar to the Y-Axis Alignment pot, but when adjusted, it rotates both the x-axis and the y-axis deflections of the trace on the face of the crt. A final adjustable level-setting control is the Geometry pot R1870, adjusted to optimize display geometry. The potential at pin 8 for the vertical shield internal to the crt is produced by zener diode VR1891 and associated components.

SIGNAL-HANDLING CIRCUITRY. The crt termination adjustment R1501 is set to match the loading characteristics of the crt's vertical deflection structure to the Vertical Output Amplifier.

LOW VOLTAGE POWER SUPPLY

The low voltages required by the instrument are produced by a high-efficiency, switching power supply. This type of supply directly rectifies and stores charge from the ac line supply; then the stored charge is switched through a special transformer at a high rate, generating the various supply voltages.

Line Rectifier

Ac line voltages of either 115 V or 230 V may provide the primary power for the instrument, depending on the setting of LINE VOLTAGE SELECTOR switch S90

(located on the instrument rear panel). Power Switch S350 applies the selected line voltage to power supply rectifier CR1011.

With the selector switch in the 115 V position, the rectifier and storage capacitors C1021 and C1022 operate as a full-wave voltage doubler. When operating in this configuration, each capacitor is charged on opposite half cycles of the ac input, and the voltages across the two capacitors in series will approximate the peak-to-peak value of the source voltage. For 230 V operation, switch S90 connects the rectifier as a conventional bridge rectifier. Both capacitors charge on both input half cycles, and the voltage across C1021 and C1022 in series will approximate the peak value of the rectified source voltage. For either configuration, the dc voltage supplied to the power supply inverter is the same.

Thermistors RT1010 and RT1016 limit the surge current when the power supply is first turned on. As current flow warms the thermistors, their resistances decrease and have little effect on circuit operation. Spark-gap electrodes E1001 and E1002 are surge-voltage protectors. If excessive source voltage is applied to the instrument, the spark-gaps conduct, and the extra current flow quickly exceeds the rating of fuse F90. The fuse then opens to protect the instrument's power supply. The EMI (electromagnetic interference) filter, inductors L1011 and L1012, capacitors C1016 and C1018, and resistors R1011, R1012, R1016 and R1018 form a line-filter circuit. This filter, along with common mode rejection transformer T1020, prevents power-line interference from entering the instrument and prevents power supply switching signals from entering the supply line.

Preregulator Control

The Preregulator Control circuit monitors the drive voltage applied to inverter output transformer T1060 and holds it at the level that produces proper supply voltages at the secondary windings.

The Preregulator Control circuit consists primarily of control IC U1030, its switching buffers, and its power supply components. The control IC senses voltage on the primary winding of T2060 and varies the "on time" of a series-switching transistor, depending on whether the sensed voltage was too high or too low. The switching transistor Q1050, rectifier CR1050, choke T1050, and capacitor C1050 form a buck-switching regulator circuit. The output voltage at W1060 is proportional to the product of the rectified line voltage on C1020-C1022 and the duty cycle of Q1050. In normal operation, Q1050 is on about one-half the time. When Q1050 is off, current flows to W1060 and T1060 through CR1050.

PREREGULATOR CONTROL POWER SUPPLY. Since the Preregulator Control network controls supply startup and preregulates the secondary supplies, an independent power source must be established for it before any of the other power supplies will operate. The independent power supply for the control circuitry is composed of Q1021, Q1022, and associated components.

Initially, when instrument power is applied, the positive plate of capacitor C1025 is charged toward the positive rectified line voltage through R1020. The voltage at the base of Q1022 follows at a level determined by the voltage divider composed of R1022, R1024, CR1023, and the load within U1030. When the voltage across C1025 reaches about +21 V, the base voltage of Q1022 reaches +6.8 V and Q1022 turns on, saturating Q1021. The +21 V on the emitter of Q1021 appears at its collector and establishes the positive voltage supply for the Preregulator IC. With Q1021 on, R1024 is placed in parallel with R1022, and both Q1022 and Q1021 remain saturated.

The +21 V level begins to drain down as the control IC draws current from C1025. If the Preregulator Control IC doesn't start the switching supply (and thus recharge C1025 and C1023 via CR1022) by the time the voltage across C1025 reaches about +8 V, Q1021 will turn off. Resistor R1024 pulls the base of Q1022 low and turns that transistor off also. (Capacitor C1025 would only discharge low enough to turn off the transistors under a fault condition.) In this event, C1025 would then charge again to +21 V, and the start sequence would repeat. Normally, the control IC will start Inverter action before the +8 V level is reached, and current is drawn through T1050 via Q1050. This induces a current in the secondary winding of T1050 via Q1050. This induces a current in the secondary winding of T1050 and charges C1025 positive via diode CR1022. The turns ratio of T1050 sets the secondary voltage at approximately +15 V; and, as long as the supply is being properly regulated, C1025 will be charged up to that level and held there.

PREREGULATOR START-UP. As the supply for the Preregulator Control IC is established, an internal switching oscillator begins to run. The oscillator generates a repetitive triangular wave (as shown in Figure 3-10) at a frequency determined primarily by R1032 and C1032. The simplified schematic of Figure 3-11 illustrates the voltage control functions of U1030.

As the Preregulator power supply turns on, capacitor C1034 charges from the +5 V reference level toward ground potential through R1034 and R1037. As it does, the voltage at pin 4 (one input of Dead-Time Comparator U1) will pass through the positive-peak value of the triangular waveform on the other input of the Dead-Time

Comparator. The comparator will then begin outputting narrow pulses that become progressively wider as the voltage on pin 4 settles to zero volts. These pulses drive switching transistor Q1050, and their slow progression from narrow to wide causes the various secondary supplies to gradually build up to their final operating levels. The slow buildup prevents a turn-on current surge that would cause the current-limit circuitry to shut down the supply.

During startup, capacitor C1072 acts as a substantial load, and a relatively large current flows in the windings of T1050 for the first few cycles of Preregulator switching. These strong current pulses ensure that storage capacitor C1066 becomes charged sufficiently to start the Inverter Drive circuit. Once the Inverter Drive stage is operating, the normal switching current through T1050 maintains the required charge on C1066. (The Inverter Drive power supply is discussed later in this description.)

Dead-Time Comparator U1 is referenced at approximately 0.1 V above the ground level at pin 4 (established when C1034 becomes fully charged) and outputs a narrow, negative-going pulse that turns off switching transistor Q1050 for a portion of each switching cycle. This off time ensures that flip-flop U1064B in the Inverter Drive circuit toggles every cycle (thereby maintaining the proper duty cycle), independent of the voltage conditions being sensed by the remainder of the voltage control circuitry.

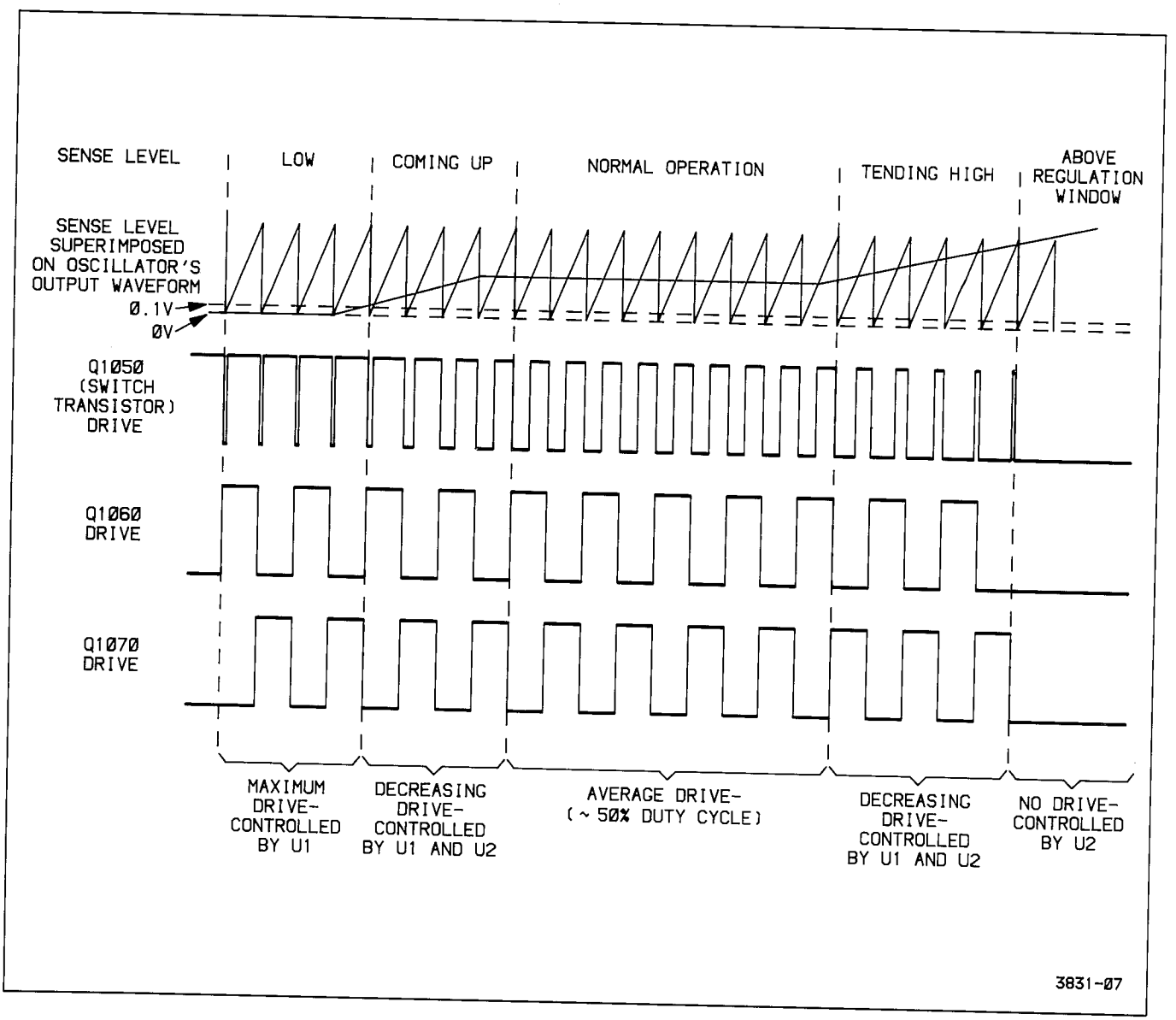


Figure 3-10. Timing relationships of the Inverter Drive signals.

PREREGULATION. Once the initial charging at power-up is accomplished, as just described, the voltage-sensing circuitry begins controlling the Inverter switching action. The actual voltage sensing is done by error amplifier U2. The level at the center tap of output transformer T1060 is applied to pin 1 and is compared to the reference established by R1045 and R1046 at pin 2. If the sensed level at pin 1 is lower than the reference level (as it will always be for the first few switching cycles), the of error-amplifier U2 will be LO. The LO, applied to the inverting input of U3, results in a long-duty-cycle drive signal to transistor Q1050 (via CR1030). Since the Inverter Drive stage will alternately turn either Q1060 or Q1070 on, relatively large current pulses will result in the primary winding of inverter output transformer T1060.

These large current pulses, over the period of a few cycles, will increase the charge on the storage capacitors on the secondary side of the transformer and will reduce the current demand on the inverter output transformer. As the demand increases, the voltage across the primary winding will increase until it reaches the point where the two inputs of U2 are at the same potential. At this point, the output of U2 (to U3) will settle to a level approximately equal to the midpoint of the triangular waveform applied to the other input of U3. The resulting drive signal has an approximate 50% duty cycle and will respond to changes in either the ac line voltage or supply load conditions. Depending on the output levels sensed, the duty cycle of the drive signal will change (sensed level rises or falls with respect to the triangular waveform) to hold the secondary supplies at their proper levels.

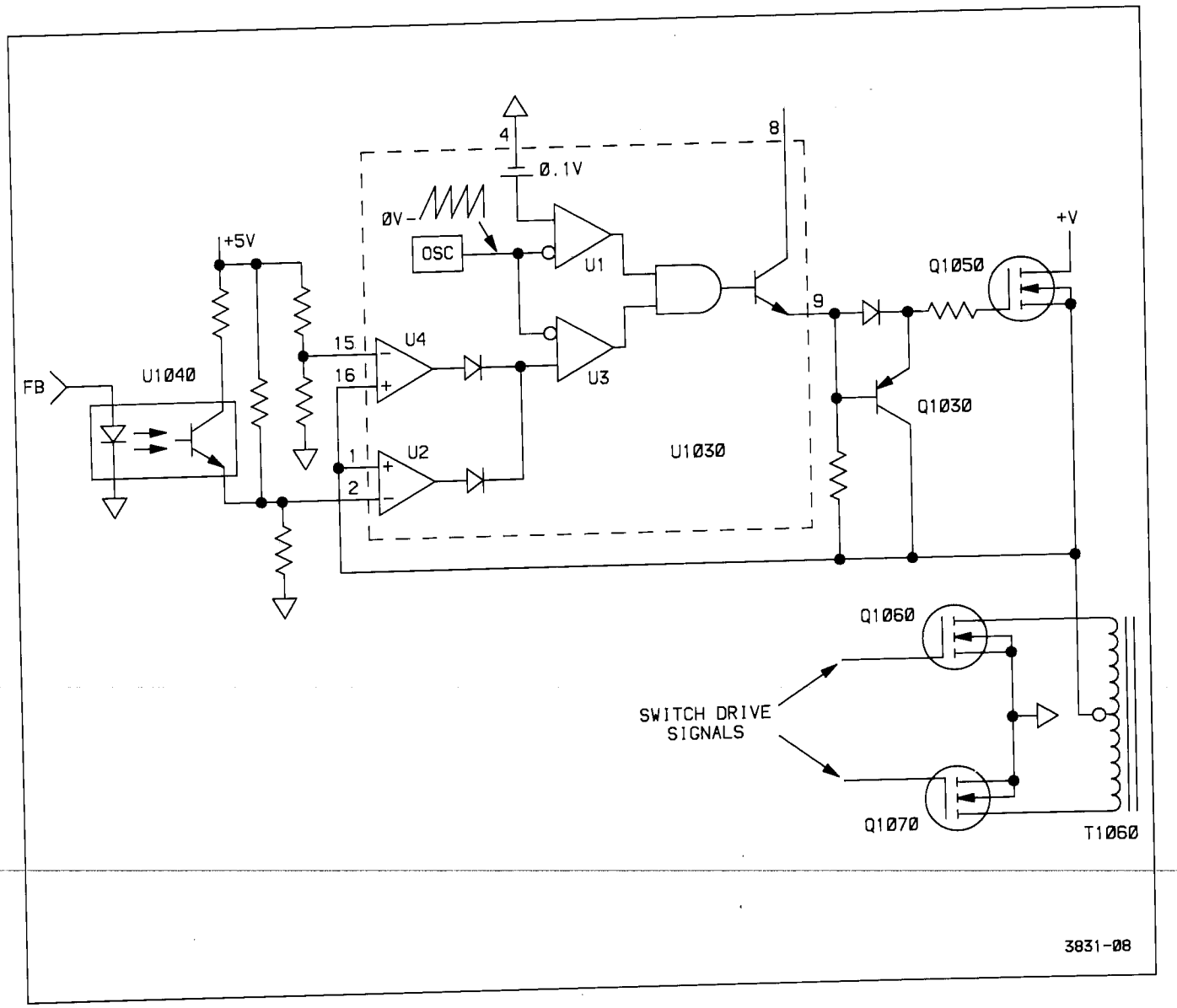


Figure 3-11. Simplified schematic of control network.

Opto-isolator U1040 and resistor R1044 form a control network that allows a voltage sensed at the feedback input (FB) to slightly alter the voltage-sense reference applied to pin 2 of U2. The FB signal is generated by the +5 V Inverter Feedback amplifier (U1371, diagram 10) and is directly related to the level of the $+5V_D$ supply line. Base drive to the shunt transistor (in opto-isolator U1040) is increased should the FB signal go below its nominal value. Additional current is shunted around R1045 (via R1044) and raises the voltage-sense reference level to error-amplifier U2. This increases the voltage applied to the primary winding of the output transformer, since U2 sensing depends on a balanced condition. Higher currents are induced in the secondary windings, and the secondary voltages begin to return to their nominal values. As the $+5V_D$ line returns to its nominal level, base drive to the shunt transistor will be reduced and the voltage in the primary winding will follow. Should the FB signal level tend too high, opposite control responses occur. Further information about the FB signal is given in the +5 V Inverter Feedback description.

Error amplifier U4 and the voltage divider composed of R1035 and R1031 provide a backup sensing circuit. Its operation is similar to that of error amplifier U2, just described, but it senses at a slightly higher level. As long as U2 is operating properly, U4 will be inactive. However, should a failure occur in the U2 sensing circuitry, the voltage on the primary winding of T1060 will rise to the sensing level at pin 15 of U4. Sense amplifier U4 will then take over, preventing a damaging over-voltage condition.

Inverter Drive

The Inverter Drive circuit performs the necessary switching to drive the inverter output transformer. Like the Preregulator Control IC, the Inverter Drive circuit requires an independent power supply, since it must be operational before any of the secondary supply voltages can be generated.

INVERTER DRIVE POWER SUPPLY. This power supply consists of Q1062, VR1062, and their associated components. As power is first applied, the initial charging current through T1050 induces a current in the transformer secondary winding (pins 8 and 9). The alternating current is rectified by the diode bridge composed of CR1062, CR1063, CR1064, and CR1065 and stored in C1066, providing power for the Inverter Drive circuitry.

When the Preregulator Control IC turns switching transistor Q1050 on for the first time, the charge stored on C1066 during the initial charging period is sufficient to properly turn on one of the current-switching transistors (either Q1060 or Q1070) for the first cycle. After that, the alternating drive signals continue to induce current into the secondary winding of T1050 to provide operating power as long as the instrument is turned on.

The current rectified by the diode bridge and stored on capacitor C1066 is regulated down to the required voltage level by R1061, VR1062, and Q1062. Zener diode VR1062 references emitter-follower Q1062 and holds the supply output at approximately +11.4 V.

INVERTER DRIVE GENERATOR. The Inverter Drive generator consists of U1062, U1064, U1066, switching transistors Q1060, Q1070 and their associated components. The circuitry alternately switches current through each leg of the output transformer (T1060) primary winding and produces the ac current required for transformer action.

Out-of-phase input signals to comparator U1062C come from two resistive voltage dividers placed in either leg of one secondary winding of T1050. The comparator detects the phase changes (crossover points) of the secondary current caused as Q1050 switches on and off. Every complete on-off cycle of Q1050 produces a positive clock at pin 14 of U1062C that toggles flip-flop U1064B. The toggling alternately turns switching transistors Q1060 and Q1070 on, each with an approximate 50% duty cycle.

Comparators U1062A and U1062B, at the Q and $\overline{Q}$ output of the flip-flop, detect the precise crossing point of the toggling drive signals and ensure that only one switching transistor will be on at any one time. These mutually-exclusive drive signals are buffered by inverters U1066A and U1066B and applied to switching transistors Q1060 and Q1070 to alternately turn them on and off at one-half the switching rate of Q1050. By alternately switching opposite ends of the primary winding to ground, the current flowing through switching transistor Q1050 will flow alternately in each half of the primary winding. This produces ac voltages at the secondary windings that are then rectified, providing the various unregulated dc supply voltages.

Current Limit

The Current Limit circuit, composed of transistor Q1040 and the associated components, limits the maximum current flow in the output transformer to about 1 ampere. Resistor R1040 (connected to the Preregulator Control IC +15 V supply) forward biases germanium diode CR1040 and applies approximately +0.3 V across the base-to-emitter junction of Q1040. Current flowing to the output transformer develops a voltage drop across R1050 that adds to the bias developed by CR1040. As the current to the transformer increases, the voltage drop across R1050 also increases until, at around 1 A, the combined voltage drop across R1050 and CR1040 forward biases transistor

Q1040. The base of Q1022 is pulled negative through R1042, and the +15 V supply for the Preregulator IC turns off (see Preregulator Control description). The power supply will try to restart itself; but, as long as the excessive-current condition persists, the current-limit circuit will keep shutting the supply down, protecting the instrument.

Rectifiers

The rectifiers convert the alternating current from the secondary windings of inverter output transformer T1060 to the various dc supply voltages required by the instrument. Rectification is done by conventional diode rectifier circuits, and filtering is done by conventional LC networks.

The +87 V unregulated supply is produced by a voltage-doubler circuit. The positive plate of C1130 at the anode of CR1132 is referenced at approximately +45 V through diode CR1131 (to the +42 V unregulated supply). As the positive half cycle from the 42 V secondary winding (actually about +45 V peak) is applied to the negative plate of C1130, the positive plate is elevated to a peak value of approximately +90 V. Diode CR1132 becomes forward biased and storage capacitor C1132 is charged to about +90 V. Following cycles replenish the charge drawn off by the loads on the +87 V supply line.

Line Signal

A sample of the ac line voltage is coupled to the Trigger circuit by transformer T1229 and provides the LINE TRIG signal to the Trigger hybrid. Transformer current is limited to a safe value by resistors R1014 and R1015 placed in series with the primary winding leads. The transformer's output characteristics are matched to the input of the Trigger circuit hybrid by R1208 and C1208.

Line Up Signal

The circuit composed of Q1029, opto-isolator U1029, and their associated components, detects when power has been applied to the instrument and the Preregulator Control power supply is functioning properly. When the rectified line voltage reaches proper operating voltage, the voltage divider composed of R1027 and R1028 forward biases Q1029. As soon as the Preregulator Control power supply turns on, current flows through R1029, Q1029, and the opto-isolator LED. The illuminated LED saturates transistor U1029 and the LINE UP signal to the Power-Up Delay circuit (diagram 1) is pulled HI, indicating that the Preregulator Control circuit should now be functioning properly.

POWER DOWN. When instrument power is turned off, the voltage across the primary storage capacitors (C1021 and C1022) begins to fall as the capacitors discharge. As the voltage drops, the bias current through R1027 to the base of Q1029 also drops until the bias voltage across R1028 reaches a point about 2 V above the average transformer drive level at pin 2 of U1029. At this point, Q1029 turns off, and the LINE UP signal to the Power-Up Delay circuit goes LO. This LO signals the Microprocessor that it should start its power down routine.

The Line Up circuit tells the Microprocessor that the primary capacitors have started discharging while there is still a stored charge (set by R1027 and R1028) about 40% in excess of that required to keep the power supply voltages in regulation. This allows the Microprocessor to complete the power-down sequence before the supplies drop below their normal operating level. Further information about the power-down sequence is given in the Microprocessor Reset Control description.

Fan Circuit

Fan motor B10 is driven by adjustable three terminal regulator U1110. The fan's speed is determined by the voltage supplied by U1110 and varies with ambient temperature.

As the ambient temperature in the cabinet increases, the resistance of thermistor RT1110 decreases causing more current to flow in R1112. This causes the voltage at pin 2 and therefore the voltage at pin 3 of U1110 to increase, and the fan motor speed increases to provide more cooling capacity.

LOW-VOLTAGE REGULATORS

The Low-Voltage Regulators remove ac noise and ripple from the various unregulated dc supply voltages. Each regulator output is automatically current limited if the output current exceeds the requirements of a normally functioning instrument. This limiting prevents any further component damage.

+10 Volt Reference

Each of the power-supply regulators control their respective outputs by comparing their output voltages to a known reference level. In order to maintain stable supply voltages, the reference voltage must itself be highly stable. The circuit composed of U1290, U1300C and associated components establish this reference.

Resistor R1400 and capacitor C1400 form an RC filter network that smooths the unregulated +15 volt supply

before it is applied to voltage-reference IC U1290. The +2.5 V output from pin 2 of U1290 is applied to the noninverting input of operational amplifier U1300C. The output of U1300C is the source of the +10 V reference level used by the various regulators. The output level is set by the voltage divider formed by R1291, R1293, and potentiometer R1292. The Volt Ref Adjust pot in the divider allows the reference level to be precisely set. Zener diode VR1292 prevents the reference from exceeding +11 volts should a failure in the reference circuitry occur.

+87 V Regulator

The +87 V Regulator is composed of Q1220, Q1221, Q1222, Q1223, U1281A, and their associated components. The circuit regulates and limits both the voltage and current of the supply output.

Initially, as power is applied, the voltage applied to pin 2 of U1281A from the voltage divider formed by R1227 and R1228 is lower than the +10 V reference level applied to pin 3. The output of U1281A is forced high, reverse biasing the base-emitter junction of Q1222 and turning it completely off. With Q1222 off, all the current through R1212 is supplied as base current to Darlington transistor pair Q1221 and Q1220, and maximum current flows in series-pass transistor Q1220. This charges up the various loads on the supply line, and the output level charges positive.

As the regulator output charges toward +87 V, the voltage divider applies a positive-going voltage to the inverting input of U1281A. When the output level reaches +87 volts, the inverting input reaches the +10 V reference at the noninverting input. The output voltage at pin 1 of U1281A will go negative and the base-emitter junction of Q1222 will be biased into the active region. As Q1222 turns on, base drive for the Darlington pair (Q1221 and pass transistor Q1220) is reduced. The output will be held at the level required (+87 V) for voltage at the two inputs of amplifier U1281A to be in balance.

Current limiting is a foldback design and is performed by Q1223 and its associated components. Under normal current demand conditions, Q1223 is off. If the regulator output current exceeds approximately 100mA (as it might if a component fails), the voltage drop across R1221 and CR1220 reaches a point that forward biases Q1223 via the bias divider formed by R1222 and R1223. As Q1223 turns on, a portion of the base-drive current to Q1221 is shunted away by Q1223. This reduces the base-drive current (and thus the output current) of series-pass transistor Q1220.

+42 V Regulator

The circuit configuration and operation of the +42 V Regulator is identical to that of the +82 V Regulator. Current limiting of the +42 V supply occurs at approximately 400 mA. Base drive to Darlington pair Q1241 and Q1240 is via R1244 and is dependent on proper operation of the +87 Volt Regulator. This dependency ensures that the relative polarities of the two supplies are never reversed (preventing semiconductor-junction damage in the associated load circuitry).

+15 V Regulator

The +15 V Regulator uses three-terminal regulator U1260 and operational amplifiers U1371A and U1371B, arranged as voltage sensors, to achieve regulation of the +15 V supply. The three-terminal regulator holds its output voltage at pin 2 at 1.25 volts more positive than the reference input level at pin 1. The voltage at the reference pin is established by current flow in either diode CR1262 or CR1263.

Resistors R1261 and R1262 at the regulator output divide the +15 V level down for comparison with the +10 V reference applied to pin 5 of operational amplifier U1371B. When the input voltage at pin 6 (supplied by the voltage divider) is lower than the +10 V reference, the output of amplifier U1371B is high and the output voltage of U1260 is allowed to rise. As the regulator output reaches +15 V, the voltage on pin 6 of U1371B approaches the level on pin 5, and the amplifier begins sinking current away from the reference pin of the three-terminal regulator via diode CR1263. This lowers the voltage on the reference pin and holds the output at +15 V.

The other voltage-sensing amplifier (U1371A) ensures that the relative polarity between the +15 V supply and the +42 V supply is maintained, preventing component damage in the load circuitry. Should the +42 V supply be pulled below +15 V (excessive loading or supply failure), the voltage at pin 3 of U1371A falls below the voltage at pin 2 and the amplifier output voltage goes low. This forward biases CR1262 and lowers the reference voltage for U1260, reducing the output voltage.

Current limiting for the +15 V supply is provided by the internal circuitry of the three-terminal regulator.

+5 V Regulator

Regulation of the +5 V supply is provided by a circuit similar to those of the +87 V and the +42 V Regulators. As long as the relative polarity between the +15 V and the +5 V supplies is maintained, base drive to Q1281 is

supplied through R1283. The current through Q1281 provides base drive for series-pass transistor Q1280.

When voltage-sense amplifier U1300B detects that the output voltage has reached +5 V, it begins shunting base-drive current away from Q1281 via CR1281 and holds the output voltage constant.

Current limiting for the +5 V supply is done by U1300A and associated components. Under normal current-demand conditions, the output of U1300A is high and diode CR1282 is reverse biased. However, should the current through the current-sense resistor R1281 reach approximately 2 A, the voltage developed across R1281 will raise the voltage at pin 2 of U1300A (via divider R1282 and R1286) to a level equal to that at pin 3. This causes the output of U1300A to go low, forward biasing CR1282. This sinks base drive current away from Q1281 and lowers the output current in series-pass transistor Q1280.

−15 V Regulator

Operation of the −15 V Regulator, composed of three-terminal regulator U1330, operational amplifier U1270C, and their associated components, is similar to that of the +15 V Regulator with the following major changes. The control voltage at the three-terminal regulator's reference pin (pin 1) is established by the current through series-resistors R1333 and R1334. The reference pin is clamped by CR1332 at about −5.6 V should a failure in the sensing network occur. (Clamping also prevents latchup of the operational amplifier during start-up of the power supply.) Finally, the sensing divider formed by R1331 and R1332 is referenced to the +10 V reference instead of ground to enable sensing of negative voltage.

−8 V Regulator

Operation of the −8 V Regulator is similar to that of the +87 V and +42 V Regulators. Due to the lower operating voltages of the −8V Regulator the common-base transistor present in both the +87 V and the +42 V is not required. Current limiting in the −8 V supply occurs at about 480 mA.

−5 V Regulator

Operation of the −5 Volt Regulator is similar to that of the +5 V Regulator. Current limiting in the −5 V supply occurs at about 2 A.

+5 V Inverter Feedback

Operational amplifier U1371C and associated components are configured as a frequency-compensated voltage-sensing network. The circuit monitors the +5 V digital power supply line from the rectifiers and provides feedback to the Preregulator Control IC (U1030) via opto-isolator U1040 (both on diagram 9). The feedback is used to slightly vary the voltage-sensing characteristics of the Preregulator Control circuitry. The feedback (FB) signal slightly varies the voltage to the Inverter output transformer and holds the output of the 5 V secondary windings at an optimum level. Output levels of the other secondary windings are related to the +5 V_D level and are also held at their optimum values. This technique minimizes power losses in the series-pass transistors and increases regulator reliability.

Power-Up Delay

The Power-Up Delay circuit, composed of Q1370, Q1376, U1371D, and the associated components, ensures that the various regulated power supplies have time to reach their proper operating voltages before signaling the Microprocessor that the power supplies are up.

When power is first applied, a LINE UP signal from the Preregulator Control circuit goes HI, indicating that the power switch has been closed and that ample supply voltage is available for driving the Inverter transformer. The HI is applied to the base of Q1370, but since the collector is not properly biased yet, no transistor current will flow. As the Inverter begins to run, the various voltages from the secondary rectifiers begin coming up to their proper levels. A +2.5 V reference voltage is applied to operational amplifier U1371D pin 12 and forces the output high, biasing Q1376 on.

Before any of the Low-Voltage Regulators may function properly, the +10 V reference voltage must be established as previously described. When the +15 V Regulator turns on, current flows through Q1370, and pin 13 of U1371D is pulled above the +2.5 V reference through divider R1370 and R1372. The output of U1371D goes low, turning off Q1376.

When power to the instrument is turned off, the LINE UP signal goes LO (as explained in the Line Up Signal description). The falling LINE UP signal turns Q1370 off and drives the output of U1371D high. The output level from U1371D turns Q1376 on and pulls the PWR UP signal to the Microprocessor LO. This LO initiates the power-down sequence used to shut down the instrument in an orderly fashion. The delay between the time that the PWR UP signal goes LO and when the regulated power supplies fall below their normal operating levels provides ample time for the Microprocessor to complete the power-down sequence.

Power Supply Shutdown

Phosphor damage can occur to the CRT if certain regulated power supply voltages are overloaded due to excessive current draw by their loads. U1300C and its associated circuitry monitor the +15 V and the +5 V Regulator supplies. The +87 V and the +42 V Regulator supplies are monitored via R1294 and R1295 respectively. If any of these regulated supplies exceed their limit, current is sourced to U1300D (pin 13). When this happens, the +10 V Reference begins to drop which in turn lowers all the regulated supplies. This causes the high voltage oscillator to shutdown preventing damage to the CRT. Q1290 and its associated circuitry allows the +10 V Reference to come up and stabilize before the shutdown circuitry is enabled. Jumper J208 is used to disconnect the shutdown circuitry for troubleshooting purposes.

POWER DISTRIBUTION

Schematic diagrams 11 and 12 illustrate the power distribution of the instrument. The connections to the labeled boxes (representing the hybrids and ICs) show the power connections to each device, while connections to non-power lines are shown by the component and schematic number. Power supply decoupling is done with traditional LRC networks as shown on the diagrams.

Several intermediate supply voltages are generated by devices shown on diagrams 11 and 12. An approximate +32 volt supply for the A and B Sweeps is developed by emitter-follower Q700 and its associated components. Zener diodes VR125 and VR225 develop approximate +6.2 volt supplies for the CH 1 and CH 2 Preamps respectively, and zener diode VR2805 establishes an approximate −6.8 volt supply for U2800 and U2805.

INTERCONNECTIONS

Schematic diagram 13 illustrates the circuit board interconnections of the instrument. Connector numbers and cabling types are shown.

THEORY OF OPERATION (SN B049999 & BELOW)

INTRODUCTION

SECTION ORGANIZATION

This section contains a functional description of the instrument circuitry. The discussion begins with an overview of the instrument functions and continues with detailed explanations of each major circuit. Reference is made to supporting schematic and block diagrams which will facilitate understanding of the text. These diagrams show interconnections between parts of the circuitry, identify circuit components, list specific component values, and indicate interrelationships with front-panel controls.

The detailed block diagram and the schematic diagrams are located in the tabbed "Diagrams" section at the rear of this manual, while smaller functional diagrams are contained within this section near their respective text. The particular schematic diagram associated with each circuit description is identified in the text, and the diagram number is shown (enclosed within a diamond symbol) on the tab of the appropriate foldout page. For optimum understanding of the circuit being described, refer to both the applicable schematic diagram and the functional block diagram.

HYBRID AND INTEGRATED CIRCUIT DESCRIPTIONS

Digital Logic Conventions

Digital logic circuits perform many functions within this instrument. The operation of these circuits is represented by specific logic symbology and terminology. Most logic-function descriptions contained in this manual use the positive-logic convention. Positive logic is a system of notation whereby the more positive of two levels is the TRUE (or 1) state; the more negative level is the FALSE (or 0) state. In the logic descriptions, the TRUE state is referred to as HI, and the FALSE state is referred to as LO. The specific voltages which constitute a HI or a LO state vary between individual devices. For specific device characteristics, refer to the manufacturer's data book.

Hybrids

Some of the circuits in this instrument are implemented in hybrid devices. The hybrids are specialized electronic devices combining thick-film and semiconductor technologies. Passive, thick-film components and active, semiconductor components are interconnected to form the circuit on a ceramic carrier. The end result is a relatively small "building block" with enhanced performance characteristics, all in one package. Hybrid circuits are shown on schematics simply as blocks with inputs and outputs. Information about hybrid functioning is contained in the related portion of the Detailed Circuit Description.

Linear Devices

The operation of individual linear integrated circuit devices is described in this section using waveforms or other graphic techniques to illustrate their operation.

BLOCK DIAGRAM

The following discussion is provided to aid in understanding the overall operation of the instrument circuitry before the individual circuits are discussed in detail. A simplified block diagram of the instrument, showing basic interconnections, is shown in Figure 3-1. The diamond-enclosed numbers in each block refer to the schematic diagram(s) at the rear of this manual in which the related circuitry is located.

BLOCK DESCRIPTION

The Low Voltage Power Supply is a high-efficiency, switching supply with active output regulation that transforms the ac source voltage to the various dc voltages required by the instrument. The High Voltage Power Supply circuit develops the high accelerating potentials required by the crt, using voltage multiplication techniques, and the DC Restorer provides interfacing for the low-potential intensity signals from the Z-Axis Amplifier to the crt control grid.

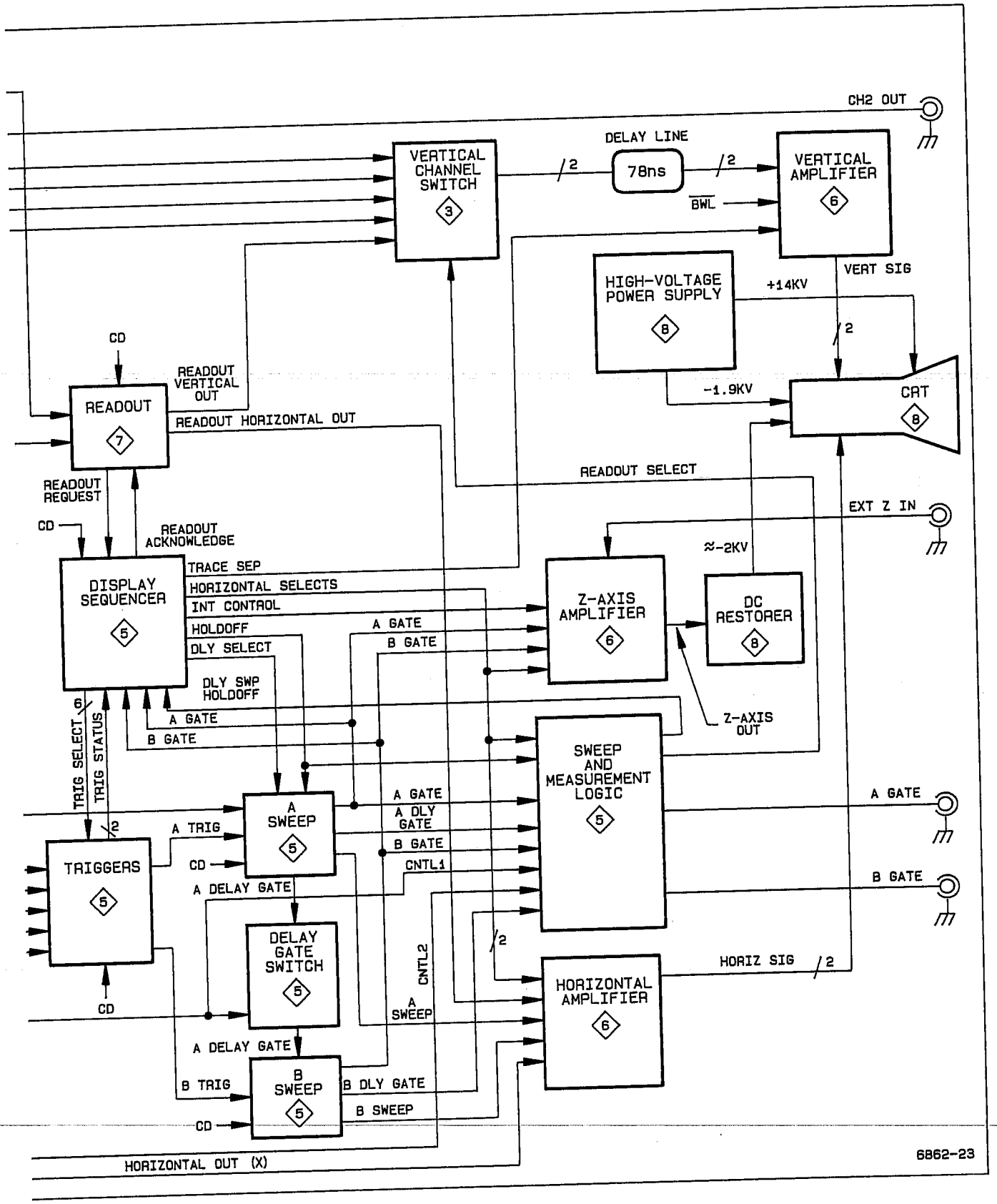


Figure 3-1. Instrument block diagram (cont).

Most of the activities of the instrument are directed by a microprocessor. The microprocessor, under firmware control (firmware is the programmed instructions contained in read-only memory that tells the processor how to operate), monitors instrument functions and sets up the operating modes according to the instructions received.

Various types of data read to and from the Microprocessor (program instructions, constants, control data, etc.) are all transferred over a group of eight bidirectional signal lines called the Data Bus. The Data Bus is dedicated solely to microprocessor-related data transfer.

Another group of signal lines, called the Address Bus, are responsible for selecting or "addressing" the memory location or device that the Microprocessor wants to communicate with. Typically, depending on the instruction being executed, the processor places an address on the Address Bus to identify the location the Microprocessor must communicate with. This address, along with some enabling logic, opens up an appropriate data path between the processor and the device or memory location via the Data Bus; and data is either read from or written to that location by the processor.

While executing the control program, the Microprocessor retrieves previously stored calibration constants and front-panel settings and, as necessary places program-generated data in temporary storage for later use. The battery backed up RAM provides these storage functions.

When power is applied to the instrument, a brief initialization sequence is performed, and then the processor begins scanning the front-panel controls. The switch settings detected and the retrieved front-panel data from the battery backed up RAM causes the processor to set various control registers and control voltages within the instrument that define the operating mode of the instrument. These register settings and voltage levels control the vertical channel selection and deflection factors, the sweep rate, the triggering parameters, the readout activity, and sequencing of the display. Loading the control data into the various registers throughout the instrument is done using a common serial data line (CD). Individual control clock signals (CC) determine which register is loaded from the common data line.

Coordination of the vertical, horizontal, and Z-Axis (intensity) components of the display must be done in real time. Due to the speed of these display changes and the precise timing relationships that must be maintained between display events, direct sequencing of the display is beyond the capabilities of the processor control. Instead, control data from the processor is sent to the Display Sequencer (a specialized integrated circuit) which responds by setting up the various signals that control the stages handling real-time display signals. The controlled stages are stepped through a predefined sequence that is determined by the control data. Typically, as the sequence is being executed, the Display Sequencer will be changing vertical signal sources, Z-Axis intensity levels, triggering sources, and horizontal sweep signal sources. The specific activities being carried out by the Display Sequencer depend on the display mode called for by the control data.

Vertical deflection for crt displays comes from one or more of the four front-panel vertical inputs and, when displaying readout information, from the Readout circuitry. Signals applied to the front-panel Channel 1 and Channel 2 inputs are connected to their respective Preamplifiers via processor-controlled Attenuator networks. Control data from the Microprocessor defining the attenuation factor for each channel is serially loaded into the Auxiliary Control Register and then strobed into the Attenuator Mag-Latch Relays in parallel. The relay switches of each Attenuator network are either opened or closed, depending on the data supplied to the Mag-Latch Relay Drivers. The relays are magnetically latched and remain as set until new control data is strobed in. The Auxiliary Control Register is therefore available, and different mode data is clocked into the register to set up other portions of the instrument.

Attenuated Channel 1 and Channel 2 input signals are amplified by their respective Preamplifiers. The gain factor for the Channel 1 and Channel 2 Preamplifiers is settable by control data from the processor. The Channel 3 and Channel 4 input signals are amplified by their respective Preamplifiers by either of two gain factors set by control bits from the Auxiliary Control Register. All four of these preamplified signals are applied to the Vertical Channel Switch where they are selected by the Display Sequencer for display when required.

Each of the vertical signals is also applied to the A and B Trigger circuitry via trigger pickoff outputs from the Preamplifier stages. Any one of the signals may be selected as the trigger SOURCE for either the A or the B Trigger circuitry as directed by the Display Sequencer. The line trigger signal provides an added trigger source for A Sweeps only. Control data from the Microprocessor is written to the Trigger circuitry to define the triggering LEVEL, SLOPE, and COUPLING criteria. When the selected trigger signal meets these requirements, a sweep can be initiated. The Trigger circuit initiates both the A Sweep and the B Sweep as required by the display mode selected.

In the case of A Sweeps, the LO state of the THO (trigger holdoff) signal from the Display Sequencer enables

the A Sweep circuit and the next A trigger initiates the sweep. For B sweeps, and in the case of intensified sweeps, the A Sweep delay gate signal (DG) enables the B Sweep circuit. Depending on the B trigger mode selected, a B Sweep will be initiated either immediately (RUN AFT DLY) or on the next B trigger signal (TRIG AFT DLY). The slope of the sweep ramp is dependent on Microprocessor-generated control data loaded into the internal control register of the A and B Sweep circuit hybrids.

Sweep signals generated by each of the Sweep hybrids are applied to the Horizontal Amplifier. The Horizontal Amplifier is directed by the Display Sequencer to select one of the sweep ramps for amplification in sequence. In the case of Readout and X-Y displays, the X-Readout and CH 1 input signals are selected to be amplified, also under direction of the Display Sequencer.

To control the display intensity, the Display Sequencer directs the Z-Axis circuit to unblank the display at the appropriate time for the sweeps and readout displays. When the display is unblanked, the Display Sequencer selects the display intensity for either waveform displays or for readout displays by switching control of the Z-Axis beam current between the front-panel INTENSITY and READOUT INTENSITY potentiometers as appropriate.

During readout displays, the vertical dot-position signal from the Readout circuitry is applied to the Vertical Amplifier via the Vertical Channel Switch. Horizontal dot-position deflection for the readout display is selected by internal switching in the Horizontal Amplifier.

The vertical, horizontal, and Z-Axis signals are applied to their respective amplifiers where they are raised to crt-drive levels. The output signals from the Vertical and Horizontal Amplifiers are applied directly to the crt deflection plates. The Z-Axis Amplifier output signal requires interfacing to the high-potential crt environment before application to the crt control grid. The necessary Z-Axis interfacing is provided by the DC Restorer circuit located on the High-Voltage circuit board. The resulting display may be of waveforms, alphanumeric readout, or a combination of both.

DETAILED CIRCUIT DESCRIPTION

INTRODUCTION

The following discussion provides detailed information concerning the electrical operation and circuit relationships of the instrument. Circuitry unique to the instrument is described in detail, while circuits common in the electronics industry are not. The descriptions are accompanied by supporting illustrations and tables. Diagrams identified in the text, on which associated circuitry is shown, are located at the rear of this manual in the tabbed foldout pages.

PROCESSOR AND DIGITAL CONTROL

The Processor and Digital Control circuitry (diagram 1) directs the operation of most oscilloscope functions by following firmware control instructions stored in memory. These instructions direct the Microprocessor to monitor the front-panel controls and to send control signals that set up the various signal processing circuits accordingly.

Microprocessor

The Microprocessor (U2140) is the center of control activities. It has an eight-bit, bidirectional data bus for data display transfer (D0 through D7) and a 16-bit address bus (A0 through A15) for selecting the source or destination of the data. Precise timing of instruction execution, addressing, and data transfer is provided by an external, crystal-controlled clock signal.

The clock signal is developed by the Microprocessor Clock stage and applied to the Microprocessor at pin 39. Using the external clock as a reference, the Microprocessor generates synchronized control output signals, R/$\overline{W}$ (read-write), E (enable), and VMA (valid memory address) that maintain proper timing relationships throughout the instrument.

Microprocessor Clock

The Microprocessor Clock stage generates a 5-MHz square-wave clock signal to the Microprocessor and a 10-MHz clock signal to portions of the Readout circuitry. Inverter U2540A acts as an oscillator with crystal Y2540 providing feedback at the resonant frequency. The required phase shift for oscillation to occur is produced by C2550, C2551, R2545, and the crystal. The RC network composed of R2543, C2640, R2541, and R2542 biases input pin 1 of U2540A in the active region and establishes approximate symmetry of the oscillator output. The signal is buffered and inverted by U2540B to provide the 10-MHz clock signal.

Flip-flop U2440A is a divide-by-two circuit that reduces the 10-MHz clock down to a 5-MHz square-wave signal used to clock the Microprocessor and the Display Sequencer. The 10-MHz clock is supplied to the Readout Board for dot timing and is also available for use with option circuitry.

Reset Control

The Reset Control circuitry ensures that, at power up, the Microprocessor begins program execution from a known point in memory and with all the processor registers in known states. It also allows the processor to reset itself when power is turned off so that the instrument powers down in a known state.

POWER UP SEQUENCE. Reset generator U2240 generates the power-up reset. As power is applied to the instrument U2240 tests the voltage at U2240 pin 7. The reset generator forces U2240 pin 5 LO, and the LO is applied to the processor $\overline{\text{RESET}}$ input (pin 40). After the SENSE input reaches its nominal voltage level, the reset condition continues to allow the microprocessor system time to reset. The reset continues for the time determined by C2350. The effect of power supply transients is reduced by C2240. After the suplies reach their nominal level and the delay period ends U2240 pin 5 goes HI. The RESET signal to the processor then goes HI to enable normal execution to begin, and the processor is directed to the starting address of the power-up routine, which it then performs.

POWER DOWN SEQUENCE. When the instrument power switch is turned off, the PWR UP signal from J251 pin 12 immediately goes LO. This LO generates the NMI (non-maskable interrupt) request to the processor on pin 6 which causes the processor to branch to the power-down routine. Under direction of that routine, the processor begins shutting down the instrument in an orderly fashion before the power supply outputs can drop below the operating thresholds. This routine disconnects the CH1 and CH2 50-Ω input terminations to protect them from accidental application of excessive voltage during storage or bench handling.

As the operating voltages are falling, the Reset circuitry must not generate a false RESET signal to the processor. Such a restart when the power supply voltages are outside their normal operating range would produce unpredictable processor operation that could alter the contents of the battery backed up RAM. When the processor has completed all the other power-down tasks, it finally sets the PWR DOWN signal HI via U2310 (diagram 2). This signal is applied to inverter U2650C at pin 11. Pin 9 of U2650C goes LO and immediately pulls pin 2 of Reset Generator U2240 LO to prevent a reset to the processor. Reset Generator U2240 immediately switches state to assert the $\overline{\text{RESET}}$ signal to the processor. The $\overline{\text{RESET}}$ signal is held LO until the power supplies have fully discharged.

For diagnostic purposes, the PWR DOWN reset signal can be disabled. Moving jumper P503 to the DIAG (diagnostic) position keeps U2240 pin 2 HI. The RESET signal is therefore held HI, and the processor can execute a free-running NOP (no operation) loop without interruption if the PWR DOWN bit is set HI while the Address Bus is incrementing.

Data Bus

Tri-state buffer U2350 is used to buffer the data signals to the Microprocessor from other devices on the bus. When not enabled, the device is switched to isolate the processor from the buffered Data Bus. Buffer U2350 is enabled via the Read-Write Latch U2440B when the processor reads data from another device on the bus.

When the processor writes data onto the bus, Octal Latch U2450 is enabled by the Read-Write Latch U2440B. When the E (enable) signal at pin 11 of U2450 is HI, processor data bits are passed asynchronously through the latch to the buffered data bus. When the E signal goes LO, data bits meeting setup times are latched into the device. The latched Q outputs provide the required drive current to the various devices on the bus and ensure that data hold times are met for correct data transfer. When the Read-Write Latch places a HI on pin 1 of U2450, latch U2450 is disabled, and the outputs are switched to their high-impedance state.

Data transfers to and from the processor may be interrupted by removing Diag/Norm Jumper P503. This forces a NOP (no operation) condition that is useful for verifying the functionality of the processor (when a data-bus device is suspected of causing a system failure) or for troubleshooting the Address Bus and Address Decode circuitry. Removing the jumper removes the operating power from both U2350 and U2450 to disconnect the Microprocessor from the buffered Data Bus. With the Data Bus disconnected, a resistor network pulls the processor Data Bus lines (D0 through D7) to a NOP (no operation) instruction. A NOP causes the Microprocessor to continuously increment through its address field. The Address Decode circuitry may then be checked to determine if it is operating properly.

Address Decode

The Address Decode circuitry generates enabling signals and strobes that allow the Microprocessor to control the various devices and circuit functions. The controlling signals are generated as a result of the Microprocessor placing specific addresses on the Address Bus. Figure 3-2 illustrates the enables and strobes generated by the Address Decode circuitry.

Address decoding is performed by a programmable array logic device, a three-line-to-eight-line decoder, and a four-line-to-sixteen-line decoder attached to the Address Bus. The five most significant address bits are decoded by U2250. This device initially separates the total addressable-memory space (64K-bytes) into thirty-two, 2K-byte blocks. Addresses in the top 32K-bytes (address bit A15 HI) select one of two read-only memories (ROM), U2160, or U2260. When the VMA (Valid Memory Address) and E (Enable) outputs from the Microprocessor go HI, the selected ROM is enabled, and the data from the selected address location is read from the ROM.

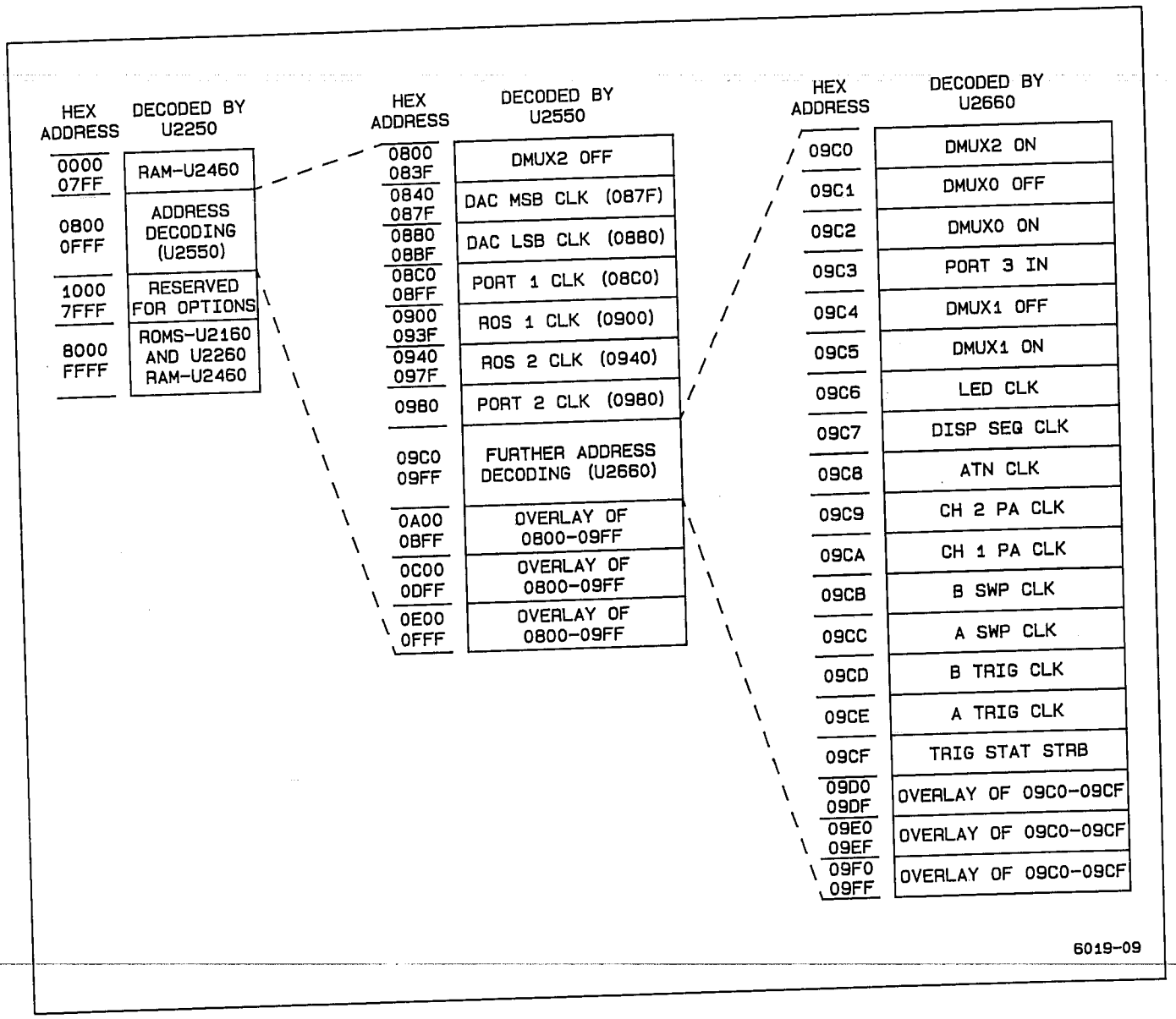


Figure 3-2. Address decoding.

The programmable array logic device also generates the $\overline{OE}$ and $\overline{WE}$ signals to the random-access memory (RAM). This RAM can be accessed with addresses 8000 to 9FFF if either PB0, PB1, or PB2 signals are HI. In this mode the ROMS, U2160 and U2260, are not accessible in this address range.

Of the bottom 32K-bytes of addresses, only the lowest 4K-bytes are further decoded. Addresses in the lowest 2K-byte block of addresses will cause U2250 to generate an enable signal to the RAM, U2460. Addresses in the next 2K-byte block of addresses will enable U2550 to do the next stage of address decoding.

The level of decoding performed by U2550 uses address bits A6, A7, and A8 to separate the addresses within the 2K-byte block of addresses 0800 thru 0FFF into 32 groups of 64 addresses. Address bits A9 and A10 are not used in the decoding scheme, so each of these 32 blocks is not uniquely identified. This results in four duplicate sections within the address block, each consisting of eight groups of 64 addresses. The upper three sections in the address space are never used; therefore, decoding by U2550 may be more simply thought of as eight groups of 64 address locations. Addresses within these eight groups generate control signals to other portions of the instrument.

The final level of address decoding is done by four-line-to-sixteen-line decoder U2660. When enabled by the Y7 output of U2550, this decoder separates the highest 64-address group decoded by U2550 into 16 individual control signals. In this level of decoding, address bits A4 and A5 are not decoded, so that the 64 possible addresses consist of four overlayed blocks of 16 addresses each.

Each of the control signals generated by the Address Decode circuitry are present only as long as the specific address defining that signal is present on the Address Bus. However, one of the addressable control signals decoded by U2550 and five of the addressable control signals decoded by U2660 are used to either set or reset flip-flops U2650A, U2650B, and U2650D. The control signals are, in effect, latched and remain present to enable multiplexers U2521, U2530, (diagram 2) and U170 (diagram 4). When enabled, these multiplexers route analog control signals from DAC (digital-to-analog converter) U2101 (diagram 2) to the various analog control circuits.

Read-only Memory (ROM)

The Read-only Memory consists of one, 128K-byte ROM or two, 64K-byte ROMs that contain operating instructions (firmware) used to control processor (and thus oscilloscope) operation. Addresses from the Microprocessor that fall within the top 32K-bytes of addressable space cause one of the two read-only memory integrated circuits to be enabled. (See Address Decode description.) Instructions are read out of the enabled ROM (or PROM) IC from the address location present on its 16 address input pin (A0 through A14, Page Select). The eight-bit data byte from the addressed locations is placed onto the Buffered Data bus (BD0 through BD7) to be read by the Microprocessor.

Random-Access Memory (RAM)

The RAM consists of integrated circuit U2460 and provides the Microprocessor with 8K-bytes of battery backed up temporary storage space for data that is developed during the execution of a routine. The RAM is enabled whenever an address in the lowest 2K-byte of addresses is placed on the Address Bus or whenever an address of 8000 thru 9FFF is placed on the Address bus with either PB0, PB1, or PB2 set HI. When writing into the RAM, the write-enable signal (WE) on pin 27 of U2460 is set LO along with the chip enable (CE1) signal on pin 20. At the same time, the output-enable (OE) on pin 22 is HI to disable the RAM output drivers. Data is then written to the location addressed by the Microprocessor. If data is to be read from the RAM, the WE signal is set HI to place the RAM in the read mode, and the OE signal is set LO to enable the output drivers. This places the data from the addressed location on the buffered Data Bus where it can be read by the Microprocessor.

The RAM also provides non-volatile storage for the calibration constants and the power-down front-panel settings. When power is applied to the instrument, the Microprocessor reads the calibration constants and generates control voltages to set up the analog circuitry. The front-panel settings that were present at power-off are recalled and the instrument is set to the operating mode previous to power-off.

Battery Circuitry

The Battery circuit composed of BT2570, R2770, CR2770, CR2370, CR2371, and C2470 provides the standby voltage necessary to maintain the contents of the CMOS RAM (U2460). The circuit composed of R2530, U2620C, R2504, and R2506 provides the microprocessor a means of monitoring the battery voltage to detect when the battery needs to be replaced.

Timing Logic

The Timing Logic circuit composed of U2440B, and U2540F generates time- and mode-dependent signals from control signals output from the Microprocessor. The enable (E) signal output from the Microprocessor is a 1.25 MHz square wave used to synchronize oscilloscope functions to processor timing.

Data applied to the Address Bus, Data Bus, and various control signals are allowed to settle (become valid) before any of the addressed devices are enabled. This is accomplished by switching the E signal HI a short time after each processor cycle begins. Inverter U2540F inverts the polarity of the delayed enable signal and enables the Address Decode stage only after the address bus has settled.

Read-Write Latch U2440B is used to delay the processor's read/write signal (R/$\overline{W}$) from the Microprocessor to meet hold-time requirements of the RAM. At the same time, it generates delayed read and write enabling signals of both polarities to meet the requirements of Buffer U2350 and Latch U2450 (in the Microprocessor Data Bus) and various other devices in the Readout circuitry (diagram 7).

When R/$\overline{W}$ goes LO for a write cycle, Read-Write Latch U2440B is reset, and Q output (pin 9) is held LO, Latch U2450 is in its transparent state at this time, and data from the Microprocessor is applied asynchronously to the buffered Data Bus. At the end of the write cycle, the R/$\overline{W}$ signal goes HI, and the reset to U2440B is removed. The E signal also goes through a negative transition, and data on the Microprocessor data bus lines is latched into U2450. The next positive transition of the 1.25-MHz E signal (1/2 E cycle after the R/$\overline{W}$ signal goes HI) clocks the HI level at U2440B pin 12 (the D input) to the Q output, and the $\overline{Q}$ output (pin 8) goes LO. The 1/2 E cycle delay between the time R/$\overline{W}$ goes HI and the time that the Q output of U2440B goes HI keeps Latch U2450 outputs on long enough to meet the data hold time for the RAM. At the end of that delay time, pin 1 of U2450 goes HI, and the Latch outputs are switched to the high-impedance state to isolate it from the buffered Data Bus.

READOUT FRAMING AND INTERRUPT TIMING. Binary counter U2640 is used to generate a readout-framing clock to the Readout circuitry and a real-time interrupt request to the Microprocessor via inverter U2540E. The readout-framing clock is a regular square-wave signal obtained from U2640 pin 12, 14 or 15 by dividing the 1.25-MHz E signal by 512 (2^9), 1024 (2^{10}), or 2048 (2^{11}). This clock tells the readout circuitry to load the next block (subframe) of readout information to be displayed. Pin 12 is for a reduced interfere mode for TV applications, pin 14 is used for retrofitability into older 2 line instruments, and pin 15 is for newer 4 line readout instruments. (See "Readout" description for further information concerning alphanumeric display.) The real-time interrupt request, which occurs every 3.3 ms, is obtained from pin 2 by dividing the E signal by 8192 (2^{13}).

When the real-time request occurs, $\overline{IRQ}$ (pin 4 of U2140) goes LO, and the processor breaks from execution of its mainline program. The Microprocessor first resets Binary Counter U2640 by setting pin 19 of U2301 (diagram 2) HI (to generate the reset), then it resets pin 19 LO to allow the counter to start again. At this time, the Microprocessor sets analog control voltages and reads trigger status from the Display Sequencer (diagram 5). When this is completed, it reverts back to the mainline program.

In addition to the analog control and trigger status update that occurs with each interrupt, on every fifth interrupt cycle, the Microprocessor also scans the front-panel potentiometers. Every tenth interrupt cycle, scanning the front-panel switches and checking the 50-Ω DC inputs for overloads is added to the previously mentioned tasks. If all the tasks are not completed at the end of one interrupt cycle, the real-time interrupt request restarts the analog updates, but as soon as those are accomplished, the Microprocessor will pick up with its additional tasks where it was before the interrupt occurred. This continues until all tasks are completed. If any pot or switch changes are detected, the Microprocessor updates the analog control voltages and the control register data to reflect those changes prior to reverting back to the mainline program instructions.

FRONT-PANEL SCANNING and ANALOG CONTROLS

The Analog Control circuitry (diagram 2), under Microprocessor control, reads the front-panel controls and sets various analog control voltages to reflect these front-panel settings. The calibration constants determined during instrument calibration and the last "stable" front-panel setup conditions are stored in battery backed up RAM. At power-on the stored front panel information is used to return the instrument to its previous state.

Hardware I/O

Data transfer from the Analog Control circuitry to the Microprocessor is via Status Buffer U2220. Data bits applied to the input pins are buffered onto the Data Bus when enabled by the Address Decode circuitry. Via the Status Buffer, the processor is able to (1) determine the settings of front- and rear-panel pots and switches, (2) determine instrument type (2445B or 2455B), (3) determine if a triggered sweep is in progress, and (4) read the contents of the Readout RAM. When disabled, the buffer outputs are switched to high impedance states to isolate them from the buffered Data Bus.

Data transfer from the Microprocessor to the Analog Control circuitry is via registers U2210 and U2310. Via register U2210, the Microprocessor is able to select the pot-scanning multiplexers, turn the trigger LED on and off, and control other hardware via serial control data and the attenuator strobe. Via register U2310, the processor controls pot selection, ROM addressing, and power down timing.

Front-Panel Switch Scanning

The Front-Panel Switches are arranged in a matrix of ten rows and five columns. Most of the row-column intersections contain a switch. When a switch is closed, one of the row lines is connected to one of the column lines through a diode. Reading of the switches is accomplished by setting a single row line LO and then checking each of the five column lines sequentially to determine if a LO is present (signifying that a switch is closed). After each of the five columns have been checked, the current row line is reset HI and the next row line is set LO for the next column scan cycle. A complete Front-Panel scan consists of all ten row lines LO in sequence and performing a five-column scan for each of the rows.

Row lines are set LO when the microprocessor writes a LO to one of the flip-flops in octal registers U2301 or U2201. The row data placed on the buffered Data Bus by the Microprocessor is clocked into the registers as two, eight-bit words by clocks from the Address Decode circuitry (DAC LSB CLK for the lower eight bits and DAC MSB CLK for the upper eight bits). All eight outputs of register U2201 and two outputs of U2301 drive the ten rows of the front-panel switch matrix (the fifth line of the matrix is not used). Series resistors in the lines limit current flow and eliminate noise problems associated with excessive current flow.

While each row is selected, the processor will scan each of the five column lines in sequence. To scan the columns, the processor increments three data select bits from U2301 that define the column to be checked. Eight-line data selector U2410 connects the associated column line to Status Buffer U2220. As each line is selected, the Microprocessor reads the Status Buffer to determine if the associated switch is open or closed.

In addition to the front-panel switches, the CAL/NO CAL jumper (P501) is checked to determine whether the instrument should be allowed to execute the calibration routines. The levels on U2410 pin 7 and 9 are read by scanning two additional columns at power-up. If the jumper is pulling the CAL bit LO, the operator will be allowed to use the calibration routines stored in firmware. If the NO CAL bit is pulled LO, the calibration routines may not be performed. If the jumper is removed, and neither bit is pulled LO, the Microprocessor is forced into a special diagnostic mode (CYCLE) used to record certain operating failures during long-term testing of the instrument. (See the "Maintenance" section for an explanation of the diagnostic modes.) Removing P501 or switching it between the CAL and NO CAL positions will not be recognized by the Microprocessor until the instrument is powered down and then turned back on.

The resistors in series with the input lines to U2410 are current-limiting resistors that protect the CMOS eight-line data selector from static discharges. The resistors connected from the input lines to the +5 V supply are pull-up resistors for the front-panel column lines.

Digital-to-Analog Converter (DAC)

DAC U2101 is used to set the various analog references in the instrument and is used to determine the settings of the front panel potentiometer. The 12-bit digital values to be converted are written to octal registers U2301 and U2201 for application to the DAC input pins. The DAC then outputs two complementary analog currents that are proportional to the digital input data. (Complementary, in this case, means that the sum of the two output currents is always equal to a fixed value.)

The maximum range of the output currents is established by a voltage-divider network composed of R2010, R2012, R2013, and R2011 conected to the positive and negative reference current inputs of the DAC (pins 14 and 15 respectively). A +10-V reference voltage applied to the DAC through R2013 sets the basic reference current. Resistor R2011 and potentiometer R2010 provide a means to adjust this current over a small range for calibration purposes. The nominal reference current is 1 mA, the DAC full-scale output current is 4 mA. The output currents flow through series resistors R2520 and R2521, connected to the +1.36-V reference, and proportional voltages result.

Pot Scanning

The Pot Scanning circuitry, in conjunction with the DAC, derives digital values for each of the various front-panel potentiometers. Scanning of the pots is accomplished by data selectors U2401, U2501, and U2601. Three bits are written to register U2310 and select the pot to be read. The bits are latched in the register and keep the pot selected until the register is reset. The Microprocessor writes a LO to the inhibit input pin (pin 6) of either U2401, U2501 or U2601 via register U2210 to enable the device. The enabled data selector connects the analog voltage at the wiper of the selected pot to comparator U2510.

Comparator U2510 compares the analog voltage of each pot to the output voltage from the DAC (pin 18). To determine the potentiometer output voltage, the processor performs a binary search routine that changes the output voltage from the DAC in an orderly fashion until it most closely approximates the voltage from the pot.

The conversion algorithm is similar to successive approximation and generates an eight-bit representation of the analog level. When the pot's value is determined, the Microprocessor stores that value in memory. Once all of the pots have been read and the initial value of each has

been stored, the processor uses a shorter routine to determine if any pot setting changes. To do this the DAC output is set to the last known value of the pot (plus and minus a small drift value), and the status bit is read to see that a HI and LO occurs. If within the limits, the processor assumes that the pot setting has not changed and scans the next pot. When the processor detects that a pot setting has changed, it does another binary search routine to find the new value of that pot.

Analog Control

The operating mode and status of the instrument requires that various analog voltages (for controlling instrument functions) be set and updated. The digital values of the controlling voltages are generated by the Microprocessor and converted by the DAC. Analog multiplexers U2521 and U2530 (on diagram 2) and U170 (on diagram 4) route the DAC voltages to sample-and-hold circuits that maintain the control voltages between updates.

The Microprocessor writes three selection bits to register U2301 that directs the DAC output to the appropriate sample-and-hold circuit and charges a capacitor (or capacitors) to the level of the DAC. When the processor disconnects the DAC voltage from the sample-and-hold circuit (by disabling the multiplexer) the capacitor(s) remains charged and holds the control voltage near the level set by the DAC. Due to the extremely high input impedance of the associated operational amplifiers, the charge on the capacitor(s) remains nearly constant between updates.

FRONT-PANEL CONTROLS

The Front Panel is the operator's interface for controlling the user-selectable oscilloscope functions. Along with the crt, it provides visual feedback to the user about the present operating state of the instrument.

Most of the Front-Panel controls (diagram 3) are "cold" controls; i.e., they are not connected directly into the signal path. Therefore, associated circuits are not influenced by the physical parameters (such as capacitance, resistance, and inductance) of the controls. In addition, translating the analog output levels of most of the potentiometers to digital equivalents allows the processor to handle the data in ways that result in a variety of enhanced control features.

To maintain the front-panel operating setup between uses of the instrument, the digitized values of the potentiometers and front-panel switch settings are stored in battery backed up RAM so that when the instrument power is turned off, these control settings are not lost. Then, when power is next applied, the instrument will power up to the same configuration as when the power was last removed (assuming the settings of the non-digitized pots and switches remain the same).

The Front-Panel Controls also allow the user to initiate and direct the diagnostic routines (and when enabled, the calibration routines) programmed into the read-only memory (ROM). These routines are explained in the Maintenance section of this manual.

Front-Panel Switches

The Front Panel Switches are arranged in a ten-row-by-five-column matrix, with each switch assigned a unique location within the matrix (see Figure 3-3). A closed switch connects a row and a column together through an isolating diode. To detect a switch closure, the switch matrix is scanned once every 32 ms (every tenth Microprocessor interrupt cycle). When scanning, the Microprocessor sequentially sets each individual row line LO. A closed switch enables the LO to be passed through the associated diode to a column line. When the processor checks each of the five column lines associated with the selected row, the LO column is detected. The intersection of the selected row and the detected column uniquely identifies the switch that is closed. Further information about switch scanning is found in the "Front-Panel Scanning" description located in the "Analog Control" discussion.

As each switch is read, the processor compares the present state of the switch to its last-known state (stored in memory) and, if the same, advances to check the next switch. When a switch is detected as having changed, the processor immediately reconfigures the setup conditions to reflect the mode change and stores the new state of the switch in memory. The detected status of the switch on each of the following scan cycles is then compared against the new stored data to determine if the switch changes again. The 32-ms delay between the time a switch is detected as having changed and the next time it is read effectively eliminates the effects of switching noise (switch bounce) that may occur after the switch is actuated.

Front-Panel Pots

The thirteen Front-Panel Potentiometers, READOUT INTENSITY, and INTENSITY are "cold" controls that control the linear functions of the instrument. (SCALE ILLUM and FOCUS are not considered part of the Front-Panel Control circuitry for the purposes of this description.) All are digitized and control their functions indirectly. Data Selectors U2401, U2501, and U2601 in the Analog Control circuitry (diagram 2) route the wiper arm voltage of the pot being read to comparator U2510 where it is compared

with the output of DAC U2101. The processor changes the DAC output until it most closely matches the output voltage of the pot, then stores the digital value of the "match". See the "Pot Scanning" description in the "Analog Control" discussion for further information on the reading of pot values.

Like the switch matrix scanning, the Front-Panel pot scanning routine is performed every 16 ms. When entered, the routine reads the settings of the "last-moved" pot and one "unmoved" pot. Each succeeding scan continues to read the last-moved pot in addition to a new unmoved pot. In this way, each pot is monitored, but most of the scan time is devoted to the pot that is still moving (needing continuous updating).

As the initial pot settings are determined, a digital representation of each value is stored in memory. The processor then checks each pot against its last-known value to determine if a pot has moved. If a pot is detected as moving, the processor executes a routine that converts the movement (displacement from last-set value) into a corresponding control voltage.

When producing the actual analog control levels, the processor can manipulate the digital values read for the various pots before sending the output data to the DAC. This allows many of the oscilloscope parameters to vary in an enhanced fashion. The pot data is manipulated by the processor in a manner that produces such features as variable resolution, continuous rotation, fine-resolution backlash, and electrically detented controls.

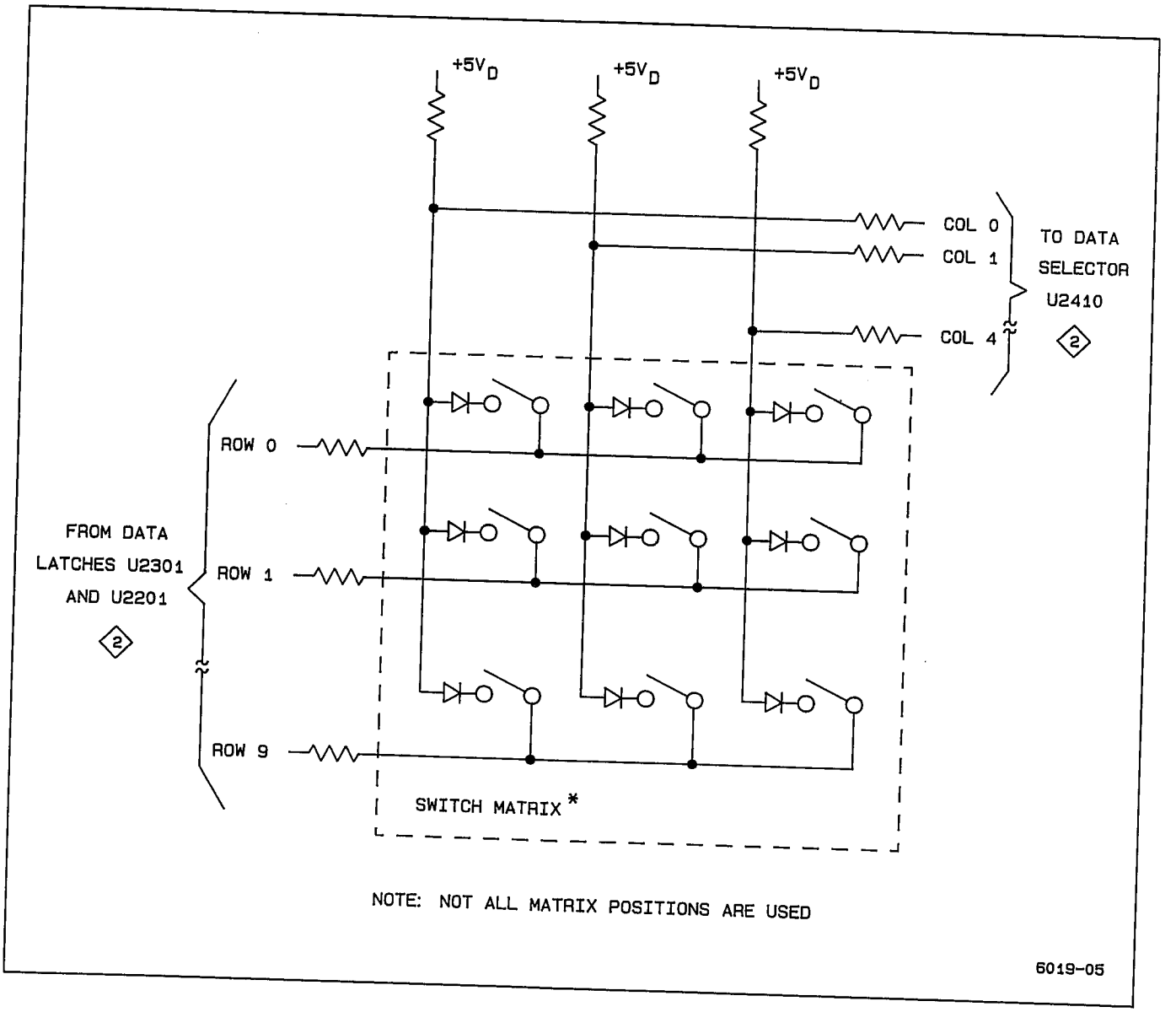


Figure 3-3. Front-panel switch matrix.

With all thirteen Front-Panel Potentiometers, READOUT INTENSITY, and INTENSITY controls, the processor reads the magnitude and direction of pot rotation and produces variable-resolution control voltages. If a pot's direction of rotation changes, the magnitude of the change from the last-set position remains small, or if it was not the last pot moved, a fine-resolution control voltage results. In the fine-resolution range, a given rotation displacement will cause a small control voltage change. The same displacement farther away from the last-set reference will cause a proportionally larger control voltage change, producing a coarse-resolution effect. If the changing pot is the last one moved and the direction of rotation remains the same, the algorithm continues from where it left off during the preceding scan; producing control voltage changes with the same increment as it was last using.

The delta reference controls (Δ REF OR DLY POS and Δ) are continuous-rotation potentiometers. They each consist of two pots ganged together with their wiper arms electrically oriented at 180° apart. As the wiper of one pot is leaving its resistive element, the wiper of the other pot comes onto its element. The Microprocessor has the ability to watch the output voltage from each wiper and when it detects that the controlling wiper is nearing the end of its range, it will switch control over to the other wiper. The routine the processor uses to watch these pots sets the associated control voltage on the basis of relative voltage changes (ΔV) that occur. Switching between the pots to change control to the opposite wiper arm is based on specific voltage levels being sensed.

Sensing specific voltage levels is also used when reading the VOLTS/DIV VAR, SEC/DIV VAR, and HOLDOFF controls. These pots have both a mechanical detent and a processor-generated electrical detent. As one of these controls is moved out of the mechanical detent position, the processor watches the analog voltage changes that occur; but the associated control voltage will not change until a specific voltage level (the electrical detent level) is reached. Once the electrical detent value is exceeded, the processor begins to vary the associated control voltage in response to further pot rotation. When returning to the mechanical position, the electrical detent level is reached first, and the variable voltage action is stopped before the mechanical detent is entered.

Front-Panel Status LEDs

Light-emitting diodes (LEDs) are used to provide visual feedback to the operator about the oscilloscope status and operating mode by backlighting front-panel nomenclature. A 48-bit status word, defining the diodes to be illuminated, is generated by the processor and then serially clocked into the six LED-Status Registers (U3001, U3002, U3003, U3004, U3005, and U3006). The registers hold the selected diodes on until the next update. Whenever the processor detects that a front-panel control has changed (and a new status display is required), a new status word is generated and applied to pin 1 of U3002. As each of the bits is clocked into the Q_A position of U3002, the preceding bit is shifted to the next register position. After 48 bits have been clocked into (and 40 bits through) U3002, all six LED-Status registers are full and contain the LED illumination pattern to be displayed to the user. A LO at any Q output of the registers illuminates the corresponding front-panel LED.

The TRIG'D LED is not driven by the LED-Status Register. It is driven by the Analog Control circuitry and illuminated whenever a triggered sweep is in progress.

ATTENUATORS AND PREAMPS

The Attenuators and Preamps circuitry (diagram 4) allows the operator to select the vertical deflection factors. The Microprocessor reads the Channel VOLTS/DIV switches and VOLTS/DIV VAR controls and then digitally switches the attenuator and sets the preamplifier gains accordingly.

CHANNEL 1 AND CHANNEL 2 ATTENUATORS

The Channel 1 and Channel 2 Attenuators are identical in operation, with corresponding circuitry in each channel performing the same function. Therefore, only the Channel 1 circuitry is described.

Input signals from the Channel 1 input connector are routed through an attenuator network by four pairs of magnetic-latch relay contacts. The position of the relays is set by Microprocessor data placed into Auxiliary Control Register U140. Relay buffer U110 provides the necessary drive current to the relays.

Four input coupling modes (1MΩ AC, GND, 1MΩ DC, and 50 Ω DC) and three attenuation factors (1X, ÷10, and ÷100) may be selected by closing different combinations or relay contacts. The three attenuation factors, along with the variable gain factors of the Vertical Preamplifier, are used together to obtain the crt deflection factors. The relays are magnetically latched and once set, remain in position until new attenuator-relay-setting data and strobes are generated. (See the "Auxiliary Control Register" description for a discussion of the relay-latching procedure.)

The 50 Ω termination resistor has a thermal sensor associated with it that produces a dc voltage (CH 1 OVL) proportional to the input power. Should the input power exceed the normal safe-operating level for the 50 Ω DC input, the termination resistor temperature will exceed the

normal operating limit and change the output voltage of the thermal sensor. The amplitude of this dc level is periodically checked via comparator U2510 and DAC U2101 (on diagram 2) and allows the Microprocessor to detect when an overload condition is present. When an overload occurs, the processor switches the input coupling to the 1 MΩ position to prevent damage to the attenuator and displays 50 Ω OVERLOAD on the crt.

Compensating capacitor C105 is adjusted at the time of calibration to normalize input capacitance of the preamplifier to the attenuator.

A probe-coding ring around the BNC input connector passes probe coding information (a resistance to ground) to the Analog Control circuitry for detection of probe attenuation factors. The readout scale factors are set to reflect the detected attenuation factor of the attached probe.

Auxiliary Control Register

The Auxiliary Control Register allows the Microprocessor to control various mode and range dependent functions of the instrument. Included in these functions are: attenuation factors, input coupling, Channel 3 and Channel 4 gains, vertical-bandwidth limiting, the X-Y display mode, and the state of the measurement PAL.

When the Microprocessor sets the input coupling mode and attenuation factors for Channel 1 and Channel 2, a series of eight, 16-bit control words are serially clocked into shift registers U140 and U150 (eight bits in each register). Each control word is used to set the position of one of the eight attenuator and coupling relays (four relays are in each attenuator assembly). Each control word will have one HI bit. This bit will correspond to the specific relay contact to be closed. Relay buffers U110 and U130A (for Channel 1) and U120 and U130B (for Channel 2) are Darlington configurations that invert the polarities of all bits. This results in a LO being applied to only the coil lead associated with the contact to be closed; all other coil leads are held HI.

To set a relay once the control word is loaded, the Microprocessor generates a ATTN STRB (attenuator strobe) to U130G pin 7 via R129 and C130. The strobe pulses the output of U130G LO for a short time. This output pulse attempts to turn on both Q130 and Q131 (relay drivers) via their identical base-bias networks. Due to the lower level from the turned on Darlington relay buffer (coupled through the associated coil diode and either CR130 or CR131 to one of the bias networks), one transistor will turn on harder as the ATTN STRB pulse begins to forward bias the transistors. The more positive collector voltage of the transistor turning on harder is fed through the bias diode (again either CR130 or CR131) to further turn off the opposite transistor. This action results in one transistor being fully on and the other one being fully off. The saturated transistor sources current through the two stacked relay coils to the LO output of either U140 or U150 (current sink) to close the selected contacts. Once set, the magnetic-latch feature will hold the relay set to this position until opposing data is clocked into the Auxiliary Control Register and strobed into the relay. All coil leads for the remaining relays are set HI, and only the selected relay will be set.

To set the seven remaining Attenuator and coupling relays, the sequence just described is repeated seven more times. Whenever the Microprocessor determines that the attenuation factor or input coupling has changed, the entire relay-setting procedure is repeated for all eight relays.

After the coupling and attenuator relays have been latched into position, the Auxiliary Control Register is free to be used for further circuit-controlling tasks. Eight more bits of control data are then clocked into U140 either to enable or disable the following functions: vertical bandwidth limiting (BWL), triggered X-Y mode (TXY), the A and B Sweep Delay Comparators (BDCA and BDCA), and slow-speed intensity limit (SIL); or to alter the Channel 3 and Channel 4 gain factors (GA3 and GA4). Four other bits are clocked into register U150: one to produce the CTC signal, one to control the scale illumination circuit during SGL SEQ display mode, and two (CNTL1 and CNTL2) to control the state of the measurement PAL, U975. The CTC control bit is used to enable a sweep-start linearity circuit in the A Sweep circuitry (diagram 5) on the 2 ns and 20 ns per division sweeps.

Analog Control Demultiplexer

When enabled by the Address Decode circuitry, Analog Control Demultiplexer U170 directs the analog levels applied to pin 3 from DAC U2101 (diagram 2) to one of six sample-and-hold circuits. In the Preamplifier circuitry, the sample-and-hold circuits maintain the VAR gain and DC Bal control-voltage levels applied to both the Channel 1 and Channel 2 Preamplifiers U100 and U200 between updates. Two of the Demultiplexers outputs direct analog levels to the Holdoff and Channel 2 Delay offset sample-and-hold circuits (diagram 5). Routing is determined by the three-bit address from register U2301 (diagram 2) applied to Demultiplexer U170 on pins 9, 10, and 11.

Channel 1 Preamplifier

Channel 1 Preamplifier U100 converts the single-ended input signal from the Channel 1 Attenuator to a differential output signal used to drive the Vertical Channel Switch. The device produces either amplification or attenuation in predefined increments, depending on the control data writ-

ten to it from the Microprocessor. The preamp also has provisions for VAR gain, vertical positioning, and a trigger signal pickoff.

The Channel 1 vertical input signal is applied to pin A of Channel 1 Preamplifier U100. Control data from the processor is clocked into the internal control register via pin 22 (CD) by the clock signal applied to pin 23 ($\overline{CC}$). The data sets the device to have an input-to-output gain ratio of 2, 4, or 10, depending on the VOLTS/DIV control setting.

Two analog control voltages set by DACs modify the differential output signal at pins 9 and 10. The front-panel Channel 1 POSITION control supplies a position signal to U100 pin 17 (via MUX U2530 and sample-and-hold U2430 and C2432) that vertically positions the Channel 1 display on the CRT. A DC Bal signal is applied to pin 2 of U100 from MUX U170 via the sample-and-hold circuit composed of U160A and C177. This DC BAL signal is a dc offset-null level that is determined during the automatic DC Bal procedure. The offset value is stored as a calibration constant in RAM and is recalled at regular intervals to set the DC Bal level, holding the Preamplifier in a dc balanced condition.

The Channel 1 VOLTS/DIV VAR control is monitored by the Microprocessor during the front-panel scanning routine. When the processor has determined where the VOLTS/DIV VAR control is positioned, it causes DAC U2101 (diagram 2) to produce a corresponding control level and routes it to the VAR gain sample-and-hold circuit composed of U160D, C179, and associated components. The control voltage at the output of U160D (pin 14) sets the variable gain of the Preamplifier.

A pickoff amplifier internal to U100 conditions the trigger signal and provides the proper signal level at pin 15 to drive the A/B Trigger Generator (U500, diagram 5). The pickoff point for the trigger signal is prior to the addition of the vertical position offset, so the position of the signal on the crt has no effect on the trigger operation. However, the pickoff point is after the DC Bal and Variable gain signals have been added to the signal so both of these functions will affect trigger operation.

Common-mode signals are rejected from the trigger signal by the circuitry composed of operation amplifier U450B and associated components. The inverting input of U450B (pin 6) is connected to the common-mode point between APO+ (pin 12) and TPO− (pin 15) of U100. Any common-mode signals present are inverted and applied to a common-mode point between R451 and R453 to cancel the signals from the differential output. A filter network composed of LR 180 and the built-in circuit board capacitor (5.6 pF) reduces trigger noise susceptibility. Trigger signals for options are obtained from J100.

The Channel 1 input signal used to provide the horizontal deflection for the X-Y displays is obtained from U100 pin 11. The components between pin 11 and the Horizontal Output Amplifier provide phase compensation of the signal. During instrument calibration, the delay produced by C115, C116, L115, R115, and variable capacitor C118 is matched to the 78-ns delay of the vertical delay line (DL100, diagram 6).

Channel 2 Preamplifier

Operation of Channel 2 Preamplifier U200 is nearly identical to that of the Channel 1 Preamplifier just described. The exceptions are that the output polarity of the Channel 2 signal may be either normal or inverted and that the signal obtained from the BPO+ output (pin 11) is conditioned differently for a different purpose than in the Channel 1 Preamplifier circuitry.

Inverting the Channel 2 signal for the CH 2 INVERT feature is accomplished by biasing on different amplifiers. The control data clocked into the internal control register from pin 22 sets up the necessary switching.

The Channel 2 BPO+ signal at U200 pin 11 provides an accurate representation of the Channel 2 signal at the rear-panel CH 2 OUT connector.

Channel 3 and Channel 4 Preamplifier

The functions provided by the Channel 3 and Channel 4 Preamplifier are similar to those provided by the Channel 1 and Channel 2 Preamplifiers. The single-ended CH 3 and CH 4 input signals are converted to differential signals, and vertical gain and vertical positioning are added to the output signals. Trigger pickoff signals are generated for both channels and are routed to the Trigger hybrid.

Channel 3 and Channel 4 gains may be either 0.1 volt per division or 0.5 volt per division. The logic levels of control bits applied to U300 pin 30 (GA3) and pin 31 (GA4) from Auxiliary Control Register U140 sets the gain of the Channel 3 and Channel 4 preamplifiers respectively. Vertical positioning of the Channel 3 and Channel 4 signals on the crt is controlled by the voltage levels applied to pin 29 (POS3) and pin 32 (POS4) from the front-panel CH 3 and CH 4 POSITION potentiometers (via MUX U2530 and sample-and-hold amplifiers U2430C and C2333 and U2430D and C2332).

Dc offsets in the output signal due to any tracking differences between the +5-V and the −5-V supply to

U300 are reduced by the tracking regulator circuit composed of U165A, Q190, and associated components. Operational amplifier U165A and Q190 is configured so that the output of voltage at the emitter of Q190 follows the −5-V supply applied to R198. This tracking arrangement ensures that the supply voltages are of equal magnitudes to minimize dc offsets in the output signals.

Scale Illumination

The Scale Illumination circuit consists of U130C, U130D, U130E, U130F, and associated components. The circuit enables the operator to adjust the illumination level of the graticule marks on the crt face plate using the SCALE ILLUM control.

Components U130C through U130F, depicted on diagram 4 as inverters, are actually Darlington transistor pairs. Figure 3-4 is a simplified illustration of the Scale Illumination circuitry, redrawn to show U130C through F as Darlington transistor pairs for the purpose of the following description.

Darlington transistors U130D and U130E control the current flow to scale-illumination lamps DS100, DS101, and DS102. Base drive current for U130D and U130E via R133 is set by the front-panel SCALE ILLUM pot R134. Voltage at the more negative end of the pot is set by the self-biasing configuration of U130F and R135. The voltage level established by these two components is two diode drops above ground (≈ 1.2 V) so that, at full counterclockwise rotation, the wiper voltage of the SCALE ILLUM pot will just match the turn-off point of U130D and U130E. The voltage at the other end of the pot is set by the collectors of U130D and U130E. As the SCALE ILLUM pot is advanced, the base drive to U130D and U130E increases, and the voltage on their collectors moves closer to ground potential. This increases the current through the scale-illumination lamps to make them brighter and produces some negative feedback to the base circuit through the SCALE ILLUM pot. Negative feedback stabilizes the base drive to U130D and U130E to hold the illumination level constant at the selected setting of the SCALE ILLUM control.

During SGL SEQ display mode, the graticule is illuminated only once during the sequence for photographic purposes. In this mode, a HI is initially written to Auxiliary Control Register U150 (bit Q_H). This turns on U130C and shunts the base drive current of U130D and U130E to ground. At the point in the sequence when the graticule should be illuminated, the processor writes a LO to bit Q_H, and Q130C is turned off. This enables U130D and U130E to turn on the lamps to the illumination level set by the SCALE ILLUM pot.

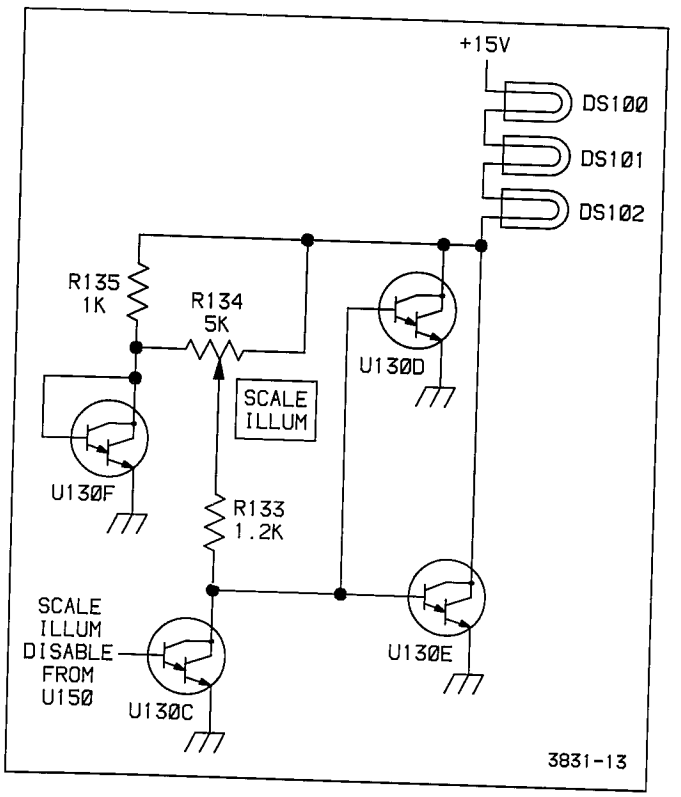


Figure 3-4. Scale illumination circuit.

DISPLAY SEQUENCER, TRIGGERS, AND SWEEPS

The Display Sequencer circuitry (diagram 5) controls and sequences the "analog-type" oscilloscope functions in real time, dependent on control data it receives from the Microprocessor. The A/B Trigger circuitry, under control of the Display Sequencer, detects when triggering requirements are met and initiates the appropriate sweep. The A Sweep and B Sweep circuits generate sweep ramps under control of the Display Sequencer when triggered by the A/B Trigger circuitry.

Display Sequencer

The Display Sequencer consists primarily of integrated circuit U650. This IC accepts analog and digital control signals from various parts of the instrument and, depending on the control data string clocked into its internal control register from the Microprocessor, will change control signals that it sends to other, signal-handling circuits.

In the course of developing waveform displays, the Display Sequencer selects one or more vertical channels, sets the trigger source, and selects the horizontal display

mode. In most cases, the trigger selection does not change after it has been set unless a front-panel trigger control is changed. An exception is that in VERT TRIGGER MODE, the trigger source tracks the sequencing of the vertical channels (unless AUTO LVL MODE, or CHOP VERTICAL MODE is also selected). Trigger source selection lines are changed only during trigger holdoff time between sweeps.

Fifty-five bits of serial data from the processor defining the instrument's operating sequence are applied to the Display Sequencer data input, pin 25. The data string is clocked into U650 to the internal control register by the processor-generated control clock applied to pin 24. The data string is organized in several fields, with each field defining the operating mode of one specific instrument function.

Display Sequencer U650 controls the various functions defined by the data fields by setting the levels of the associated control lines. The functions and controlling signal lines for each function are as follows:

VERTICAL DISPLAY SELECTION. CH 1, CH 2, CH 3, CH 4, ADD, and Readout Y signals are selected by the $\overline{VS1}$, $\overline{VS2}$, $\overline{VS3}$, and $\overline{VS4}$ control signals. See the Vertical Channel Switch description for further information.

HORIZONTAL DISPLAY SELECTION. A Sweep, B Sweep, CH 1 (for X-Y displays) and Readout X are selected by the $\overline{HSA}$ and $\overline{HSB}$ control signals. See the Horizontal Output Amplifier description for further information.

TRIGGER SOURCE SELECTION. CH 1, CH 2, CH 3, CH 4, ADD, Line, and a sample of the vertical output signal (for calibration purposes only) are selectable as the Trigger SOURCE by the $\overline{SR0A}$, $\overline{SR1A}$, $\overline{SR2A}$, $\overline{SR0B}$, $\overline{SR1B}$, and $\overline{SR2B}$ control lines (pins 28, 27, 29, 32, 31, and 30 respectively). See the A/B Trigger description for further information.

TRIGGER HOLDOFF. Sweep recovery time and the circuit initialization time required when front-panel controls are changed are controlled by the THO (trigger holdoff) signal.

DELTA TIME (Δt) DELAY SELECTION. DLY REF 0 or DLY REF 1 is selected by the $\overline{DS}$ (delay select) signal.

TRIGGER and SWEEP ACTIVITY (STATUS). The activity of the Trigger and Sweep circuits, as indicated by the $\overline{SGA}$, $\overline{SGB}$, $\overline{TSA}$, and $\overline{TSB}$ lines, is reported to the Microprocessor via the TSO (trigger status output) line when clocked by the $\overline{TSS}$ (trigger status strobe) signal.

INTENSITY CONTROL. The readout intensity, display intensity, and display intensity compensation are controlled by the BRIGHT output level.

DISPLAY BLANKING. Display blanking for CHOP VERTICAL MODE, Readout transitions, and front-panel control changes is controlled by the BLANK output.

READOUT CONTROL. The vertical selection, horizontal selection, and intensity controls are all set to their readout modes either at the end of an A Sweep ($\overline{SGA}$ goes HI) or in response to a readout request ($\overline{ROR}$) from the Readout circuitry (diagram 7). While in the readout mode, the BLANK control signal is driven by the readout blank ($\overline{ROB}$) input signal on pin 5 (also from the Readout circuitry). The readout active line ($\overline{ROA}$, pin 6), when set LO, tells the Readout circuitry that readout dots may be displayed if necessary. The $\overline{ROA}$ signal is always set LO at the start of the trigger holdoff time following sweeps, and it is held there until the holdoff time is almost over. This allows the majority of holdoff time to be used for displaying readout dots. The Display Sequencer will switch the $\overline{ROA}$ signal back to HI before the end of holdoff so that the readout display does not interfere with display of the vertical signal at the triggering event.

TRACE SEPARATION. Vertical separation between the A Sweep trace and the B Sweep traces (for alternate horizontal sweep displays), and between the reference B Sweep trace and the delta B Sweep trace (when delta time is selected in B Sweep only mode), is enabled by the TS1+TS2 output.

X10 HORIZONTAL MAGNIFICATION. Horizontal X10 magnification is controlled by the $\overline{MAG}$ output.

CALIBRATOR TIMING. The 5-Hz to 5-MHz drive signal to the Calibrator circuitry is provided by the CT output.

DELAY GATE OPERATION. Analog Switches U850B and U850C select the delay references for each sweep. Depending on the display mode and point in the display sequence, the DS control signal (U650 pin 40) routes one of the two analog delay references through U850B and U850C to the two sweep hybrids. The selected reference level is compared against the changing sweep ramp voltages to generate the delay gates that control each sweep's functions.

After an A Sweep has been initiated by a trigger, a delay gate circuit within U700 compares the A Sweep

ramp voltage to the selected delay reference. When the sweep ramp reaches the delay reference level, the DG (delay gate) output goes LO, enabling the B trigger portion of U500 and B Sweep hybrid U900. Then, when B triggering occurs (for TRIG AFT DLY mode), the A/B Trigger hybrid sets the $\overline{TGB}$ (trigger gate B) signal LO, initiating the B Sweep. In RUN AFT DLY mode, however, the $\overline{TGB}$ signal to U900 is held LO, and the B Sweep is initiated at the end of the A Sweep delay time when the A Sweep delay gate goes LO.

STATUS MONITORING. As the Display Sequencer controls the display system in real time, it continually monitors the trigger and sweep operations and updates the internal trigger status register accordingly. The Microprocessor checks the contents of this register every 3.3 ms to determine the current status of the trigger and sweep circuitry. The Microprocessor reads the trigger status register by generating a series of trigger status strobe ($\overline{TSS}$) pulses (U650 pin 19) to serially clock the contents of the register out to the TSO (trigger status output) line and onto the Data Bus (via Status Buffer U2220 on diagram 2). The system status information obtained by this check is used for AUTO LVL triggering, AUTO free-run triggering, detecting the completion of all sweeps in a SGL SEQ display, automatic measurement functions, and during instrument calibration.

INTENSITY CONTROL. The Display Sequencer controls the intensity for both sweep and readout displays. The analog levels at pins 22 and 23 determine the basic intensity level of the displays. Two internally generated DAC currents (developed by multiplying the IREF current at pin 20 by two processor-generated numbers stored internally) are added to the basic intensity level currents to produce the display intensity seen on the crt (see Table 3-1). The two DAC currents added to the INTENSITY current are dependent on sweep speed, number of channels being displayed, and whether or not the X10 MAG feature is in use. These added currents increase crt beam current and hold the display intensity somewhat constant under the varying display conditions. The resulting current is applied to Z-Axis Amplifier U950 (diagram 6) from the BRIGHT output of the Display Sequencer (pin 21).

To produce the intensified zone on the A Sweep trace for A intensified by B Sweep displays, an additional current is added to the crt drive signal by the Z-Axis Amplifier during the concurrence of the $\overline{SGAZ}$ and $\overline{SGBZ}$ (sweep gate A and B z-axis) signals.

The readout intensity (ROI) level, controlled from the front-panel READOUT INTENSITY pot (via MUX U2530 and sample-and-hold U2630A and C2732). The Microprocessor increases readout intensity when the pot is rotated

Table 3-1
Intensity Control

Type of Display	Horizontal Selects		Resulting Current at BRIGHT Output
	$\overline{HSA}$	$\overline{HSB}$	
X/Y	LO	LO	DI (display intensity) only
A Sweep	LO	HI	DI + A SWP DAC current
B Sweep	HI	LO	DI + B SWP DAC current
Readout	HI	HI	ROI (readout intensity) only

either direction from center. Minimum readout intensity current occurs at the midpoint of the READOUT INTENSITY pot rotation. The Microprocessor also detects to which side of center the READOUT INTENSITY control is set. Depending on the status received, the processor sets up the Readout circuitry (diagram 7) to display either all of the readout information or just the "delta type" readouts.

Blanking of the crt display during CHOP VERTICAL MODE displays or when switching between dot positions in the readout displays is controlled by the Display Sequencer's BLANK output (pin 3). When the signal is LO, the crt z-axis is turned on to the selected intensity level; when HI, the crt display is blanked.

READOUT CONTROL. The readout request signal ($\overline{ROR}$), the readout active signal ($\overline{ROA}$), and the readout blank signal ($\overline{ROB}$) control readout displays. During the first part of the holdoff time, up until one or two holdoff ramps before holdoff time ends (dependent on the sweep rate), the Display Sequencer sets the $\overline{ROA}$ signal line LO. While the $\overline{ROA}$ line is LO, the Readout circuitry may display readout character dots if necessary. During readout displays, the horizontal and vertical select signals ($\overline{HSA}$, $\overline{HSB}$, $\overline{VS1}$, $\overline{VS2}$, $\overline{VS3}$, and $\overline{VS4}$) are all set HI. This deselects the waveform-related sweep and deflection signals and gives display control to the Readout circuitry. While readout information or cursors are being displayed, the BLANK output signal (pin 3) is controlled by the readout blank ($\overline{ROB}$) signal from the Readout circuitry, and the readout intensity (ROI) signal pin (pin 23) controls the BRIGHT output level.

During holdoff, the Display Sequencer always sets the readout active ($\overline{ROA}$) line LO. As previously described, setting the $\overline{ROA}$ signal LO allows the Readout circuitry to display readout dots. In some settings of the SEC/DIV switch, with adequate trigger rates, holdoff time is provided for the Readout circuitry to display all the readout information without causing noticeable display flicker.

In those cases where the holdoff time is insufficient to prevent flicker, a portion of the Readout circuitry will request display control by setting the readout request ($\overline{\text{ROR}}$) signal LO. The Display Sequencer recognizes all readout requests immediately and switches the horizontal and vertical select lines to the readout display mode. The Readout circuitry displays one readout dot and then resets the readout request HI to switch back to the display of waveforms. Readout requests occur as required during sweep times to keep the readout display up to date. (See "Readout" description for further information).

TRACE SEPARATION. The TRACE SEP feature is used to position the alternate B Delayed Sweep trace downward from the A Sweep when Alternate Horizontal Display Mode (TURN-ALT) is active. It is also used when either the Δt or $1/\Delta t$ measurement function is used with B Sweep only displays. In the latter case, the TRACE SEP control vertically positions the trace(s) associated with the Δ control.

When the Display Sequencer determines that trace separation should be active, the LO TSIN level at pin 7 is routed to pins 9 and 8, the TS1 and TS2 outputs (connected together). This LO output turns off transistor Q600 (diagram 6), thereby enabling the trace separation voltage from the front-panel TRACE SEP pot (via MUX U2530 and sample-and-hold U2630C and C2631) to be applied to pin 42 of Vertical Output Amplifier U600. To disable the trace separation function, the Display Sequencer sets the TS1 + TS2 control line HI, turning on Q600 and shunting the trace separation signal to ground.

X10 MAG SELECT. The $\overline{\text{MAG}}$ (sweep magnifier) output (pin 39) drives the magnifier control input (pin 14) of Horizontal Output hybrid U800 and the select input (pin 9) of analog switch U860C (diagram 6). Analog switch U860C routes a magnifier gain-control voltage to the Horizontal Amplifier to set the horizontal gain for the X10 magnified displays.

CH 2 DELAY OFFSET. The $\overline{\text{VS2}}$ (vertical select, channel 2) output applied to analog switch U860B at pin 10 routes a calibrated offset voltage from sample-and-hold buffer U165D to both sweep hybrids when the Channel 2 vertical signal is being displayed. The offset voltage is used to eliminate the apparent propagation delay between the Channel 2 and the Channel 1 (or CH 2 and either one of the other channels). A step in the calibration procedure allows use of the front-panel Channel 2 Delay Offset feature to be either enabled or disabled. When enabled, the Channel 2 offset may be adjusted up to ± 500 ps (with respect to Channel 1) using the Δ control.

CALIBRATOR TIMING. The Calibrator timing signal (CT) from the Display Sequencer is generated by an internal counter. The counter divides the 5-MHz clock input at pin TC (timing clock) by a value that is a function of sweep speed. The resulting square-wave output signal drives the Calibrator circuit. For ease of sweep rate verification, the Calibrator signal provides a display of five complete cycles on the crt at sweep speeds from 100 ms per division to 0.1 μs per division. Below 100 ms per division, the Calibrator output frequency remains at 5 Hz; and above 0.1 μs per division, the Calibrator frequency remains at 5 MHz.

When chopping between vertical channels, the Display Sequencer adds a 200-ns skew at the end of some sweeps to desynchronize the chop frequency from the sweep speed (to prevent the sweep from locking onto the chop frequency). Due to this, the Calibrator signal has an irregular pulse repetition characteristic between sweeps. This will not be apparent when observing the Calibrator signal on the instrument crt since the skew is synchronized to the sweep, but may be observed when the Calibrator output signal is used with other instrumentation. The skew can be eliminated by setting the instrument to SGL SEQ Mode (to shut off the sweeps).

Holdoff Circuitry

The holdoff circuit, used to delay the start of a sweep until all circuits have recovered from the previous sweep, is made up of U165C, Q154, Q155, and associated components. Operational Amplifier U165C and capacitor C180 form a sample-and-hold buffer used to set the charging current for holdoff-ramp integrating capacitor C171. A control voltage from digital-to-analog converter (DAC) U2201 (diagram 2) via multiplexer U170 (diagram 4) is stored on C180. The stored voltage level sets the base voltage for both Q154 and Q155 via amplifier U165C. Transistors Q154 and Q155 form a current-mirror with nearly equal collector currents. Transistor Q154 is a current-to-voltage converter that provides negative feedback to U165C, setting loop gain. Transistor Q155 acts as a constant-current source that charges integrating capacitor C171, producing a linear holdoff ramp.

A comparator circuit in U650 detects when the ramp crosses a predefined threshold voltage (approximately +3 V). When the threshold is reached, pin 10 of U650 (HRR) goes LO and the integrating capacitor is discharged. At that same time, an internal counter that keeps track of the holdoff ramp cycles is incremented. The ramps continue to be generated and reset until the holdoff ramp counter has counted the number of ramp cycles defined by the sweep-rate-dependent holdoff data field stored in the Display Sequencer control register. At all sweep speeds except 5 ns per division, the count is at least two holdoff ramp cycles. The front-panel variable HOLDOFF control affects holdoff time by varying the HOLDOFF control voltage to

U165C (from the DAC), changing the charging rate of integrating capacitor C171.

When holdoff time requirements are met (determined by the number of ramps counted), the Display Sequencer sets the THO (trigger holdoff) signal LO. This enables both the A Sweep hybrid (U700) and the A Trigger circuitry in U500. The Trigger circuit begins monitoring the selected trigger source line and, when a triggering event is detected that meets the triggering requirements defined by the stored control data, initiates the A Sweep and sets the $\overline{TSA}$ (trigger status, A Sweep) line to Display Sequencer U650 LO (indicating that the A Sweep has been triggered).

As the A Sweep circuit (U700) responds to the trigger, it sets the $\overline{SGA}$ (sweep gate A) line LO (via U980A) indicating that an A Sweep is in progress. After the sweep has run to completion, U700 sets the $\overline{SGA}$ line HI signaling the end of sweep. The Display Sequencer then sets the THO line HI, resetting A/B Trigger hybrid U500 and A Sweep hybrid U700 in preparation for the next sweep.

Triggers

The A/B Trigger hybrid (U500) and associated circuitry select the triggering signal source for each horizontal sweep as directed by the Display Sequencer. When the proper triggering criteria to initiate a sweep are detected, a triggering gate signal is produced to start the selected sweep.

Control data from the processor defining trigger mode, coupling, and slope parameters for each trigger is clocked into two storage registers internal to U500 by the A TRIG CLK signal on pin 23 ($\overline{CCA}$) and the B TRIG CLK signal on pin 47 ($\overline{CCB}$). The Display Sequencer selects the A trigger source with the $\overline{SR0A}$, $\overline{SR1A}$, and $\overline{SR2A}$ signal lines; the B trigger source is selected using the $\overline{SR0B}$, $\overline{SR1B}$, and $\overline{SR2B}$ signal lines. Table 3-2 illustrates trigger source selection.

To initiate the A Sweep, the trigger hybrid compares the selected signal to the analog trigger level input at pin 13, the TLA (trigger level A). B trigger signals are compared to the TLB (trigger level B) signal at pin 37 when trigger B Sweeps are required. When the proper trigger signal is detected, U500 outputs a trigger gate ($\overline{TGA}$ or $\overline{TGB}$) to the appropriate sweep circuit to initiate that sweep.

When an A Sweep is initiated, the trigger-status line ($\overline{TSA}$) (trigger status A, U500 pin 20) goes LO to signal the Display sequencer that a trigger has occurred. Until the sweep is completed, the $\overline{TGA}$ signal on pin 18 (or $\overline{TGB}$ signal on pin 42 for B Sweeps) remains LO. After the A Sweep is completed, the A Sweep Gate ($\overline{SGA}$) from A Sweep hybrid U700 (via U980A) will go HI, causing the Display Sequencer to set its THO (trigger holdoff) line (pin 13) HI. This resets the sweep hybrid and the trigger hybrid in preparation for the next trigger event.

Table 3-2
Trigger Source Selection

Select Inputs			Trigger Source
$\overline{SR2A(B)}$	$\overline{SR1A(B)}$	$\overline{SR0A(B)}$	
H	H	L	CH 1
H	L	H	CH 2
H	L	L	ADD
L	H	L	CH 3
L	L	H	CH 4
H	H	H	LINE (or BWLB)[a]

[a]During calibration routines from the Diagnostic Monitor.

The B Trigger Holdoff input (THOB, U500 pin 39) is held HI (keeping the B Trigger reset) until the A Sweep Delay Gate (DG, U700 pin 41) goes LO (see the following A Sweep description). When DG goes LO, the B Trigger portion of U500 is enabled. The B Sweep Trigger functions in a manner similar to that of the A Sweep Trigger just described. During a parametric measurement, the THOB line may be driven by either A Sweep Delay Gate, or BHO from the measurement PAL, U975. If CNTL1 is LO, THOB is driven by A Sweep Delay Gate through the buffer transistor Q741. If CNTL1 is HI, Q741 is held off by Q742 and THOB is driven by BHO.

A Sweep

When properly triggered, the A Sweep circuit generates linear sweep ramps of selectable slopes. When amplified, these ramp signals horizontally sweep the crt beam across the face of the crt. The A Sweep circuitry consists of U700, Q709, Q710, Q741, U910B, U980A, and associated components.

The A Sweep ramp signal is derived by charging one of several selectable capacitors from a programmable constant-current source. Capacitor selection depends on the sweep-rate-dependent control data (CD) on pin 29 that is clocked into A Sweep hybrid U700 by the A SWP CLK on pin 28 ($\overline{CC}$). This sweep-rate data causes some internal logic to select either hybrid-mounted capacitors CT0 or CT1 or capacitor C708 at the CT2 (timing capacitor two) pin. An additional capacitor, C709, may be selected (via Q709 and Q710) if the control data asserts the TCS (timing capacitor select) signal on pin 9. TCS will be HI for A

Sweep speeds slower than 1 ms per division. Capacitor C707 and associated circuitry form a linearity compensation circuit.

The constant current to charge the selected capacitor is derived from the DAC-controlled voltage, A TIM REF (A timing reference), generated on the Control Board. The ITREF input (U700 pin 24) is held at zero volts by an internal programmable current-mirror circuit at that input (see Figure 3-5). The A TIM REF voltage is applied to the current mirror via series resistors R723 and R724 to establish the input reference current (ITREF). The output of this current mirror is related to the input reference current by a multiple "M" that is set by a control data field stored in the internal control register of U700. The derived output current (M x ITREF) is connected to another programmable current-mirror circuit, U910B, external to the hybrid. The output of U910B provides the actual charging current and is a control-data-selected multiple of the M x ITREF current.

At the time of calibration, the processor will vary the ITREF input current until the slope of the output ramp for specific current-mirror/timing capacitor combinations is precisely set. The values of A TIM REF at these settings allow the processor to precisely calculate the characteristics of the current-mirror circuits at their various multiplication factors and the charging characteristics of the timing capacitors. These values are stored as calibration constants in nonvolatile memory (RAM U2460, diagram 1).

Once the calibration constants are set, any setting of the SEC/DIV switch causes the Microprocessor to recall the associated calibration constants from RAM. The processor then calculates the proper value of A TIM REF based on the selected timing capacitor and the current-mirror multiplication factors.

If the SEC/DIV VAR control is out of the calibrated detent position, the processor will decrease the A TIM REF voltage from the maximum, in-detent value by an amount proportional to the position setting of the VAR control. At the maximum, fully counterclockwise setting of the VAR control, the ITREF current is one-third that of the normal, in-detent current.

For A Sweep hybrid U700 to initiate a sweep at the selected rate, the $\overline{\text{AUXTRIG}}$ (auxiliary trigger) input (pin 3), the THO (trigger holdoff) line from the Display Sequencer (on pin 1), and the $\overline{\text{TRIG}}$ (trigger) line from the trigger hybrid (on pin 2) must all be LO. With these three inputs LO, the A SWEEP ramp begins, and the sweep gate ($\overline{\text{SG}}$) output (pin 45) goes LO. The buffered sweep gate signal ($\overline{\text{SGA}}$) at the output of U975 returns to the Display Sequencer through R981 to indicate that the A Sweep is active. The sweep gate signal is used by various other circuits for their timing activities and is held LO until the A SWEEP ramp ends. The buffered (negative) sweep gate is inverted and routed to the rear-panel A GATE output connector via U975.

Diodes CR752 and CR753 and associated components form a charging network that permits delaying the timing of the end-of-A-Sweep gate signal ($\overline{\text{SGAZ}}$) for B Sweep displays. For normal A Sweep operation with the $\overline{\text{SGBZ}}$ signal HI, the SGAZ signal will end quickly, since the capacitance associated with Z-Axis hybrid U950 input (diagram 6) will be charged positively through both R753 and R754. For B Sweep operation ($\overline{\text{SGBZ}}$ is LO), the end of the SGAZ gate signal will be delayed slightly (with respect to the normal sweep gate) since charging of the Z-Axis input capacitance will be at a slower rate through R754 only. This allows more of the B Sweep to be displayed than would otherwise be possible.

The A Sweep Delay Gate (DG) signal acts as the trigger holdoff (THO) signal for the B Sweep and the B Trigger circuitry. It is generated by comparing the A SWEEP ramp voltage to the selected delay reference (DR) level from analog switch U850C. As the ramp voltage crosses the delay reference level, the delay gate (DG) output signal goes LO, removing the HI THO level to the B

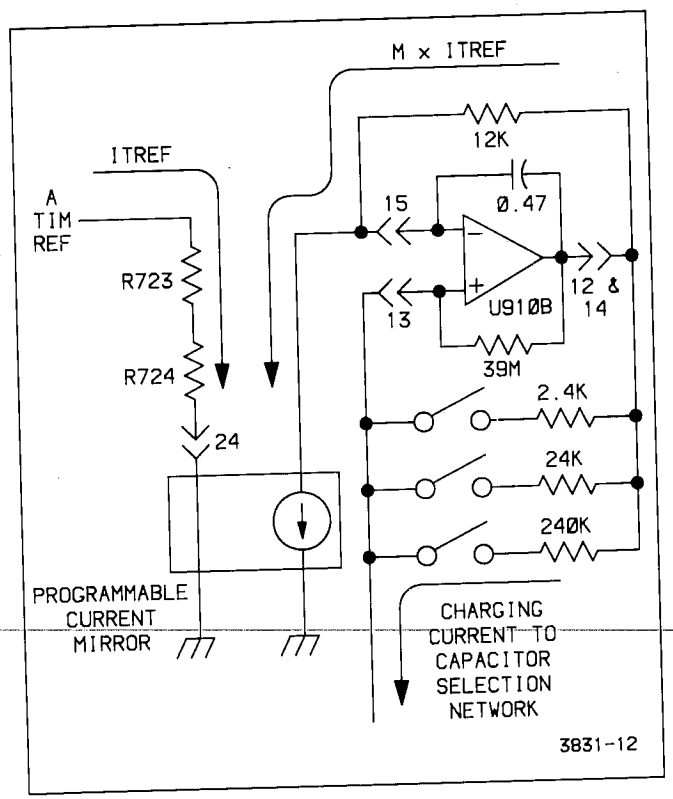


Figure 3-5. Sweep generator.

Sweep. This enables the B Sweep to run immediately in RUN AFT DLY B Trigger Mode or, when in TRIG AFT DLY B Trigger Mode, enables the B Sweep to run when a B triggering event occurs.

The BDCA (A Sweep bypass-delay comparator) input (U700 pin 39) is a data bit from Auxiliary Control Register U140 (diagram 4) that, when HI, sets the A Sweep DG output LO at the beginning of the A Sweep. This enables the B Sweep to run immediately at the start of the A Sweep and is used for calibration purposes and for options.

The capacitive load (part of the etched-circuit board) at the RDA (retrace delay adjust) input (U700 pin 4) is used to delay the retrace of the sweep until the Z-Axis drive is fully turned off in response to the SGAZ gate going HI. This delay prevents any part of the retrace from being seen.

B Sweep

Operation of B Sweep hybrid U900 is similar to that just described for the A Sweep with the following exceptions: the THO input (and thus sweep enabling) is controlled by the A Sweep hybrid or the Measurement PAL, U975, and not the Display Sequencer (see the preceding A Sweep description). The timing capacitor select output, TCS, is not used, and only three timing capacitors are selectable (two on the B Sweep hybrid at CT0 and CT1 and one externally at CT2).

Calibrator

The Calibrator circuit, composed of Q550, U165B, U550A, B, C, and D, and associated components, generates a square wave output of precise amplitude and frequency characteristics. The CALIBRATOR signal provided at the front-panel output connector is useful for adjusting probe compensation and verifying VOLTS/DIV, SEC/DIV, and Δt (delta time) calibration. Output frequency is controlled by the Display Sequencer and is set to display five cycles across the ten crt graticule divisions at sweep speed settings from 100 ns per division to 100 ms per division. This feature allows quick and easy verification of the sweep rates. The Calibrator circuitry is essentially a voltage regulator that is alternately switched on and off, producing the square-wave output signal.

When the timing signal (CT) from the Display Sequencer to the base of U550D is LO, U550C (configured as a diode) is forward biased, shunting bias current away from Q550, keeping it turned off. When transistor Q550 is off, the front-panel CAL OUT connector is pulled to ground potential through R558, setting the lower limit of the CALIBRATOR output signal.

As the CAL signal goes from LO to HI, the emitter of U550D is pulled HI to reverse bias U550C. Bias current for Q550 is established, and the transistor is turned on. The voltage at the emitter of Q550 rises to a level of +2.4 volts, determined by the voltage regulator composed of U165B, U550A, U550B, and associated components. This regulated level is applied to the front-panel CALIBRATOR connector through a voltage-divider network composed of R557 and R558. This produces an output voltage of 400 mV with an effective output impedance of 50 Ω.

Since the frequency of the CALIBRATOR signal is controlled by the same divider chain that controls operation of the vertical chopping rate, the intentional 200-ns shift added to the chop signal at the end of some sweeps (to desynchronize the chopping rate from the sweep rate) shows up on the CALIBRATOR signal as an irregular-width pulse. This shift is not apparent when viewing the CALIBRATOR signal on the instrument providing the signal (since the skew occurs during sweep-retrace time), but it should be taken into account when using the CALIBRATOR signal with other instrumentation. The skew can be eliminated from the signal by setting the instrument TRIGGER MODE to SGL SEQ (to shut off the sweeps).

PARAMETRIC MEASUREMENTS

The VOLTS Parametric Measurement is made using the same methods and circuitry that is used in the Auto Level trigger mode to find the peak voltages. The accuracy of the VOLTS measurement is based on the accuracy of the trigger level and the DC balance of the instrument.

All of the time-based Parametric Measurements use the A and B Sweep gates and delay gates as the basis for the measurements.The measurement PAL, U975, controls the signal flow while in the Parametric mode. The measurement flip-flop, U980B, reports the state of a variety of conditions to the SLIC through the $\overline{\text{SGB}}$ line. The SLIC data is read by the processor system and used to compute the desired measurement.

VERTICAL CHANNEL SWITCH AND OUTPUT AMPLIFIERS

The Vertical Channel Switch (diagram 6) selects the signal source for vertical deflection of the crt beam. The Vertical, Horizontal, and Z-Axis output amplifiers provide the signal amplification necessary to drive the crt.

Vertical Channel Switch

The Vertical Channel Switch consists of hybrid Channel Switch U400, that selects one of the vertical signals for application to the Vertical Output Amplifier, and a combined switch/amplifier circuit that converts the single-ended readout vertical signal into a differential signal for application to the Channel Switch.

Table 3-3
Vertical Display Selection

Select Inputs				Vertical Display
$\overline{VS1}$	$\overline{VS2}$	$\overline{VS3}$	$\overline{VS4}$	
L	H	H	H	CH 1
H	L	H	H	CH 2
L	L	H	H	ADD
H	H	L	L	CH 3
H	H	H	L	CH 4
H	H	H	H	Readout (Y)

Channel selection is controlled by the Display Sequencer $\overline{VS1}$ through $\overline{VS4}$ signals applied to the vertical channel selection pins (pin 24, pin 25, pin 13, and pin 14 respectively). (See Table 3-3 for the Vertical Display Selection.) When a vertical select line is LO, the associated input signal pins are connected to the differential output (+OUT, pin 11 and −OUT, pin 3). The CH 5 input signal (Readout Vertical) is added to the output whenever both the $\overline{VS3}$ and $\overline{VS4}$ select signals are HI but will only contain readout information when the readout select logic (U975A and U975C) detects that the Display Sequencer has set both the Horizontal Select signals ($\overline{HSA}$ and $\overline{HSB}$) HI (readout selected).

READOUT SWITCH/AMPLIFIER. Transistors U485A, U485B, U485C, U485D, and U475C, along with their associated components, make up an analog switch circuit that routes either the readout vertical signal at the base of U485A or the ground reference at the base of U485C to the output at the emitter of U475C. The signal selected depends on the complementary voltages applied to the emitter junctions of the two emitter-coupled transistor pairs, U485A and B and U485C and D. The selection voltages are developed by voltage-divider networks on the complementary logic outputs of U975A and U975C.

When readout information is to be displayed, the horizontal select inputs to U980B and U980C go HI and the output of NAND-gate U975C goes LO. The LO applied to the divider network of R498, R484, and R471 pulls the anode of CR484 low enough to reverse bias it. This forward biases the emitter-coupled pair U485A and B via R483. NAND-gate U975A inverts the LO and applies a HI to the junction of R497 and R485. The HI forward biases CR485, and the emitters of U485C and D are pulled to a level in excess of +2 V, reverse biasing the transistor pair. With U485C and D reverse biased, the ground reference level at the base of U485C is isolated from the output, while the readout vertical information is allowed to pass through the forward-biased transistor pair.

When readout information is not being displayed, a HI is present at the output of NAND-gate U975C. The HI forward biases CR484 and, when inverted by U975A, reverse biases CR485. With the biasing conditions reversed, the transistor pair of U485C and D becomes forward biased and U485A and B becomes reversed biased. The ground reference level present at the base of U485C is coupled to the output, while the readout vertical signal is isolated.

The output signal (either the readout vertical signal or the ground reference level) is applied to the CH5+ input of Channel Switch U400 via R495 and R412. The inverting amplifier circuit composed of U475A, U475B, U475D, and associated components inverts the readout vertical signal for application to the CH5− input. The amplifier is an inverting unity-gain configuration with transistors U475A and U475B connected as an emitter-coupled pair. The base of U475A is referenced to ground through R482. The base of U475B is pulled to the same level by the negative feedback from emitter-follower U475D through R478. The noninverted signal is applied to the base of U475B through R492 and will attempt to increase or decrease the current to the base of U475B, depending on the amplitude and polarity of the signal. However, the negative feedback from the collector of U475B (via U475D and R478) will hold the base of U475B at the ground reference level. The feedback current through R478 develops a voltage drop across R478 that is equal in amplitude but opposite in polarity to the noninverted vertical readout signal. The inverted readout signal is applied to the Channel Switch on pin 2 (CH5−) via R476 and R402.

The HF ADJ (high-frequency adjust) potentiometer R417 (2455B only) and resistor R416 (connected to pin 16) adjust the high-frequency response of the Channel Switch hybrid.

Vertical Output Amplifier

Vertical Output Amplifier U600 is a hybrid device that provides the final amplification of the selected vertical signal, raising it to the level required to drive the crt deflection plates. Vertical deflection signals from the Vertical Channel Switch are delayed approximately 78 ns by Delay Line DL100. This delay allows the Sweep and Z-Axis circuits to turn on before the triggering event begins vertical deflection of the crt beam, thereby permitting the operator to view the triggering event. The bridged-T network, composed of inductors and capacitors built into the circuit board, corrects phase-distortion introduced by the delay line. The RLC networks connected between the output pins of U400 are adjusted during calibration to obtain the correct overall high-frequency response of the vertical deflection system. The vertical signal from the Delay Line is applied to pins 10 and 3 of U600. The RL network connected between pins 8 and 5 (COMPA and COMPB) of U600 compensates the signal for the skin-effect losses associated with the delay line.

Amplifier gain and vertical centering are adjusted by R638 and R639 respectively, primarily to match the

amplifier hybrid to the crt installed in the instrument. The Dynamic Centering circuit sinks an intensity-dependent correction current away from the vertical centering input at pin 39. The correction signal holds the vertical centering stable over a wide range of varying display intensities. Readout jitter adjustment pot R618 is used to minimize thermal distortion in the output amplifier to reduce jitter in the display readout.

The vertical output signal at pins 28 and 33 of U600 (OUT A and OUT B) is applied to the vertical deflection plates of the crt (diagram 8) via L628 and L633. The deflection plates form a distributed-deflection structure that is terminated by a hybrid resistor network. One element of the terminating network is an adjustment potentiometer used to match the network impedance to that of the crt.

BANDWIDTH LIMITING. Bandwidth limiting coils L644 and L619, along with capacitors built into U600, form a three-pole filter used to roll off high-frequency response of the Vertical Output amplifier above 20 MHz. To limit the vertical bandwidth, the $\overline{\text{BWL}}$ (bandwidth limit) input to U600 (pin 16) is pulled LO. It may be set LO either by the BWL control data bit from Auxiliary Control Register U140 (diagram 4) when the operator selects the Bandwidth Limit feature or automatically by the output of NAND-gate U975A in the Vertical Channel Switch circuitry (via CR616) when the readout is being displayed.

TRACE SEPARATION. The voltage applied to the TS (trace separation) input of U600 (pin 42) is used to offset the output levels to vertically shift the position of the trace on the crt. During normal sweep displays, TS1 + TS2 signal applied to the base of Q600 by the Display Sequencer (diagram 5) is HI, and the transistor is turned on. The TRACE SEP level at the junction of R642 and CR600 is shunted to ground, and no offsetting at the output signal will occur. For those displays in which trace separation should occur, the Display Sequencer switches the base of Q600 to ground level to turn off the transistor. The trace separation level set by front-panel TRACE SEP control R3190 (via MUX U2530 and sample-and-hold circuit U2630C and C2631) is applied to the TS input of U600, and a corresponding offset of the displayed trace will occur.

BEAM FIND. As an aid in locating off-screen or overscanned displays, the instrument is provided with a beam-finding feature. When the front-panel BEAM FIND button is pushed, the beam-find input pin (BF, pin 15) of U600 will be pulled HI. While BF is HI, the dynamic range of Vertical Output Amplifier U600 is reduced, and all deflected traces will be held to within the vertical limits of the crt graticule.

Also, the activation of the BEAM FIND switch is detected by the microprocessor during its normal Front-Panel Switch Scanning. When detected, this generates a User Request SRQ if option 10 is installed.

OUTPUT PROTECTION CIRCUIT. A current-limit circuit composed of transistors Q623 and Q624 protects the Vertical Output Amplifier from a short-circuited output or a bias-loss condition. Either of these fault conditions will cause excessive current to flow into pins 30 and 31 of U600. Current in FET Q624 is limited to the IDSS current, so the voltage at pins 24, 30 and 31 will drop. This decreases the forward bias on pass-transistor Q623 and lowers the voltage at pin 23 of U600 enough to provide some degree of protection for the device.

Horizontal Amplifier

The Horizontal Amplifier circuitry consists of a Horizontal Output Amplifier U800, a unity-gain buffer amplifier made up of the five transistors in U735, and associated components.

UNITY-GAIN BUFFER AMPLIFIER. The amplifier circuit composed of U735A, B, C, D, and E along with their associated components, form a unity-gain amplifier that buffers the ramp signal from A Sweep Generator U700 to the Horizontal Output Amplifier. Transistors U735C and D form a differential pair with the negative excursion of their emitters limited to −5 V (clamped by U735E). Negative feedback from the collector of U735C to its base is via emitter-followers U735A and B (in parallel) which drive the A Sweep input (pin 18, A+) of Horizontal Output Amplifier U800.

HORIZONTAL OUTPUT AMPLIFIER. Integrated circuit U800 provides the final amplification of the selected horizontal-deflection signal required to drive the crt. One of the single-ended input signals applied to the four input pins is converted to a differential-output signal at the output pins of the amplifier. The four deflection signals to U800 are: the A sweep (pin 18, A+), the B Sweep (pin 16, B+), the Readout Horizontal signal (pin 17, RO) and the Channel 1 signal (used for horizontal deflection of the X-Y displays) at pin 20, the X+ input pin. Signal selection is done by an internal channel switch and is controlled by the $\overline{\text{HSA}}$ (horizontal select A) and $\overline{\text{HSB}}$ (horizontal select B) signals from the Display Sequencer (see Table 3-4).

Switching between unmagnified (X1) gain and magnified (X10 gain) is also controlled by signals from the Display Sequencer. For normal horizontal deflection, the $\overline{\text{MAG}}$ signal on pin 14 of U800 is HI, and the gain of the output amplifier produces normal sweep deflection. Precise X1 deflection gain is set by adjusting X1 Gain pot R860.

Table 3-4
Horizontal Display Selection

Control Level		Selected Signal
$\overline{HSA}$	$\overline{HSB}$	
H	H	Readout (X)
H	L	B Sweep Ramp
L	H	A Sweep Ramp
L	L	X Input (from CH 1)

When the X10 MAG feature is selected, amplifier gain for the magnified sweeps is increased by a factor of 10. The $\overline{MAG}$ signal from the Display Sequencer goes LO when magnified sweep is to be displayed. This switches the amplifier gain and switches analog switch U860C from the X1 position to the X10 position. Amplifier gain in the magnified mode is adjusted by adding or subtracting a small bias current using X10 Gain control R850. Dc offsets in the amplifier and crt are compensated for, using Horiz Centering pot R801 to precisely center the display. An intensity-dependent position correction signal, used to hold the horizontal centering stable over a wide range of varying display intensities, is also added at this point by the Dynamic Centering circuitry.

Timing and linearity of the sweep is affected by the amplifier transient response; and Trans Resp pot R802, connected to pin 2, is adjusted during calibration for optimum accuracy of the high-speed sweeps.

As with the Vertical Output Amplifier, the Beam Find feature reduces the dynamic range of the Horizontal Output Amplifier. While the front-panel BEAM FIND button is pressed in, a HI is placed on U800 pin 15 via pull-up resistor R615, and the horizontal deflection is reduced, moving horizontally off-screen displays to within the graticule viewing area.

Z-Axis Amplifier

Z-Axis Amplifier U950 turns the crt beam off and on at the desired intensity levels as the oscilloscope goes through its display sequence. The BRIGHT (brightness) signal applied to U950 pin 44 from the Display Sequencer U650 (diagram 5) is amplified to the level required to drive the crt control grid (via the DC Restorer circuitry) and sets the crt beam intensity. The BLANK input signal applied to U950 pin 5, also from the Display Sequencer, blanks the trace during sweep retrace, chop switching, and readout blanking by reducing the VZ OUT signal to a blanked level. Sweep gate z-axis signals ($\overline{SGAZ}$ and $\overline{SGBZ}$) from the A Sweep and B Sweep hybrids (U700 and U900) respectively, (diagram 5) are applied to the Z-Axis Amplifier on pins 4 and 3. These signals turn the beam current on and off for the related displays and, when used in conjunction with the BLANK signal on pin 5, enable the sweeps to be blanked while still allowing the Readout circuitry to blank and unblank the crt for the readout displays.

Control signals applied to U950 pin 48, pin 2, and pin 1 ($\overline{HSA}$, $\overline{HSB}$, and TXY respectively) switch some internal logic circuitry to enable or disable different input signals for the various types of displays. Table 3-5 illustrates the effects of the various input signals on the output signal for different combinations of $\overline{HSA}$, $\overline{HSB}$, and $\overline{TXY}$.

The Z-Axis hybrid has an internal limiter circuit that prevents the crt from being damaged during high-intensity, high-repetition-rate displays. A signal representative of the intensity setting and the sweep repetition rate is integrated on C957 and results in a control level at pin 7 of U950 used to limit intensity of the crt beam. Maximum Grid drive is controlled by R949 on U950 pin 9.

Focus tracking for intensity (VZ OUT) level changes is provided by the VQ OUT (quadrapole output voltage) signal at pin 22 of U950. The VQ OUT signal varies the focusing voltages (and thus the focusing strength) of two quadrapole lenses in the crt (diagram 8). The VQ OUT signal is related to the VQ OUT level exponentially and provides the greatest auto-focus control at high intensity levels. Gain of the VQ OUT signal is set by the High-Drive Focus adjustment, R1842. The VQ OUT signal also drives the Dynamic Centering circuit and holds the display position stable during wide-range intensity level changes.

Dynamic Centering

The circuit composed of U3401, U3402, and associated components generates compensating signals to offset positioning effects that occur in the crt when the intensity is varied over a wide range. The VQ OUT signal from Z-Axis Amplifier U950 is exponentially proportional to the display intensity and dynamically controls the intensity-dependent offsets.

Dynamic Centering adjustment pots R3401 and R3407 set the gain and polarity of the signals at their related outputs by varying the current in the emitter circuit of one of two emitter-coupled pairs of transistors. Adjusting the bias level, at either pin 4, above $\simeq -10.6$ volts (determined by R3410 and R3411 at the complementary inputs, pins 1) will generate an inverted signal, while adjusting the

Table 3-5
Blanking and Intensity Control Selection

Control Inputs			Intensity Affected By	Blanking Affected By	Typical Display
TXY	$\overline{HSA}$	$\overline{HSB}$			
X[a]	H	H	BRIGHT (RO level)	BLANK	Readout
X	H	L	BRIGHT, Z EXT	BLANK, $\overline{SGAZ}$, $\overline{SGBZ}$	Delayed Sweep
X	L	H	BRIGHT, $\overline{SGBZ}$,Z EXT	BLANK, $\overline{SGAZ}$	Main Sweep
L	L	L	BRIGHT, $\overline{SGBZ}$, Z EXT	BLANK	X-Y
H	L	L	BRIGHT, $\overline{SGBZ}$, Z EXT	BLANK, $\overline{SGAZ}$	X-Y

[a]X=State doesn't matter.

bias levels below −10.6 volts will cause a noninverted signal. Amplitude of the resulting signal is dependent on how far from the −10.6-volt reference the bias is set. The output signal is added or subtracted from the position voltage applied to the Vertical and Horizontal Output Amplifiers. Both pots are adjusted so that position shifts due to display intensity variations are minimized.

READOUT

The Readout circuitry (diagram 7) is responsible for displaying the alphanumeric readout characters in the crt. An eight-bit character code specifying each character (or cursor segment) to be displayed is written from the Microprocessor to a corresponding location in the Character RAM U2920 (a 2K-x-8-bit, random access memory integrated circuit). Each of the following 128 locations in the RAM, address locations 0 through 63 for the first and fourth readout lines and 128 through 191 for the second and third readout lines, corresponds to one of the 128 possible character locations in the crt readout display (see Figure 3-6). The next 128 RAM locations, address locations 64 through 127 for the first and fourth readout lines and 192 through 255 for the second and third readout lines, are used to store cursor segment information for the display of the ΔV and Δt measurement cursors. The eight-bit character code written to each location in RAM points to a block of addresses in Character ROM U2930. This block in the ROM contains the dot-position information for the specific character to be displayed at the associated crt position.

Each character is made up of zero (for a space character) or more dots displayed in an eight-wide by sixteen-high dot matrix. Specific blocks of ROM addresses contain all the X-Y offset coordinates for the dots in a particular character in the readout. The coordinates are referenced to the lower-left corner of the character dot matrix. Each individual data byte in the block of ROM addresses contains both the X and the Y coordinates for one dot of the associated character.

To display a character, a combination of the character position on the crt (the RAM address) and the byte of X-Y position data from Character ROM U2930 (relative to that character position) is applied to Horizontal and Vertical DAC (digital-to-analog converters) circuits, U2910 and U2905 respectively. In these circuits, the X-Y position data is converted to analog deflection signals used to position each dot in the crt readout display. Each of the position bytes are read from the block of ROM defining the character under control of the readout timing and sequencing circuitry. The resulting dots, when displayed in sequence, form the character at the proper location on the crt.

Readout I/O

The Readout I/O circuitry, composed of U2860, U2865, U2960, and associated components, provides the interface between the Microprocessor and the Readout board. Two types of data, Readout mode data and character data, are written to the Readout board serially via data bus line BD0.

STORING A CHARACTER. Displaying a character starts with serially clocking 16 character data bits into a 16-bit shift register formed by registers U2960 and U2860. The $\overline{ROST}$ strobe (readout strobe one) from the Address Decode circuitry (diagram 1) is the clocking signal. The first eight bits of the loaded data indicate the character to be displayed, while the last eight select the location on the crt that the character is to be displayed.

On positive-going transitions of the $\overline{ROST}$ strobe, the data bit present on the BD0 data line is shifted into the first latch of character address register U2960. The following negative-going edges of the $\overline{ROST}$ strobe are inverted

by U2965A to produce a positive transition that shifts the data bit present at U2960 pin 9 (Q_{SH}) into U2860. After 15 $\overline{ROS1}$ strobes have occurred, seven bits of character data are latched into U2860, and the eighth character bit and seven of the character address bits are latched into character address register U2960 (though they have not been shifted into their correct positions for addressing the RAM).

At this point, the last character bit remains to be shifted into the registers, but the operating mode must be set up first to ensure correct operation upon shifting in the final bit. The eight bits of mode data are shifted into the mode control register U2865 by the $\overline{ROS2}$ strobe. Bit Q_4 ($\overline{WRITE}$), along with the $\overline{ROS2}$ and the R/ $\overline{W}$ DLYD signal are applied to the RAM enabling circuitry and determine when new character information will be written into the Character RAM. With U2865 loaded with the mode data, a final $\overline{ROS1}$ strobe clocks the eighth bit of character data from U2960 to U2860 on the negative edge, and the positive edge of the strobe clocks the eighth character address bit into U2960.

With control bit Q_4 from U2865 LO, the outputs of U2860 are enabled and the eight bits of character data (CD0 through CD7) are written in parallel into the Character RAM at the location selected by the eight-bit address from U2960. Register U2960 is enabled only when the Readout is not displaying characters (the REST signal at pin 15 of U2960 is HI).

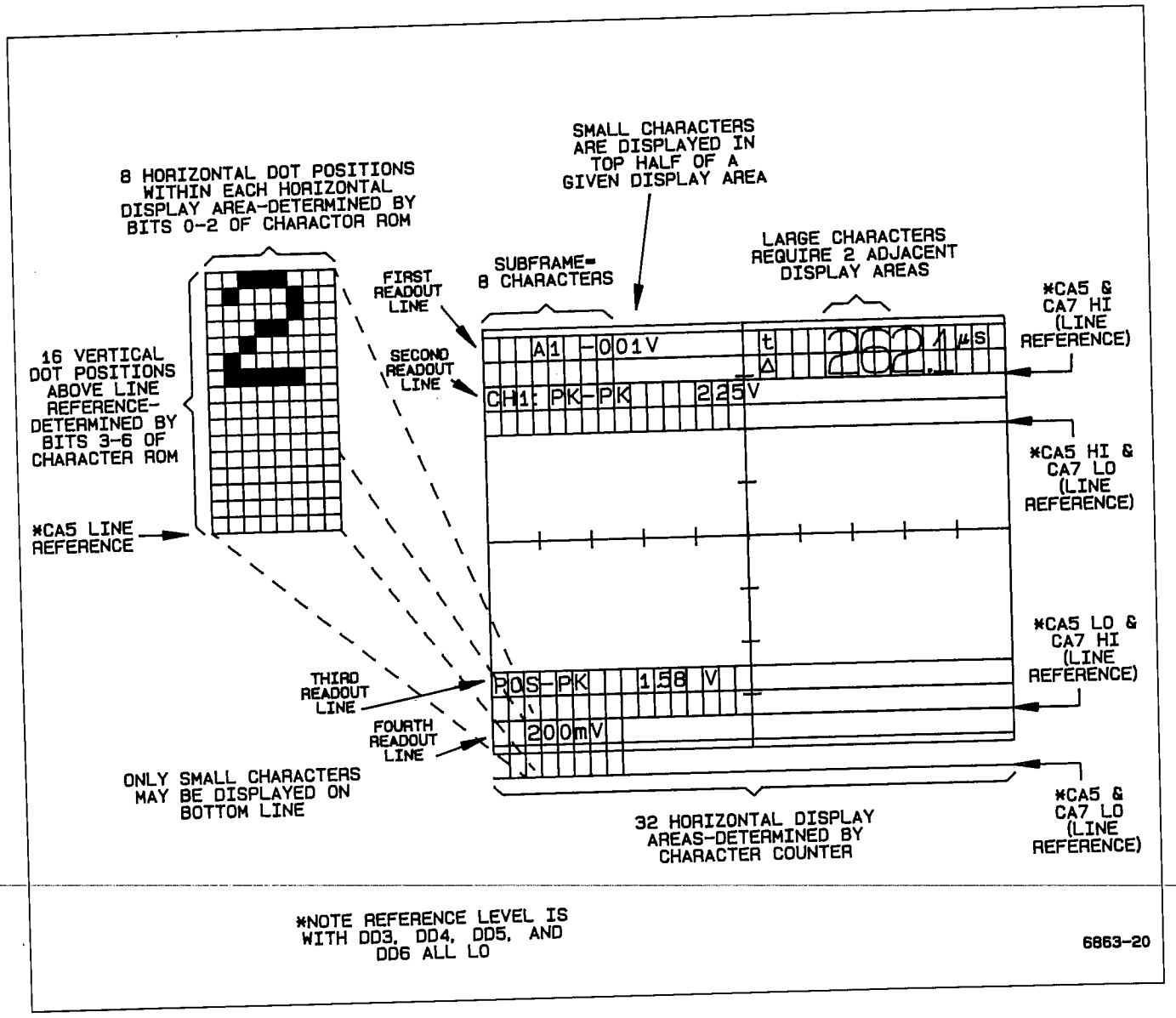


Figure 3-6. Developing the readout display.

The character data register U2860 also provides a means for the Microprocessor to read data from the Character RAM for partial verification of Readout circuit operation (during the power-up tests). The eight bits of parallel data from the Character RAM location selected by character address register U2960 are loaded into U2860 by setting bit Q_3 of mode control register U2865 LO. Inverter U2965C converts the LO to a HI and applies it to character-register U2860 at pin 1. The HI on pin 1, in combination with the fixed HI on pin 19 of U2860, switches the character register to the Parallel Load mode. The next positive transition of the $\overline{ROST}$ strobe loads the eight data bits placed on the CD0 through CD7 bus lines into the register in parallel. Bit Q_3 is then returned HI, and the next positive transition of the $\overline{ROST}$ strobe shifts the Q_A bit to pin 8 (Q_A'), the RO DO (readout data out) line. Seven more $\overline{ROST}$ strobes shift the remaining seven bits of character data out onto the RO DO line to Status Buffer U2220 (diagram 2) to be read, one at a time, by the processor.

Character RAM

Character RAM U2920 provides temporary storage of the readout character selection data. This character data is organized as 256 eight-bit words that define the character that should be displayed at any given readout position on the crt. Cursor information is also stored in U2920 when cursors are to be displayed.

RAM locations may be addressed either from the Readout I/O stage by character address register U2960, as previously described, or by the Character Counter stage. Each of the following 128 address locations corresponds to a specific readout location on the crt. Address locations 0 through 63 correspond to the first and fourth readout lines and 128 through 191 to the second and third readout lines. The next 128 address locations store cursor information. Address locations 64 through 127 correspond to the first and fourth readout line storage and 192 through 255 to the second and third readout line storage. The eight bits of data written to one of these locations from the Readout I/O stage is a code that identifies the specific character (or cursor segment) that should be displayed at the associated crt location. After the display data is written into the RAM, the Character Counter is allowed to address the RAM, incrementing through the RAM address field. The eight-bit character codes for each display location are output to Character ROM U2930 in sequence.

Character Counter

The Character Counter stage consists of two four-bit counters (both within U2940) cascaded together to form an eight-bit counter and tristate buffer U2935 which drives the RAM address lines.

As the Character Counter addresses each RAM location (the counter also determines the character screen location), a sequence of "dot display cycles" is performed in which the individual dots that make up the character are positioned on the crt and turned on. The $\overline{EOCH}$ (end of character) signal applied to U2855A prevents the counter from incrementing until all dots of the character have been displayed. As the last dot of a character is addressed, the $\overline{EOCH}$ bit at pin 2 of U2855A goes LO. The next $\overline{GETDOT}$ pulse increments U2940 (via U2855A), and the next RAM location is addressed to start the display of the next character. Space characters have the $\overline{EOCH}$ bit set LO for the first "dot" of the character and merely advance the Counter to the next character address without displaying any dots. See the Character ROM description for further explanation of the $\overline{EOCH}$ bit.

Character ROM

Character ROM U2930 contains the horizontal and vertical dot-position information for all of the possible characters (or cursor segments) that may be displayed. The eight bits of character data from the Character RAM are applied to the eight most-significant address inputs (A4 through A11) of the Character ROM and select a block of dot-positioning data unique to the character to be displayed. The Dot Counter increments the four least-significant address lines (A0 through A3), causing the ROM to output a sequence of eight-bit words, each defining a dot position for the selected character.

The three least-significant bits of a ROM dot-data word (DD0 through DD2) select one of eight horizontal positions for the dot within an eight-by-sixteen character matrix (see Figure 3-6). The next four bits (DD3 through DD6) define the vertical position of the dot within the matrix. These dot-data bits are applied to the Horizontal and Vertical Character DACs, where they are converted to the analog voltages used to position the dot on the crt.

The last dot-data bit DD7 is the $\overline{EOCH}$ (end of character) bit and, when LO, indicates that the last dot of the character is addressed. It is used to reset the Dot Counter (via U2855B) and enables the Character Counter to be incremented (via U2855A) after the last dot of a character has been displayed.

Two servicing jumpers, J401 and J402, have been provided to disable the Character ROM and force the DD7 bit ($\overline{EOCH}$) LO. In certain instances, these two conditions may be useful when troubleshooting the Readout circuitry. To prevent damage to the ROM output circuitry, J402 should only be installed after J401 is installed (to disable the ROM).

Dot Counter

The Dot Counter consists of two four-bit counters (both within U2870), OR-gate U2835A, inverter U2980D, and inverting input AND-gate U2855B. It sequences through a

block of addresses containing dot-position data for a selected character. The Dot Counter is incremented when a dot is finished (via Inverter U2980D) by the $\overline{\text{GETDOT}}$ signal from the Dot Cycle Generator.

The counter increments through the block of dot-position data until the last byte of the block is encountered (last dot). This last data byte has the $\overline{\text{EOCH}}$ (end of character) bit (DD7) set LO. The dot is positioned and displayed in the normal manner, but when the $\overline{\text{GETDOT}}$ signal occurs for the next dot display cycle, the $\overline{\text{EOCH}}$ bit is latched into U2905 and generates the $\overline{\text{EOCH1}}$ (end of character, delayed one dot) signal at U2905 pin 15. With $\overline{\text{EOCH}}$ and $\overline{\text{EOCH1}}$ both LO, the HI reset pulse produced at pin 4 of NOR-gate U2855B resets the counter and, except for space characters, the $\overline{\text{EOCH}}$ bit returns HI. As the reset is removed from the Dot Counter, it is reenabled for display of the next character. For space characters, the $\overline{\text{EOCH}}$ bit will be detected as a LO when the first dot is read from the Character ROM, and the Character Counter will advance to the next character on the next rising edge of $\overline{\text{GETDOT}}$.

Counter U2870 and OR-gate U2835A enable characters of more than 16 dots to be displayed. Since most of the readout characters are small, using 16 dots or less, efficient data storage is achieved by storing the dot-position data as 16 consecutive bytes. For displaying these smaller characters, the least significant four bits from U2870 are sufficient to address the 16 possible dot-position bytes.

When larger characters (up to 32 dots) are to be displayed, an additional bit of counter data must be used to address the ROM. This fifth bit comes from U2870 pin 3 and is ORed by U2835A with bit CD0 from the Character RAM. The block address for these larger characters always has bit CD0 set LO, so the counter bit from U2870 pin 3 is in control of the ROM address line at pin 4 of U2930. When displaying these larger characters, the dot count goes beyond 16 dots before the $\overline{\text{EOCH}}$ bit is set LO. On the seventeenth character, the fifth counter bit (pin 3 of U2870) will go HI to address the next 16-byte block of character data in ROM U2930. The lower four bits of the DOT Counter then sequence through this additional block in the normal manner until the $\overline{\text{EOCH}}$ bit is encountered, resetting the counter.

Horizontal DAC

The Horizontal DAC generates the voltages used to horizontally position dots of the readout display on the crt. Five data bits (CA0 through CA4) from the Character Counter stage position a character to the correct column in the display (32 possible columns across the crt), while three data bits from Character ROM U2930 (DD0 through DD2) horizontally position the dots within the eight-by-sixteen character matrix (see Figure 3-6).

The eight bits of position data are written to the permanently enabled DAC each time a new dot is requested by the Dot Cycle Generator. The $\overline{\text{GETDOT}}$ signal applied to pin 11 (Chip Select) enables the DAC to be written into, and the falling edge of the 5-MHz clock applied to pin 12 (Write) writes the data at the eight DAC input pins into an internal latch. The voltage at the DAC output pin changes to reflect the data present in the latch.

Vertical Character DAC

The function of Vertical Character DAC U2905 is similar to that of the Horizontal DAC just described. It is responsible for vertically positioning each character dot on the crt. The Vertical DAC circuit is made up of six, D-type flip-flops (contained within U2905) and an accompanying resistor weighting network. The outputs of the flip-flops source different amounts of current to a summing node through a resistor weighting network.

The six data bits are latched into U2905 on the rising edge of the $\overline{\text{GETDOT}}$ signal. Two bits of character address data (CA5 and CA7) from the Character Counter switches the vertical display position between the four readout display lines. When the display is to be in the bottom line, bit CA5 is set LO. With CA5 LO, zener diode VR2925 is biased off and a small current is sourced to the summing node via R2925. Vertical position above this reference is determined by dot data bits DD3 through DD6. When the top line is to be displayed, the CA5 bit is set HI, biasing VR2925 on. A larger current is now sourced into the summing node via R2925 and enough voltage is developed across R2926 to move the display to the top row of the crt. The CA7 bit is used to offset the top and bottom readout display lines to form the center two readout display lines. As before, the individual dots are then positioned above this reference level by dot data bits DD3 through DD6.

Mode Select Logic and Analog Channel Switch

The Mode Select Logic circuitry is composed of analog switches U2800 and U2805, buffers U2820A and B, gates U2810A, B, C, and D, U2900B and C, and part of U2905. It controls the readout display mode by selecting which deflection signals should drive the Horizontal and Vertical Deflection Amplifiers during a readout display. Five display modes are decoded by the Mode Select Logic: character display, vertical cursor 0, vertical cursor 1, horizontal cursor 0, and horizontal cursor 1.

For normal character displays, cursor select bit CA6 on U2800 pin 1 is LO. This LO signal passes through analog switch U2800 and is latched into U2905 when the $\overline{\text{GETDOT}}$ request from the Dot Cycle Generator goes HI. This latched LO selects the character display mode by

Table 3-6
Readout Display Mode Selection

Control Bits				Mode Selected	Horizontal Signal	Vertical Signal
CA6 (Cursor Select)	DD5	DD4	DD3			
L	X[a]	X	X	Character Display	Horiz DAC	Vert DAC
H	L	H	L	Vert Cursor 1	Horiz DAC	DLY REF 1
H	L	H	H	Horiz Cursor 1	DLY REF 1	Horiz DAC
H	H	L	L	Vert Cursor 0	Horiz DAC	CURSOR 0
H	H	L	H	Horiz Cursor 0	CURSOR 0	Horiz DAC
H	L	L	X	Return to character display Mode		

[a]X = State doesn't matter.

forcing the outputs of U2900B and C and U2810A and B HI. The HI outputs of U2900B and C applied to the select input pins of analog switch U2805 cause the Horizontal DAC output signal applied to U2805 pin 11 to be routed to the Horizontal Amplifier (diagram 6) via buffer U2820B. The same HI logic levels cause NOR-gates U2810C and D to produce a LO at their outputs. This causes analog switch U2800 to route the Vertical DAC output signal applied to pin 12 to the Vertical Output Amplifier (also diagram 6) via buffer U2820A.

For cursor displays, cursor select bit CA6 goes HI. This HI is routed through analog switch U2800 and latched into U2905 when $\overline{\text{GETDOT}}$ next goes HI. This produces a HI at U2905 pin 16, enabling the Mode Select Logic to decode output bits DD3, DD4, and DD5 (from U2905) to determine which of the four possible cursor modes is selected (see Table 3-6). Once one of the cursor modes is entered, analog switch U2800 routes a fixed HI from pin 5, pin 2, or pin 4 to U2905 to keep the Mode Select Logic enabled. Character display mode is reentered only when return-to-character-mode data is decoded (DD4 and DD5 both LO). When that occurs, U2800 routes the CA6 bit to U2905 and, if the bit is LO, the cursor display mode is halted.

CURSOR DEVELOPMENT. Cursors are displayed in short sections, alternating between both vertical positions (for the delta voltage cursors) or both horizontal positions (for the delta time cursors). When displaying delta voltage cursors, the CURSOR 0 level is routed to the Vertical Amplifier by analog switch U2800. This level determines the vertical position of one of the voltage cursors. Horizontal-positioning voltages for one segment of the cursor are routed from Horizontal DAC through analog switch U2805 and buffer U2820B to horizontally position each of the dots making up the cursor segment. DLY REF 1 is then used to vertically position the second cursor, and the Horizontal DAC positions each of the dots for that cursor segment. The cycle is repeated until all segments of both cursors are displayed.

Delta time cursor displays are similar in that the CURSOR 0 and DLY REF 1 signals are used to position the cursors. In this case, however, analog switch U2805 selects the CURSOR 0 and DLY REF 1 signals alternately to position the cursors horizontally, and the Horizontal DAC output is routed via analog switch U2800 and buffer U2820A to vertically position the dots within each cursor segment.

Refresh Prioritizer

The Refresh Prioritizer circuitry consists of U2850A and B, U2950A, U2990A, and U2985. It keeps track of how well the Readout circuitry is doing in displaying all the required readout information and maintains the overall refresh rate. Since the readout display must remain flicker-free and at a constant intensity over the entire sweep rate range, various modes of displaying readout information are provided. The Refresh Prioritizer keeps track of the display status and enables the various readout-display modes as required to produce minimal interference with the displayed waveform trace(s).

Ideally, readout information should be displayed only when the oscilloscope is not trying to display waveform traces. These times occur before a trace commences, after a trace is completed, or between consecutive traces. Displaying in this mode corresponds to "priority one" in

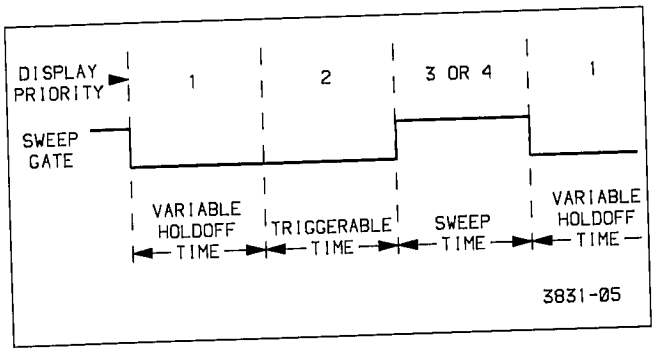


Figure 3-7. Readout display priorities.

Figure 3-7 and causes no interference with the displayed waveforms. If the Readout circuitry is able to display all the required readout dots during the holdoff time between sweeps, the prioritizer U2985 will turn off the Dot Start Governor until the next subframe of readout information is to be displayed. When the sweep times are either too fast to finish a readout display during holdoff (at 5 ns per division no identifiable holdoff time exists) or too slow to allow flicker-free readout, readout display modes other than priority one are initiated.

The next most desirable time for dots to be displayed is during "triggerable" time: that time between sweeps when the oscilloscope is waiting for a sweep trigger event to occur. This is designated priority two and may cause slight interference on the leading edge of the displayed trace if a dot is being displayed when the actual trigger occurs.

Finally, the least desirable dot display time is during a waveform trace display. This display time is designated either priority three or priority four. (Priority four indicates a higher demand of display time.) In priorities three and four, dot displays occur during the main portion of the waveform display. However, the waveform blanking associated with these displays is relatively random in nature and is usually not noticeable.

To start a readout display, the ROSFRAME (readout subframe) request from the Timing Logic (diagram 1) clocks the Q output of flip-flop U2850A HI. ROSFRAME is a periodic clocking signal used to hold the overall refresh rate constant and occurs at regular intervals, regardless of the state of the display.

As the Dot Cycle Generator runs, it resets half of U2830 in the Dot Timer at somewhat irregular intervals with the STARTDOT signal (via inverter U2890A). The Dot Timer then starts a timing sequence, and the rising edge of the REFRESH signal from U2830 pin 4 clocks the latched ROSFRAME request from U2850A pin 5 to the Q output (pin 9) of flip-flop U2850B. This HI, applied to the S1 input (pin 10) of prioritizer U2985, sets it up to increment with the next REFRESH clock applied to its clock input (pin 11). The LO $\overline{Q}$ output of U2850B (pin 8) applied to the reset input of U2850A resets the latched ROSFRAME request. See Figure 3-8 for an illustration of the timing sequence involved.

The next REFRESH clock increments the display priority to one by clocking a HI to the Q_D output (pin 12) of prioritizer shift register U2985. (Table 3-7 illustrates the operation of U2985.) The same clock latches the now LO ROSFRAME request at U2850B pin 12 to the Q output (pin 9), where it is applied to the S1 input (pin 10) of prioritizer U2985. The LO on the S1 input of the prioritizer will remain until another ROSFRAME request from the Timing Logic occurs, and the encoded priority at the output pins of U2985 will remain as it is presently set.

As each of the consecutive dots of the readout frame are displayed, the Dot and Character Counters increment until all dots of the subframe have been displayed (eight characters). As the Character Counter increments to address the next character of the display (first character of the next frame), the fourth bit of counter U2940 goes HI and sets the S0 input (pin 9) of prioritizer U2985 HI via exclusive-OR-gate U2990A. The Dot Timer then clocks the prioritizer with a REFRESH clock on pin 11 of U2985, and the priority is decremented back to zero (indicating that the subframe is completed). The next ROSFRAME request starts the process over again to display the next subframe of readout display. The sequence just described is the priority one display mode and is used when holdoff time between sweeps allows all dots of the subframe to be displayed before the next ROSFRAME request occurs.

If a second ROSFRAME request occurs before the Character Counter indicates the end of the subframe (to decrement the prioritizer back to zero), input S1 of U2985 will be set HI (while the S0 input pin remains LO) and the

Table 3-7
Operation of Prioritizer Shift Register

Select Inputs		Mode
S0	S1	
H	H	Parallel Load
H	L	L → Q_A (decrease priority)
L	H	H → Q_D (increase priority)
L	L	Hold Data

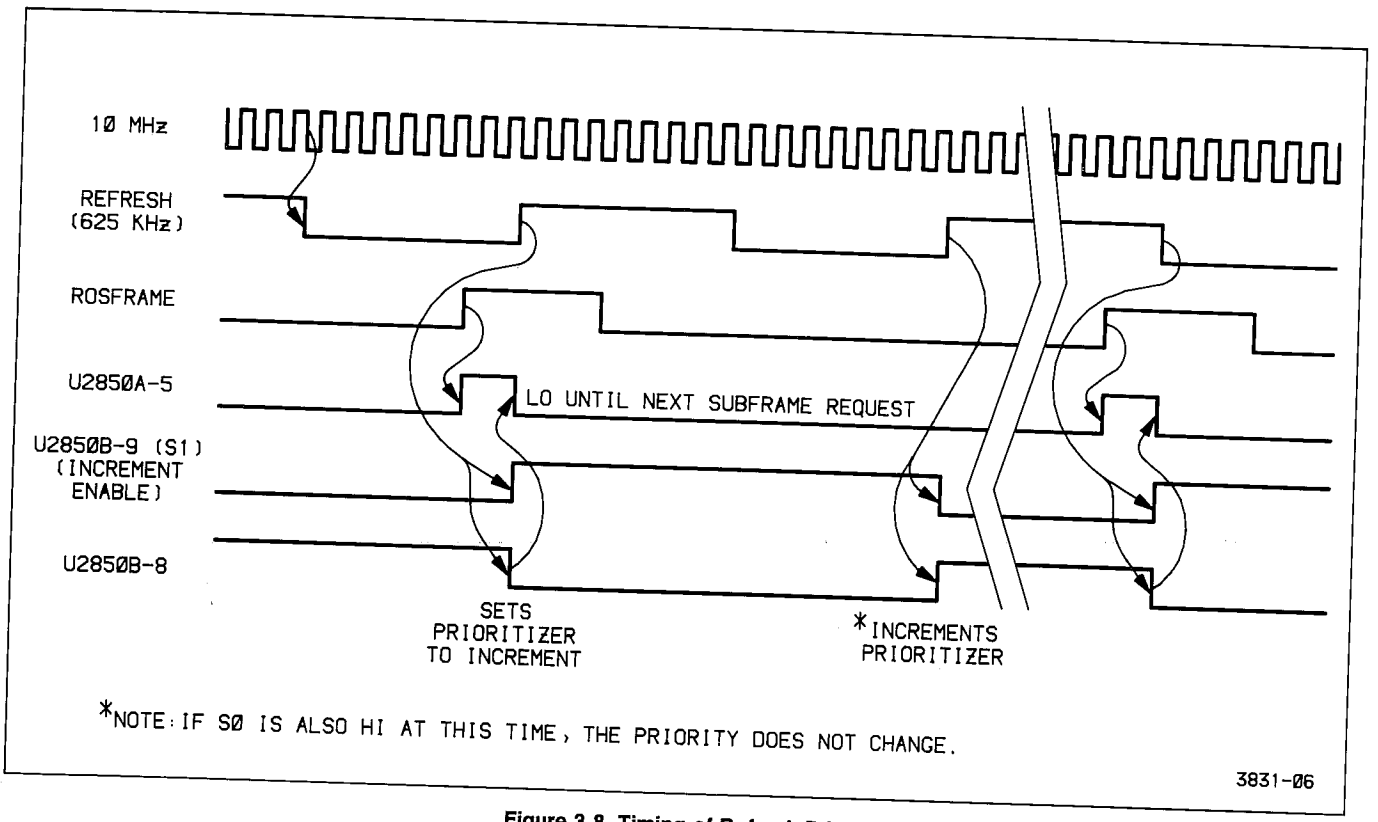


Figure 3-8. Timing of Refresh Prioritizer.

Prioritizer will increment to priority two (outputs Q_C and Q_D go HI) on the next STARTDOT cycle. If this display priority still is inadequate to complete the subframe display before the next ROSFRAME request occurs, priority two will be incremented up to priority three, or even to priority four should the condition persist. Priority four is operationally the same as priority three, but it is used to keep the readout circuitry continuously displaying readout data on through the next subframe, thus allowing the display to catch up. If priority four is in effect, the next decrement that occurs at the end of a subframe only returns the prioritizer to priority three, not to priority two.

The circuit composed of flip-flop U2950A and exclusive-OR-gate U2990A enables either edge of the CA3 bit to decrement the priority of the display when a subframe is completed. Either a negative or positive transition on pin 2 of U2990A will cause the output at pin 3 go HI since the Q output of U2950A is still at the opposite level. The HI from U2990A indicates that the end of the present subframe has occurred, and it sets up the prioritizer to decrement with the next REFRESH clock. At the same time that the prioritizer decrements, the changed level of the CA3 bit is clocked through U2950A and causes the output of exclusive-OR-gate U2990A to return LO until the next subframe is completed.

If the subframe is completed (S0 on U2985 goes HI) when a ROSFRAME request is also pending (S1 is also HI), U2985 does a parallel load, reloading the present priority back into the prioritizer. Since, in this case, the subframe display was completed at the same rate as the ROSFRAME request occurred, the readout display priority is not changed.

Dot Start Governor

The Dot Start Governor detects the display priority from the Refresh Prioritizer and initiates dot-display cycles as the appropriate conditions are met. The conditions tested include display priority, sweep gate completion, dot completion, readout control status, and the readout active enable from the Display Sequencer.

When the readout board status line (ACTIVE/$\overline{\text{ADDRESSABLE}}$) is HI (signifying display) and the REST line goes HI to indicate that the dot cycle is complete, AND-gate U2970C generates a HI at pin 8 (DOTOK) to signal that a new dot display is allowed. The HI from U2970C enables most of the gating in the Dot Start Governor. If the Refresh Prioritizer has encoded a display priority of either one or two, the output of exclusive-OR-gate U2990B is HI. When DOTOK from U2970C goes HI to enable a dot display, the LO reset from pin 6 of U2970B to pin 1 of flip-flop U2880A is removed. Now, when the A Sweep gate ($\overline{\text{SGA}}$) goes HI (beginning of Holdoff), the HI at the D input of U2880A is clocked to the Q output and the $\overline{Q}$ output at pin 6 will go LO, requesting display of a

priority one or two dot. This LO dot request is propagated through U2885B, U2890D, U2890B, and U2890C and sets the STARTDOT signal LO. STARTDOT going LO resets Dot Cycle Generator shift register U2995 and counter U2830B of the Dot Timer. Resetting the Dot Cycle Generator shift register causes the REST signal from U2995 pin 13 to go to a LO, removing the HI DOTOK signal at U2970C pin 8. As DOTOK goes LO, STARTDOT at pin 8 of U2890C goes HI to start the DOT Cycle Generator. At the same time the reset to U2880A is asserted via U2970B and the dot request is removed. Both the Dot Timer and the Dot Cycle Generator are now enabled and start the first dot-display cycle during holdoff time.

After the Display Sequencer U650 (diagram 5) has time to respond to the end of the sweep gate, it sets the readout active signal ($\overline{\text{ROA}}$) to pin 10 of U2880A LO. This sets pin 9 of U2855C LO, and the signal is propagated through U2885C, U2965C and D, and U2890B, as before, resetting the Dot Timer and the Dot Cycle Generator. REST then goes LO as before and starts the Dot Cycle Generator and Dot Timer. This cycle continues, displaying one dot per cycle (except for the first non-displayed dot of a character which is automatically initiated by $\overline{\text{EOCH2}}$, until the Display Sequencer determines that the readout time is over (sets $\overline{\text{ROA}}$ HI) or until the display priority is decremented to zero.

When a display priority of three or four exists, the output of U2990B will be LO, and U2970B, U2880A, and the associated logic gates following it will not be able to initiate a dot cycle. In either of these display priorities, U2970D, U2835C, U2980A, U2965B, and flip-flop U2950B detect the higher priority and generate a readout request signal ($\overline{\text{ROR}}$) to the Display Sequencer. The LO from U2950B pin 8 propagates through U2890B and U2890C to initiate a STARTDOT cycle. When the Display Sequencer recognizes that the readout request signal is LO, it will perform the mode-dependent setup functions necessary to give display control to the Readout Board and will then set the $\overline{\text{ROA}}$ (readout active) line LO. The LO will be clocked into U2880B, and the Dot Cycle Generator will generate a $\overline{\text{GETDOT}}$ signal, resetting the readout request from flip-flop U2950B. Only one dot is displayed for each readout request.

A similar readout display request will be generated when priority-two-or-higher displays are required when sweep gates are not present (dot display during triggerable time after holdoff). This condition is detected by NAND-gate U2885A. AND-gate U2970D allows a readout request to be generated when in the interfere mode. This mode is only invoked during a single-sequence waveform display and ensures that all of the selected sweep combinations are displayed once, followed by a complete readout frame (for the purpose of crt photography).

Dot Cycle Generator

The Dot Cycle Generator, composed of shift register U2995, flip-flop U2880B, and associated gating circuitry, generates time-related signals for the following purposes: unblanking the crt to display a dot; requesting the next byte of dot data in preparation for displaying the next dot; and reenabling itself to repeat the tasks, via the Dot Start Governor (dependent on the display priority).

The timing relationships of the Dot Cycle Generator output signals are controlled by shift register U2995. When the Dot Start Governor initiates a STARTDOT cycle as previously described, the STARTDOT signal initially goes LO, resetting all the Q outputs of U2995 LO and setting the Q output of flip-flop U2880B to a HI. The STARTDOT signal is then returned HI, and the Dot Timer counter U2830A and shift register U2995 are enabled. The shift register begins to consecutively shift HI logic levels to its Q output pins with each 5-MHz clock from the Dot Timer. After approximately 400 ns, pin 5 (Q_C) of the shift register will go HI. The HI at Q_C propagates through exclusive-OR-gate U2990D and AND-gate U2970A to unblank the crt by setting the readout blanking signal ($\overline{\text{ROB}}$) HI.

When the Q_F output of U2995 goes HI (1 μs after STARTDOT), the output of U2990D goes LO and the output of U2990C goes HI. The LO from U2990D propagates through U2970A to blank the crt ($\overline{\text{ROB}}$ goes LO) and to clock flip-flop U2880B via NAND-gate U2980C. The $\overline{\text{ROA}}$ (readout active) level from the Display Sequencer (diagram 5) is clocked from the D input (pin 12) of U2880B to the Q output; and, if LO (indicating that the readout circuitry had control of the crt when unblanking occurred; thus the dot was displayed), the output of U2980B is set HI. With three HI levels applied to NAND-gate U2885C, a $\overline{\text{GETDOT}}$ request is generated to get the next byte of dot-position data for display. The next 5-MHz clock sets the Q_G output of U2995 HI, and the output of U2990C goes LO, removing the LO $\overline{\text{GETDOT}}$ signal.

At 1.4 μs after STARTDOT goes HI, U2995 pin 13 (Q_H) goes HI to produce the REST signal, indicating that the current dot cycle is complete and the Dot Cycle Generator is at REST. If the readout ACTIVE/ $\overline{\text{ADDRESSABLE}}$ mode bit at U2970C pin 10 is still HI, the REST signal going HI produces a HI DOTOK signal (next dot is allowed) at pin 8. This HI applied to pin 10 of U2890C, along with any of the possible dot requests from the Dot Start Governor, will initiate another STARTDOT cycle for the next dot of the display. As long as the Display Sequencer holds the readout active line ($\overline{\text{ROA}}$) LO, U2885B, U2890D, and U2890B of the Dot Start Governor will automatically initiate dot cycles as soon as the previous one ends (REST goes HI), until the Refresh Prioritizer is decremented to zero.

When the last dot of the character is called from the Character ROM, the $\overline{\text{EOCH}}$ bit (DD7) applied to latch U2905 at pin 18 (in the Vertical Character DAC circuitry) is LO. At the end of that dot display cycle, the $\overline{\text{GETDOT}}$ signal (going HI) clocks the LO $\overline{\text{EOCH}}$ bit into latch U2905 and increments character counter U2940. The latched bit becomes the $\overline{\text{EOCH1}}$ signal (end of character, delayed one dot request) and is applied to U2885A, along with the already LO $\overline{\text{EOCH}}$ bit, to reset Dot Counter U2870. The least-significant bits to the Character ROM address pins (A0 through A4) are then zeros, and the first dot of the next character is addressed. The Horizontal and Vertical DACs don't write this first dot position data into their registers until the end of the next $\overline{\text{GETDOT}}$ signal. That same $\overline{\text{GETDOT}}$ signal also clocks $\overline{\text{EOCH1}}$ into U2905 which becomes $\overline{\text{EOCH2}}$ at pin 16 (end of character, delayed by two dot requests). $\overline{\text{EOCH2}}$ is applied to NAND-gate U2980C and disables the gate prior to the time the Dot Cycle Generator attempts to unblank the crt for the first dot display; thus the first dot of a character is never displayed.

Disabling the unblanking path for the first dot of each character in the manner just described allows the more radical voltage changes between characters to settle before the actual display of the next character begins. When the dot data for one of these undisplayed dots also has the $\overline{\text{EOCH}}$ bit set LO, it is a space character, and the display is advanced to the next character.

Dot Timer

The Dot Timer, composed of U2890A and U2830, generates three, time-related signals used to synchronize the display and maintain the proper sequencing of the individual character dots.

The two least-significant bits of the Dot Timer, from U2830 pins 11 and 10, are reset at the beginning of a dot cycle by a LO STARTDOT signal applied to the reset input of the counter via U2890A. As the dot-display cycle begins, the STARTDOT signal returns HI and the Dot Timer begins counting in a binary fashion. The 10-MHz clock applied to pin 13 is divided by two to produce the 5-MHz clocking signal at output pin 11. The 5-MHz clock sequences the Dot Cycle Generator through the various phases of the dot-display cycle. The REFRESH output signal from U2830 pin 4 updates the Refresh Prioritizer as each subframe is displayed.

A third clock, from U2830 pin 6, occurs at approximately 8-μs intervals and allows any pending dot requests to generate a $\overline{\text{ROR}}$ signal to the Display Sequencer via flip-flop U2950B. (Readout request generation is described in the Dot Start Governor discussion.)

HIGH VOLTAGE POWER SUPPLY AND CRT

The High-Voltage Supply and CRT circuit (diagram 8) provides the voltage levels and control circuitry for operation of the cathode-ray tube (crt). The circuitry consists of the High Voltage Oscillator, the High Voltage Regulator, the Cathode Supply, the Anode Multiplier, the DC Restorer, Focus Amplifiers, the CRT and the various CRT Control circuits.

High-Voltage Oscillator

The High-Voltage Oscillator transforms power obtained from the −15 volt unregulated supply to the various ac levels necessary for the operation of the crt circuitry. The circuit consists of transformer T1970, switching transistor Q1981, and associated circuitry. The low-voltage oscillations set up in the primary winding of T1970 are raised by transformer action to high-voltage levels in the secondary windings. These ac secondary voltages are applied to the DC Restorer, the Cathode Supply, and the anode multiplier circuits.

Oscillation occurs due to the positive feedback from the primary winding (pin 4 to pin 5) to the smaller base-drive winding (pin 3 to pin 6) for transistor Q1981. The frequency of oscillation is about 50 kHz, and is determined primarily by the resonant frequency of the transformer.

When power is first applied, the High-Voltage Regulator circuit detects that the negative crt cathode voltage is too positive and pulls pin 2 of transformer T1970 negative. The negative level forward biases transistor Q1981 via the base-drive winding of the transformer. Current begins to flow in the primary winding through transistor Q1981, inducing a magnetic field around the transformer primary winding. The increasing magnetic field induces a current in the base-drive winding that further increases the base drive to the transistor. This in-phase feedback causes current in Q1981 to increase until the primary winding current reaches its maximum value. As the rate of change of the primary current peaks and then reverses, the induced magnetic field begins to decay. This decreases the base-drive current and begins turning Q1981 off.

As Q1981 is beginning to turn off, the magnetic field around the primary winding continues to collapse at the resonant frequency rate of the transformer. This induces into the base-drive winding a voltage that completely turns off the transistor. The collapsing magnetic field goes to zero, then builds in the opposite direction to a maximum before collapsing again (resonant flywheel effect). This sequence of events occurs repetitively as the circuit continues to oscillate.

The oscillating magnetic field in the primary winding couples power into the secondary windings of the transformer. The amplitude of the voltages induced in the secondary windings is a function of the turns ratios of the transformer windings.

High-Voltage Regulator

The High-Voltage Regulator consists of U1956A and B and associated components. It monitors the crt Cathode Supply voltage and varies the bias point of the switching transistor in the High Voltage Oscillator to hold the Cathode Supply voltage at the nominal level. Since the output voltages at the other secondary winding taps are related by turns ratios to the Cathode Supply voltage, all voltages are held in regulation.

When the Cathode Supply voltage is at the proper level (−1900 V), the current through R1945 and the 19-MΩ resistor internal to High Voltage Module U1830 holds the voltage developed across C1932 at zero volts. This is the balanced condition and sets base drive in Q1981 via integrator U1956A and voltage-follower U1956B. Varying base drive to Q1981 holds the secondary voltages in regulation.

If the Cathode Supply voltage level tends too positive, a slightly positive voltage will develop across C1932. This voltage causes the outputs of integrator U1956A and voltage-follower U1956B to move negative. The negative shift charges capacitor C1951 to a different level, around which the induced feedback voltage at the base-drive winding will swing. The added negative bias causes Q1981 to turn on earlier in the oscillation cycle, and a stronger current pulse is induced in the secondary windings. The increased power in the secondary windings increases the secondary voltages until the Cathode Supply voltage returns to the balanced condition (zero volts across C1932). Opposite action occurs should the Cathode Supply voltage tend too negative.

Cathode Supply

The Cathode Supply circuit is composed of a voltage-doubler and an RC filter network contained within High-Voltage Module U1830. This supply produces the −1900 V accelerating potential applied to the CRT cathode and the −900 V slot lens voltage. The −1900 V supply is monitored by the High Voltage Regulator to maintain the regulation of all voltages from the High Voltage Oscillator.

The alternating voltage (950 V peak) from pin 10 of transformer T1970 is applied to a conventional voltage-doubler circuit at pin 7 of the High Voltage Module. On the positive half cycle, the input capacitor of the voltage doubler (0.006 μf) is charged to −950 V through the forward-biased diode connected to ground at pin 9 of the module (charging path is through the diode, so stored charge is negative). The following negative half cycle adds its ac component (−950 V peak) to this stored dc value and produces a total peak voltage of −1900 V across the capacitor. This charges the 0.006-μf storage capacitor (connected across the two doubler diodes) through the second diode (now the forward-biased diode) to −1900 V. Two RC filters follow the voltage doubler to smooth out the ac ripple. A resistive voltage divider across the output of the filter network provides the −900-V slot lens potential.

Anode Multiplier

The Anode Multiplier circuit (also contained in High Voltage Module U1830) uses voltage multiplication to produce the +14 kV CRT anode potential. Circuit operation is similar to that of the voltage-doubler circuit of the Cathode Supply.

The first negative half-cycle charges the 0.001-μf input capacitor (connected to pin 8 of the High Voltage Module) to a positive peak value of +2.33 kV. The following positive half cycle adds its positive peak amplitude to the voltage stored on the input capacitor and boosts the charge on the second capacitor of the multiplier (and those following) to +4.66 kV. Following cycles continue to boost up succeeding capacitors to values 2.33 kV higher than the preceding capacitor until all six capacitors are fully charged. This places the output of the last capacitor in the multiplier at +14 kV above ground potential. Once the multiplier reaches operating potential, succeeding cycles replenish current drawn from the Anode Multiplier by the crt beam. The 1-MΩ resistor in series with the output protects the multiplier by limiting the anode current to a safe value.

Focus Amplifier

The Focus Amplifier, in conjunction with the auto-focus circuitry of Z-Axis hybrid U950 (diagram 6), provides optimum focus of the crt beam for all settings of the front-panel INTENSITY control. The Focus Amplifier itself consists of two shunt-feedback amplifiers composed of Q1851, Q1852, and associated components. The outputs of the amplifiers set the operating points of a horizontally converging quadrapole lens and a vertically converging quadrapole lens within the crt. The convergence strength of each lens is dependent on the electric field set up between the lens elements.

Since the bases of Q1851 and Q1852 are held at constant voltages (set by their emitter potentials), changing the position of the wiper arms of the ASTIG and FOCUS pots changes the amount of current sourced to the base junctions through R1856 and R1857 respectively. This changes the base-drive currents and produces different

output levels from the Focus Amplifiers; that, in turn, changes the convergence characteristics of the quadrapole lenses.

Initially, at the time of adjustment, the FOCUS and ASTIG potentiometers are set for optimum focus of the crt beam at low intensity. After that initial adjustment, the ASTIG pot normally remains as set, and the FOCUS control is positioned by the user as required when viewing the displays. When using the FOCUS control, transistor Q1852 is controlled as described above; however, an additional current is also supplied to the base node of Q1851 from the FOCUS pot through R1855. This additional current varies the base-drive current to Q1851 and provides tracking between the two lenses as the FOCUS control is adjusted during use of the instrument.

The convergence strengths of the quadrapole lenses also dynamically track changes in the display intensity. The VQ OUT signal, applied to the crt at pins 5 and 6, is exponentially related to the VZ OUT (intensity) signal driving the crt control grid and increases the strength of the lenses more at higher crt beam currents. (A higher beam current requires a stronger lens to cause an equal convergence of the beam.)

DC Restorer

The DC Restorer provides crt control-grid bias and couples both the dc and the low-frequency components of the Z-Axis drive signal to the crt control grid. This circuit allows the Z-Axis Amplifier to control the display intensity by coupling the low-voltage Z-Axis drive signal (VZ OUT)

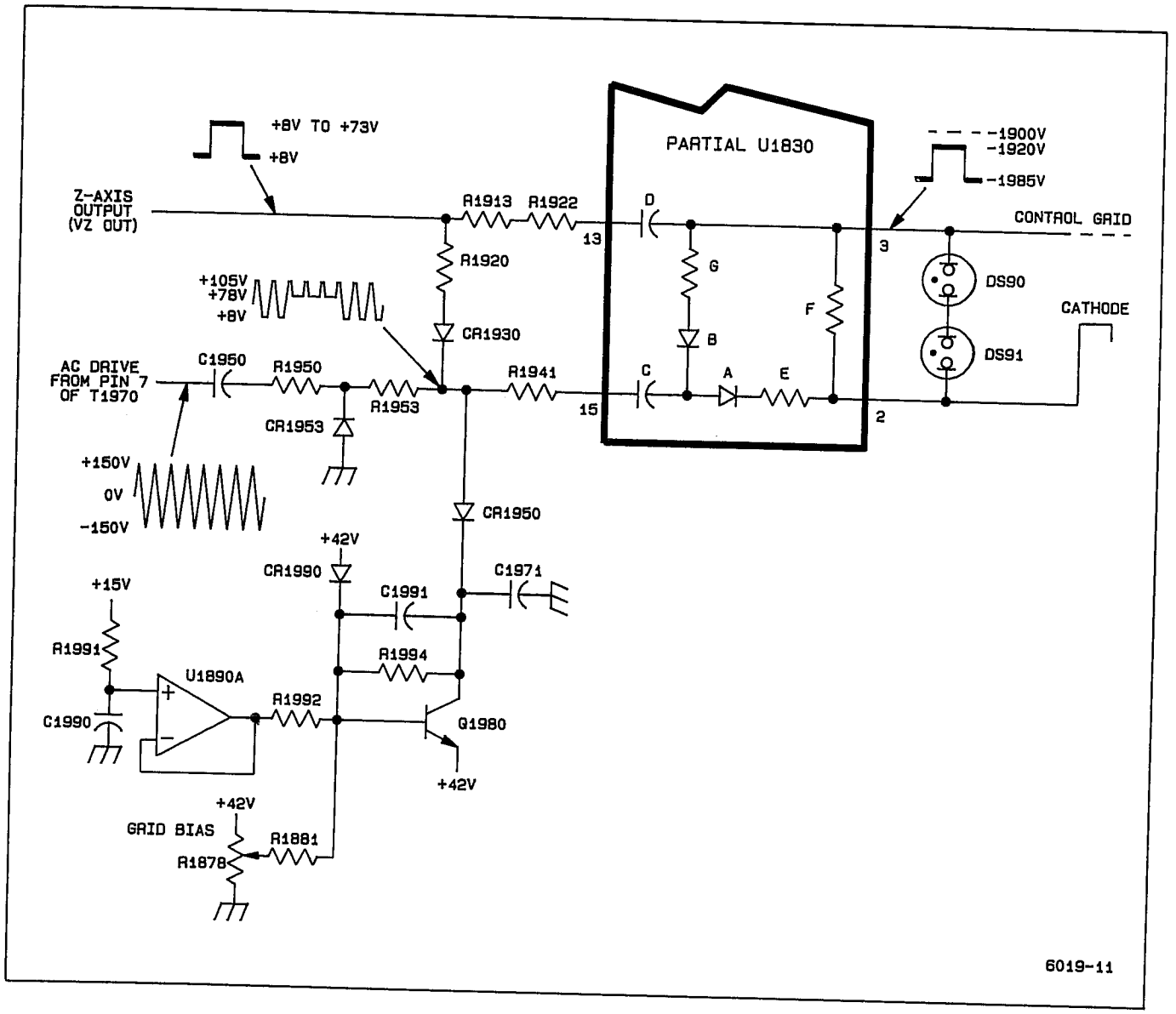


Figure 3-9. Dc restorer circuit.

to the elevated crt control-grid potential (about −1.9 kV).

The DC Restorer circuit (Figure 3-9) operates by impressing the crt grid bias setting and the Z-Axis drive signal on an ac voltage waveform. The shaped ac waveform is then coupled to the crt control grid through a coupling capacitor that restores the dc components of the signal.

GRID BIAS LEVEL. An ac drive voltage of approximately 300 V peak-to-peak is applied to the DC Restorer circuit from pin 7 of transformer T1970. The negative half cycle of the sinusoidal waveform is clipped by CR1953, and the positive half cycle (150 V peak) is applied to the junction of CR1930, CR1950, and R1941 via R1950 and R1953. Transistor Q1980, operational amplifier U1890A, and associated components form a voltage clamp circuit that limits the positive swing of the ac waveform at the junction.

Transistor Q1980 is configured as a shunt-feedback amplifier, with C1991 and R1994 as the feedback elements. The feedback current through R1994 develops a voltage across the resistor that is positive with respect to the +42.6 V on the base of the transistor. The value of this additive voltage plus the diode drop across CR1950 sets the upper clamping threshold. Grid Bias potentiometer R1878 sinks varying amounts of current away from the base node of the transistor and thus sets the feedback current through R1994. The adjustment range of the pot can set the nominal clamping level between +71 V and +133 V.

When the amplitude of the ac waveform is below the clamping threshold, series diode CR1950 will be reverse biased and the ac waveform is not clamped. During the time the diode is reverse biased, transistor Q1980 is kept biased in the active region by the charge retained on C1971 from the previous cycle. As the amplitude of the ac waveform at the junction of CR1930 and CR1950 exceeds the voltage at the collector of Q1980, diode CR1950 becomes forward biased, and the ac waveform is clamped at that level. Any current greater than that required to maintain the clamp voltage will be shunted to the +42 V supply by transistor Q1980.

Operational amplifier U1890A sinks a time-dependent variable current away from the base node of Q1980 that modifies the crt control-grid bias during the first few minutes of instrument operation. The circuit compensates for the changing drive characteristics of the crt as it warms up.

At power-up, capacitor C1990 begins charging through R1991 toward the +15 V supply. The output of U1890A follows the rising voltage on pin 3; and after about ten minutes (for all practical purposes), it reaches +15 V. As the output voltage slowly increases, the charging current through R1992 causes the Grid Bias voltage to gradually lower about ten volts from its power-on level. The charge on C1990 dissipates slowly; therefore, if instrument power is turned off and then immediately back on again, the output of U1890A will still be near the +15 V limit rather than starting at zero volts as when the crt was cold.

Z-AXIS DRIVE LEVEL. The variable-level Z-Axis signal (VZ OUT) establishes the lower clamping level of the ac waveform applied to the High Voltage Module. When the amplitude of the waveform drops below the Z-Axis signal, CR1930 becomes forward biased, and the ac waveform is clamped to the Z-Axis signal level. The VZ OUT level may vary between +8 V and +75 V, depending on the setting of the front-panel INTENSITY and READOUT INTENSITY controls.

The ac waveform, now carrying both the grid-bias information and the Z-Axis drive information, is applied to a DC Restorer circuit in the High Voltage Module where it is raised to the high-voltage levels of the crt control grid.

DC RESTORATION. The DC Restorer circuit in the High Voltage Module is referenced to the crt cathode voltage via a connection within U1830. Capacitor C (in Figure 3-9), connected to pin 15 of U1830, initially charges to a level determined by the difference between the Z-Axis signal level and the crt cathode potential. The Z-Axis signal sets the level on the positive plate of capacitor C through R1920, CR1930, and R1941; the level on the negative plate is set by the crt cathode voltage through resistor E and diode A. Capacitor D is charged to a similar dc level through resistors F, R1922, and R1913.

When the ac waveform applied to pin 15 begins its transition from the lower clamped level (set by the Z-Axis signal) towards the upper clamped level (set by the Grid Bias potentiometer), the charge on capacitor C increases. The additional charge is proportional to the voltage difference between the two clamped voltage levels.

When the ac waveform begins its transition from the upper clamped level back to the lower clamped level, diode A becomes reverse biased. Diode B becomes forward biased, and an additional charge proportional to the negative excursion of the ac waveform (difference between the upper clamped level and the lower clamped level) is added to capacitor D through diode B and resistor G. The amount of change added to capacitor D depends on the setting of the front-panel INTENSITY control, as it sets the lower clamping level of the ac waveform. This added charge determines the potential of the control grid with respect to the crt cathode.

The potential difference between the control grid and the cathode controls the beam current and thus the display intensity. With no Z-Axis signal applied (INTENSITY control off), capacitor D will be charged to its maximum negative value, since the difference between the two clamped voltage levels is at its maximum value. This is the minimum intensity condition and reflects the setting of the Grid Bias potentiometer. During calibration, the Grid Bias pot is adjusted so that the difference between the upper clamping level (set by the Grid Bias pot) and the "no signal" level of the Z-Axis drive signal (VZ OUT) produces a control grid bias that barely shuts off the crt electron beam.

As the INTENSITY control is advanced, the amplitude of the square-wave Z-Axis signal increases accordingly. This increased signal amplitude decreases the difference between the upper and lower clamped levels of the ac waveform, and less charge is added to capacitor D. The decreased voltage across capacitor D decreases the potential difference between the control grid and the cathode, and more crt beam current is allowed to flow. Increased beam current increases the crt display intensity.

During the periods that capacitor C is charging and discharging, the control-grid voltage is held stable by the long-time-constant discharge path of capacitor D through resistor F. Any charge removed from capacitor D during the positive transitions of the ac waveform will be replaced on the negative transitions.

The fast-rise and fast-fall transitions of the Z-Axis signal are coupled to the crt control grid through capacitor D. This ac-coupled fast-path signal quickly sends the crt electron beam to the new intensity level, then the slower DC Restorer path "catches up" to handle the dc and low-frequency components of the Z-Axis drive signal.

Neon lamps DS90 and DS91 prevent arcing inside the crt should the control grid potential or cathode potential be lost for any reason.

CRT Control Circuits

The CRT Control circuits provide the various potentials and signal attenuation factors that set up the electrical elements of the crt. The control circuitry is divided into two separate categories: (1) level setting and (2) signal handling. The level setting circuitry produces voltages and current level necessary for the crt to operate, while the signal-handling portion is associated with changing crt signal levels.

LEVEL-SETTING CIRCUITRY. Operational amplifier U1890B, transistor Q1980, and associated components form an edge-focus circuit that sets the voltages on the elements of the third quadrapole lens. The positive lens element is set to its operating potential by Edge Focus adjustment pot R1864 (via R1897). This voltage is also divided by R1893 and R1982 and applied to the non-inverting input of U1890B to control the voltage on the other element of the lens.

The operational amplifier and transistor are configured as a feedback amplifier, with R1891 and R1990 setting the stage gain. Gain of the amplifier is equal to the attenuation factor of divider network R1893 and R1892, so total overall gain of the stage from the wiper of R1864 to the collector of Q1890 is unity. The offset voltage between lens elements is set by the ratio of R1891 and R1990 and the +10 V reference applied to R1990. This configuration causes the two voltages applied to the third quadrapole lens to track each other over the entire range of Edge Focus adjustment pot R1864.

Other adjustable level-setting circuits include Y-Axis Alignment pot R1848, used to rotate the beam alignment after vertical deflection. This adjustment controls the amount of current through the Y-Axis Alignment coil around the neck of the crt and is set to produce precise perpendicular alignment between x- and y-axis deflections. The TRACE ROTATION adjustment R975 is a front-panel screwdriver-adjustable control. The effect of the adjustment is similar to the Y-Axis Alignment pot, but when adjusted, it rotates both the x-axis and the y-axis deflections of the trace on the face of the crt. A final adjustable level-setting control is the Geometry pot R1870, adjusted to optimize display geometry. The potential at pin 8 for the vertical shield internal to the crt is produced by zener diode VR1891 and associated components.

SIGNAL-HANDLING CIRCUITRY. The crt termination adjustment R1501 is set to match the loading characteristics of the crt's vertical deflection structure to the Vertical Output Amplifier.

LOW VOLTAGE POWER SUPPLY

The low voltages required by the instrument are produced by a high-efficiency, switching power supply. This type of supply directly rectifies and stores charge from the ac line supply; then the stored charge is switched through a special transformer at a high rate, generating the various supply voltages.

Line Rectifier

Ac line voltages of either 115 V or 230 V may provide the primary power for the instrument, depending on the setting of LINE VOLTAGE SELECTOR switch S90

(located on the instrument rear panel). Power Switch S350 applies the selected line voltage to power supply rectifier CR1011.

With the selector switch in the 115 V position, the rectifier and storage capacitors C1021 and C1022 operate as a full-wave voltage doubler. When operating in this configuration, each capacitor is charged on opposite half cycles of the ac input, and the voltages across the two capacitors in series will approximate the peak-to-peak value of the source voltage. For 230 V operation, switch S90 connects the rectifier as a conventional bridge rectifier. Both capacitors charge on both input half cycles, and the voltage across C1021 and C1022 in series will approximate the peak value of the rectified source voltage. For either configuration, the dc voltage supplied to the power supply inverter is the same.

Thermistors RT1010 and RT1016 limit the surge current when the power supply is first turned on. As current flow warms the thermistors, their resistances decrease and have little effect on circuit operation. Spark-gap electrodes E1001 and E1002 are surge-voltage protectors. If excessive source voltage is applied to the instrument, the spark-gaps conduct, and the extra current flow quickly exceeds the rating of fuse F90. The fuse then opens to protect the instrument's power supply. The EMI (electromagnetic interference) filter, inductors L1011 and L1012, capacitors C1016 and C1018, and resistors R1011, R1012, R1016 and R1018 form a line-filter circuit. This filter, along with common mode rejection transformer T1020, prevents power-line interference from entering the instrument and prevents power supply switching signals from entering the supply line.

Preregulator Control

The Preregulator Control circuit monitors the drive voltage applied to inverter output transformer T1060 and holds it at the level that produces proper supply voltages at the secondary windings.

The Preregulator Control circuit consists primarily of control IC U1030, its switching buffers, and its power supply components. The control IC senses voltage on the primary winding of T2060 and varies the "on time" of a series-switching transistor, depending on whether the sensed voltage was too high or too low. The switching transistor Q1050, rectifier CR1050, choke T1050, and capacitor C1050 form a buck-switching regulator circuit. The output voltage at W1060 is proportional to the product of the rectified line voltage on C1020-C1022 and the duty cycle of Q1050. In normal operation, Q1050 is on about one-half the time. When Q1050 is off, current flows to W1060 and T1060 through CR1050.

PREREGULATOR CONTROL POWER SUPPLY. Since the Preregulator Control network controls supply startup and preregulates the secondary supplies, an independent power source must be established for it before any of the other power supplies will operate. The independent power supply for the control circuitry is composed of Q1021, Q1022, and associated components.

Initially, when instrument power is applied, the positive plate of capacitor C1025 is charged toward the positive rectified line voltage through R1020. The voltage at the base of Q1022 follows at a level determined by the voltage divider composed of R1022, R1024, CR1023, and the load within U1030. When the voltage across C1025 reaches about +21 V, the base voltage of Q1022 reaches +6.8 V and Q1022 turns on, saturating Q1021. The +21 V on the emitter of Q1021 appears at its collector and establishes the positive voltage supply for the Preregulator IC. With Q1021 on, R1024 is placed in parallel with R1022, and both Q1022 and Q1021 remain saturated.

The +21 V level begins to drain down as the control IC draws current from C1025. If the Preregulator Control IC doesn't start the switching supply (and thus recharge C1025 and C1023 via CR1022) by the time the voltage across C1025 reaches about +8 V, Q1021 will turn off. Resistor R1024 pulls the base of Q1022 low and turns that transistor off also. (Capacitor C1025 would only discharge low enough to turn off the transistors under a fault condition.) In this event, C1025 would then charge again to +21 V, and the start sequence would repeat. Normally, the control IC will start Inverter action before the +8 V level is reached, and current is drawn through T1050 via Q1050. This induces a current in the secondary winding of T1050 via Q1050. This induces a current in the secondary winding of T1050 and charges C1025 positive via diode CR1022. The turns ratio of T1050 sets the secondary voltage at approximately +15 V; and, as long as the supply is being properly regulated, C1025 will be charged up to that level and held there.

PREREGULATOR START-UP. As the supply for the Preregulator Control IC is established, an internal switching oscillator begins to run. The oscillator generates a repetitive triangular wave (as shown in Figure 3-10) at a frequency determined primarily by R1032 and C1032. The simplified schematic of Figure 3-11 illustrates the voltage control functions of U1030.

As the Preregulator power supply turns on, capacitor C1034 charges from the +5 V reference level toward ground potential through R1034 and R1037. As it does, the voltage at pin 4 (one input of Dead-Time Comparator U1) will pass through the positive-peak value of the triangular waveform on the other input of the Dead-Time

Comparator. The comparator will then begin outputting narrow pulses that become progressively wider as the voltage on pin 4 settles to zero volts. These pulses drive switching transistor Q1050, and their slow progression from narrow to wide causes the various secondary supplies to gradually build up to their final operating levels. The slow buildup prevents a turn-on current surge that would cause the current-limit circuitry to shut down the supply.

During startup, capacitor C1072 acts as a substantial load, and a relatively large current flows in the windings of T1050 for the first few cycles of Preregulator switching. These strong current pulses ensure that storage capacitor C1066 becomes charged sufficiently to start the Inverter Drive circuit. Once the Inverter Drive stage is operating, the normal switching current through T1050 maintains the required charge on C1066. (The Inverter Drive power supply is discussed later in this description.)

Dead-Time Comparator U1 is referenced at approximately 0.1 V above the ground level at pin 4 (established when C1034 becomes fully charged) and outputs a narrow, negative-going pulse that turns off switching transistor Q1050 for a portion of each switching cycle. This off time ensures that flip-flop U1064B in the Inverter Drive circuit toggles every cycle (thereby maintaining the proper duty cycle), independent of the voltage conditions being sensed by the remainder of the voltage control circuitry.

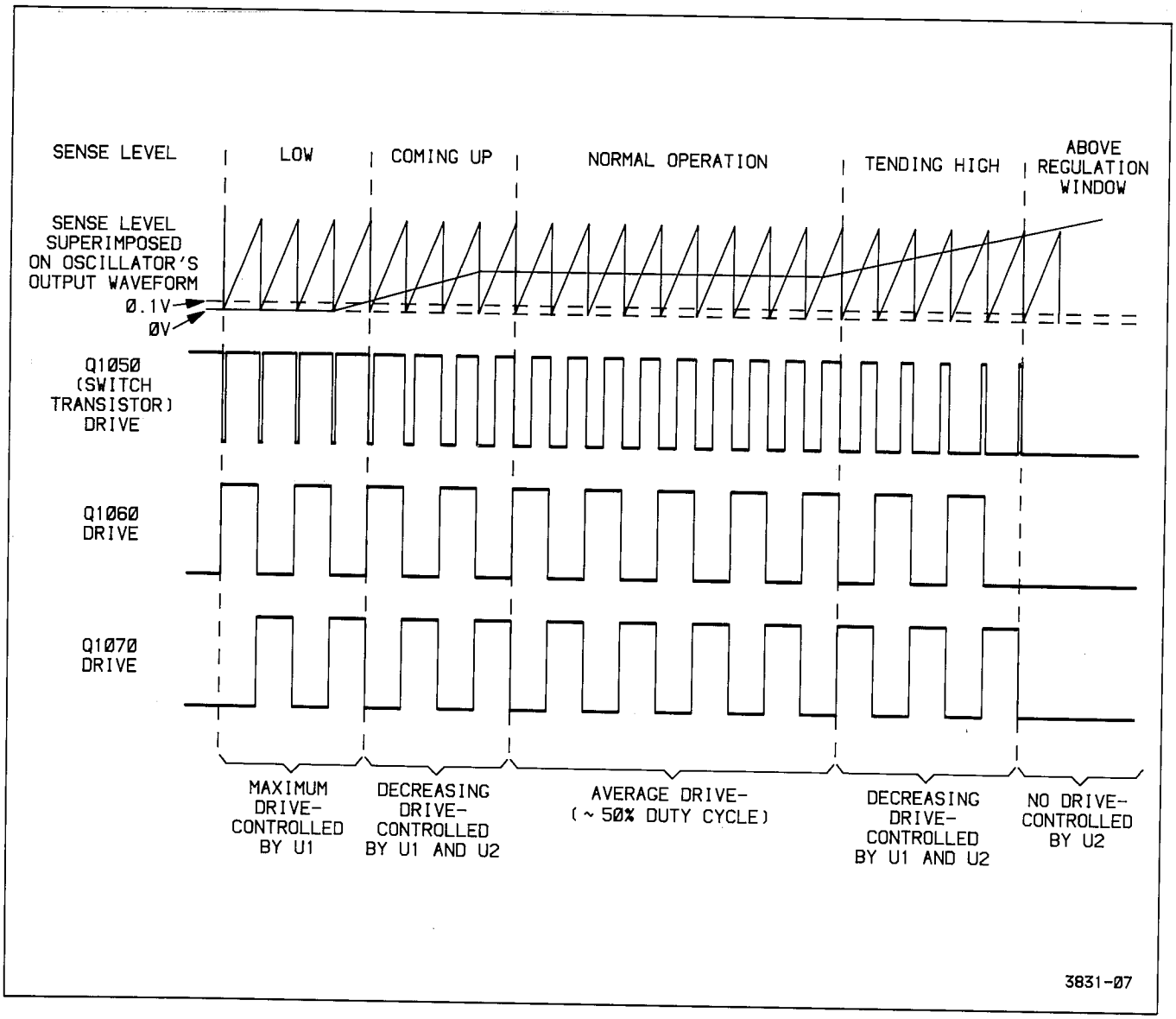


Figure 3-10. Timing relationships of the Inverter Drive signals.

PREREGULATION. Once the initial charging at power-up is accomplished, as just described, the voltage-sensing circuitry begins controlling the Inverter switching action. The actual voltage sensing is done by error amplifier U2. The level at the center tap of output transformer T1060 is applied to pin 1 and is compared to the reference established by R1045 and R1046 at pin 2. If the sensed level at pin 1 is lower than the reference level (as it will always be for the first few switching cycles), the of error-amplifier U2 will be LO. The LO, applied to the inverting input of U3, results in a long-duty-cycle drive signal to transistor Q1050 (via CR1030). Since the Inverter Drive stage will alternately turn either Q1060 or Q1070 on, relatively large current pulses will result in the primary winding of inverter output transformer T1060.

These large current pulses, over the period of a few cycles, will increase the charge on the storage capacitors on the secondary side of the transformer and will reduce the current demand on the inverter output transformer. As the demand increases, the voltage across the primary winding will increase until it reaches the point where the two inputs of U2 are at the same potential. At this point, the output of U2 (to U3) will settle to a level approximately equal to the midpoint of the triangular waveform applied to the other input of U3. The resulting drive signal has an approximate 50% duty cycle and will respond to changes in either the ac line voltage or supply load conditions. Depending on the output levels sensed, the duty cycle of the drive signal will change (sensed level rises or falls with respect to the triangular waveform) to hold the secondary supplies at their proper levels.

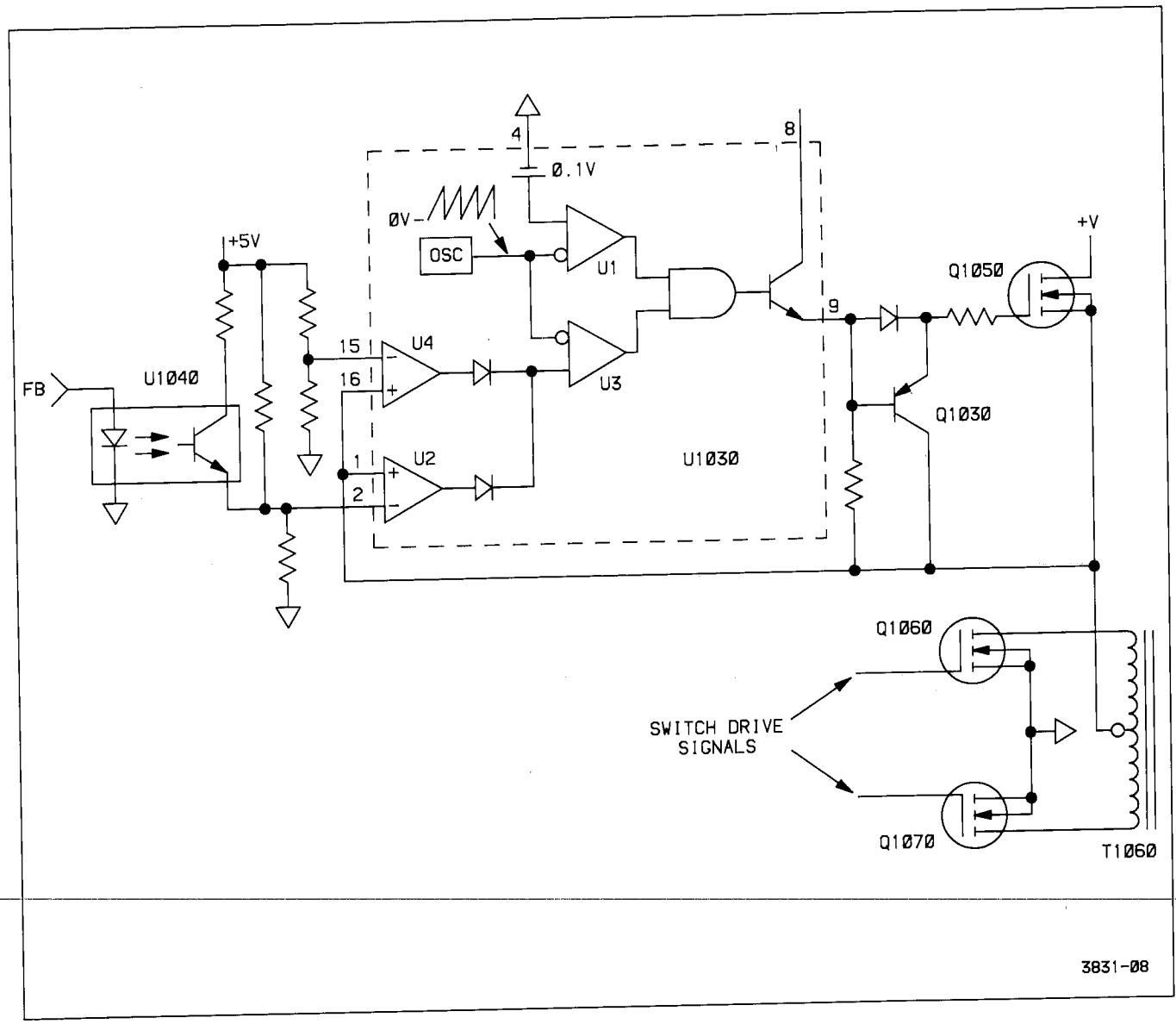


Figure 3-11. Simplified schematic of control network.

Opto-isolator U1040 and resistor R1044 form a control network that allows a voltage sensed at the feedback input (FB) to slightly alter the voltage-sense reference applied to pin 2 of U2. The FB signal is generated by the +5 V Inverter Feedback amplifier (U1371, diagram 10) and is directly related to the level of the $+5V_D$ supply line. Base drive to the shunt transistor (in opto-isolator U1040) is increased should the FB signal go below its nominal value. Additional current is shunted around R1045 (via R1044) and raises the voltage-sense reference level to error-amplifier U2. This increases the voltage applied to the primary winding of the output transformer, since U2 sensing depends on a balanced condition. Higher currents are induced in the secondary windings, and the secondary voltages begin to return to their nominal values. As the $+5V_D$ line returns to its nominal level, base drive to the shunt transistor will be reduced and the voltage in the primary winding will follow. Should the FB signal level tend too high, opposite control responses occur. Further information about the FB signal is given in the +5 V Inverter Feedback description.

Error amplifier U4 and the voltage divider composed of R1035 and R1031 provide a backup sensing circuit. Its operation is similar to that of error amplifier U2, just described, but it senses at a slightly higher level. As long as U2 is operating properly, U4 will be inactive. However, should a failure occur in the U2 sensing circuitry, the voltage on the primary winding of T1060 will rise to the sensing level at pin 15 of U4. Sense amplifier U4 will then take over, preventing a damaging over-voltage condition.

Inverter Drive

The Inverter Drive circuit performs the necessary switching to drive the inverter output transformer. Like the Preregulator Control IC, the Inverter Drive circuit requires an independent power supply, since it must be operational before any of the secondary supply voltages can be generated.

INVERTER DRIVE POWER SUPPLY. This power supply consists of Q1062, VR1062, and their associated components. As power is first applied, the initial charging current through T1050 induces a current in the transformer secondary winding (pins 8 and 9). The alternating current is rectified by the diode bridge composed of CR1062, CR1063, CR1064, and CR1065 and stored in C1066, providing power for the Inverter Drive circuitry.

When the Preregulator Control IC turns switching transistor Q1050 on for the first time, the charge stored on C1066 during the initial charging period is sufficient to properly turn on one of the current-switching transistors (either Q1060 or Q1070) for the first cycle. After that, the alternating drive signals continue to induce current into the secondary winding of T1050 to provide operating power as long as the instrument is turned on.

The current rectified by the diode bridge and stored on capacitor C1066 is regulated down to the required voltage level by R1061, VR1062, and Q1062. Zener diode VR1062 references emitter-follower Q1062 and holds the supply output at approximately +11.4 V.

INVERTER DRIVE GENERATOR. The Inverter Drive generator consists of U1062, U1064, U1066, switching transistors Q1060, Q1070 and their associated components. The circuitry alternately switches current through each leg of the output transformer (T1060) primary winding and produces the ac current required for transformer action.

Out-of-phase input signals to comparator U1062C come from two resistive voltage dividers placed in either leg of one secondary winding of T1050. The comparator detects the phase changes (crossover points) of the secondary current caused as Q1050 switches on and off. Every complete on-off cycle of Q1050 produces a positive clock at pin 14 of U1062C that toggles flip-flop U1064B. The toggling alternately turns switching transistors Q1060 and Q1070 on, each with an approximate 50% duty cycle.

Comparators U1062A and U1062B, at the Q and $\overline{Q}$ output of the flip-flop, detect the precise crossing point of the toggling drive signals and ensure that only one switching transistor will be on at any one time. These mutually-exclusive drive signals are buffered by inverters U1066A and U1066B and applied to switching transistors Q1060 and Q1070 to alternately turn them on and off at one-half the switching rate of Q1050. By alternately switching opposite ends of the primary winding to ground, the current flowing through switching transistor Q1050 will flow alternately in each half of the primary winding. This produces ac voltages at the secondary windings that are then rectified, providing the various unregulated dc supply voltages.

Current Limit

The Current Limit circuit, composed of transistor Q1040 and the associated components, limits the maximum current flow in the output transformer to about 1 ampere. Resistor R1040 (connected to the Preregulator Control IC +15 V supply) forward biases germanium diode CR1040 and applies approximately +0.3 V across the base-to-emitter junction of Q1040. Current flowing to the output transformer develops a voltage drop across R1050 that adds to the bias developed by CR1040. As the current to the transformer increases, the voltage drop across R1050 also increases until, at around 1 A, the combined voltage drop across R1050 and CR1040 forward biases transistor

Q1040. The base of Q1022 is pulled negative through R1042, and the +15 V supply for the Preregulator IC turns off (see Preregulator Control description). The power supply will try to restart itself; but, as long as the excessive-current condition persists, the current-limit circuit will keep shutting the supply down, protecting the instrument.

Rectifiers

The rectifiers convert the alternating current from the secondary windings of inverter output transformer T1060 to the various dc supply voltages required by the instrument. Rectification is done by conventional diode rectifier circuits, and filtering is done by conventional LC networks.

The +87 V unregulated supply is produced by a voltage-doubler circuit. The positive plate of C1130 at the anode of CR1132 is referenced at approximately +45 V through diode CR1131 (to the +42 V unregulated supply). As the positive half cycle from the 42 V secondary winding (actually about +45 V peak) is applied to the negative plate of C1130, the positive plate is elevated to a peak value of approximately +90 V. Diode CR1132 becomes forward biased and storage capacitor C1132 is charged to about +90 V. Following cycles replenish the charge drawn off by the loads on the +87 V supply line.

Line Signal

A sample of the ac line voltage is coupled to the Trigger circuit by transformer T1229 and provides the LINE TRIG signal to the Trigger hybrid. Transformer current is limited to a safe value by resistors R1014 and R1015 placed in series with the primary winding leads. The transformer's output characteristics are matched to the input of the Trigger circuit hybrid by R1208 and C1208.

Line Up Signal

The circuit composed of Q1029, opto-isolator U1029, and their associated components, detects when power has been applied to the instrument and the Preregulator Control power supply is functioning properly. When the rectified line voltage reaches proper operating voltage, the voltage divider composed of R1027 and R1028 forward biases Q1029. As soon as the Preregulator Control power supply turns on, current flows through R1029, Q1029, and the opto-isolator LED. The illuminated LED saturates transistor U1029 and the LINE UP signal to the Power-Up Delay circuit (diagram 1) is pulled HI, indicating that the Preregulator Control circuit should now be functioning properly.

POWER DOWN. When instrument power is turned off, the voltage across the primary storage capacitors (C1021 and C1022) begins to fall as the capacitors discharge. As the voltage drops, the bias current through R1027 to the base of Q1029 also drops until the bias voltage across R1028 reaches a point about 2 V above the average transformer drive level at pin 2 of U1029. At this point, Q1029 turns off, and the LINE UP signal to the Power-Up Delay circuit goes LO. This LO signals the Microprocessor that it should start its power down routine.

The Line Up circuit tells the Microprocessor that the primary capacitors have started discharging while there is still a stored charge (set by R1027 and R1028) about 40% in excess of that required to keep the power supply voltages in regulation. This allows the Microprocessor to complete the power-down sequence before the supplies drop below their normal operating level. Further information about the power-down sequence is given in the Microprocessor Reset Control description.

Fan Circuit

Fan motor B10 is driven by adjustable three terminal regulator U1110. The fan's speed is determined by the voltage supplied by U1110 and varies with ambient temperature.

As the ambient temperature in the cabinet increases, the resistance of thermistor RT1110 decreases causing more current to flow in R1112. This causes the voltage at pin 2 and therefore the voltage at pin 3 of U1110 to increase, and the fan motor speed increases to provide more cooling capacity.

LOW-VOLTAGE REGULATORS

The Low-Voltage Regulators remove ac noise and ripple from the various unregulated dc supply voltages. Each regulator output is automatically current limited if the output current exceeds the requirements of a normally functioning instrument. This limiting prevents any further component damage.

+10 Volt Reference

Each of the power-supply regulators control their respective outputs by comparing their output voltages to a known reference level. In order to maintain stable supply voltages, the reference voltage must itself be highly stable. The circuit composed of U1290, U1300C and associated components establish this reference.

Resistor R1400 and capacitor C1400 form an RC filter network that smooths the unregulated +15 volt supply

before it is applied to voltage-reference IC U1290. The +2.5 V output from pin 2 of U1290 is applied to the noninverting input of operational amplifier U1300C. The output of U1300C is the source of the +10 V reference level used by the various regulators. The output level is set by the voltage divider formed by R1291, R1293, and potentiometer R1292. The Volt Ref Adjust pot in the divider allows the reference level to be precisely set. Zener diode VR1292 prevents the reference from exceeding +11 volts should a failure in the reference circuitry occur.

+87 V Regulator

The +87 V Regulator is composed of Q1220, Q1221, Q1222, Q1223, U1281A, and their associated components. The circuit regulates and limits both the voltage and current of the supply output.

Initially, as power is applied, the voltage applied to pin 2 of U1281A from the voltage divider formed by R1227 and R1228 is lower than the +10 V reference level applied to pin 3. The output of U1281A is forced high, reverse biasing the base-emitter junction of Q1222 and turning it completely off. With Q1222 off, all the current through R1212 is supplied as base current to Darlington transistor pair Q1221 and Q1220, and maximum current flows in series-pass transistor Q1220. This charges up the various loads on the supply line, and the output level charges positive.

As the regulator output charges toward +87 V, the voltage divider applies a positive-going voltage to the inverting input of U1281A. When the output level reaches +87 volts, the inverting input reaches the +10 V reference at the noninverting input. The output voltage at pin 1 of U1281A will go negative and the base-emitter junction of Q1222 will be biased into the active region. As Q1222 turns on, base drive for the Darlington pair (Q1221 and pass transistor Q1220) is reduced. The output will be held at the level required (+87 V) for voltage at the two inputs of amplifier U1281A to be in balance.

Current limiting is a foldback design and is performed by Q1223 and its associated components. Under normal current demand conditions, Q1223 is off. If the regulator output current exceeds approximately 100mA (as it might if a component fails), the voltage drop across R1221 and CR1220 reaches a point that forward biases Q1223 via the bias divider formed by R1222 and R1223. As Q1223 turns on, a portion of the base-drive current to Q1221 is shunted away by Q1223. This reduces the base-drive current (and thus the output current) of series-pass transistor Q1220.

+42 V Regulator

The circuit configuration and operation of the +42 V Regulator is identical to that of the +82 V Regulator. Current limiting of the +42 V supply occurs at approximately 400 mA. Base drive to Darlington pair Q1241 and Q1240 is via R1244 and is dependent on proper operation of the +87 Volt Regulator. This dependency ensures that the relative polarities of the two supplies are never reversed (preventing semiconductor-junction damage in the associated load circuitry).

+15 V Regulator

The +15 V Regulator uses three-terminal regulator U1260 and operational amplifiers U1371A and U1371B, arranged as voltage sensors, to achieve regulation of the +15 V supply. The three-terminal regulator holds its output voltage at pin 2 at 1.25 volts more positive than the reference input level at pin 1. The voltage at the reference pin is established by current flow in either diode CR1262 or CR1263.

Resistors R1261 and R1262 at the regulator output divide the +15 V level down for comparison with the +10 V reference applied to pin 5 of operational amplifier U1371B. When the input voltage at pin 6 (supplied by the voltage divider) is lower than the +10 V reference, the output of amplifier U1371B is high and the output voltage of U1260 is allowed to rise. As the regulator output reaches +15 V, the voltage on pin 6 of U1371B approaches the level on pin 5, and the amplifier begins sinking current away from the reference pin of the three-terminal regulator via diode CR1263. This lowers the voltage on the reference pin and holds the output at +15 V.

The other voltage-sensing amplifier (U1371A) ensures that the relative polarity between the +15 V supply and the +42 V supply is maintained, preventing component damage in the load circuitry. Should the +42 V supply be pulled below +15 V (excessive loading or supply failure), the voltage at pin 3 of U1371A falls below the voltage at pin 2 and the amplifier output voltage goes low. This forward biases CR1262 and lowers the reference voltage for U1260, reducing the output voltage.

Current limiting for the +15 V supply is provided by the internal circuitry of the three-terminal regulator.

+5 V Regulator

Regulation of the +5 V supply is provided by a circuit similar to those of the +87 V and the +42 V Regulators. As long as the relative polarity between the +15 V and the +5 V supplies is maintained, base drive to Q1281 is

supplied through R1283. The current through Q1281 provides base drive for series-pass transistor Q1280.

When voltage-sense amplifier U1300B detects that the output voltage has reached +5 V, it begins shunting base-drive current away from Q1281 via CR1281 and holds the output voltage constant.

Current limiting for the +5 V supply is done by U1300A and associated components. Under normal current-demand conditions, the output of U1300A is high and diode CR1282 is reverse biased. However, should the current through the current-sense resistor R1281 reach approximately 2 A, the voltage developed across R1281 will raise the voltage at pin 2 of U1300A (via divider R1282 and R1286) to a level equal to that at pin 3. This causes the output of U1300A to go low, forward biasing CR1282. This sinks base drive current away from Q1281 and lowers the output current in series-pass transistor Q1280.

−15 V Regulator

Operation of the −15 V Regulator, composed of three-terminal regulator U1330, operational amplifier U1270C, and their associated components, is similar to that of the +15 V Regulator with the following major changes. The control voltage at the three-terminal regulator's reference pin (pin 1) is established by the current through series-resistors R1333 and R1334. The reference pin is clamped by CR1332 at about −5.6 V should a failure in the sensing network occur. (Clamping also prevents latchup of the operational amplifier during start-up of the power supply.) Finally, the sensing divider formed by R1331 and R1332 is referenced to the +10 V reference instead of ground to enable sensing of negative voltage.

−8 V Regulator

Operation of the −8 V Regulator is similar to that of the +87 V and +42 V Regulators. Due to the lower operating voltages of the −8V Regulator the common-base transistor present in both the +87 V and the +42 V is not required. Current limiting in the −8 V supply occurs at about 480 mA.

−5 V Regulator

Operation of the −5 Volt Regulator is similar to that of the +5 V Regulator. Current limiting in the −5 V supply occurs at about 2 A.

+5 V Inverter Feedback

Operational amplifier U1371C and associated components are configured as a frequency-compensated voltage-sensing network. The circuit monitors the +5 V digital power supply line from the rectifiers and provides feedback to the Preregulator Control IC (U1030) via opto-isolator U1040 (both on diagram 9). The feedback is used to slightly vary the voltage-sensing characteristics of the Preregulator Control circuitry. The feedback (FB) signal slightly varies the voltage to the Inverter output transformer and holds the output of the 5 V secondary windings at an optimum level. Output levels of the other secondary windings are related to the +5 V_D level and are also held at their optimum values. This technique minimizes power losses in the series-pass transistors and increases regulator reliability.

Power-Up Delay

The Power-Up Delay circuit, composed of Q1370, Q1376, U1371D, and the associated components, ensures that the various regulated power supplies have time to reach their proper operating voltages before signaling the Microprocessor that the power supplies are up.

When power is first applied, a LINE UP signal from the Preregulator Control circuit goes HI, indicating that the power switch has been closed and that ample supply voltage is available for driving the Inverter transformer. The HI is applied to the base of Q1370, but since the collector is not properly biased yet, no transistor current will flow. As the Inverter begins to run, the various voltages from the secondary rectifiers begin coming up to their proper levels. A +2.5 V reference voltage is applied to operational amplifier U1371D pin 12 and forces the output high, biasing Q1376 on.

Before any of the Low-Voltage Regulators may function properly, the +10 V reference voltage must be established as previously described. When the +15 V Regulator turns on, current flows through Q1370, and pin 13 of U1371D is pulled above the +2.5 V reference through divider R1370 and R1372. The output of U1371D goes low, turning off Q1376.

When power to the instrument is turned off, the LINE UP signal goes LO (as explained in the Line Up Signal description). The falling LINE UP signal turns Q1370 off and drives the output of U1371D high. The output level from U1371D turns Q1376 on and pulls the PWR UP signal to the Microprocessor LO. This LO initiates the power-down sequence used to shut down the instrument in an orderly fashion. The delay between the time that the PWR UP signal goes LO and when the regulated power supplies fall below their normal operating levels provides ample time for the Microprocessor to complete the power-down sequence.

POWER DISTRIBUTION

Schematic diagrams 11 and 12 illustrate the power distribution of the instrument. The connections to the labeled boxes (representing the hybrids and ICs) show the power connections to each device, while connections to non-power lines are shown by the component and schematic number. Power supply decoupling is done with traditional LRC networks as shown on the diagrams.

Several intermediate supply voltages are generated by devices shown on diagrams 11 and 12. An approximate +32 volt supply for the A and B Sweeps is developed by emitter-follower Q700 and its associated components. Zener diodes VR125 and VR225 develop approximate +6.2 volt supplies for the CH 1 and CH 2 Preamps respectively, and zener diode VR2805 establishes an approximate −6.8 volt supply for U2800 and U2805.

INTERCONNECTIONS

Schematic diagram 13 illustrates the circuit board interconnections of the instrument. Connector numbers and cabling types are shown.

PERFORMANCE CHECK AND FUNCTIONAL VERIFICATION PROCEDURE

INTRODUCTION

This procedure is used to verify proper operation of instrument controls and to check the instrument's performance against the requirements listed in the "Specification" (Section 1). This procedure verifies instrument function and may be used to determine need for readjustment. These checks may also be used as an acceptance test and as a preliminary troubleshooting aid.

Removing the wrap-around cabinet is not necessary to perform this procedure. All checks are made using the operator accessible front- and rear-panel controls and connectors.

Within the procedure, steps to verify proper operation of an instrument control or function that is not specified in the "Specification" section begin with the word "VERIFY". These functions ARE NOT specifications and should not be interpreted as such. Steps to check performance specifications begin with the word "CHECK".

PREPARATION

Test equipment items 1 through 18 listed in Table 4-1 are required to perform this procedure. The specific pieces of equipment required to perform the checks within each section are listed at the beginning of that section. The item numbers in parenthesis next to each piece of equipment refer to the numbered equipment list of Table 4-1. Items 19 through 25 are used only for instrument calibration (see the Adjustment Procedure in Section 5).

Before performing this procedure, ensure that the LINE VOLTAGE SELECTOR switch is set for the ac power source being used (see "Preparation for Use" in Section 2). Connect the instrument to be checked and the test equipment to an appropriate power source. Turn the instrument on and ensure that no error message is displayed on the CRT. If an error message is present, have the instrument repaired or calibrated by a qualified service technician before performing this procedure.

The procedure is divided into sections to permit functional and performance verifications of individual sections of the instrument without performing the entire procedure. Perform all steps within a section, both in the sequence presented and in their entirety, to ensure that control settings are correct for the following step.

When performing partial procedures, the Initial Control Settings at the first of the section should be set up first; then make any changes noted at the first of the subsection to be performed. When performing the procedures in sequence, merely change those controls that have changed from the previous step.

NOTE

In order to see a channel's VOLTS/DIV setting, the channel must be selected using the VERTICAL MODE switches.

On instruments with Option 06 or 09 (CTT) installed: selecting Intensified, Alternate, or B Horizontal Mode will automatically enable the Counter/Timer/ Trigger option for precision Delay, Delta Time, and 1/Delta Time measurements. Several sections of the Performance Verification Procedure specify various delay settings for B Trigger in either the RUN AFTER DELAY, TRIGGERED AFTER DELAY, or TRIG Δ DELAY mode. Procedure steps involving delay settings that the CTT option will affect have alternate instructions listed.

Table 4-1
Test Equipment Required

Item and Description	Minimum Specification	Use	Example of Applicable Test Equipment
1. Variable Power Supply	Variable output voltage: 0 V to +16 V.	Check 50 Ω input overload switching.	TEKTRONIX PS 503A.
2. Leveled Sine-Wave Generator (Primary)	Frequency: 250 kHz to 250 MHz. Output: 0 V to 5 V. Reference frequency: 50 kHz.	Check CTT and Trigger.	TEKTRONIX SG 503.
3. Calibration Generator	Fast-rise, low aberration amplitudes: to 1 V. Rise time: 1 ns or less. Repetition rate: 1 kHz to 100 kHz. Precision amplitudes: 0.01 V to 50 V ±0.25%.	Signal source for gain and transient response.	TEKTRONIX PG 506. CG-5001
4. Leveled Sine-Wave Generator (Secondary)	Frequency: 245 kHz to 500 MHz. Output: 0.5 V to 4.0 V. Reference frequency: 50 kHz.	Check bandwidth and triggering.	TEKTRONIX SG 504 with Leveling head. SG503 (To 260MHz)
5. Function Generator	Repetition rate: 60 Hz to 1 MHz. Output to 15 V p-p.	Check triggers and coupling.	TEKTRONIX FG 501A. FG503
6. Time-Mark Generator	Markers: 2 ns to 5 s in a 1-2-5 sequence. Marker accuracy: ±0.1%. For CTT checks accuracy: ±0.00005%.	Check horizontal timing and CTT.	TEKTRONIX TG 501. CTT requires TG501 Option 01. CG-5001
7. Oscilloscope with P6137 10X Standard Accessory Probe	Bandwidth: 150/250 MHz. General Purpose.	Check power supply ripple and output signals. Troubleshooting.	TEKTRONIX 2445B/2455B. 2236/475 200MHz 100 MHz
8. T-Connector (2 required)	Impedance: 50 Ω. Connectors: BNC.	Signal interconnection.	TEKTRONIX Part Number 103-0030-00.
9. Precision BNC Cable	Impedance: 50 Ω. Connectors: BNC. Length: 36 in.	Signal interconnection.	TEKTRONIX Part Number 012-0482-00.
10. BNC Cable (4 required)	Impedance: 50 Ω. Connectors: BNC. Length: 43 in.	Signal interconnection.	TEKTRONIX Part Number 012-0057-01.
11. Dual-Input Coupler	Connectors: BNC female-to-dual-BNC male.	Signal interconnection.	TEKTRONIX Part Number 067-0525-02.
12. Termination (2 required)	Impedance: 50 Ω. Connectors: BNC.	Signal interconnection.	TEKTRONIX Part Number 011-0049-01.
13. Adapter	Subminiature probe-tip-to-BNC.	Signal interconnection.	TEKTRONIX Part Number 013-0195-00.
14. Adapter	BNC female-to-BNC female.	Signal interconnection.	TEKTRONIX Part Number 103-0028-00.
15. Adapter	Connectors: BNC female-to-dual banana.	Signal interconnection.	TEKTRONIX Part Number 103-0090-00.

Table 4-1 (cont)

Item and Description	Minimum Specification	Use	Example of Applicable Test Equipment
16. Attenuator	Attenuation factor: 2X. Impedance: 50 Ω. Connectors: BNC.	Signal attenuation.	TEKTRONIX Part Number 011-0069-02.
17. Attenuator	Attenuation factor: 5X. Impedance: 50 Ω. Connectors: BNC.	Signal attenuation.	TEKTRONIX Part Number 011-0060-02.
18. Attenuator	Attenuation factor: 10X. Impedance: 50 Ω. Connectors: BNC.	Signal attenuation.	TEKTRONIX Part Number 011-0059-02.
19. Digital Multimeter (DMM)	DC volts range to +20 V. Accuracy: ±0.2%.	Check power supplies and CALIBRATOR.	TEKTRONIX DM 502A.
20. Low-Capacitance Alignment Tool	Shaft length: 2 in.	Adjust variable resistors and capacitors.	TEKTRONIX Part Number 003-0675-00.
21. 1X Probe	Attenuation: 1X. Bandpass: <20 MHz.	Check power supply ripple.	TEKTRONIX P6101-01.
22. Normalizer	Input resistance: 1 MΩ. Input capacitance: 15 pf.	Check input capacitance.	TEKTRONIX Part Number 067-0537-00.
23. Tunnel Diode Pulser	Rise time: 125 ps or less.	Check transient response.	TEKTRONIX Part Number 067-0681-01.
24. Pulse Generator (2 required)	Frequency: 10 MHz. Pulse width: 50 ns. Pulse width accuracy: 5%. Positive trigger input, 1 V to 5 V into 50 Ohms. Positive trigger output, 1 V into 50 Ohms. Variable pulse duration.	CTT Checks.	TEKTRONIX PG502 Pulse Generator. PG508
25. Adapter (2 required)	Connectors: BNC male-to-dual-binding.	CTT Checks.	TEKTRONIX Part Number 103-0035-00.
26. Adapter	BNC-to-probe-tip.	Signal interconnection.	TEKTRONIX Part Number 013-0227-00.

VERTICAL

Equipment Required (see Table 4-1)

- Power Supply (Item 1)
- Primary Leveled Sine-Wave Generator (Item 2)
- Calibration Generator (Item 3)
- Secondary Leveled Sine-Wave Generator (Item 4)
- 10X Probe (supplied with 2445B/2455B) (Item 7)
- Precision 50 Ω BNC Cable (Item 9)
- 50 Ω BNC Cable (Item 10)
- Dual-Input Coupler (Item 11)
- 50 Ω BNC Termination (Item 12)
- Subminiature Probe Tip-to-BNC Adapter (Item 13)
- BNC Female-to-BNC Female Adapter (Item 14)
- BNC Female-to-Dual Banana Adapter (Item 15)
- 2X Attenuator (Item 16)
- 5X Attenuator (Item 17)
- 10X Attenuator (Item 18)
- 1X Probe (Item 21)
- BNC-to-probe-tip Adapter (Item 26)

INITIAL CONTROL SETTINGS

Control settings not listed do not affect the procedure.

a. Set:

NOTE

Select channels to set VOLTS/DIV.

VOLTS/DIV

CH 1 and CH 2	1 V
CH 1 and CH 2 VAR	In detent
CH 3 and CH 4	0.1V

VERTICAL MODE

CH 1	On
CH 2, CH 3, CH 4, ADD, and INVERT	Off
CHOP/ALT VERTICAL MODE	ALT
20 MHz BW LIMIT	Off

Input Coupling

CH 1 and CH 2	1 MΩ GND

HORIZONTAL

A SEC/DIV	10 ms (knob in)
SEC/DIV VAR	In detent
X10 MAG	Off
TRACE SEP	Fully CW

Delta

Δt and ΔV	Off (press and release until associated readout is off)
TRACKING	Off

TRIGGER

HOLDOFF	Fully CCW
LEVEL	Midrange
SLOPE	+ (plus)
A/B TRIG	SELECT A
MODE	AUTO LVL
SOURCE	VERT
COUPLING	DC

NOT TESTED

1. Verify CH 1 and CH 2, 50 Ω OVERLOAD protection.

a. Connect the Power Supply to the CH 1 OR X input connector via a 50 Ω BNC cable and a BNC female-to-dual banana adapter.

b. Using the CH 1 VERTICAL POSITION control, position the trace on the bottom horizontal graticule line.

c. Change CH 1 Input Coupling to 1 MΩ DC.

d. Turn the Power Supply on.

e. Adjust the Power Supply output level until the CH 1 trace rises to 1 division above the center graticule line (+5 V).

f. Change CH 1 Input Coupling to 50 Ω DC.

g. VERIFY—For a period of one minute, the readout display does not indicate any overload condition (50 Ω OVERLOAD).

h. Change the CH 1 VOLTS/DIV control to 5 V and the CH 1 Input Coupling to 1 MΩ DC.

i. Increase the Power Supply output level until the CH 1 trace rises to the center graticule line (+20 V).

CAUTION

To prevent damage to the input circuitry when in 50 Ω DC, the 20 V source must not be applied to the CH 1 OR X or CH 2 input connectors for longer than 20 seconds. If the automatic OVERLOAD switching does not occur within 20 seconds, turn the Power Supply off immediately.

j. Set the CH 1 Input Coupling to 50 Ω DC.

k. VERIFY—Within 20 seconds after CH 1 input coupling is set to 50 Ω DC, the readout display indicates "50 Ω OVERLOAD", the CH 1 Input Coupling changes to 1 MΩ GND automatically, and the trace returns to the bottom horizontal graticule line.

l. Turn the Power Supply Off.

m. Disconnect the Power Supply from CH 1 input.

n. Clear the OVERLOAD condition by pressing the upper CH 1 Input Coupling button.

o. VERIFY—The CH 1, 1 MΩ DC indicator is lit and the readout display no longer indicates 50 Ω OVERLOAD.

p. Set the VERTICAL MODE buttons to display Channel 2 and repeat parts a through o to verify 50 Ω OVERLOAD protection for Channel 2.

2. Check CH 1 and CH 2 Low-Frequency AC Coupling.

a. Set:

NOTE

Select channels to set VOLTS/DIV.

VOLTS/DIV

CH 1 and CH 2	100 mV

VERTICAL MODE

CH 1	On
CH 2	Off

Input Coupling

CH 1 and CH 2	1 MΩ GND

HORIZONTAL

A SEC/DIV	10 ms (knob in)

b. Connect the CALIBRATOR output signal to the CH 1 OR X input connector using a 1X probe. .4V

c. Position the ground-reference trace 2 divisions below the center horizontal graticule line.

d. Set the CH 1 Input Coupling to 1 MΩ DC.

e. CHECK—Displayed signal is vertically centered and has an amplitude of 3.88 to 4.12 divisions.

f. Set the Channel 1 Input Coupling to the upper 1 MΩ GND position.

g. Using the CH 1 POSITION control, align the trace with the center horizontal graticule line.

h. Set the Channel 1 Input Coupling to 1 MΩ AC.

i. CHECK—Displayed signal is a tilted square wave, 4.36 to 5.37 divisions in amplitude, vertically centered on the graticule.

j. Move the probe to the Channel 2 input connector.

k. Set the VERTICAL MODE buttons to deselect Channel 1 and display Channel 2.

NOTE

Instruments with TV OPTION 05 have a TV CLAMP feature that is enabled by pushing the upper CH 2 INPUT COUPLING button while in AC COUPLING. The letters TVC appear in the top right readout when this mode is selected. Push the lower CH 2 INPUT COUPLING button to return to normal AC coupling.

l. Repeat parts c through i for Channel 2.

m. Disconnect the test setup.

3. Check CH 1 and CH 2 VOLTS/DIV, CH 2 INVERT, ΔV and TRIGGER LEVEL Readout Accuracies, Variable VOLTS/DIV, Vertical Linearity, and ADD.

a. Set:

NOTE

Select channels to set VOLTS/DIV.

VOLTS/DIV

CH 1 and CH 2	2 mV

VERTICAL MODE

CH 1	On
CH 2	Off
CH 1 and CH 2	1 MΩ DC
BW LIMIT	On

HORIZONTAL

A SEC/DIV	1 ms (knob in)
ΔV	On (press and release for a ΔV readout)

TRIGGER

MODE	AUTO
COUPLING	DC

NOTE

The instrument must have had at least 20 minutes warmup prior to performing the following steps.

b. Momentarily press and hold both the CH 1 and CH 2 upper Input Coupling buttons until a moving dot display replaces the normal signal. This performs a DC Balance of CH 1 and CH 2 and the readout displays DC BALANCE IN PROGRESS.

c. When the signal and readout displays automatically return to normal, set the CH 1 and CH 2 Input Coupling to 1 MΩ DC.

d. Connect the Calibration Generator to the CH 1 OR X input connector via a 50 Ω BNC cable. Do not use a termination.

e. CHECK—CH 1 and CH 2 VOLTS/DIV, ΔV, and TRIGGER LEVEL readout accuracies as follows:

1. Set VOLTS/DIV control to the first position listed in Table 4-2.

2. Set the Calibration Generator STD AMPLITUDE output level to the corresponding Standard Amplitude Input Level in Table 4-2.

NOTE

To properly verify TRIGGER LEVEL Readout Accuracy, the Calibration Generator's STD AMPLITUDE output must have rising and falling transition times (10% to 90%) > 20 ns. No overshoot should appear on the waveform.

3. Verify that the generator output meets the requirements noted above.

4. Use the VERTICAL POSITION control to set the bottom of the signal 2 divisions below graticule center.

5. Rotate the Δ REF OR DLY POS control to align the reference cursor with the bottom of the waveform.

6. Rotate the Δ control to align the delta cursor with the top of the signal display.

7. CHECK—Vertical Deflection Accuracy (measured against the graticule) and ΔV Readout Accuracy are within the limits listed in Table 4-2.

8. Set the TRIGGER LEVEL control at the most positive voltage that produces a barely triggered, jittering display for each position (+ and −) of SLOPE.

9. CHECK—The A Trigger Level readings are within the limits given in the +Peak column of Table 4-2.

10. Set the TRIGGER LEVEL control at the most negative voltage that produces a barely triggered, jittering display for each position (+ and −) of SLOPE.

11. CHECK—The A Trigger Level readings are within the limits given in the −Peak column of Table 4-2.

12. Set the TRIGGER LEVEL for a stable display.

13. Pull the SEC/DIV knob out.

14. Set:

B TRIGGER

MODE	TRIG AFT DLY
SOURCE	VERT
COUPLING	DC
SLOPE	+

NOTE

On CTT instruments, rotate the Δ REF OR DLY POS control for the specified delay. As the control is rotated, the readout delay value will be followed by the word "SET". This value shows the approximate delay. A few seconds after control movement has stopped, the word "SET" will disappear and the readout delay value as measured by the CTT will appear. This is normal operation and not cause for concern.

15. Adjust Δ REF OR DLY POS control for a delay reading of 0.000 ms.

16. Set the TRIGGER LEVEL control to the most positive voltage that produces an intensified point on the waveform display for each position (+ and −) of SLOPE.

17. CHECK—The B Trigger Level readings are within the limits given in the +Peak column of Table 4-2.

18. Set the TRIGGER LEVEL control to the most negative voltage that produces an intensified point on the waveform display for each position (+ and −) of SLOPE.

19. CHECK—The B Trigger Level readings are within the limits given in the −Peak column of Table 4-2.

NOTE

On CTT instruments, repeat sections 16-19 for TRIG Δ DLY trigger mode using the +Peak and −Peak columns of Table 4-2.

20. Push the SEC/DIV knob in.

21. Change the VOLTS/DIV to the next position listed in Table 4-2.

22. Set the Calibration Generator to the corresponding signal amplitude setting.

23. Press and release the ΔV pushbutton to obtain the ΔV readout display.

24. Repeat subparts 4 through 23 of part e for each VOLTS/DIV setting listed in Table 4-2.

25. Set the TRIGGER COUPLING to NOISE REJ.

26. Set the CH 1 VOLTS/DIV to 50 mV.

27. Set the Calibration Generator STD AMPLITUDE output level to 0.2 V.

28. CHECK—Trigger Level Readout is within the limits given in Table 4-2 for NOISE REJ Coupling.

f. Return the TRIGGER COUPLING to DC.

g. Set the CH 1 VOLTS/DIV and the Calibration Generator output level to produce a vertical signal display 5 divisions in amplitude.

h. CHECK—Display amplitude reduces to 2 divisions or less when the VOLTS/DIV VAR control (of the channel under test) is rotated fully CCW. Return the VOLTS/DIV VAR control to its maximum CW (detent) position.

i. Set the Calibration Generator output level and VERTICAL POSITION controls for a 2-division display vertically centered on the graticule. Use the CH 1 VAR control if necessary to obtain the correct display amplitude.

j. Set the VERTICAL POSITION control to align the top edge of the display with the top graticule line.

k. CHECK—Signal display amplitude is 1.9 to 2.1 divisions.

l. Set the VERTICAL POSITION control to align the bottom edge of the signal display with the bottom graticule line.

m. CHECK—Signal display amplitude is 1.9 to 2.1 divisions.

Table 4-2
Accuracy Limits
CH 1, CH 2 INVERT, and Delta Volts Readouts

VOLTS/DIV Switch Setting CH 1 and CH 2	Standard Amplitude Input Level	Vertical Deflection Accuracy (±2% (in divisions)	Delta Volts Readout Accuracy (limits) 1.25% +.03 div	Limits of Trigger LEVEL Readout			
				DC Coupling		NOISE REJ Coupling	
				+Peak	−Peak	+Peak	−Peak
2 mV	10 mV	4.90 to 5.10	9.81 mV to 10.20 mV	8.0 mV to 12.0 mV	+1.7 mV to −1.7 mV		
5 mV	20 mV	3.92 to 4.08	19.6 mV to 20.4 mV	16.8 mV to 23.2 mV	+2.6 mV to −2.6 mV		
10 mV	50 mV	4.90 to 5.10	49.0 mV to 50.9 mV	44 mV to 56 mV	+4.5 mV to −4.5 mV		
20 mV	0.1 V	4.90 to 5.10	98.1 mV to 102.0 mV	89 mV to 111 mV	+8.0 mV to −8.0 mV		
50 mV	0.2 V	3.92 to 4.08	196 mV to 204 mV	178 mV to 222 mV	+16 mV to −16 mV	148 mV to 252 mV	+46 mV to −46 mV
100 mV	0.5 V	4.90 to 5.10	490 mV to 509 mV	0.450 V to 0.550 V	+0.035 V −0.035 V		
200 mV	1.0 V	4.90 to 5.10	0.981 V to 1.020 V	0.90 V to 1.10 V	+0.07 V to −0.07 V		
500 mV	2.0 V	3.92 to 4.08	1.96 V to 2.04 V	1.78 V 2.22 V	0.16 V to −0.16 V		
1.0 V	5.0 V	4.90 to 5.10	4.90 V to 5.09 V	4.50 V to 5.50 V	+0.35 V to −0.35 V		
2.0 V	10.0 V	4.90 to 5.10	9.81 V to 10.2 V	9.0 V to 11.0 V	+0.7 V to −0.7 V		
5.0 V	20.0 V	3.92 to 4.08	19.6 V to 20.4 V	17.8 V to 22.2 V	+1.6 V to −1.6 V		

n. Set the CH 1 and CH 2 Input Coupling to 50 Ω DC.

o. Connect the Calibration Generator to the CH 1 OR X input connector via a 50 Ω BNC cable. Do not use a termination.

p. Check CH 1 and CH 2 VOLTS/DIV 50 Ω Coupling accuracy as follows:

1. Set VOLTS/DIV control to the first position listed in Table 4-3.
2. Set the Calibration Generator STD AMPLITUDE output level to the corresponding Standard Amplitude Input Level in Table 4-3.
3. Use the VERTICAL POSITION control to set the bottom of the signal 2 divisions below graticule center.
4. CHECK—Vertical Deflection Accuracy (measured against the graticule) is within the limits listed in Table 4-3.
5. Change the VOLTS/DIV to the next position listed in Table 4-3.
6. Set the Calibration Generator to the corresponding signal amplitude setting.
7. Repeat subparts 3 through 6 of part p for each VOLTS/DIV setting listed in Table 4-3.
8. Set CH 1 and CH 2 Input Coupling to 1 MΩ DC.

q. Move the test signal to CH 2 and set the VERTICAL MODE controls to display CH 2.

r. Return the CH 1 VOLTS/DIV VAR control to the calibrated detent position.

s. Repeat parts e through p for CH 2.

t. Return the CH 2 VOLTS/DIV VAR control to the calibrated detent position.

u. Rotate the Δ REF OR DLY POS control CCW until the cursor stops moving.

v. CHECK—Cursor is aligned with the bottom graticule line within ±0.2 division.

w. Rotate the Δ control CW until the cursor stops moving.

x. CHECK—Cursor is aligned with the top graticule line within ±0.2 division. Push ΔV to turn off cursors.

y. Turn the INVERT function on, and obtain a 5-division signal as explained in part g.

z. VERIFY—A down-arrow symbol appears to the left of the CH 2 VOLTS/DIV readout.

aa. CHECK—Display amplitude is between 4.9 divisions and 5.1 divisions in amplitude (5 divisions ±2%). Turn the INVERT function off when finished.

bb. Connect a 5 V standard-amplitude signal from the Calibration Generator to the CH 1 OR X and CH 2 input connectors via a 50 Ω BNC cable and a Dual-Input Coupler.

cc. Set:

VOLTS/DIV

CH 1 and CH 2	2 V

VERTICAL MODE

CH 1 and CH 2	Off
ADD	On
VAR	In Detent

dd. CHECK—Vertical deflection amplitude is 4.9 to 5.1 divisions.

ee. VERIFY—A + (plus) symbol appears to the left of the CH 2 VOLTS/DIV readout.

ff. CHECK—Signal amplitude reduces to 0.2 division or less when CH 2 INVERT is on.

Table 4-3
Accuracy Limits
CH 1 and CH 2 VOLTS/DIV 50 Ω Coupling

VOLTS/DIV Setting CH 1 and CH 2	Standard Amplitude Input Level	Vertical Deflection Accuracy ±3% in Divisions
2 mV	20 mV	4.85 to 5.15
5 mV	50 mV	4.85 to 5.15
10 mV	0.1 V	4.85 to 5.15
20 mV	0.2 V	4.85 to 5.15
50 mV	0.5 V	4.85 to 5.15
100 mV	1.0 V	4.85 to 5.15
200 mV	2.0 V	4.85 to 5.15
500 mV	5.0 V	4.85 to 5.15
1.0 V	10.0 V	4.85 to 5.15
2.0 V[a]		
5.0 V[a]		

[a]**Not checked. Attempting to check would exceed Maximum Input Voltage.**

gg. Set:

VERTICAL MODE

CH 3	On
CH 1, CH 2, CH 4, ADD, and INVERT	Off

hh. Move the Dual-Input Coupler to the CH 3 and CH 4 input connectors.

ii. CHECK—VOLTS/DIV and TRIGGER LEVEL Readout accuracies for both setting-input level combinations listed in Table 4-4 as in subparts 4 through 23 of part e.

jj. Set the Calibration Generator output level and VERTICAL POSITION controls for a 2-division display vertically centered on the graticule.

kk. Set the VERTICAL POSITION control to align the top edge of the display with the top graticule line.

ll. CHECK—Signal display amplitude is 1.9 to 2.1 divisions.

mm. Set the VERTICAL POSITION control to align the bottom edge of the signal display with the bottom graticule line.

nn. CHECK—Signal display amplitude is 1.9 to 2.1 divisions.

oo. Set the VERTICAL MODE buttons to disable CH 3 and display CH 4.

pp. Repeat parts ii through nn for CH 4.

qq. Disconnect the test setup.

Table 4-4
CH 3 and CH 4 Accuracy Limits

VOLTS/DIV Switch Setting CH 3 and CH 4	Standard Amplitude Signal Input Level	Vertical Deflection Accuracy (±10% in divisions)	Trigger LEVEL Readout When Barely Triggered at the Indicated Peak	
			+Peak	−Peak
0.1 V	0.5 V	4.50 to 5.50	0.455 V to 0.545 V	±0.03 V
0.5 V	2.0 V	3.60 to 4.40	1.82 V to 2.18 V	±0.12 V

4. Check Channel 2 Delay.

a. Set:

VERTICAL

CH 1 and CH 2 MODE	On
CH 3 and CH 4 MODE	Off
20 MHz BW LIMIT	Off
CH 1 and CH 2 Input Coupling	50 Ω DC
CH 1 and CH 2 VOLTS/DIV	10 mV

HORIZONTAL

A SEC/DIV	1 μs (knob in)

TRIGGER

SOURCE	CH 1

b. Connect a 100 kHz, fast-rise, positive-going signal from the Calibration Generator to the CH 1 OR X and the CH 2 input connectors via a 50 Ω BNC cable, a 5X attenuator and a Dual-Input Coupler.

c. Set the output level of the Calibration Generator for an approximate 5-division, vertically-centered display for both channels.

d. Use either the CH 1 or CH 2 VAR control to match signal amplitude between both channels.

e. Set:

A SEC/DIV	10 ns (knob in)
X10 MAG	On

f. Use the Horizontal POSITION control to move the rising edges of the CH 1 and CH 2 displays to graticule center.

g. Pull the SEC/DIV knob out to activate the CH 2 DLY feature.

NOTE

If the readout displays "CH 2 DLY DISABLED" instead of "CH 2 DLY—TURN Δ" the delay matching feature has been disabled and the remainder of this subsection cannot be performed. In this case, proceed to subsection 5 below.

h. CHECK—Δ control will position the CH 2 display 0.5 division or more (500 ps) to either side of the CH 1 display.

i. Superimpose the rising edges of the pulses using the Δ control.

j. Turn X10 MAG off and push in the SEC/DIV knob.

k. Disconnect the test setup.

5. Check Vertical Bandwidth—All Channels.

a. Set:

HORIZONTAL

A SEC/DIV	50 μs (knob in)

TRIGGER

SOURCE	VERT

NOTE

Select channels to set VOLTS/DIV.

VOLTS/DIV

CH 1 and CH 2	20 mV
CH 3 and CH 4	0.1 V

VAR

CH 1 and CH 2	Calibrated (in detent)

VERTICAL MODE

CH 1	On
CH 2, CH 3, and CH 4	Off

Input Coupling

CH 1 and CH 2	50 Ω DC

b. Connect the output of the Primary Leveled Sine-Wave Generator to the CH 1 OR X input connector via a precision 50 Ω BNC cable and any combination of the 10X, 5X, or 2X Attenuators needed to reduce the signal amplitude to the level called out in the next step.

c. Set the generator output level for a 6-division display at the reference frequency, then change the generator output to 100 MHz for the 2445B or to 200 MHz for the 2455B.

d. CHECK—Signal display amplitude is 4.25 divisions or greater while sweeping the generator frequency from 100 MHz to 150 MHz for the 2445B or from 200 MHz to 250 MHz for the 2455B.

e. Set the VOLTS/DIV to 0.5 V and repeat parts c and d.

f. Set the VOLTS/DIV to 1 V and the generator output level for a 4-division display at the reference frequency, then change the generator frequency to 100 MHz for the 2445B or to 200 MHz for the 2455B.

g. CHECK—Signal display amplitude is 2.82 divisions or greater while sweeping the generator frequency from 100 MHz to 150 MHz for the 2445B or from 200 MHz to 250 MHz for the 2455B.

h. Move the signal to CH 2 input connector and set the VERTICAL MODE to disable CH 1 and display CH 2.

i. CHECK—Repeat parts c through g for CH 2.

j. Set the VERTICAL MODE to display CH 3 only.

k. Attach the standard-accessory 10X probe (supplied with the instrument) to the CH 3 input connector and the probe tip to the CALIBRATOR terminal.

l. Set the SEC/DIV (knob in) to 1 ms.

m. Adjust probe compensation for the best flat top on the square-wave signal display.

n. Disconnect the probe tip from the CALIBRATOR terminal. Remove the grabber tip from the probe, unscrew and remove the plastic barrel, and connect the probe to the output of the Primary Sine-Wave Generator via a BNC-to-probe-tip adapter.

o. Set the SEC/DIV to 50 μs (knob in).

p. Set the generator output for a 4-division display at the reference frequency, then change the generator frequency to 100 MHz for 2445B or to 200 MHz for the 2455B.

q. CHECK—Signal display amplitude is 2.82 divisions or greater while sweeping the generator frequency from 100 MHz to 150 MHz for the 2445B or from 200 MHz to 250 MHz for the 2455B.

r. Move the signal to CH 4 and set the VERTICAL MODE to display CH 4 only.

s. CHECK—Repeat parts k through q for CH 4.

t. Disconnect the test setup.

6. Check Common Mode Rejection Ratio (CMRR).

a. Set:

NOTE

Select channels to set VOLTS/DIV.

VOLTS/DIV

CH 1 and CH 2	10 mV
CH 1 and CH 2 VAR	In detent

VERTICAL

CH 1, ADD, and INVERT	On
CH 2, CH 3, and CH 4	Off

Input Coupling

CH 1 and CH 2	50 Ω DC

HORIZONTAL

A SEC/DIV	50 μs (knob in)

TRIGGER

SOURCE	CH 1

b. Connect a reference frequency signal from the Primary Leveled Sine-Wave Generator to the CH 1 OR X and CH 2 input connectors via a 50 Ω BNC cable, a 5X attenuator, and a Dual-Input Coupler.

c. Set the generator output level for an 8-division display of the reference signal on CH 1.

d. Adjust either the CH 1 VAR control or the CH 2 VAR control for a minimum ADD display amplitude while leaving the other control in the calibrated detent (whichever provides the best CMRR).

e. Set the generator frequency to 50 MHz.

f. Set the A SEC/DIV to 20 ns.

g. CHECK—ADD display amplitude is 0.4 division or less (discount trace width).

h. Set ADD and INVERT Off and rotate the CH 1 and CH 2 VAR controls CW to their calibrated detent positions.

i. Disconnect the test setup.

7. Check Channel Isolation.

a. Set:

VERTICAL MODE	
CH 1, CH 2, CH 3, and CH 4	On
CHOP/ALT	ALT
Input Coupling	
CH 1 and CH 2	50 Ω DC
VOLTS/DIV	
CH 1 and CH 2	100 mV
CH 3 and CH 4	0.1 V
TRIGGER	
SOURCE	CH 1
Horizontal	
A SEC/DIV	20 ns (knob in)

b. Connect the Primary Leveled Sine-Wave Generator to the CH 1 OR X input connector via a 50 Ω BNC cable.

c. Set the generator frequency to 100 MHz and adjust the output level for an 8-division display.

d. CHECK—Amplitude of each trace other than CH 1 is 0.08 division or less (discount trace width).

e. Move the signal to the CH 2 input connector and change the TRIGGER SOURCE to CH 2.

f. CHECK—Amplitude of each trace other than CH 2 is 0.08 division or less (discount trace width).

g. Add a 50 Ω BNC termination to the BNC cable and move the signal to CH 3.

h. Set the TRIGGER SOURCE to CH 3 and adjust the generator output for a signal display amplitude of 8 divisions.

i. CHECK—Amplitude of each trace other than CH 3 is 0.16 division or less (discount trace width).

j. Move the signal to CH 4 input connector and set TRIGGER SOURCE to CH 4.

k. CHECK—Amplitude of each trace other than CH 4 is 0.16 division or less (discount trace width).

l. Connect the generator to the CH 1 OR X input connector. Remove the 50 Ω termination.

m. Set the TRIGGER SOURCE to CH 1.

n. Set the generator output frequency to 150 MHz for the 2445B or to 250 MHz for the 2455B. Set the generator output level for an 8-division display.

o. CHECK—Amplitude of each trace other than CH 1 is 0.16 division or less (discount trace width).

p. Move the signal to the CH 2 input connector and set the TRIGGER SOURCE to CH 2.

q. CHECK—Amplitude of each trace other than CH 2 is 0.16 division or less (discount trace width).

r. Disconnect the test setup.

8. Set CH 1 and CH 2 DC Balance.

NOTE

For an accurate DC Balance setting, the instrument MUST be allowed to warm up for 20 minutes before performing the following steps.

a. Press both the CH 1 and CH 2 upper Input Coupling buttons for approximately 1 second, then release them.

b. VERIFY—DC BALANCE IN PROGRESS is in top line of readout. A flashing dot may be displayed. The display returns to normal in approximately 15 seconds.

c. VERIFY—There is less than 0.2 division + 0.5 mV vertical trace shift between adjacent settings of the CH 1 and CH 2 VOLTS/DIV as they are rotated through each of their positions.

d. VERIFY—There is less than 0.2 division vertical trace shift between the CH 3 and CH 4 VOLTS/DIV settings.

e. VERIFY—There is less than 1.0 division vertical trace shift as the CH 1 and CH 2 VOLTS/DIV VAR controls are rotated fully CCW.

f. VERIFY—There is less than 0.5 division vertical trace shift when the INVERT button is pressed.

g. Return the VERTICAL VAR controls to their detent positions and turn the CH 2 INVERT function off.

9. Check CH 2 SIGNAL OUT and Cascaded Operation.

a. Set:

NOTE

Temporarily select CH 2 to set CH 2 VOLTS/DIV.

VERTICAL

CH 1	On
CH 2, CH 3, and CH 4	Off
20 MHz BW LIMIT	On

VOLTS/DIV

CH 1 and CH 2	2 mV

Input Coupling

CH 1 and CH 2	1 MΩ DC

HORIZONTAL

A SEC/DIV	200 μs (knob in)

TRIGGER

MODE	AUTO LVL
SOURCE	VERT
COUPLING	HF REJ

b. Connect a 1 kHz, 1 mV standard-amplitude signal from the Calibration Generator to the CH 2 input connector via a 50-Ω BNC cable.

c. Connect the CH 2 signal from the rear-panel CH 2 SIGNAL OUT connector to the CH 1 OR X input connector via a precision 50-Ω BNC cable.

d. CHECK—Display amplitude is 4.5 to 5.5 divisions (discount trace width).

e. Set CH 2 Input Coupling to GND and align the CH 1 trace with the center graticule line.

f. CHECK—Trace noise is 1.2 divisions peak to peak or less.

g. Set CH 1 Input Coupling to GND and align the trace with the center graticule line.

h. Return CH 1 Input Coupling to 1 MΩ DC.

i. Set the CH 1 VOLTS/DIV to 10 mV.

j. CHECK—Trace is within 2 divisions of the ground reference set above (discount trace width).

k. Disconnect the test setup.

10. Check BW Limit Operation.

a. Set:

VERTICAL

CH 1	Off
CH 2	On
BW LIMIT	On
CH 2 VOLTS/DIV	10 mV
CH 2 Input Coupling	50 Ω

HORIZONTAL

A SEC/DIV	50 μs (knob in)

b. Connect the Primary Leveled Sine-Wave Generator output to the CH 2 input connector via a precision 50 Ω BNC cable.

c. Set the generator frequency to 50 kHz and adjust the output level for a 6-division display on the CRT.

d. Gradually increase the generator output frequency until the display amplitude decreases to 4.24 divisions.

e. CHECK—Generator frequency is between 13 MHz to 24 MHz.

f. Turn BW LIMIT off.

g. Disconnect the test setup.

TRIGGERING

Equipment Required (see Table 4-1)

- Primary Leveled Sine-Wave Generator (Item 2)
- Secondary Leveled Sine-Wave Generator (Item 4)
- Function Generator (Item 5)
- 10X Probe (supplied with 2445B/2455B) (Item 7)
- T-Connector (2 required) (Item 8)
- Precision 50 Ω BNC Cable (Item 9)
- 50 Ω BNC Cable (4 required) (Item 10)
- Dual-Input Coupler (Item 11)
- 50 Ω BNC Termination (2 required) (Item 12)
- Subminitiare Probe Tip-to-BNC Adapter (Item 13)
- 10X Attenuator (Item 18)
- Adapter (Item 25) (2 Required)

INITIAL CONTROL SETTINGS

Control settings not listed do not affect the procedure.

a. Set:

NOTE

Select channels to set VOLTS/DIV.

VOLTS/DIV

CH 1	100 mV
CH 2	500 mV
CH 1 and CH 2 VAR	In detent
CH 3 and CH 4	0.5 V

VERTICAL MODE

CH 1	On
CH 2, CH 3, CH 4, ADD and INVERT	Off
CHOP/ALT	ALT
20 MHz BW LIMIT	Off

Input Coupling

CH 1 and CH 2	1 MΩ DC

HORIZONTAL

A SEC/DIV	2 μs (knob in)
SEC/DIV VAR	In detent
X10 MAG	Off
TRACE SEP	Fully CW

Delta

Δt and ΔV	Off (press and release until associated readout is off)
TRACKING	Off

Trigger

HOLDOFF	B ENDS A (fully CW)
LEVEL	Midrange
SLOPE	+ (plus)
MODE	AUTO LVL
SOURCE	VERT
COUPLING	DC

1. Check A and B Triggers.

NOTE

The Trigger Level Readout Accuracies are checked in the Vertical Performance Checks.

a. Refer to Table 4-5 to determine what the A Trigger requirements are and at what frequencies various checks are made.

b. Using a 50-Ω BNC cable, connect one of the following test generators to the CH 1 input connector. Select the generator that produces the proper frequency range for the conditions being tested as called out in Tables 4-5 and 4-6. When using the leveled sine-wave generators (items 2 and 3 below), the output must be terminated into 50Ω (either the 50-Ω input coupling or a 50-Ω termination may be used).

1. Function Generator (60 Hz, 30 kHz and 80 kHz).

2. Primary Leveled Sine-Wave Generator (50 MHz and 250 MHz).

3. Secondary Leveled Sine-Wave Generator (300 MHz and 500 MHz).

NOTE

To obtain signal amplitudes less than 1 division, first set the signal for either 4, 5, or 10 times the specified amplitude; then reduce the amplitude by a factor of 4, 5, or 10 by increasing the VOLTS/DIV settings as necessary.

c. For each combination listed in the table, set the generator Test Frequency and the oscilloscope TRIGGER COUPLING as indicated, performing the following steps to verify the Triggering levels in each setup.

d. Set the VOLTS/DIV and the generator output level to obtain the test signal amplitude indicated for the particular combination being tested. When checking channel 1 and channel 2 500 MHz triggering, also adjust the VOLTS/DIV VAR for the correct input level.

e. Set the A SEC/DIV and the X10 MAG to obtain a well-defined display of the test signal.

NOTE

Normally, unless trigger sensitivity is very close to the specified limits, it is sufficient to check each of the indicated frequency-coupling combinations listed in the table in Channel 1 only; checks for Channels 2, 3 and 4 need only be done in DC COUPLING (to verify signal path).

f. CHECK—For a stable triggered display (unless otherwise indicated) for each of the Test Frequency-TRIGGER COUPLING combinations listed in Table 4-5. When testing the 150-MHz 2445B or 250-MHz 2455B triggering, check that trigger jitter is < 100 ps (0.2 division at 5 ns/div with X10 MAG), with 5 divisions of signal and TRIGGER LEVEL adjusted for minimum jitter.

Table 4-5
CH 1 or CH 2 Triggering Conditions

Test Frequency	Minimum Vertical Display Levels at Which Triggering Should Occur				
	TRIGGER COUPLING				
	DC	NOISE REJ	HF REJ	LF REF	AC
60 Hz	a	a	a	No Trigger, Freeruns	0.35 Div
30 kHz	a	a	0.5 Div	a	a
80 kHz	a	a	a	0.5 Div	a
50 MHz	0.35 Div	1.2 Div	No Trigger, Freeruns at 1.2 Div	0.5 Div	0.35 Div
250 MHz 2445B only	1.5 Div	4.5 Div	No Trigger, Freeruns at 3.0 Div	1.5 Div	1.5 Div
300 MHz 2445B only	1.0 Div	3.0 Div	No Trigger, Freeruns at 3.0 Div	1.0 Div	1.0 Div
500 MHz 2455B only	1.5 Div	4.5 Div	a	1.5 Div	1.5 Div

[a]Not necessary to check.

g. Press the ADD button to select the function and press the CH 1 button to turn off the CH 1 display.

h. Repeat the DC TRIGGER COUPLING tests of Table 4-5 while in the ADD mode, adding 0.5 DIV to the 2455B 300- and 500-MHz amplitudes and adding 0.5 DIV to the 2445B 150- and 250-MHz amplitudes.

i. Move the signal to the CH 2 input connector and repeat step h for CH 2.

j. Press the CH 2 button to select the channel and press the ADD button to turn off the ADD display.

k. Repeat the DC TRIGGER COUPLING tests of Table 4-5 while in CH 2 mode.

l. If trigger sensitivity is close to the specified limits given in steps c through k above, test all of the frequency-coupling combinations given in Table 4-5 for CH 2.

m. Move the test signal to CH 3 and CH 4 in turn and repeat parts c through f using Table 4-6.

Table 4-6
CH 3 or CH 4 Triggering Conditions

Test Frequency	Minimum Vertical Display Levels at Which Triggering Should Occur				
	TRIGGER COUPLING				
	DC	NOISE	HF REJ	LF REF	AC
60 Hz	a	a	a	No Trigger, Freeruns	0.18 Div
30 kHz	a	a	0.25 Div	a	a
80 kHz	a	a	a	0.25 Div	a
50 MHz	0.18 Div	0.6 Div	No Trigger, Freeruns at 0.6 Div	0.18 Div	0.18 Div
300 MHz	0.5 Div	1.5 Div	No Trigger, Freeruns at 1.5 Div	0.5 Div	0.5 Div
500 MHz	0.75 Div	2.25 Div	a	0.75 Div	0.75 div

[a]Not necessary to check.

n. Set:

TRIGGER

MODE	AUTO
LEVEL	Fully clockwise

o. Pull the SEC/DIV knob out and set the B SEC/DIV 1 setting (CW) faster than the A SEC/DIV setting, then push the SEC/DIV knob back in.

NOTE

On CTT instruments, rotate the Δ REF OR DLY POS control for the specified delay. As the control is rotated, the readout delay value will be followed by the word "SET". This value shows the approximate delay. A few seconds after control movement has stopped, the word "SET" will disappear and the readout delay value as measured by the CTT will appear. This is normal operation and not cause for concern.

p. Verify that the CRT readout displays DLY and not Δt. If Δt is displayed, press the Δt button in and release it to select the DLY function. When DLY is displayed, rotate the Δ REF OR DLY POS control CCW until the readout display indicates zero delay. (The display will indicate DLY?, which is normal.)

q. Press the A/B TRIG button to select the B TRIGGER.

r. Set B TRIGGER MODE to TRIG AFT DLY and adjust TRIGGER LEVEL for a stable signal display.

s. Repeat parts a through m for B TRIGGER, changing the SEC/DIV and X10 MAG as required to maintain a well-defined display.

t. Disconnect the test setup.

2. Check Composite Triggering.

a. Set:

VERTICAL

CH 1, CH 2, CH 3, and CH 4	On
ADD	Off
CHOP/ALT	ALT

Input Coupling

CH 1 and CH 2	1 MΩ DC

TRIGGER

A/B TRIG	A
MODE	NORM
SOURCE	CH 1
COUPLING	DC

HORIZONTAL

A SEC/DIV	10 μs (knob in)

b. Connect the Primary Leveled Sine Wave Generator to the CH 1 and CH 2 inputs via a 50-Ω BNC cable and a Dual-Input Coupler.

c. Set the Primary Leveled Sine Wave Generator Generator for a 50-kHz, 1.35-division display for CH 1 and CH 2.

d. Connect the Primary Leveled Sine-Wave Generator to the CH 3 input connector using a 50-Ω BNC cable and a 50-Ω termination.

e. Set TRIGGER SOURCE to CH 3.

f. Set the generator output level for a 0.7-division display at the reference frequency (50 kHz).

g. Connect the Secondary Leveled Sine-Wave Generator to the CH 4 input using a BNC cable and a 50-Ω termination.

h. Set TRIGGER SOURCE to CH 4.

i. Set the generator output level for a 0.7-division display at the reference frequency.

j. Set TRIGGER SOURCE to VERT.

k. CHECK—Display will trigger as the TRIGGER LEVEL control is rotated through its range.

l. Pull the SEC/DIV knob out, rotate it to 5 μs, and push it back in.

m. Press the A/B TRIG button and set the B TRIGGER MODE to TRIG AFT DLY.

n. Set B TRIGGER SOURCE to VERT.

NOTE

On CTT Instruments, rotate the Δ REF OR DLY POS control for the specified delay. As the control is rotated, the readout delay value will be followed by the word "SET". This value shows the approximate delay. A few seconds after control movement has stopped, the word "SET" will disappear and the readout delay value as measured by the CTT will appear. This is normal operation and not cause for concern.

o. Rotate the Δ REF OR DLY POS control CCW until the delay readout indicates DLY? 0.00 μs.

p. CHECK—Display will trigger as the TRIGGER LEVEL control is rotated through its range.

q. Rotate the SEC/DIV knob back to 10 μs (knob in).

r. Disconnect the test setup.

3. Check Trigger Noise Rejection—All Channels.

a. Set:

NOTE

Select channels to set VOLTS/DIV.

VOLTS/DIV

CH 1	5 mV
CH 2	50 mV
CH 3 and CH 4	0.1 V

VERTICAL

CH 1	On
CH 2, 3, 4	Off

Input Coupling

CH 1 and CH 2	1 MΩ DC

HORIZONTAL

A SEC/DIV	10 μs (knob in)

TRIGGER

MODE	AUTO LVL
SOURCE	VERT

b. Connect the Function Generator to the CH 1 input via a 50-Ω BNC cable and a 10X attenuator.

c. Set the Function Generator output frequency and level for a 50-kHz, 4-division display.

d. Set the CH 1 VOLTS/DIV to 50 mV.

e. Set the TRIGGER COUPLING to NOISE REJ.

f. CHECK—Display will not trigger (freeruns).

g. Pull the SEC/DIV knob out, rotate it to 5 μs and push it back in.

h. Press the A/B TRIG button to select the B TRIGGER.

i. Set the TRIGGER MODE to B TRIG AFT DLY.

j. Set TRIGGER COUPLING to NOISE REJ.

k. CHECK—Display will not trigger for any setting of the LEVEL control.

l. Rotate the SEC/DIV back to 10 μs (knob in).

m. Move the input signal to CH 2, CH 3, and CH 4 in turn, selecting each channel as the display source. Repeat parts f through k for each channel.

4. Check Slope Selection and Verify Line Trigger.

a. Set:

VERTICAL

CH 1 VOLTS/DIV	5 V
CH 1 Input Coupling	1 MΩ DC
CH 1	On
CH 2, 3, 4	Off

HORIZONTAL

A SEC/DIV	2 ms (knob in)
X10 MAG	Off

TRIGGER

MODE	AUTO
SOURCE	LINE
COUPLING	AC

CAUTION

In the next part, DO NOT connect the probe ground lead to the ac power source.

b. Attach the 10X probe to the CH 1 OR X input connector and connect the probe tip to the ac power source.

c. CHECK—Display can be triggered in both the + (plus) and − (minus) positions of the SLOPE switch using the TRIGGER LEVEL control and that the displayed slope agrees with the selected slope.

d. CHECK—Display phase shifts slightly as the TRIGGER COUPLING is changed from AC to DC.

e. Disconnect the test setup.

HORIZONTAL

Equipment Required (see Table 4-1)

- Primary Leveled Sine-Wave Generator (Item 2)
- Calibration Generator (Item 3)
- Time-Mark Generator (Item 6)
- T-Connector (Item 8)
- Precision 50-Ω BNC Cable (Item 9)
- BNC Cable 50-Ω (2 required) (Item 10)
- Dual Input Coupler (Item 11)
- Pulse Generator (Item 24)

INITIAL CONTROL SETTINGS.

Control settings not listed do not affect the procedure.

a. Set:

NOTE

Select channels to set VOLTS/DIV.

VERTICAL

CH 1 and CH 2	
VOLTS/DIV	0.5 V
CH 1 VAR	In detent
CH 3 and CH 4	
VOLTS/DIV	0.1 V
CH 1	On
CH 2, CH 3, CH 4, ADD, and INVERT	Off
CHOP/ALT	ALT
20 MHz BW LIMIT	Off
CH 1 and CH 2 Input Coupling	50 Ω DC

HORIZONTAL

A SEC/DIV	200 ns (knob in)
SEC/DIV VAR	In detent
X10 MAG	Off
TRACE SEP	Fully CW

Delta

ΔV and Δt	Off (press and release until associated readout is off)
TRACKING	Off

TRIGGER

HOLDOFF	B ENDS A
LEVEL	Midrange
SLOPE	+ (plus)
MODE	AUTO LVL
SOURCE	VERT
COUPLING	DC

1. Check Horizontal Display Modes (A, A INTEN, ALT, and B).

a. Use a 50-Ω BNC cable to connect 200 ns time markers from the Time-Mark Generator to the CH 1 OR X input connector.

b. Adjust the TRIGGER LEVEL control as necessary for a stable signal display.

c. Pull the SEC/DIV knob out and set the B TRIGGER MODE to RUN AFT DLY.

NOTE

On CTT instruments, rotate the Δ REF OR DLY POS control for the specified delay. As the control is rotated, the readout delay value will be followed by the word "SET". This value shows the approximate delay. A few seconds after control movement has stopped, the word "SET" will disappear and the readout delay value as measured by the CTT will appear. This is normal operation and not cause for concern.

d. Set the Δ REF OR DLY POS control for a DLY readout of approx 1000 ns.

e. VERIFY—An intensified zone appears on the displayed signal near graticule center. The INTENSITY control may need adjustment.

f. Rotate the Δ REF OR DLY POS control to center the intensified zone on one of the time markers near graticule center.

g. Set the B SEC/DIV to 50 ns (knob out).

h. Rotate the TRACE SEP control CCW to separate the the A and B sweep displays.

i. CHECK—The B sweep is displayed with the A sweep.

j. Push the SEC/DIV knob in.

k. CHECK—Only the B sweep is displayed.

2. Check A and B Timing, A Cursor Accuracies, and A Cursor Range.

a. Set:

HORIZONTAL

A SEC/DIV	10 ns (knob in)
TRACE SEP	Fully CW

Delta

Δt	On (press and release for Δt display)

b. Select 10 ns time markers from the Time-Mark Generator and adjust the TRIGGER LEVEL control for a stable display.

c. Use the Horizontal POSITION control to align the 2nd time marker with the 2nd vertical graticule line (2nd from the left edge of the display).

NOTE

The 2 ns and the 5 ns time markers are sinusoidal. Use either the rising or falling zero-crossings as alignment points.

d. Align the Δ REF OR DLY POS cursor with the 2nd time marker and align the Δ cursor with the 10th time marker.

e. CHECK—The A Sweep timing and cursor readout accuracies are within limits given in Tables 4-7 and 4-8.

NOTE

If the 2nd and 10th time markers are within 0.06 division of the 2nd and 10th vertical graticule lines for unmagnified sweeps and within 0.1 division for magnified sweeps, the sweep timing accuracy is conservatively within limits. When the timing accuracy is checked at each sweep speed, note any SEC/DIV setting at which the timing error exceeds the 0.06-division limit. Check these sweep speeds against the major-division time-interval limits given in Table 4-8.

NOTE

For SEC/DIV setting of 10 ns, the time-marker period is greater than 1 division when the sweep is magnified. At 1 ns per division (SEC/DIV setting of 10 ns with X10 MAG), input the signal through a dual input coupler to CH1 and CH2. Select CH1, CH2, and CH2 INVERT. Set the CH 1 and CH 2 VOLTS/DIV settings for a 6 division signal. Center the waveforms. Check for 4 cycles between the 2nd and 10th vertical graticule lines (within 0.1 division) at the intersections of the waveforms.

f. Repeat parts c, d and e for each A SEC/DIV-time marker combination given in Table 4-7 for both unmagnified and magnified sweeps.

Table 4-7
Settings for A and B Timing Accuracy Checks and A and B Cursor Accuracy Limits

±.86%

SEC/ DIV Setting	Unmagnified		X10	
	Time Markers	Limits of Δt Cursor Readout	Time Markers	Limits of Δt Cursor Readout
10 ns	10 ns	79.30 ns to 80.70 ns	2 ns 2 Div/cycle	7.89 ns to 8.11 ns ±.86 (4 cycles)
20 ns	20 ns	158.60 ns to 161.40 ns	2 ns	15.78 ns to 16.22 ns ±.86
50 ns	50 ns	396.5 ns to 403.5 ns	5 ns	39.45 ns to 40.55 ns
100 ns	0.1 μs	793.0 ns to 807.0 μs	10 ns	78.90 ns to 81.10 ns
200 ns	0.2 μs	1586.0 ns to 1614.0 ns	20 ns	157.80 ns to 162.20 ns
500 ns	0.5 μs	3965 ns to 4035 ns	50 ns	394.5 ns to 405.5 ns
1 μs	1 μs	7.930 μs to 8.070 μs	0.1 μs	789.0 ns to 811.0 ns
2 μs	2 μs	15.860 μs to 16.140 μs	0.2 μs	1578.0 ns to 1622.0 ns
5 μs	5 μs	39.65 μs to 40.35 μs	0.5 μs	3945 ns to 4055 ns
10 μs	10 μs	79.30 μs to 80.70 μs	1 μs	7.890 μs to 8.110 μs
20 μs	20 μs	158.60 μs to 161.40 μs	2 μs	15.780 μs to 16.220 μs
50 μs	50 μs	396.5 μs to 403.5 μs	5 μs	39.45 μs to 40.55 μs
100 μs	100 μs	793.0 μs to 807.0 μs	10 μs	78.90 μs to 81.10 μs
200 μs	200 μs	1586.0 μs to 1614.0 μs	20 μs	157.80 μs to 162.20 μs
500 μs	500 μs	3965 μs to 4035 μs	50 μs	394.5 μs to 405.5 μs
1 ms	1 ms	7.930 ms to 8.070 ms	100 μs	789.0 μs to 811.0 μs
2 ms	2 ms	15.860 ms to 16.140 ms	200 μs	1578.0 μs to 1622.0 μs
5 ms	5 ms	39.65 ms to 40.35 ms	500 μs	3945 μs to 4055 μs
10 ms	10 ms	79.30 ms to 80.70 ms	1 ms	7.890 ms to 8.110 ms
20 ms	20 ms	158.60 ms to 161.40 ms	2 ms	15.780 ms to 16.220 ms
50 ms	50 ms	396.5 ms to 403.5 ms	5 ms	39.45 ms to 40.55 ms
A SEC/DIV ONLY	(B Sweep does not have these sweep speeds)			
100 ms	0.1 s	793.0 ms to 807.0 ms	10 ms	78.90 ms to 81.10 ms
200 ms	0.2 s	1578.0 ms to 1622.0 ms	20 ms	157.00 ms to 163.00 ms
500 ms	0.5 s	3945 ms to 4055 ms	50 ms	392.5 ms to 407.5 ms

Table 4-8
Horizontal Timing Accuracy Checked Against the Graticule

	Over Any									
	1 Div	**2 Div**	**3 Div**	**4 Div**	**5 Div**	**6 Div**	**7 Div**	**8 Div**	**9 Div**	**10 Div**
Time-marker Accuracy (X10 MAG Off)	±0.07 Div	±0.07 Div	±0.08 Div	±0.09 Div	±0.10 Div	±0.10 Div	±0.11 Div	±0.12 Div	±0.12 Div	±0.13 Div
Time-marker Accuracy (X10 MAG On) (Exclude first 0.5 division of sweep rate)	±0.07 Div	±0.08 Div	±0.1 Div	±0.11 Div	±0.12 Div	±0.13 Div	±0.14 Div	±0.16 Div	±0.17 Div	±0.18 Div
As Measured Against These Time-Marker Pairs (X10 MAG Off Only)	1-2	1-3	1-4	1-5	1-6	1-7	1-8	1-9	1-10	1-11
	2-3	2-4	2-5	2-6	2-7	2-8	2-9	2-10	2-11	
	3-4	3-5	3-6	3-7	3-8	3-9	3-10	3-11		
	4-5	4-6	4-7	4-8	4-9	4-10	4-11			
	5-6	5-7	5-8	5-9	5-10	5-11				
	6-7	6-8	6-9	6-10	6-11					
	7-8	7-9	7-10	7-11						
	8-9	8-10	8-11							
	9-10	9-11								
	10-11									

g. Rotate the Δ REF OR DLY POS control CCW until the cursor stops moving.

h. CHECK—Δ REF OR DLY POS cursor aligns with the 1st graticule line within 0.2 division.

i. Rotate the Δ control CW until the cursor stops moving.

j. CHECK—Δ cursor aligns with the 11th graticule line within 0.2 division.

k. Set the A SEC/DIV to 10 ns.

l. Rotate the Δ REF OR DLY POS and the Δ controls to precisely superimpose the cursors near the 2nd graticule line.

m. CHECK—Δt readout indicates a difference of 0.30 ns or less.

n. Rotate the Δ REF OR DLY POS and the Δ controls to precisely superimpose the cursors near the 10th graticule line.

o. CHECK—Δt readout indicates a difference of 0.30 ns or less.

p. Set:

HORIZONTAL

A SEC/DIV	20 ns
B SEC/DIV	10 ns (knob in)
X10 MAG	Off

TRIGGER

B TRIGGER MODE	RUN AFT DLY

Delta

Δt	On (Push Δt for Δt cursors)

NOTE

To easily maintain the A SWP and B SWP difference while checking timing and cursor accuracy, use the following method:

1. Set TRACK/INDEP off. Pull B SEC/DIV knob out.

2. Push and hold the TRACK/INDEP button while pushing the B SEC/DIV knob in. This will lock the A and B sweeps together with A sweep one speed slower than B sweep.

3. Press Δt button to obtain Δt cursors.

q. CHECK—The B sweep timing and cursor accuracy as in parts b through f, making sure that the A SEC/DIV is set slower than the B SEC/DIV.

3. Check Delta Time Accuracy using the Delayed Sweep.

a. Set:

HORIZONTAL

A SEC/DIV	20 ns
B SEC/DIV	10 ns (knob out)
X10 MAG	On
TRACK/INDEP	Off

Delta

Δt	Off (DLY readout)

TRIGGER

MODE	AUTO LVL
SOURCE	VERT
COUPLING	DC
SLOPE	+ (plus)
LEVEL	As required for a stable display
B TRIG MODE	RUN AFT DLY

NOTE

Certain time marks from the TG 501 (and other Time-Mark Generators) will vary in width and may be displaced in time. This will happen in a repeatable sequence and is caused by the loading and interaction of the 2, 5, and 10 dividers. This is most noticeable with 10 ns, 20 ns, and 50 ns markers. The following procedure will use the above markers to set up the proper references but the 5 ns markers will be used to make the actual measurement. Close inspection of apparent jitter or mistrigger of the time marks will show the trigger point to be stable with the apparent jitter to be variable with unique combinations of trigger holdoff and sweep speed. This is normal behavior with this type of signal and is not an instrument defect.

It is not necessary to count the number of marks given in the tables. Switching to 10 ns, 20 ns, or 50 ns markers as required and then to 5 ns will show the proper 5 ns mark to be used.

For CTT instruments, use the following setup of the CTT while performing the Delta Time performance check.

1. Press the MEASURE button to enter MENU mode.

2. Select COUNTER ("4") from menu.

3. Select PERIOD ("2") from menu.

The CTT period readout will appear on the left side of the upper line of readout. The word "SET" will appear next to the readout delay value. This denotes the indirect measurement mode of Delta Time, simulating a non-CTT scope.

b. Set the Time-Mark Generator for 20-ns markers. Adjust the Vertical VOLTS/DIV as required for a display of 3 to 6 divisions.

c. Adjust the Δ REF OR DLY POS control for a readout display of DLY 21.25 ns.

d. Adjust the Horizontal POSITION control CW until the trace stops moving, then CCW to display the leading edge of the 2nd B sweep time marker near the graticule center. This becomes the reference point for the following procedure. Set the Time-Mark Generator to 5 ns and adjust the Vertical VOLTS/DIV and Trigger LEVEL as required.

e. Press and release the Δt button to obtain the Δt display. Push in the SEC/DIV knob for B SWP only. Rotate the Δ control for a readout display of Δt −20.0 ns. If the time marks are not superimposed, adjust the Δ control to do so.

f. CHECK—Δt readout is within the limits listed in Table 4-9 for the 1st 5 ns time marker; then check that the 5th through 37th time markers are within the given limits as the Δ control is rotated CW to superimpose the time marker on the reference time marker.

NOTE

Correct time marks to superimpose on the reference marker can be easily found by noting the Delta Time Readout.

g. Set:

HORIZONTAL

A SEC/DIV	50 ns
B SEC/DIV	10 ns (knob out)
X10 MAG	ON

Delta

Δt	Off (DLY readout)

Table 4-9
Delta Time Display Accuracy

Time-Marker Period and A SEC/DIV Switch Setting	B SEC/DIV Switch Setting	Marker Superimposed using the Δ (Delta) Control	Delta Time Readout Accuracy Limits
20 ns	1 ns[a]	1st	−19.55 ns to −20.45 ns
		5th	−0.40 ns to 0.40 ns
		9th	19.55 ns to 20.45 ns
		13th	39.50 ns to 40.50 ns
		17th	59.45 ns to 60.55 ns
		21st	79.40 ns to 80.60 ns
		25th	99.30 ns to 100.70 ns
		29th	119.25 ns to 120.75 ns
		33rd	139.20 ns to 140.80 ns
		37th	159.15 ns to 160.85 ns
50 ns	1 ns[a]	1st	−49.2 ns to −50.8 ns
		21st	49.2 ns to 50.8 ns
		91st	398.1 ns to 401.9 ns

[a]5 ns with X10 MAG on (button in).

h. Set the Time-Mark Generator for 50-ns time markers and adjust the Δ REF OR DLY POS control for a readout display of DLY 53.2 ns.

i. Position the leading edge of the 1st B-sweep time marker near graticule center using the Horizontal POSITION control. Set the Time-Mark Generator to 5 ns and adjust the Vertical VOLTS/DIV and Trigger LEVEL as required.

j. Press and release the Δt button to obtain a Δt display. Push in the SEC/DIV knob for B sweep only. Adjust the Δ control for a readout display of Δt −50.00 ns. If the time markers are not superimposed, adjust the Δ control to do so.

k. CHECK—Δt readout is within the limits listed in Table 4-9 for the 1st 5 ns time marker; then check that the 21st and 91st time markers are within the given limits as the Δ control is rotated CW to superimpose each time marker on the reference time marker.

l. Set:

Delta

TRACK/INDEP	TRACK

HORIZONTAL

A SEC/DIV	100 ns
B SEC/DIV	10 ns (knob out)
X10 MAG	On

m. Select 0.1-μs time markers from the Time-Mark Generator.

n. Adjust the Δ and Δ REF OR DLY POS controls for a Δt readout display of 800.0 ns.

o. Adjust the Horizontal POSITION control to align the leading edge of the 2nd time marker on the A sweep with the 2nd vertical graticule line.

p. Rotate the TRACE SEP control CCW to separate the traces.

q. Adjust the Δ REF OR DLY POS control to intensify the 2nd and 10th time markers (of the A sweep) and display the leading edges of the displayed B-sweep time markers in the center area of the graticule.

r. VERIFY—The horizontal distance between the leading edges of the B-sweep time markers is within the conservative guideline listed in Table 4-10. If this guideline is met, accuracy between each marker is ensured, and the following CHECK step need not be performed.

NOTE

To easily maintain the A SWP and B SWP difference while testing Delta Time, use the following method:

1. Starting with the 1-μs test in Table 4-10 (X10 MAG Off), turn TRACKING off.
2. Press and hold the TRACKING button, then push the SEC/DIV knob in. This will lock the sweeps together at that difference.
3. Pull the SEC/DIV knob out.

The fastest sweep speed at which the X100 difference is maintained is with an A SEC/DIV of 1 μs and a B SEC/DIV of 10 ns, after which only the A-sweep speed will change with the SEC/DIV knob. Push TRACKING to unlock this setup.

s. CHECK—The horizontal distance between the leading edges of the B-sweep time markers is within the specified limits given in Table 4-10. The limit given is for separation between the 2nd and 10th marker; however, separation between the 2nd marker and each succeeding marker should also be checked, calculating the limits from the specification as listed at the top of the table.

t. Repeat part r (and s if necessary) for each combination of A SEC/DIV, B SEC/DIV, and X10 MAG settings listed in Table 4-10. The Δt readout should be set to indicate eight times the A SEC/DIV setting. At the slowest sweep speeds, the B SEC/DIV knob can be pushed in (in B Sweep only) to increase the display repetition rate.

Table 4-10
Delayed Sweep Delta Time Accuracy

		Displayed Separation of Delayed Time Markers (for 2nd and 10th markers)	
A SEC/DIV and TIME Markers	**B SEC/DIV as Displayed on Readout**	**Conservative Guideline (divisions)**	**Specified Limit: (0.3% time) interval +0.1% of full scale-divisions + 200 ps**
0.1 μs	1 ns[a]	2.4	3.4
0.2 μs	2 ns[a]	2.4	3.4
0.5 μs	5 ns[a]	2.4	3.4
1 μs	10 ns[b]	2.4	3.4
2 μs	20 ns	2.4	3.4
5 μs	50 ns	2.4	3.4
10 μs	100 ns	2.4	3.4
20 μs	200 ns	2.4	3.4
50 μs	500 ns	2.4	3.4
0.1 ms	1 μs	2.4	3.4
0.2 ms	2 μs	2.4	3.4
0.5 ms	5 μs	2.4	3.4
1 ms	10 μs	2.4	3.4
2 ms	20 μs	2.4	3.4
5 ms	50 μs	2.4	3.4
10 ms	100 μs	2.4	3.4
20 ms	200 μs	2.4	3.4
50 ms	500 μs	2.4	3.4
0.1 s	1 ms	2.4	3.4
0.2 s	2 ms	6.4	7.4
0.5 s	5 ms	6.4	7.4

[a]X10 MAG On.

[b]For remainder of table, turn X10 MAG Off.

PARAMETRIC MEASUREMENTS CHECK

Initial Control Settings.

Control settings not listed do not affect the procedure.

VERTICAL MODE

CH 1	On
CH 2, 3, 4	Off

Input Coupling

CH 1	50 Ω DC

1. Check Timing Accuracy

NOTE

All Parametric timing measurements are derived from the same timing ramps as the period measurements. Verification of the period measurements provides verification of all timing measurements.

a. Connect Time Mark generator to CH 1 OR X input of the oscilloscope under test.

b. For each entry in Table 4-11:

1. Set Time Mark generator as indicated.
2. Press MEASURE.
3. Select FREQ from menu.
4. Verify resulting period measurement is within limits shown in Table 4-11.

NOTE

If the 50 ns period is out of limits shown on Table 4—11, perform step 2 (50 ns Timing Accuracy Verification) below.

c. Disconnect Time Mark generator.

2. 50 ns Timing Accuracy Verification

NOTE

Some Time Mark generators have jitter at the 50 ns setting which may produce an erroneous period reading. Use the following procedure to verify the 50 ns period measurement.

Table 4-11
Parametric Measurement Period Checks

Time Mark Setting	Minimum Period	Maximum Period	Time Mark Setting	Minimum Period	Maximum Period
2 ns	1.49 ns	2.51 ns	20 µs	19.90 µs	20.10 µs
5 ns	4.48 ns	5.52 ns	50 µs	49.75 µs	50.25 µs
10 ns	9.45 ns	10.55 ns	100 µs	99.50 µs	100.5 µs
20 ns	19.40 ns	20.40 ns	200 µs	199.0 µs	201.0 µs
50 ns	49.25 ns	50.75 ns[a]	500 µs	497.5 µs	502.5 µs
100 ns	99.0 ns	101.0 ns	1 ms	995.0 µs	1.005 µs
200 ns	198.5 ns	201.5 ns	2 ms	1.990 ms	2.010 ms
500 ns	497.0 ns	503.0 ns	5 ms	4.975 ms	5.025 ms
1 µs	994.5 µs	1.005 µs	10 ms	9.950 ms	10.05 ms
2 µs	1.989 µs	2.011 µs	20 ms	19.90 ms	20.10 ms
5 µs	4.975 µs	5.025 µs	50 ms	49.75 ms	50.25 ms[b]
10 µs	9.950 µs	10.05 µs	100 ms	99.50 ms	100.5 ms[b]

[a]If the 50 ns setting is not within the limits given, perform step 2 (50 ns Timing Accuracy Verification).
[b]For this setting, change MINFREQ to 10 Hz.

NOTE

This procedure need only be performed if the 50 ns reading from step 1 above was outside the limits listed in Table 4-11.

a. Connect Primary leveled sine-wave generator (item 2) to CH 1 OR X input of the oscilloscope under test and the test oscilloscope using a T-connector.

b. Set frequency for 20 MHz.

c. Adjust generator output amplitude for at least a 200 mV peak- peak display on the test oscilloscope.

d. Using the counter in the test oscilloscope, measure period of signal.

e. Press MEASURE then select FREQ on the oscilloscope under test.

f. Verify that the oscilloscope under test reads a period that is within 0.5% + 0.5 ns of the value measured by the counter on the test oscilloscope.

3. Verify Positive and Negative Peak Volts Measurements

a. Set CH 1 OR X input coupling to 1 MΩ.

b. Set CH 1 VOLTS/DIV to 50 mV.

c. Set A SEC/DIV to 500 μs.

d. Connect the + fast rise output of the Calibration Generator to the CH 1 OR X input via a 50-Ω BNC cable.

e. Adjust Calibration Generator amplitude for a 4 division 1 kHz display.

f. Measure VOLTS by pressing MEASURE and then selecting VOLTS.

g. CHECK—POS-PK reading is 0.0 mV ± 5 mV.

h. Connect the − fast rise output of the Calibration Generator to the CH 1 OR X input via a 50-Ω BNC cable.

i. Repeats steps e and f for − fast rise connected to CH 1.

j. CHECK—NEG-PK reading is 0.0 mV ± 5 mV.

k. Disconnect fast rise Generator.

4. Verify Average and Peak-Peak Volts Measurements

a. Connect standard-amplitude calibration Generator to CH 1 OR X input via a BNC T-Connector (item 8) and a 50-Ω cable.

b. For each entry in Table 4-12:

1. Measure VOLTS by pressing MEASURE and then selecting VOLTS.

2. Verify PK-PK reading is within limits specified.

3. Connect the BNC T-Connector via a 50-Ω cable and BNC to dual banana adapter to the Digital Multimeter (item 19).

4. Select appropriate DMM voltage range and note voltage reading.

5. Verify AVG reading is within limits specified.

NOTE

To insure accurate VOLT measurements it is necessary to disconnect the DMM input from the BNC T-Connector at the standard-amplitude Generator output PRIOR to selecting a VOLTS measurement. Re-connect meter when VOLTS measurements are completed.

c. Disconnect calibration generator from CH 1 OR X input and connect to CH 2 OR Y input.

d. Select only CH 2 for display.

e. Repeat step b for CH 2.

f. Disconnect test setup.

Table 4-12
Parametric Measurement Volts Checks

Calibration Generator Setting	Min[a] PK-PK	Max[a] PK-PK	AVG[a]
20 mV	14 mV	26 mV	Within ±(5% of DM501A reading + 5.6 mV)
50 mV	43 mV	57 mV	Within ±(5% of DM501A reading + 5.6 mV)
0.1 V	90 mV	110 mV	Within ±(5% of DM501A reading + 5.6 mV)
0.2 V	185 mV	215 mV	Within ±(5% of DM501A reading + 6.5 mV)
0.5 V	470 mV	530 mV	Within ±(5% of DM501A reading + 6.5 mV)
1 V	0.945 V	1.055 V	Within ±(5% of DM501A reading + 6.5 mV)
2 V	1.89 V	2.10 V	Within ±(5% of DM501A reading + 15 mV)
5 V	4.74 V	5.25 V	Within ±(5% of DM501A reading + 15 mV)
10 V	9.49 V	10.50 V	Within ±(5% of DM501A reading + 15 mV)
20 V	19.0 V	21.0 V	Within ±(5% of DM501A reading + 100 mV)
50 V	47.5 V	52.5 V	Within ±(5% of DM501A reading + 100 mV)

[a]Disconnect DMM prior to selecting VOLTS measurement.

COUNTER/TIMER/TRIGGER CHECKS

This section contains the portion of the Option 06 (Counter/Timer/Trigger) performance check procedure that directly affects operation of the horizontal timing modes. If your instrument does not contain this option, continue with the Horizontal checks.

Test equipment listed in Table 4-1 is required to perform this procedure. To assure accurate measurements, it is important that test equipment used for making these checks meet or exceed the specifications described in Table 4-1 for CTT checks.

INITIAL CONTROL SETTINGS

Control settings not listed do not affect the procedure.

NOTE

Select channels to set VOLTS/DIV.

VERTICAL

CH 1 and CH 2	
VOLTS/DIV	500 mV
CH 1 and CH 2 VAR	In detent
CH 3 and CH 4	
VOLTS/DIV	0.1 V
CH 1	On
CH 2, CH 3, CH 4,	
ADD, and INVERT	Off
CHOP/ALT	ALT
20 MHz BW LIMIT	Off
CH 1 and CH 2	
Input Coupling	50 Ω DC

HORIZONTAL

A SEC/DIV	10 ns (knob in)
SEC/DIV VAR	In detent
X10 MAG	Off
TRACKING	Off
TRACE SEP	Fully CW

Delta

Δt and ΔV	Off (press and release until associated readout is Off)

TRIGGER

HOLDOFF	Fully CCW
A and B LEVEL	Midrange
SLOPE	+ (plus)
A MODE	AUTO LVL
SOURCE	VERT
COUPLING	DC
B MODE	RUN AFT DLY

CTT and WR Options

MENU Functions	OFF

1. Check Maximum Input Frequency at Minimum Sensitivity

a. Connect the leveled sinewave generator's output via a 50-Ω cable to the CH 1 input connector.

b. Set generator to produce a 150-MHz, 4-division display.

c. Press the MEASURE button to enter MENU mode.

d. Select COUNTER ("4") from menu.

e. Select FREQ ("1") from menu.

f. Press the upper Trigger MODE button to reinitialize the auto-trigger level.

g. CHECK—Reading is between 149 MHz and 151 MHz and is stable.

2. Check Minimum Sensitivity at 50 MHz

a. Set the generator to produce a 50.0-MHz, 1.3-division display.

b. Press the upper Trigger MODE button to reinitialize the auto-trigger level.

c. CHECK—Reading is between 49.9 MHz and 50.1 MHz and is stable.

d. Disconnect the test equipment from the instrument.

3. Check Frequency Accuracy

a. Connect the time-mark generator output via a 50-Ω cable to the CH 1 input connector.

b. Set the generator to produce 10-ns time markers four divisions in amplitude using CH 1 VOLTS/DIV and VAR VOLTS/DIV.

c. Press the upper Trigger MODE button to reinitialize the auto-trigger level.

d. CHECK—Reading is between 99.9995 MHz and 100.0005 MHz.

4. Check Minimum Input Frequency

a. Set the time-mark generator to produce 2-s time markers.

b. Set:

CH 1 VOLTS/DIV	100 mV
A SEC/DIV	50 ms (knob in)
A TRIGGER MODE	NORM

c. Adjust the A Trigger LEVEL control for a stable trigger.

d. CHECK—Reading is between 499.9975 mHz and 500.0025 mHz.

e. Disconnect the test equipment from the instrument.

5. Check Delay Time

a. Set:

CH 1 VOLTS/DIV	500 mV
CH 1 Input Coupling	GND
A SEC/DIV	20 ns (knob in)
A TRIGGER MODE	AUTO

b. Connect the output of the time-mark generator via a 50-Ω cable to the positive trigger input of the pulse generator.

c. Connect the output of the pulse generator via a 50-Ω cable to the CH 1 input connector.

d. Set the time-mark generator to produce 20-ns time markers.

e. Set the pulse generator to produce a positive 5-ns pulse when externally triggered.

f. Adjust the CH 1 POSITION control to center the CH 1 display.

g. Set the CH 1 Input Coupling to 50 Ω DC.

h. Adjust the pulse generator to produce a 5-division peak-to-peak display, centered about ground.

i. Push INIT @50%.

j. Pull out the SEC/DIV knob.

k. Press the A/B TRIG button.

l. Set the B Trigger:

SLOPE	+ (plus)
MODE	TRIG AFT DLY
SOURCE	VERT
COUPLING	DC

m. Adjust the B Trigger LEVEL for a readout of 0.00 V.

n. Turn the Δ REF OR DLY POS control counterclockwise until the intensified zone stops moving to the left.

o. CHECK—Reading is either 59.5 ns to 60.5 ns or 69.5 ns to 70.5 ns.

6. Check Delta Time Accuracy

a. Press MEASURE button.

b. Select <MORE> ("8") from menu.

c. Select CONFIGURE ("5") from menu.

d. Select RESOLUTION ("4") from menu.

e. Select 10 ps ("4") from menu.

f. Set the A AND B SEC/DIV to 1 μs (knob out).

g. Press A/B TRIG to access the B TRIGGER controls.

h. Press the lower Trigger MODE button to enter TRIG AFT DLY mode.

i. Set the time-mark generator to produce 1-μs time markers.

j. Set the pulse generator to produce a positive 0.5-μs pulse when externally triggered.

k. Press and release the Δt button until the Delta Time readout appears.

l. Turn the Δ control to intensify the rising edge of the second square wave.

m. Turn the Δ REF OR DLY POS control to intensify the rising edge of the second square wave.

n. CHECK—That the averaged Δt reading is between +0.00005 μs and −0.00005 μs.

o. Turn the Δ control to intensify the rising edge of the eleventh square wave.

p. CHECK—Averaged Δt reading is between 8.99990 μs and 9.00010μs.

q. Set the A AND B SEC/DIV to 100 μs (knob out).

r. Set the time-mark generator to produce 0.1-ms time markers.

s. Set the pulse generator to produce a positive 50-μs pulse when externally triggered.

t. Turn the Δ control to intensify the rising edge of the eleventh square wave.

u. Turn the Δ REF OR DLY POS control to intensify the rising edge of the second square wave.

v. CHECK—Reading is between +899.996 μs and +900.004 μs.

w. Press MEASURE button.

x. Select <MORE> ("8") from menu.

y. Select CONFIGURE ("5") from menu.

z. Select RESOLUTION ("4") from menu.

aa. Select AUTO ("1") from menu.

7. Verify Delay-By-Events

a. Set the A SEC/DIV to 100 μs (knob in).

b. Set the A Trigger SLOPE to − (minus).

c. Press the Δt button until the Δt display disappears.

d. Press the MEASURE button.

e. Select <MORE> ("8") from menu.

f. Select DLY-BY-EVENTS ("1") from menu.

g. Select B-SWP ("5") from menu.

h. Select ATRG-STRT ("2") from menu.

i. Select DLY-BY-B ("3") from menu.

j. Select RUN ("8") from menu.

k. Pull out the SEC/DIV knob.

l. Use the Δ REF OR DLY POS and the Δ controls to set the number of delaying events to 1.

m. VERIFY—that the intensified zone moves to each succeeding rising edge as the delaying event count is changed to 2, 3, 4, and 5.

8. Check Logic Trigger

a. Set the A AND B SEC/DIV to 20 ns (knob out).

b. Set the time-mark generator to produce 0.1 μs time markers.

c. Set the pulse generator to produce a positive 5-ns pulse when externally triggered.

d. Set the B Trigger MODE to TRIG AFT DLY.

e. Set the B Trigger SOURCE to CH 1.

f. Press the MEASURE button.

g. Select <MORE> ("8") from menu.

h. Select LOGIC-TRIGGER ("4") from menu.

i. Select A:A-AND-B ("1") from menu.

j. Push in the SEC/DIV knob.

k. Adjust the B Trigger LEVEL for a readout of 0.00 V.

l. Press the A/B TRIG button to illuminate an A Trigger MODE indicator.

m. Adjust the A Trigger LEVEL for a readout of 1.00 V.

n. Set the CH 1 Input Coupling to GND.

o. Turn the CH 1 POSITION control to align the trace with the center horizontal graticule line; do not readjust the CH 1 POSITION control during the remainder of this step.

p. Set the CH 1 Input Coupling to 50 Ω DC.

q. Set X10 MAG on.

r. Turn the Horizontal POSITION control to align the rising edge of the first displayed signal with the intersection of the second vertical graticule and the center horizontal graticule lines.

s. Set the pulse generator to produce a 2-ns pulse when externally triggered.

t. Increase the duration of the pulse until a stable display is obtained.

u. CHECK—Width of the pulse measured at the center horizontal graticule line is less than 4 ns.

v. Set X10 MAG off.

w. Press the upper Trigger MODE button.

x. Press the lower Trigger MODE button.

y. Press the upper Trigger MODE button.

z. Disconnect the test equipment from the instrument.

9. Verify Trigger Delta Delay

a. Connect the leveled sinewave generator's output via a 50-Ω cable to the CH 1 input connector. Set the A SEC/DIV to 10 μs. Set the Horizontal POSITION to midrange.

b. Set the generator for a 50-kHz, 6-division display.

c. Press the Trigger SLOPE button to illuminate the + SLOPE indicator.

d. Press the MEASURE button to enter MENU mode.

e. Select COUNTER ("4") from menu.

f. Select PERIOD ("2") from menu.

g. Press the upper Trigger MODE button to reinitialize the auto-trigger level.

h. Turn the SEC/DIV to 5 μs.

i. Pull out the SEC/DIV knob.

j. Press the A/B TRIG button for B Trigger MODE. Set B Trigger MODE to RUN AFTER DELAY.

k. Adjust the Δ REF OR DLY POS control for a delay of 5.00 μs.

l. Press the lower Trigger MODE button once.

m. Press the SLOPE button to select + SLOPE if necessary.

n. Press the lower Trigger MODE button once to select TRIG Δ DLY.

o. Press the Trigger SLOPE button to illuminate the – SLOPE.

p. Adjust the Δ control for a Δt reading of approximately 0.00 μs. The word **"SET"** will appear while making the adjustment.

q. VERIFY—There are two intensified zones on the displayed waveform.

r. VERIFY—The intensified zone moves on the falling edge of the waveform while adjusting the Trigger LEVEL control.

s. Press the lower Trigger MODE button to select TRIG AFT DLY.

t. VERIFY—The intensified zone moves on the rising edge of the waveform while adjusting the Trigger LEVEL control.

u. Disconnect the test equipment from the instrument.

HORIZONTAL (cont)

4. Check Delay Jitter.

a. Set:

TRACKING	Off
A SEC/DIV	1 ms
B SEC/DIV	500 ns (knob out)
B TRIG	RUN AFT DLY

b. Connect Time-Mark Generator to the CH 1 OR X input. Select 1-ms time markers from the Time-Mark Generator.

c. Align the intensified zones with the 10th time marker using the Δ REF OR DLY POS and Δ controls. Superimpose the zones to obtain a Δt readout display of 0.000 ms.

d. Push in the SEC/DIV knob and adjust TRACE SEP to separate the traces.

e. CHECK—For 0.8 divisions or less of horizontal jitter on the rising edge of both time markers.

5. Check SEC/DIV VAR Range and Accuracy.

a. Set:

A SEC/DIV	10 ms (knob in)
SEC/DIV VAR	In detent
Δt	Off (press and release to eliminate Δt readout)
HOLDOFF	B ENDS A

b. Select 10-ms time markers from the Time-Mark Generator and adjust the Time-Mark Generator variable timing control for exactly 1 time marker per division. Note the variable timing % error on the Time-Mark Generator.

c. Adjust the SEC/DIV VAR control for a sweep-speed readout (on bottom line of readout) of 20 ms and adjust the Time-Mark Generator variable timing control for exactly 2 time markers per division.

d. CHECK—The Time-Mark Generator variable timing % of error has changed 2% or less from the reading noted in part b.

e. Adjust the SEC/DIV VAR control fully CCW.

f. CHECK—Sweep speed readout displays 30.0 ms.

g. Set the Time-Mark Generator variable timing control for exactly 3 time markers per division.

h. CHECK—The Time-Mark Generator variable timing % of error has changed 2% or less from the reading noted in part b. Set SEC/DIV VAR to detent.

i. Set:

SEC/DIV VAR	CW (in detent)
A SEC/DIV	50 ms
B SEC/DIV	10 ms (knob in)
Δt	Off (DLY readout)
B TRIGGER MODE	RUN AFT DLY
Δ REF OR DLY POS	Zero delay

NOTE

On CTT instruments, rotate the Δ REF OR DLY POS control for the specified delay. As the control is rotated, the readout delay value will be followed by the word "SET", denoting the indirect measurement mode. A few seconds after control movement has stopped, the word "SET" will disappear and the readout will display a direct measurement from the CTT.

j. Repeat parts b through h for the B Sweep.

k. Disconnect the test setup.

6. Check X-Axis Gain.

a. Set:

NOTE

Select channels to set VOLTS/DIV.

VERTICAL

CH 1 and CH 2 VOLTS/DIV	10 mV
CH 2	On
CH 1, CH 3, CH 4, ADD, and BW LIMIT	Off
CH 1 Input Coupling	1 MΩ DC
CH 2 Input Coupling	1 MΩ GND

HORIZONTAL

SEC/DIV VAR	CW (in detent)
SEC/DIV	X-Y (knob in)

b. Connect a 50-mV standard-amplitude signal from the Calibration Generator to the CH 1 OR X input connector via a 50-Ω BNC cable.

c. CHECK—Signal display amplitude is 4.9 to 5.1 horizontal divisions.

d. Disconnect the test setup.

7. Check X-Axis Bandwidth.

a. Set the CH 1 Input Coupling to 50 Ω DC.

b. Connect a 50-kHz signal from the Primary Leveled Sine-Wave Generator to the CH 1 OR X input connector via a precision 50-Ω BNC cable.

c. Set the generator output for a 6-division horizontal display.

d. Change the generator frequency to 3 MHz.

e. CHECK—Signal display is greater than 4.2 horizontal divisions.

8. Check X-Y Phase Differential.

a. Set the Primary Leveled Sine-Wave Generator for a 1-MHz, 6-division horizontal display.

b. Set the CH 2 VERTICAL MODE off. CH 1 displays automatically.

c. Use the CH 1 VERTICAL POSITION control to vertically center the display on the graticule.

d. CHECK—Ellipse opening is 0.1 division or less, measured horizontally.

e. Set the CH 2 VERTICAL MODE on.

f. Set the generator for a 2-MHz, 6-division horizontal display.

g. Set the CH 2 VERTICAL MODE off.

h. CHECK—Ellipse opening is 0.3 division or less, measured horizontally.

i. Set the CH 2 VERTICAL MODE on.

9. Check X-Axis Low-Frequency Linearity.

a. Set the Primary Leveled Sine-Wave Generator and the CH 1 POSITION control for a 50-kHz, 2-division horizontal display centered on the graticule.

b. Use the CH 1 POSITION control to align the left edge of the signal with the left side vertical graticule line.

c. CHECK—Signal display is 1.8 to 2.2 divisions, measured horizontally.

d. Use the CH 1 POSITION control to position the right edge of the signal on the right side vertical graticule line.

e. CHECK—Signal display is 1.8 to 2.2 divisions, measured horizontally.

f. Disconnect the test setup.

CALIBRATOR, EXTERNAL Z-AXIS AND GATE OUTPUTS

Equipment Required (see Table 4-1)

- Calibration Generator (Item 3)
- Time-Mark Generator (Item 6)
- Oscilloscope with 10X Probe (Item 7)
- 50 Ω BNC T-Connector (Item 8)
- 50 Ω BNC Cables (2 required) (Item 10)

INITIAL CONTROL SETTINGS

Control settings not listed do not affect the procedure.

a. Set:

VERTICAL

CH 1 and CH 2	On
CH 3, CH 4, ADD, and INVERT	Off
CHOP/ALT	CHOP
20 MHz BW LIMIT	Off
CH 1 VOLTS/DIV	10 mV
CH 2 VOLTS/DIV	500 mV
CH 1 and CH 2 VAR	In detent
CH 1 Input Coupling	1 MΩ DC
CH 2 Input Coupling	50 Ω DC

HORIZONTAL

A SEC/DIV	1 ms (knob in)
SEC/DIV VAR	In detent
X10 MAG	Off

Delta

ΔV and Δt	Off (press and release until associated readout is Off)

TRIGGER

HOLDOFF	B ENDS A (fully CW)
LEVEL	Midrange
SLOPE	+ (plus)
MODE	AUTO LVL
SOURCE	CH 1
COUPLING	DC

1. Check CALIBRATOR Repetition Rate.

NOTE

Refer to the Adjustment Procedure to check the accuracy of the CALIBRATOR output levels.

a. Connect a 10X probe from the CALIBRATOR terminal to the CH 1 OR X input connector.

b. Connect 1-ms time markers from the Time-Mark Generator to the CH 2 input connector via a 50-Ω BNC cable.

c. Adjust the CH 2 VOLTS/DIV for several divisions of marker display.

d. CHECK—Horizontal drift for any time marker is 1 division or less per second (10 seconds or more for 1 marker to drift 10 horizontal divisions).

e. Set the CH 2 VERTICAL MODE off.

f. CHECK—1 cycle is displayed per 2 horizontal divisions for each A SEC/DIV setting from 100 ms to 100 ns.

g. Disconnect the test setup.

2. Check External Z-Axis Operation.

a. Set:

INTENSITY	Fully clockwise
A SEC/DIV	1ms
CH 1 VOLTS/DIV	500 mV

b. Connect a 1-kHz, 2-V standard-amplitude signal from the Calibration Generator to the CH 1 OR X input connector and the rear-panel EXT Z-AXIS input connector using a 50-Ω BNC T-Connector and two 50-Ω BNC cables.

c. CHECK—The positive portion of the 4-division signal display is blanked out.

d. Disconnect the test setup and adjust the CRT INTENSITY as desired.

3. Check A and B GATE Outputs and Verify TRIGGER HOLDOFF.

a. Set:

A SEC/DIV	100 μs
B SEC/DIV	50 μs (knob in)
Δt	Off (DLY readout)
TRIGGER MODE	AUTO
TRIGGER HOLDOFF	Minimum (CCW)
Δ REF OR DLY POS	Zero DLY readout

NOTE

On CTT instruments, rotate the Δ REF OR DLY POS control for the specified delay. As the control is rotated, the readout delay value will be followed by the word "SET", denoting the indirect measurement mode. A few seconds after control movement has stopped, the word "SET" will disappear and the readout will display the direct measurement from the CTT.

b. Connect a test oscilloscope to the A GATE OUT connector (located on the instrument rear panel) via a 50-Ω BNC cable.

c. CHECK—Test oscilloscope displays a signal with a high level between 2.4 V and 5 V and a low level between 0 V and 0.4 V.

d. VERIFY—Duration of the high level is between 1 ms and 1.2 ms.

e. VERIFY—Duration of the low level is between 80 μs and 150 μs.

f. VERIFY—Duration of the low level increases to at least 10 times the time measured in part e when the HOLDOFF control is rotated to the maximum CW position but not in the detent.

g. Move the 50-Ω BNC cable from the A GATE OUT connector to the B GATE OUT connector.

h. CHECK—Test oscilloscope displays a signal with a high level between 2.4 V and 5 V and a low level between 0 V and 0.4 V.

i. VERIFY—Duration of the high portion of the signal is between 500 μs and 600 μs.

j. Disconnect the test setup.

ADDITIONAL FUNCTIONAL VERIFICATION

Equipment Required (see Table 4-1)

10X Probe supplied with Oscilloscope (Item 7)

INITIAL CONTROL SETTINGS

Control settings not listed do not affect the procedure.

a. Set:

NOTE

Select channels to set VOLTS/DIV.

VERTICAL

CH 1 and CH 2 VOLTS/DIV	0.1 V
CH 1 and CH 2 VAR	In detent
CH 3 and CH 4 VOLTS/DIV	0.1 V
CH 1, CH 2, CH 3, CH 4, ADD, and INVERT	Off
CHOP/ALT	ALT
20 MHz BW LIMIT	Off
CH 1 and CH 2 Input Coupling	1 MΩ DC

HORIZONTAL

A SEC/DIV	1 ms (knob in)
SEC/DIV VAR	In detent
X10 MAG	Off
TRACE SEP	Fully CW

Delta

ΔV and Δt	Off (press and release until associated readout is Off)
TRACKING	Off

TRIGGER

HOLDOFF	B ENDS A (fully CW)
LEVEL	Midrange
SLOPE	+ (plus)
A/B TRIG Select	A
MODE	AUTO
SOURCE	VERT
COUPLING	DC

1. Verify ALT, CHOP, and ADD Modes and TRACE SEP.

a. VERIFY—CH 1 trace is visible with no VERTICAL MODE buttons selected.

b. Press the CH 2 VERTICAL MODE button.

c. VERIFY—CH 1 trace is not displayed and the CH 2 trace is displayed.

d. Press the CH 1 VERTICAL MODE button.

NOTE

Separate the traces by approximately 1 division using the VERTICAL POSITION controls. Do not position either trace precisely at graticule center.

e. VERIFY—Both the CH 1 and the CH 2 traces are displayed.

f. Press the ADD button.

g. VERIFY—A third trace (ADD) is displayed.

h. Press the CH 3 VERTICAL MODE button.

i. VERIFY—The CH 3 trace is added to the display.

j. Press the CH 4 VERTICAL MODE button.

k. VERIFY—The CH 4 trace is added to the display.

l. Set the SEC/DIV controls to 50 ms (knob in).

m. VERIFY—5 traces are alternately displayed in the following sequence: CH 1, CH 2, ADD, CH 3, CH 4.

n. Set the TRIGGER MODE to SGL SEQ.

o. VERIFY—After the current sequence of traces is complete, no further traces are displayed.

p. Set the TRIGGER SOURCE to LINE.

q. Press and release the lower TRIGGER MODE button.

r. VERIFY—Each time the lower TRIGGER MODE button is pressed and released, the 5 signal traces appear once (in sequence), the readout display flashes once and the scale illumination flashes on and off.

s. Set the TRIGGER MODE to AUTO LVL and press the CHOP button.

t. VERIFY—The 5 traces appear to be displayed simultaneously.

u. Set:

TRIGGER SOURCE	CH 4
A SEC/DIV	20 μs
B SEC/DIV	10 μs (knob out)
CHOP/ALT	ALT
TRACE SEP	CCW until traces are separated

v. VERIFY—An alternate B sweep trace appears for each A sweep trace (10 traces total).

2. Verify BEAM FIND Operation.

a. Set:

HORIZONTAL

POSITION	Midrange
A SEC/DIV	1 ms (knob in)
X10 MAG	On

VERTICAL

CH 1	On
CH 2, CH 3, CH 4 and ADD	Off
POSITION	Midrange

b. Press and hold the BEAM FIND button.

c. VERIFY—The trace is less than 10 divisions long and remains in the graticule area as the CH 1 POSITION control and the Horizontal POSITION control are rotated through their complete ranges.

d. Release the BEAM FIND button and set the VERTICAL POSITION and Horizontal POSITION controls to midrange.

3. Check Probe Encoding.

NOTE

Refer to instrument "Operators Manual", for the positioning of the readout display information.

a. Set:

CH 1, CH 2, CH 3, CH 4 VERTICAL MODE	On
CH 1 and CH 2 VOLTS/DIV	100 mV
CH 3 and CH 4 VOLTS/DIV	0.1

b. Connect the standard accessory 10X probe (encoded) to the CH 1 input connector.

c. CHECK—CH 1 readout changes from 100 mV to 1 V.

d. Move the probe to CH 2 and repeat part c for that channel.

e. Move the probe to CH 3.

f. CHECK—Readout changes from 0.1 V to 1 V.

g. Move the probe to CH 4 and repeat part f for that channel.

h. Short a probe coding ring to ground.

NOTE

If using a P6137 probe, press probe ID button.

i. CHECK—Readout changes to ID for that channel and the trace jumps up approximately 0.5 divisions.

j. Repeat steps h and i for each VERTICAL channel.

WORD RECOGNIZER CHECKS

Equipment Required (see Table 4-1)

10X Probe supplied with Oscilloscope (Item 7)
T-connectors (Item 8)
BNC Cables (Item 10)
Pulse Generators (Item 24)
Adapter (Item 25)
Adapter (Item 26)

INITIAL CONTROL SETTINGS

Control settings not listed do not affect the procedure.

NOTE

Select channels to set VOLTS/DIV.

a. Set:

VERTICAL MODE

CH 1, CH 2, CH 3, CH 4 On

VOLTS/DIV

CH 1 and CH 2	2 V
CH 3 and CH 4	0.1 V
Input Coupling	50 Ω DC

HORIZONTAL

A and B SEC/DIV	200 ns (knob in)

Delta

Δt and ΔV	Off (press and release until associated readout is off)

TRIGGER

SOURCE	CH 1
MODE	AUTO LVL

b. Connect the + trigger output of pulse generator # 1 via a 50-Ω cable to the + trigger input of pulse generator # 2.

c. Connect the output of pulse generator # 1 via a 50-Ω cable and T-connector to the CH 1 input connector. Use the T-connector at the CH 1 input.

d. Connect the output of pulse generator # 2 via a 50-Ω cable and T-connector to the CH 2 input connector. Use the T-connector at the CH 2 input.

e. Connect the Word Recognizer probe to the P6407 input connector at the rear of the instrument.

f. Connect a BNC-male-to-dual-binding post adaptor to the T-connector on the CH 1 input, and connect another BNC-male-to-dual-binding post adaptor to the T-connector on the CH 2 input.

g. Connect a 4-inch bare wire (suitable for connecting a scope probe) to the red binding post of the adaptor connected to the CH 1 input.

h. Connect a 4-inch bare wire (suitable for connecting a scope probe) to the red binding post of the adaptor connected to the CH 2 input.

i. Connect a 2-inch bare wire (suitable for connecting a scope probe) to the black binding post of the adaptor connected to the CH 2 input.

j. Connect both ground leads from the Word Recognizer probe to the bare wire on the black binding post on the CH 2 input.

k. Connect the CH 3 input to the WORD RECOG OUT connector using the instrument X10 probe and a BNC-to-probe-tip adaptor.

l. Set pulse generator # 1 to produce a positive 0.5-μs pulse every 1 μs.

m. Set pulse generator # 2 to produce a positive 400-ns pulse when it receives an external trigger.

NOTE

The lowest point of the HI must not be lower than 2.0 V.

n. Set both pulse generators to produce pulses of +0.6 V LO and +2.0 V HI.

o. Press the MEASURE button.

p. Select <MORE> ("8") from menu.

q. Select LOGIC-TRIG ("4") from menu.

r. Select B:WORD-REC ("6") from menu.

s. Press A/B TRIG button to select B Sweep Triggers.

1. If you wish to change the word recognizer display radix:
   a. Press the MEASURE button.
   b. Select <MORE> ("8") from menu.
   c. Select CONFIGURE ("5") from menu.
   d. Select WR-RADIX ("5") from menu.
   e. Select HEX, OCTAL, or BINARY from menu.
   f. Press RECALL to exit menu.

t. Connect the clock (C) input of the Word Recognizer to the wire on the red binding post of the CH 1 input.

u. Connect the Q and W0-W15 inputs of the Word Recognizer to the wire on the red binding post of the CH 2 input.

v. Set the A SEC/DIV to 20 ns (knob in).

2. Check Data Setup Time

a. For each test setup described in Table 4-13:

1. Vary (increase) the pulse duration of pulse generator # 2 until the active edge of the CH 2 signal falls about 10 ns after the trigger edge of the CH 1 signal.
2. CHECK—CH 3 is not displaying a signal.
3. Vary (decrease) the pulse duration of pulse generator # 2, moving the active edge of the CH 2 signal to the left until CH 3 displays a stable signal.
4. Press the Δt button.
5. Turn the Δ REF OR DLY POS control to align the delta reference cursor with the first edge of the CH 2 signal.
6. Turn the Δ control to align the delta cursor with the first edge of the CH 1 signal.
7. CHECK—Reading is ≤25 ns.
8. Press the Δt button.

Table 4-13
Data Setup Time Checks

Polarity		Word Recognizer Word Definition	A TRIGGER SLOPE
Pulse Generator # 1	# 2		
+	+	↓—0-0000	−
+	−	↓—1-FFFF	−
−	−	↑—1-FFFF	+
−	+	↑—0-0000	+

3. Check Data Hold Time

a. For each test setup described in Table 4-14:

1. Vary the pulse duration of pulse generator # 2 until the first edge of the CH 2 signal falls about 10 ns after the trigger edge of the CH 1 signal.

2. CHECK—A stable signal is displayed on CH 3.

3. Vary the pulse duration of pulse generator # 2, moving the first edge of the CH 2 signal to the left until CH 3 no longer displays a stable signal.

4. Press the Δt button.

5. Turn the Δ REF OR DLY POS control to align the delta reference cursor with the first edge of the CH 2 signal.

6. Turn the Δ control to align the delta cursor with the first edge of the CH 1 signal.

7. CHECK—Reading is >4 ns.

Table 4-14
Data Hold Time Checks

Polarity		Word Recognizer Word Definition	A TRIGGER SLOPE
Pulse Generator # 1	# 2		
+	+	↓—1-FFFF	–
+	–	↓—0-0000	–
–	–	↑—0-0000	+
–	+	↑—1-FFFF	+

4. Check Minimum Clock Pulse Width

a. Set pulse generator # 1 to produce a 5-ns positive pulse every 1 μs.

b. Press the A/B TRIG button to select A Trigger MODE.

c. Press the upper Trigger MODE button to reinitialize the auto-trigger level.

d. Press the A/B TRIG button.

e. For each test setup described in Table 4-15:

1. If there is not a stable signal displayed on CH 3, (>0.6 V amplitude), vary (increase) the pulse duration of pulse generator # 1 until CH 3 displays a stable signal.

2. Press the Δt button.

3. Turn the Δ REF OR DLY POS control to align the delta reference cursor with the leading edge of the CH 1 pulse.

4. Turn the Δ control to align the delta cursor with the trailing edge of the CH 1 pulse.

5. CHECK—Reading is ≤20 ns.

6. Press the Δt button.

Table 4-15
Minimum Clock Pulse Width Checks

Polarity		Word Recognizer Word Definition	A TRIGGER SLOPE
Pulse Generator # 1	# 2		
+	+	↑—X-XXXX	+
–	+	↓—X-XXXX	–

5. Check Delay From Selected Edge to WORD RECOG OUT.

a. Set:

VERTICAL

CH3 and CH 4	On
CH 1, CH 2, ADD, and INVERT	Off
CH 3 VOLTS/DIV	0.1 V (1 V with X10 probe attached)

HORIZONTAL

A and B SEC/DIV	20 ns

b. Connect another instrument X10 probe to the CH 4 input connector and the probe tip to the wire on the red binding post of the CH 1 input.

c. Set pulse generator # 1 to produce a 50-ns positive pulse every 10 μs. (CH 4 Display.)

d. Set the A Trigger SOURCE to CH 4.

e. For each test setup described in Table 4-16:

1. Press the Δt button.
2. Turn the Δ REF OR DLY POS control to align the delta reference cursor with the active edge of the CH 4 signal.
3. Turn the Δ control to align the delta cursor with the rising edge of the CH 3 signal.
4. CHECK—Reading is ≤55 ns.
5. Press the Δt button.

6. Check Word Recognition Delay

a. Set pulse generator # 1 to produce a positive 0.5-μs pulse every 1 μs.

b. Disconnect the C input of the Word Recognizer from the wire on the red binding post of the CH 1 input.

c. Connect the Q and W0-W15 inputs of the Word Recognizer to the wire on the red binding post of the CH 1 input.

d. For each test setup described in Table 4-17:

1. Press the Δt button. Turn the Δ REF OR DLY POS control to align the delta reference cursor with the first edge of the CH 4 signal.
3. Turn the Δ control to align the delta cursor with the rising edge of the CH 3 signal.
4. CHECK—Reading is ≤140 ns.
5. Press the Δt button.

e. Disconnect the probe on the CH 4 input.

Table 4-16
Delay From Selected Edge to WORD RECOG OUT Checks

Polarity		Word Recognizer Word	A TRIGGER SLOPE
Pulse Generator			
# 1	# 2	Definition	
+	+	↑—X-XXXX	+
−	+	↓—X-XXXX	−

Table 4-17
Word Recognition Delay

Polarity		Word Recognizer Word	A TRIGGER SLOPE
Pulse Generator			
# 1	# 2	Definition	
+	+	X—1-FFFF	+
−	+	X—0-0000	−

7. Check Data Input Coincidence

a. Set:

CH 2 and CH 3	On
CH 4	Off
A SEC/DIV	50 ns (knob in)
SOURCE	CH 2
SLOPE	– (minus)

b. Set pulse generator # 1 to produce a positive 0.5-μs pulse every 1 μs.

c. Set pulse generator # 2 to produce a negative 5-ns pulse when it receives an external trigger.

d. Set the A SEC/DIV to 20 ns (knob in).

e. Set the Word Definition of the Word Recognizer probe to BX0 0000.

f. Connect the Q and W0-W15 inputs of the Word Recognizer to the wire on the red binding post of the CH 2 input.

g. Press the A/B TRIG button to select A Trigger MODE.

h. Press the upper Trigger MODE button to reinitialize the auto-trigger level.

i. Vary (increase) the pulse duration of pulse generator # 2 until further increase makes the CH 3 display stable (>0.6 V amplitude).

j. Press the Δt button.

k. Turn the Δ REF OR DLY POS control to align the delta reference cursor with the falling edge of the CH 2 signal.

l. Turn the Δ control to align the delta cursor with the rising edge of the CH 2 signal.

m. CHECK—Reading is $\geq$20 ns and $\leq$85 ns.

n. Press the Δt button.

o. Disconnect the test setup.

ADJUSTMENT PROCEDURE

INTRODUCTION

IMPORTANT-PLEASE READ BEFORE USING THIS PROCEDURE

The "Adjustment Procedure" is used to restore optimum performance or return the instrument to conformance with its "Performance Requirements" as listed in the "Specification" (Section 1). As a general rule, these adjustments should be performed every 2000 hours of operation or once a year if used infrequently.

PARTIAL PROCEDURES

This procedure is divided into subsections to permit calibration of individual sections of the instrument whenever complete instrument calibration is not required. To perform a partial procedure, first set the instrument as directed in the Initial Setup Conditions at the beginning of the section, then make any changes called for within the procedure. Perform all steps within a subsection, both in the sequence presented and in their entirety to ensure that control settings will be correct for the following steps.

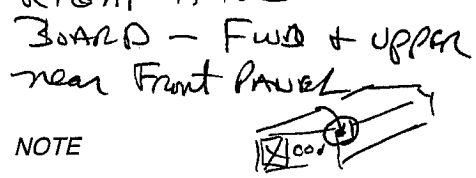
The adjustments in CAL 01, 02, 03, 06, 07 and 09 should be performed in numerical sequence; i.e., CAL 01 should be done before CAL 02, CAL 02 should be done before CAL 03, etc. CAL 04, 05, and 08 are independent of adjustments made in the other calibration routines. Performing partial procedures when setting the automatic calibration constants (i.e., only one or two of the CAL steps) is not recommended and should only be done if the calibration constants set in the preceding steps are known to be correct.

PREPARATION FOR ADJUSTMENT

It is necessary to remove the cabinet to do the Adjustment Procedure. See the cabinet removal instructions in the Maintenance section of this manual, Section 6.

All test equipment items required to do the complete Adjustment Procedure are described in Table 4-1 at the beginning of Section 4, Performance Check Procedure. The specific items of equipment needed to do each subsection in this procedure are listed at the beginning of that subsection.

BEFORE YOU BEGIN:



NOTE

When performing any of the automatic calibration routines (CAL 01 through CAL 08), the CAL/NO CAL jumper P501 must be moved to its CAL position (between pins 2 and 3) before turning the power on. When the desired calibration has been performed, return the jumper to its NO CAL position.

a. Turn instrument Power on.

NOTE

The instrument MUST have a 20-minute warmup period before making any adjustments. Performing the adjustment procedure while the temperature is drifting may cause erroneous calibration settings.

POWER SUPPLIES AND DAC REF ADJUSTMENT

Equipment Required (see Table 4-1)

Oscilloscope With 10X Probe (Item 7)
Digital Multimeter (DMM) (Item 19)
Alignment Tool (Item 20)
1X Probe (Item 21)

See **ADJUSTMENT LOCATIONS 1** *and* **ADJUSTMENT LOCATIONS 4**

at the back of this manual for test point and adjustment locations.

NOTE

If the instrument displays "DIAGNSTIC. PUSH A/B TRIG TO EXIT" at power on, one of the power-up tests has failed. If the error message on the bottom line of the display is "TEST 04 FAIL xx" where "xx" is 01, 10 or 11, stored calibration data is in error, and the instrument should be recalibrated. If this is the case, pressing the A/B TRIG button will force entry to the normal operating mode; however, the accuracy of any measurement taken could be in error.

If any other error message occurs, the failure is probably not related to calibration. In this case, the instrument should be repaired before attempting calibration.

Initial Control Settings.

Control settings not listed will not affect the procedure.

VERTICAL VOLTS/DIV

CH 2	100 mV
CH 3 and CH 4	100 mV
CH 1 and CH 2 VAR	In detent

VERTICAL MODE

CH 1	On
CH 2, CH 3, CH 4	Off
ADD, INVERT, and BW LIMIT	Off
ALT/CHOP	ALT

VERTICAL POSITION

CH 1	Midrange

Input Coupling

CH 1 and CH 2	1 MΩ DC

Horizontal

SEC/DIV	X-Y (knob in)
SEC/DIV VAR	In detent
POSITION	Midrange

TRIGGER

MODE	AUTO LVL
SOURCE	VERT
COUPLING	DC
SLOPE	+ (plus)
LEVEL	Midrange
HOLDOFF	In detent

Delta

ΔV and Δt	Off (press and release until readout display disappears)
INTENSITY	Visible display
READOUT INTENSITY	Visible display (CW from OFF)
SCALE ILLUM	Fully CCW
FOCUS	Defocused dot

1. Check/Adjust Power Supply DC Levels, Regulation, and Ripple (R1292).

a. Connect the Digital Multimeter (DMM) negative lead to chassis ground. Connect the positive lead to the first test point listed in Table 5-1 (all test points are on the Main Board).

b. CHECK—That the reading is within the limits given in Table 5-1.

c. ADJUST—Volt Ref Adj (R1292) for a DMM reading of precisely 10.00 V. The adjustment is accessible through a hole in the top cover plate.

Table 5-1
Power Supply Voltage and Ripple Tolerances

Power Supply	Test Point (+ Lead)	Reading	Total p-p Ripple	p-p Ripple at Two Times Line Frequency
+10 V	J119-4	+9.99 to +10.01	100 mV	1 mV
+87 V	J119-8	+85.26 to +88.74	80 mV	5 mV
+42.4 V	J119-9	+41.55 to +43.25	80 mV	2 mV
+15 V	J119-6	+14.775 to +15.225	15 mV	11 mV
Digital +5 V	J119-2	+4.85 to +5.15	150 mV	30 mV
Analog +5 V	J119-12	+4.925 to +5.075	15 mV	1 mV
−5 V	J119-5	−4.965 to −5.035	15 mV	1 mV
−8 V	J119-11	−7.88 to −8.12	100 mV	1 mV
−15 V	J119-1	−14.775 to −15.225	10 mV	2 mV

d. Repeat parts a and b for the other test points listed in Table 5-1.

e. Disconnect the DMM.

f. Set the test oscilloscope as follows:

Sweep Speed	5 ms/div
CH 1 Input Coupling	1 MΩ AC
Vertical controls	To display CH 1
Trigger controls	Line source, triggered display
Volts/Division	2 mV
BW Limit	20 MHz

g. Using a 1X probe, connect the test oscilloscope probe ground lead to chassis ground. Connect the probe tip to the first test point listed in Table 5-1.

h. CHECK—Ripple at two times the line frequency and the total peak-to-peak ripple do not exceed the values given in Table 5-1.

i. Repeat part h for each test point in Table 5-1.

j. Disconnect the test oscilloscope.

2. Adjust DAC Ref (R2010)

a. Set:

A SEC/DIV	100 μs
Δt	On (Δt readout)

NOTE

The objective of this step is to make the total range of the DAC output voltage (sum of the CCW and CW readings) equal to 2.5 V.

b. Connect the digital multimeter (DMM) negative lead to the chassis ground. Connect the positive lead to pin 13 of J119 (on the Main Board).

c. Set the DMM to measure approximately 1.5 Vdc.

d. Rotate the Δ control CCW until the DMM reading remains at a constant value (approximately −1.250 V). Note the reading.

e. Rotate the Δ control CW until the DMM reading remains at a constant value (approximately +1.250 V). Note the reading.

f. Add the absolute values of the readings noted in parts d and e together (approximately 2.500 V).

g. Subtract the total in part f from 2.500 V, then divide the difference by two.

h. ADJUST—DAC Ref (R2010 on the Control Board) to add the (signed) number obtained in part g to the reading obtained in part e.

i. Repeat parts d through h as necessary to obtain a total DAC range of 2.500 V.

2445B AND 2455B CRT ADJUSTMENTS

Equipment Required (see Table 4-1)

Leveled Sine-Wave Generator (Item 2)

50 Ω BNC Cable (Item 10)

Alignment Tool (Item 20)

See **ADJUSTMENT LOCATIONS 1,** **ADJUSTMENT LOCATIONS 2,** *and* **ADJUSTMENT LOCATIONS 4**

at the back of this manual for test point and adjustment locations.

NOTE

When performing the following automatic cal steps, initial setting of the front-panel controls is not required.

1. Adjust GRID BIAS (R1878)

a. Simultaneously press in and hold the Δt and the ΔV push buttons, then press and hold the SLOPE button. Hold all three buttons in for approximately one second, then release them.

b. CHECK—Top line of the readout display says: "DIAGNSTIC. PUSH A/B TRIG TO EXIT".

NOTE

The "menu" of calibration, test, and exercise routines are in a loop that may be scrolled through in single steps, either forward or backward. Pressing the upper or lower TRIGGER MODE push buttons respectively increments or decrements the menu position by one. As each routine is selected, its name appears in the lower left corner of the readout display.

c. Scroll to CAL 08.

NOTE

In this procedure, pressing the upper TRIGGER COUPLING button increments the routine to the next step. Pressing the lower TRIGGER COUPLING button will return to the previous step.

d. Press and release the upper TRIGGER COUPLING button to initiate the routine.

e. Set SCALE ILLUMINATION control (front panel) full CCW (Off).

f. ADJUST—Grid Bias (R1878) if necessary to obtain an X-Y dot near center screen.

g. Position the X-Y dot adjacent to a dot in the lower row of readout dots using CH 1 and CH 2 position controls.

h. ADJUST—Grid Bias (R1878) to match the intensity of the X-Y dot to the readout dots. (Defocusing the display may give better resolution.)

i. Press and release the upper TRIGGER COUPLING button to advance to the next step.

2. Check Grid Bias Adjustment

a. Set SCALE ILLUMINATION control (front panel) full CCW (Off).

b. CHECK—A dim X-Y dot is visable near graticule center.

c. Set INTENSITY control (front panel) full CCW (Off).

d. CHECK—The dot is no longer visable with the INTENSITY Off.

NOTE

If the dot is not present in the first part of the check or does not fully disappear during the second part of the check; the Grid Bias adjustment step should be repeated. To repeat the Grid Bias Adjust step, press the lower TRIGGER COUPLING button once to return to the Grid Bias Adjustment step and repeat step 1 above.

e. Press and release the upper TRIGGER COUPLING button to advance to the next step.

3. Adjust TRACE ROTATION (Front Panel), Y–AXIS (R1848), FOCUS (Front Panel), ASTIG (Front Panel), and GEOMETRY (R1870).

a. Using the CH 1 Vertical POSITION control, align the trace with the center horizontal graticule line.

b. Position one of the Δt cursors to the center vertical graticule line using either the Δ or the ΔREF OR DLY POS control.

c. ADJUST—INTENSITY control to align the trace with the center horizontal graticule line.

d. ADJUST—TRACE ROTATION control (front panel) to align the trace with the center horizontal graticule line.

e. ADJUST—Y-Axis Alignment (R1848) to align the Δt cursor with the center vertical graticule line.

f. Repeat parts d and e as necessary for the best aligned display.

NOTE

Y-Axis and TRACE ROTATION will remain adjusted and are not interactive of the following adjustments.

g. ADJUST—ASTIG control (front panel), in conjunction with the FOCUS control (front panel) for the sharpest possible display near the center graticule area.

h. Position the Δt cursors on (or within 0.2 division of) the first and eleventh vertical graticule lines using the ΔREF OR DLY POS and Δ controls.

NOTE

Adjust X1 Horizontal Gain (R860) if necessary to position the Δt cursors as described in step h above. If the Horizontal Gain (R860) is adjusted, it will be necessary to perform CAL 01 to restore optimum adjustment.

i. ADJUST—Geometry (R1870) for minimum curvature of both Δt cursors.

j. ADJUST—READOUT INTENSITY control (front panel) to the OFF position.

k. Using the CH 2 Vertical POSITION control, set the CH 2 trace off screen.

l. Connect a 50 kHz, 8-division signal from the Leveled Sine-Wave Generator to the CH 1 input connector via a 50 Ω BNC cable.

m. Center the display on the graticule. Set INTENSITY control as necessary for a well defined display.

n. ADJUST—Edge Focus (R1864), FOCUS control (front panel), and ASTIG control (front panel) for the most uniform focus over the entire display.

NOTE

Slight interaction between Geometry, Edge Focus, and Focus, and Astigmatism is normal. To achieve optimum edge focus it may be necessary to slightly compromise the Geometry adjustment.

o. Disconnect the Sine-Wave Generator from the CH 1 input.

p. ADJUST—READOUT INTENSITY control to display Δt cursors and readout information.

q. CHECK—Readout characters remain focused.

r. REPEAT—Parts i through q as necessary to obtain optimum focus.

s. Press and release the upper TRIGGER COUPLING button to advance to the next step.

4. Adjust HIGH DRIVE FOCUS (R1842)

a. Connect a 10 MHz, 6-division signal from the Leveled Sine-Wave Generator to the CH 1 input connector via a 50 Ω BNC cable.

b. Center the display on the graticule.

c. ADJUST—Horizontal POSITION control to view the sweep start.

d. ADJUST—High Drive Focus (R1842) for the best overall focus of the trace.

NOTE

Do not disconnect the Sine-Wave Generator from the CH 1 input.

e. Press and release the upper TRIGGER COUPLING button to advance to the next step.

5. Adjust HORIZONTAL DYNAMIC CENTERING (R3401)

a. Center the display on the graticule.

b. ADJUST—Horizontal Dynamic Centering (R3401) for minimum horizontal display shift as the INTENSITY control (front panel) is repeatedly changed from minimum to maximum trace intensity.

NOTE

Disregard any vertical shift of the waveform during the adjustment.

c. Disconnect the Sine-Wave Generator from the CH 1 input.

d. Press and release the upper TRIGGER COUPLING button to advance to the next step.

6. Adjust VERTICAL DYNAMIC CENTERING (R3407)

a. ADJUST—Vertical Dynamic Centering (R3407) for minimum vertical deflection of the intensified zone with respect to the trace.

NOTE

Correct adjustment will align the intensified zone with the trace such that a single horizontal trace results with no vertical deflection difference between the trace and the intensified zone.

b. Press and release the upper TRIGGER COUPLING button to conclude CAL 08.

CH 1 AND CH 2 INPUT CAPACITANCE, AND VERTICAL READOUT JITTER ADJUSTMENTS

Equipment Required

Calibration Generator (Item 3)
50 Ω BNC Cable (Item 10)
50 Ω Termination (Item 12)
Alignment Tool (Item 20)
Normalizer (Item 22)

See **ADJUSTMENT LOCATIONS 3** *and* **ADJUSTMENT LOCATIONS 4** *at the back of this manual for test point and adjustment locations.*

Initial Control Settings.

Control settings not listed do not affect the procedure.

VERTICAL VOLTS/DIV

CH 1 and CH 2	100 mV
CH 1 and CH 2 VAR	In detent

Input Coupling

CH 1 and CH 2	1 MΩ DC

VERTICAL MODE

CH 1	On
CH 2, CH 3, CH 4	Off
ADD, INVERT, and BW LIMIT	Off
ALT/CHOP	ALT

VERTICAL POSITION

CH 1	Midrange

Horizontal

A SEC/DIV	100 μs (knob in)
SEC/DIV VAR	In detent
Horizontal POSITION	Midrange

TRIGGER

MODE	AUTO LVL
SOURCE	VERT
COUPLING	DC
SLOPE	+ (plus)
LEVEL	Midrange
HOLDOFF	In detent

Delta

ΔV and Δt	Off (press and release until readout display disappears)
INTENSITY	Left of center
READOUT INTENSITY	As required for a visible display
SCALE ILLUM	Fully CCW
FOCUS	Best focused display

1. Adjust CH 1 and CH 2 Input Capacitance (C105 and C205).

NOTE

The objective of this adjustment is to match the input capacitance of the 50 mV per division position of the VOLTS/DIV switches to the 0.1 mV per division position. The front corner of an input square-wave signal is used to indicate when the capacitances are matched.

a. Connect a 1 kHz square-wave signal from the Calibration Generator high-amplitude output to the CH 1 OR X input connector via a 50 Ω BNC cable, a 50 Ω termination, and a normalizer. Adjust the generator output level for a 6-division signal vertically centered on the graticule.

b. Set the normalizer for a square front corner over approximately the first 40 μs (0.4 division) of the positive portion of the waveform.

c. Change the CH 1 VOLTS/DIV switch to the 50 mV position and adjust the generator for a 6-division signal display.

d. ADJUST—The CH 1 50 mV C Adj (C105 on the Main Board) for the same waveform front corner noted in part b.

e. Repeat parts b through d until no change is observed in the waveform front corner when the CH 1 VOLTS/DIV switch is alternated between the 50 mV and 0.1 V positions. When switching between positions, reestablish the reference display amplitude at each position, and observe the square-wave front corner to make the comparison.

f. Move the input signal to CH 2 and change the VERTICAL MODE to display CH 2 only. Adjust the generator amplitude for a 6-division signal amplitude.

g. Set the normalizer for a square front corner over approximately the first 40 μs (0.4 division) of the positive portion of the waveform.

h. Change the CH 2 VOLTS/DIV switch to the 50 mV position and adjust the generator for a 6-division display.

i. ADJUST—The CH 2 50 mV C Adj (C205 on the Main Board) for the same waveform front corner noted in part g.

j. Repeat parts g through i until no change is observed in the waveform front corner when the CH 2 VOLTS/DIV switch is alternated between the 50 mV and 0.1 V positions. When switching between positions, reestablish the reference signal amplitude at each position, and observe the square-wave front corner to make the comparison.

k. Disconnect the test setup.

2. Adjust Vertical Readout Jitter (R618).

NOTE

If the previous step was not performed, first set up the Initial Control Settings before, then proceed as follows.

a. Set:

VERTICAL

CH 1 Input Coupling	50 Ω DC
CH 1 VERTICAL MODE	On
CH 2, CH 3, and CH 4 VERTICAL MODE	Off

Horizontal

A SEC/DIV	500 μs (knob in)

b. Press and release the ΔV button to obtain a ΔV display.

c. Use the Δ REF OR DLY POS control to position one cursor 3 divisions above graticule center. Use the Δ control to position the other cursor 3 divisions below graticule center.

d. Connect a 1 kHz, fast-rise signal from the Calibration Generator to the CH 1 OR X input connector via a 50 Ω BNC cable.

e. Set the generator output level for an 8-division display.

f. Use the CH 1 Vertical and Horizontal POSITION controls to center the CH 1 display on the graticule.

g. ADJUST—Vertical Readout Jitter (R618) for minimum vertical jitter of the readout characters and cursors.

h. Disconnect the test setup.

AUTOMATIC CALIBRATION CONSTANTS, HORIZONTAL AND VERTICAL GAIN, CENTERING, AND TRANSIENT RESPONSE ADJUSTMENTS

NOTE

Within the following procedures, the calibration constants for timing, vertical gain, trigger level, transient response, and parametric measurements are generated by the system microprocessor and are stored in nonvolatile memory. The adjustments in CAL 01, 02, 03, 06, 07, and 09 should be performed in sequence; i.e., CAL 01 should be done before CAL 02, CAL 02 should be done before CAL 03, etc. Performing partial procedures (i.e., only one or two of the CAL steps) is not recommended and should only be done if the calibration constants that would have been set in the preceding steps are known to be correct.

The CAL functions are available only if the CAL/NO CAL jumper (P501 on the Control Board) is in the CAL position (between pins 2 and 3) when power is turned on. When the automatic calibration procedures are completed, return the jumper to the NO CAL position to prevent entry into the calibration routines.

Equipment Required (see Table 4-1)

Calibration Generator (Item 3)
Time-Mark Generator (Item 6)
Oscilloscope (Item 7)
50 Ω BNC Cable (Item 10)
Dual Input Coupler (Item 11)
5X Attenuator (Item 17)
Digital Multimeter (DMM) (Item 19)
Alignment Tool (Item 20)
Tunnel Diode Pulser (Item 23)

See **ADJUSTMENT LOCATIONS 4**

at the back of this manual for test point and adjustment locations.

INITIAL CONTROL SETTINGS

CAL/NO CAL jumper	CAL position (between pins 2 and 3) prior to turning on power

NOTE

When performing the automatic CAL steps, the initial setup of front-panel controls is not required.

CAL 01—HORIZONTAL

1. Check/Adjust Horizontal Timing, X1 Gain (R860), X10 Gain (R850), Hrz Ctr (R801), and Trans Resp (R802).

a. Simultaneously press in and hold the Δt and the ΔV push buttons, then press and hold the SLOPE switch. Hold all three switches in for approximately one second, then release them.

b. CHECK—Top line of the readout display says: **"DIAGNSTIC. PUSH A/B TRIG TO EXIT".**

NOTE

The "menu" of calibration, test, and exercise routines is in a loop that may be scrolled through in single steps, either forward or backward. Pressing the upper or lower TRIGGER MODE switch and releasing it respectively increments or decrements the menu position by one. As each routine is selected, its name appears in the lower left corner of the readout display.

When performing a calibration step, touch only the specific control or controls called out in the procedure. Movement of other controls may cause erroneous calibration results.

c. Scroll to CAL 01.

CAUTION

Upon entering CAL 01, the Input Coupling is automatically set to 50 Ω DC and the 50-Ω OVERLOAD protection is disabled. Before starting the procedure, make sure any 50-Ω OVERLOAD condition has been cleared.

In this procedure, pressing the upper TRIGGER COUPLING switch stores the current calibration parameter being set and increments the routine to the next step (except where otherwise noted).

d. Connect the DMM, set to measure approximately 500 mV, to the CALIBRATOR output.

e. Press and release the upper TRIGGER COUPLING switch.

NOTE

The CALIBRATOR output will go to its LO level on odd CAL steps and to its HI level on even steps.

f. CHECK—Readout indicates ADJUST Δ (step) 0, CH 1 PROBE TO TP800 ON MAIN BD.

g. Connect a P6137 probe from CH 1 to TP800, at rear of main board near readout connector.

h. ADJUST—Δ REF to center signal near displayed cursors, and ADJUST—Δ control to join traces.

NOTE

INTENSITY control may need to be adjusted to view signal.

i. Press and release the upper TRIGGER COUPLING switch.

j. CHECK—CALIBRATOR output voltage is 0 mV ±1 mV.

k. Disconnect the X10 probe.

l. CHECK—Readout indicates ADJUST Δ (step) 1, 100 μs (for A Sweep), and 1 μs (for B Sweep).

NOTE

The readout prompts the operator by showing the control to be moved (upper left corner), the autocal step number (upper right corner), the A-Sweep speed (bottom right center), and the B-Sweep speed (bottom right corner) as set up by the routine. An example (from step l above) is:

ADJ Δ *1*

100 μs *1 μs*

m. Connect the Time-Mark Generator, set for 0.1-ms time markers, to the CH 1 OR X input connector via a 50-Ω BNC cable.

n. Set:

VOLTS/DIV	As needed for a convenient signal display amplitude
TRACE SEP	As needed to separate the A and B Sweeps
CH 1 POSITION	As needed to view both A and B Sweeps
Horizontal POSITION	Position start of trace at the left graticule line

NOTE

In the following calibration routine some sequential pairs of steps are iterative, i.e., the earlier step is recalled if an adjustment is made in the later step. Occasionally, on the earlier of some of these pairs, the readout may indicate "LIMIT" before the correct control setting is reached. If this occurs, proceed to the next AUTOMATIC CAL step. After the adjustment at the next step is performed, the previous step will automatically be recalled, and the adjustment may be performed in the normal manner.

o. ADJUST—Δ REF OR DLY POS and Δ controls to align both the intensified zones with the 6th time marker (near graticule center) and to superimpose the delayed B-Sweep time markers. Press and release the upper TRIGGER COUPLING switch.

p. CHECK—CALIBRATOR output voltage is between 398 mV and 402 mV of the reading noted in part j. Disconnect the DMM when through.

q. CHECK—Readout indicates ADJ Δ (step) 2, 100 μs (for A Sweep), and 1 μs (for B Sweep).

r. ADJUST—Δ REF OR DLY POS control to intensify the 2nd time marker, and ADJUST—Δ control to intensify the 10th time marker. Superimpose the delayed B-Sweep time markers within 0.2 division.

s. Press and release the upper TRIGGER COUPLING switch.

t. CHECK—Readout indicates ADJ Δ (step) 3, 300 μs (for A Sweep), and 1 μs (for B Sweep).

u. ADJUST—Δ REF OR DLY POS control to intensify the 4th time marker, and ADJUST—Δ control to intensify the 28th time marker. Superimpose the delayed B-Sweep time markers within 1.2 divisions.

v. Press and release the upper TRIGGER COUPLING switch. If the adjustment in step 3 was changed, step 2 will be recalled; otherwise step 4 will be initiated.

w. CHECK—Readout indicates ADJ Δ (step) 4, 100 μs (for A Sweep), and 1 μs (for B Sweep). Set the Time-Mark Generator for 5-μs time markers.

x. ADJUST—Δ control CCW until no further movement of the B-Sweep display occurs. Note the position of the 1st time marker, then adjust the Δ control CW until the 2nd time marker moves to the left and aligns with the position just noted.

NOTE

Movement of the Δ REF control at this point will adversely affect the calibration.

y. Press and release the upper TRIGGER COUPLING switch. Set the Time-Mark Generator for 10-μs time markers.

z. CHECK—Readout indicates X1, X10, HRZ CTR, (step) 5, and 10 μs (for A Sweep) and two vertical cursors appear on the display.

aa. ADJUST—X1 Gain (R860) and Hrz Ctr (R801) to align the two cursors with the 2nd and 10th vertical graticule lines, then adjust X10 Gain (R850) for 1 time marker per division.

bb. Press and release the upper TRIGGER COUPLING switch. Set the Time-Mark Generator for 10-ms time markers.

cc. CHECK—Readout indicates ADJ Δ (step) 6, 10 ms (for A Sweep), and 100 μs (for B Sweep).

dd. ADJUST—Δ REF OR DLY POS control to intensify the 2nd time marker, and ADJUST—Δ control to intensify the 10th time marker. Superimpose the delayed B-Sweep time markers within 0.2 division.

ee. Press and release the upper TRIGGER COUPLING switch. Set the Time-Mark Generator for 1-μs time markers.

ff. For each step in Table 5-2, do the following:

1. Adjust the Δ REF OR DLY POS and Δ controls, as necessary, to intensify the indicated time marks on the A Sweep and superimpose the displayed B-Sweep markers within the listed limits.

2. Press and release the upper TRIGGER COUPLING switch.

NOTE

If the Δ control is adjusted at step 9, 12 or 14, the previous step will be repeated.

Table 5-2
Horizontal Timing

Step Number	Time-Marker Period	ΔREF Marker	Δ Marker	Superposition Tolerance In Divisions
7	1 μs	2	10	0.2
8	2 μs	2	10	0.2
9	2 μs	4	28	1.2
10	10 μs	2	10	0.2
11	50 μs	2	10	0.2
12	50 μs	4	28	1.2
13	0.5 μs	2	10	0.2
14	0.5 μs	4	28	1.2
15	0.1 μs	2	10	0.2
16[a]	20 ns	2	10	0.2[a]

[a]Note the amplitude of the time markers. Use the Δ control to adjust for one time mark per division on the A sweep. Select 5-ns time markers. Adjust CH 1 VOLTS/DIV to obtain at least a 3-division signal. Use the Δ control to superimpose the B-Sweep markers within 0.2 divisions. Select 20-ns time markers and return the VOLTS/DIV to the amplitude noted above. Press upper TRIGGER COUPLING switch to store the calibration constant.

gg. Set the TRACE SEP full CW.

hh. Connect the Time Mark Generator output to CH 1 of both the IUT (instrument under test) and the bench scope via a BNC "T" and two 50-Ω BNC cables. Connect B GATE OUT of IUT to CH 2 of bench scope via a 50-Ω BNC cable.

ii. Set bench scope to view CH 1, TRIGGER SOURCE to CH 2, and CH 1 and CH 2 input coupling to 50 Ω.

jj. For each step in Table 5-3 (except step 28), adjust the Δ control for roughly the listed number of markers over the center 8 divisions, then superimpose markers on bench scope. Manually set SEC/DIV setting of bench scope to keep a usable time mark. Use IUT DELAY POS to bring markers on the bench scope screen. Some sweep speeds might require adjusting holdoff to see both markers. When markers are superimposed, press and release the upper TRIGGER COUPLING switch. If the Δ control is adjusted at step 18, 20, 23 or 25, the previous step will be repeated. At step 28, adjust Trans Resp (R802 on the Main Board) as indicated.

NOTE

Change the CH 1 VOLTS/DIV switch setting as necessary to maintain adequate signal display amplitude.

Table 5-3
Horizontal Timing

Step No.	Bench Scope Time/DIV	Time-Marker Period	Markers Over 8 Divisions	Bench Scope Superposition Tolerance in Divisions
17	200 ns and X10 (20 ns)	1 μs	8	0.2
18	200 ns and X10 (20 ns)	1 μs	24	1.2
19	500 ns and X10 (50 ns)	2 μs	8	0.2
20	500 ns and X10 (50 ns)	2 μs	24	1.2
21	2 μs and X10 (200 ns)	10 μs	8	0.2
22	10 μs and X10 (1 μs)	50 μs	8	0.2
23	10 μs and X10 (1 μs)	50 μs	24	1.2
24	100 ns and X10 (10 ns)	500 ns	8	0.2
25	100 ns and X10 (10 ns)	500 ns	24	1.2
26	20 ns and X10 (2 ns)	100 ns	8	0.2
27	20 ns and X10 (2 ns)[a]	20 ns	8	0.2
28[b]	na	na	na	na
29	200 μs and X10 (20 μs)	1 ms	8	0.2
30[b]	na	na	na	na
31[c]	na	10 ns	8	na
32[ce]	na	10 ns	8	na
33[de]	na	2 ns	4	na
34[de]	na	2 ns	4	na

[a]Use the Δ control to adjust for approximately 1 Time-Marker per division. Set Time Mark Generator for 5 ns markers. Adjust the Δ control to superimpose the displayed bench scope display. The bench scope holdoff may require adjustment.
[b]Steps not used.
[c]Adjust volt/div for > 3 division amplitude. Adjust Δ for 1 time marker per division over the center 8 divisions.
[d]Adjust volt/div for 1 to 4 division amplitude. Adjust Δ for 1 time marker per 2 divisions over the center 8 divisions. Note end of sweep timing over the center 8 divisions (Horizontal Position control CCW). Return Horizontal Position control CW to locate beginning of sweep. Some compromise of the Δ adjustment may be necessary to obtain best timing accuracy over the center 8 divisions at the start and end of sweep.
[e]Steps 32, 33, and 34 are for 2445B instruments with serial numbers B011133 and above.

NOTE

If the remainder of the Adjustment Procedure will not be performed (in totality), readjustment of Horizontal Readout Jitter (R805) may be necessary (see subsection 3) if the X1 Gain (R860) or the X10 Gain (R850) was changed.

kk. Disconnect the test setup.

CAL 02—VERTICAL

2. Check/Adjust Vertical Preamplifier Gain, Gain (R638), and Vertical Centering (R639).

NOTE

If the previous step (CAL 01) was not performed, the adjustments in this subsection should only be performed if those constants that would have been set in CAL 01 are known to be correct.

a. Set the front-panel INTENSITY control at midrange.

b. Scroll to CAL 02.

c. Press and release the upper TRIGGER COUPLING switch. The instrument will automatically increment through steps 100 to 110.

d. CHECK—Readout indicates CH 1 VAR, CH2 POS, (step) 111, and 500 mV.

NOTE

The readout prompts the operator by showing the controls to be moved (upper left corner and upper center), the autocal step number (upper right corner), the amplitude of signal to be applied to either the CH 1 or CH 2 connectors (lower left corner), and any other scope function that is enabled. An example (from step d above) is:

CH1 VAR CH2 POS 111
500 mV

e. Connect a 0.5-V, standard-amplitude signal from the Calibration Generator to the CH 1 OR X input connector via a 50-Ω BNC cable.

f. Use the CH 2 POSITION control to vertically position the trace to within 1 division of the center graticule line.

g. ADJUST—CH 1 POSITION and VOLTS/DIV VAR controls to obtain a 10-division horizontal signal. Press and release the upper TRIGGER COUPLING switch.

NOTE

When step 111 is performed, step 112 is also automatically done. No indication of step 112 will be shown unless a LIMIT error is indicated.

NOTE

In the following steps, if the "LIMIT" message appears, it probably indicates that the TRIGGER COUPLING (step) switch was moved before the required signal was applied. Press and release the lower TRIGGER COUPLING switch, verify that the correct signal is applied, then press and release the upper TRIGGER COUPLING switch.

h. CHECK—First step number listed in Table 5-4 appears in the readout.

i. Apply the corresponding standard-amplitude signal from the Calibration Generator, then press and release the upper TRIGGER COUPLING switch.

Table 5-4
Vertical Calibration Signals

Autocal Step Readout Display	Standard-Amplitude Signal to Apply
113, 114[a]	0.5 V
115	0.2 V
116	0.1 V
117	50 mV
118	20 mV
119	1 V
120	10 V

[a]When step 113 is performed, step 114 is also automatically done. No indication of step 114 will be shown unless a LIMIT error is encountered.

j. Repeat steps h and i for each step-signal combination listed in Table 5-4.

k. Move the signal to the CH 2 input connector.

l. CHECK—Readout indicates CONNECT SIGNAL TO CH 2, (step) 121, 500 mV, 500 mV, and BWL.

m. Set the Calibration Generator for a 500-mV standard-amplitude signal, then press and release the upper TRIGGER COUPLING switch.

NOTE

When step 121 is performed, step 122 is also automatically done. No indication of step 122 will be shown unless a LIMIT error is indicated.

n. CHECK—First step number listed in Table 5-5 appears in the readout.

o. Apply the corresponding standard-amplitude signal, then press and release the upper TRIGGER COUPLING switch.

p. Repeat steps n and o for each step-signal combination listed in Table 5-5.

Table 5-5
Vertical Calibration Signals

Autocal Step Readout Display	Standard-Amplitude Signal to Apply
123, 124[a]	0.5 V
125	0.2 V
126	0.1 V
127	50 mV
128	20 mV Limit
129	1 V
130, 131[b]	10 V

[a]When step 123 is performed, step 124 is also automatically done. No indication of step 124 will be shown unless a LIMIT error is encountered.

[b]When step 130 is performed, step 131 is also automatically done. No indication of step 131 will be shown unless a LIMIT error is encountered.

q. CHECK—Procedure automatically steps through steps 132-141.

r. CHECK—Readout indicates CONNECT SIGNAL TO CH 1, 50 mV, and BWL.

s. Move the signal to the CH 1 OR X input connector and set the Calibration Generator for a 50-mV standard-amplitude signal, then press and release the upper TRIGGER COUPLING switch. Wait approximately 10 seconds for automatic calibration of the ΔV cursors.

t. CHECK—Readout indicates VERT CENTER GAIN.

u. ADJUST—Gain (R638) for precisely 5 divisions between the two horizontal cursors.

v. ADJUST—Vertical Centering (R639) to center the cursors on the graticule (align the cursors with the dotted 0% and 100(%) graticule lines).

w. Press and release the upper TRIGGER COUPLING switch. The microprocessor continues calibrating the vertical.

CAL 03—TRIGGERING

3. Check/Adjust Triggering.

NOTE

If the previous steps (CAL 01 and CAL 02) were not performed, the adjustments in this subsection should only be performed if those constants that would have been set in CAL 01 and CAL 02 are known to be correct and if a DC Balance has been performed after a 20-minute warmup period.

a. Scroll to CAL 03.

b. Press and release the upper TRIGGER COUPLING switch.

c. CHECK—Procedure automatically steps from 200 through 214 and stops at 215.

d. CHECK—Readout indicates CH 1, 500 mV, and (step) 215.

NOTE

The readout prompts the operator by showing to which connector the input signal should be applied (upper left corner), the amplitude of that signal (upper center), and the autocal step number (upper right corner). An example (from step d above) is:

CH1 500 mV 215

e. Connect a 0.5-V standard-amplitude signal from the Calibration Generator to the CH 1 OR X input connector via a 50-Ω BNC cable.

f. Press and release the upper TRIGGER COUPLING switch.

g. CHECK—Readout indicates CH 1, 500 mV, and (step) 216.

h. Press and release the upper TRIGGER COUPLING switch.

i. CHECK—Readout indicates CH 2, 500 mV, and (step) 217.

j. Move the signal to the CH 2 input connector. Press and release the upper TRIGGER COUPLING switch.

k. CHECK—Readout indicates CH 3, 500 mV, and (step) 218.

l. Move the signal to the CH 3 input connector. Press and release the upper TRIGGER COUPLING switch.

m. CHECK—Readout indicates CH 3, 2V, and (step) 219.

n. Change the generator output level to 2 V, then press and release the upper TRIGGER COUPLING switch.

o. CHECK—Readout indicates CH 4, 500 mV, and (step) 220.

p. Move the signal to the CH 4 input connector and change the generator output level to 0.5 V. Press and release the upper TRIGGER COUPLING switch.

q. CHECK—Readout indicates CH 4, 2V, and (step) 221.

r. Change the generator output level to 2 V, then press and release the upper TRIGGER COUPLING switch.

s. Disconnect the test setup.

CAL 04—CH 2 DELAY ENABLE/DISABLE

4. Check/Adjust CH 2 Delay Enable/Disable.

a. Scroll to CAL 04.

b. Press and release the upper TRIGGER COUPLING switch to initiate the routine.

c. CHECK—Readout alternately indicates "ENABLED" and "DISABLED" each time the upper TRIGGER COUPLING switch is pressed and released.

d. Leave the readout display indicating "ENABLED". Press and release the A/B TRIG button to exit the routine.

e. Connect a 100-kHz, positive-going signal from the Calibration Generator fast-rise output to the CH 1 OR X and CH 2 input connectors via a 50-Ω BNC cable, a 5X attenuator, and a Dual-Input Coupler.

f. Set:

VERTICAL

CH 1 and CH 2	On
CH 1 and CH 2 VOLTS/DIV	10 mV
CH 1 and CH 2 Input Coupling	50 Ω DC

HORIZONTAL

A SEC/DIV	10 ns (knob in)

TRIGGER

SOURCE	CH 1
MODE	AUTO LVL
COUPLING	DC
SLOPE	+ (plus)

g. Set the generator amplitude for a 3- to 5-division display amplitude. Use the CH 1 and CH 2 POSITION controls to vertically overlay the traces near the center of the graticule area.

h. Set the Horizontal POSITION control to set the rising edge of the signals near the center vertical graticule line.

i. Press and release the X10 MAG button to obtain a magnified display.

j. Pull out the SEC/DIV knob.

k. CHECK—That the readout indicates **"CH 2 DLY-TURN Δ"** and that the Δ control will move the leading edge of the CH 2 trace at least 0.5 division to either side of the CH 1 trace.

l. ADJUST—Δ control to superimpose the leading edges.

m. Push in the SEC/DIV knob.

NOTE

If the CH 2 Delay Adjust feature is to be disabled for normal instrument use, perform the following steps; otherwise, proceed to CAL 05.

n. Reenter the Diagnostic Monitor by pressing the ΔV and Δt buttons simultaneously (hold them in), then press and hold the TRIGGER SLOPE button. Release the buttons after about 1 second.

o. Scroll to CAL 04.

p. Press and release the upper TRIGGER COUPLING switch until the readout indicates "DISABLED."

q. Press and release the A/B TRIG button to return to normal operating mode.

5. Check/Adjust Hours On and Power On/Off cycles.

a. Scroll to CAL 05.

b. Press and release the upper TRIGGER COUPLING switch to initiate the routine.

c. CHECK—Readout indicates **HRS ON xxx PWR ON/OFF xxx** and **ΔΔ REF HRS Δ PWR PUSH MAG 10/1.**

d. Press and release the lower TRIGGER SOURCE and then press and release the lower TRIGGER MODE to reset HRS ON and PWR ON/OFF to zero.

NOTE

HRS ON and PWR ON/OFF can be set to any value from 0-99999 with the Δ REF and Δ controls.

e. Press and release the lower TRIGGER COUPLING switch to exit routine.

CAL 06-VERTICAL TRANSIENT RESPONSE

6. Check/Adjust Vertical Transient Response

NOTE

If CAL 02 was not performed, the adjustments in this subsection should only be performed if those constants that would have been set in CAL 02 are known to be correct.

a. Scroll to CAL 06.

b. Press and release the upper TRIGGER COUPLING button to initiate the routine.

c. CHECK—Readout indicates ADJ Δ (step) 1, 10 mV, 100 ns.

d. Connect the high-amplitude output of the Calibration Generator to the CH 1 OR X input connector via a 50 ohm BNC cable, a Tunnel Diode Pulser, and a 5X attenuator.

e. Set the generator Period switch to 100 kHz, and set the generator amplitude control to maximum.

f. Rotate the pulser Trigger control CW (from a fully CCW position) until a stable pulse first appears on the graticule. Over adjustment of the pulser Trigger control will lead to erroneous transient response adjustment. Display amplitude will be approximately 5 divisions. The oscilloscope TRIGGER LEVEL control may need to be adjusted to obtain a stable trigger.

NOTE

As a guide when performing the following adjustments, optimum performance is achieved when the CH 1 and CH 2 step response aberrations are ≤ 4% over the first 10 ns of the pulse when using 10 mV/division deflection factors (≤ 0.2 division on a 5-division signal).

g. Press and release the upper TRIGGER COUPLING button twice to advance to step 3.

h. CHECK—Readout indicates ADJ Δ (step) 3, 10 mV, 10 ns.

i. ADJUST—Trans Resp Adjustments C403, R410, (R417 on 2455B instruments) , and Δ for flattest corner over first 5 ns. The total system will tune up best if the indicator cursor is in the 8th or 9th horizontal division.

NOTE

Inductor L403 is a selectable component chosen to match transient response characteristics of the Vertical system. If spreading the coil turns will not correct the front corner overshoot, a smaller value coil should be installed. Likewise, a larger coil can be installed to raise the front corner. The proper coils to use are:

90 nH-5	*turn*	*inductor*	*Part No.*	*108-0620-00*
80 nH-4	*turn*	*inductor*	*Part No.*	*108-0552-00*
60 nH-3	*turn*	*inductor*	*Part No.*	*108-0420-00*
45 nH-2	*turn*	*inductor*	*Part No.*	*108-0578-00*

j. Turn A SEC/DIV VAR control CCW and ADJUST CRT termination (R1501) for flattest waveform over the first 0.2 division.

k. Set SEC/DIV VAR to detent.

l. Press and release the upper TRIGGER COUPLING button.

m. CHECK—Readout indicates ADJ Δ (step) 4, 10 mV, 100 ns.

n. Connect the high amplitude generator, Tunnel Diode Pulser, 5X attenuator combination to CH 2 input via a 50 ohm BNC cable.

NOTE

Pressing the lower TRIGGER COUPLING button at any step of CAL 06 will return to step 1. By then pressing the upper TRIGGER COUPLING button repeatedly, the routine can be advanced to the desired step. This is useful for cal steps 1, 2, 3, and 4 which may require some compromise of adjustments.

o. ADJUST—Δ for the flattest waveform.

NOTE

Some compromise may be necessary between step 3 and 4 for the flattest corner over first 5 ns.

p. Press and release the lower TRIGGER COUPLING button to return to step 1.

q. Connect the high amplitude generator, Tunnel Diode Pulser, 5X attenuator combination to CH 1 input via a 50 ohm BNC cable.

r. ADJUST—Trans Resp adjustments (R605, R403, C404, C601, and R1501) for the flattest response in the first 100 ns.

s. Press and release the upper TRIGGER COUPLING button.

t. CHECK—Readout indicates ADJ Δ (step) 2, 10 mV, 100 ns.

u. Connect the high amplitude generator, Tunnel Diode Pulser, 5X attenuator combination to CH 2 input via a 50 ohm BNC cable.

NOTE

Some compromise may be necessary between step 1 and 2 for the flattest response in the first 100 ns.

v. Press and release the lower TRIGGER COUPLING button to return to step 1.

w. Disconnect the Calibration Generator and connect the Secondary Leveled Sine-Wave Generator head to the CH 1 input via a 10X attenuator.

x. Set the generator for a 6-division display at the reference frequency.

y. Change the generator output frequency to 150 MHz for the 2445B or to 250 MHz for the 2455B.

z. CHECK—Display amplitude is between 4.4 divisions and 6 divisions while the generator frequency is changed from 150 MHz to 100 MHz for the 2445B or from 250 MHz to 150 MHz for the 2455B. This bandwidth provides optimum performance of the Vertical system.

aa. Press and release the upper TRIGGER COUPLING switch.

bb. Check—Readout indicates ADJ Δ (step) 2, 10 mV, 10 ns.

cc. Connect the Secondary Leveled Sine-Wave Generator head to the CH 2 input via a 10X attenuator. Repeat steps x through aa for CH 2.

dd. Connect the high amplitude generator, Tunnel Diode Pulser, 5X attenuator combination to CH 1 OR X input via a 50 Ohm BNC cable.

ee. Check—Readout indicates ADJ Δ (step) 3, 10 mV, 10 ns.

ff. ADJUST—Trans Resp adjustments R410, C403, (R417 on the 2455B instrument) and the Δ control for best response if necessary.

gg. Disconnect the Tunnel Diode pulser and connect the fast rise output of the Calibration Generator to CH 1 OR X via a 5X attenuator and a 50 ohm BNC cable. Adjust generator amplitude for a 5 division display. Note the amount of roll up or roll down in the first 3 ns. This difference represents the error between the Tunnel Diode pulser (reference) and the fast rise generator output.

hh. Press and release the upper TRIGGER COUPLING switch (step 4). Move the test signal to CH 2 and ADJUST amplitude for 5 division signal.

ii. ADJUST—Δ until CH 2 waveform best matches that noted in step ff above.

jj. Press and release the upper TRIGGER COUPLING switch (step 5). Connect the test signal to CH 1 through 2X, 2.5X, and 5X attenuators. ADJUST Δ for best front corner.

kk. Press and release the upper TRIGGER COUPLING switch (step 6). Remove 2.5X attenuator. ADJUST Δ for best corner.

NOTE

The 5 mV response should have a 4-5% front corner spike to maintain correct bandwidth.

ll. Press and release the upper TRIGGER COUPLING switch (step 7). Remove 5X attenuator. ADJUST generator for a 5 division signal. ADJUST Δ for best corner.

mm. Press and release the upper TRIGGER COUPLING switch (step 8). Adjust generator for 5 division signal. ADJUST Δ for best corner.

nn. Press and release the upper TRIGGER COUPLING switch (step 9). Adjust high amplitude generator for 5 division signal. ADJUST Δ for a front corner spike of 6 to 7%. This is necessary to have the 10X bandwidth (0.1V - 0.5V) be similar to the 10 mV bandwidth.

NOTE

Generator amplitude for the 500 mV step will be approximately 2 divisions and the amplitude for the 1 V step will be approximately 1 division.

oo. Press and release the upper TRIGGER COUPLING switch (step 10). Remove 2X attenuator. ADJUST Δ for best corner. Continue through cal step 12 as above.

pp. Press and release the upper TRIGGER COUPLING switch. Steps 13 and 14 are automatically calibrated. Connect test signal to CH 2 via 2X, 2.5X, and 5X attenuators and adjust for 5 division signal. ADJUST Δ for best corner.

qq. Repeat steps kk through oo for CH 2 (steps 16–22).

rr. Steps 23 and 24 are automatically calibrated after step 22.

ss. Disconnect the generator from the CH 2 input.

tt. CHECK—Readout indicates VERT CENTER GAIN.

uu. ADJUST —Gain (R638) and Vertical Centering (R639) to align the cursors with the dotted 0% and 100% graticule markings.

vv. Press and release the upper TRIGGER COUPLING switch to conclude the calibration routine.

CAL 07—READOUT CENTERING AND GAIN

7. Check/adjust Readout Centering and Gain (R2918, R2931)

a. Scroll to CAL 07.

b. Press and release the upper TRIGGER COUPLING switch to initiate the routine.

c. CHECK—Readout displays large 8 characters moving in the top line and BWL characters moving in the bottom line.

d. ADJUST—Readout Centering (R2918) and Gain (R2931) so characters remain just inside the graticule area.

e. Press and release the lower TRIGGER COUPLING switch.

CAL 09—PARAMETRIC MEASUREMENTS

NOTE

At the end of this calibration procedure, move the Cal/No-Cal jumper (P501) to the No-Cal position (between pins 1 and 2).

8. Adjust Parametric measurements

NOTE

If CAL 01, 02, and 03 were not performed, the adjustments in this subsection should only be performed if those constants that would have been set in CAL 01, 02 and 03 are known to be correct.

"Limit" messages that appear during this calibration are generally due to A or B Sweeps, A or B Gates, or the measurement PAL, U975.

a. Scroll to CAL 09.

b. Press and release the upper TRIGGER COUPLING button to initiate the routine.

c. CHECK—Readout indicates (step)1, CAL 09, 100 ns.

NOTE

The readout prompts the operator by showing the autocal step number (upper right corner) and Time-Marker Generator setting (lower right corner).

d. Connect the Time-Mark Generator, set for 0.1 μs time markers, to the CH 1 OR X input connector via a 50 ohm BNC cable.

e. Press and release the upper TRIGGER COUPLING button to calibrate the step.

f. CHECK—Readout indicates (step)2, CAL 09, 100 ns.

g. For the remaining steps in Table 5-6, do the following:

1. Set the Time-Marker Generator output for markers corresponding to the Step Number.
2. Press and release the upper TRIGGER COUPLING button to calibrate the step.

Table 5-6
Parametric Measurement Calibration

Autocal Step Readout Display	Time Markers to Apply	Autocal Step Readout Display	Time Markers to Apply
2	0.1 μs	10	50 μs
3	0.2 μs	11	0.1 ms
4	0.5 μs	12	0.2 ms
5	1 μs	13	0.5 ms
6	2 μs	14	1 ms
7	5 μs	15	2 ms
8	10 μs	16	5 ms
9	20 μs	17[a]	0.2 ms

[a]At the conclusion of step 17 calibration, the instrument returns to the Diagnostic readout display. Disconnect the Time-Mark Generator at this time.

h. Steps 18 through 28 are automatically calibrated by the system processor.

i. CHECK—Calibration is concluded and the instrument returns to the Diagnostic menu.

j. Disconnect the Time-Mark generator.

9. Adjust Bandwidth Limit

a. Set:

Vertical

CH 1 POSITION	Midrange
CH 1 MODE	On
CH 2, CH 3, and CH 4 MODE	Off
20 MHz BW LIMIT	On
CH 1 VOLTS/DIV	10 mV
CH 1 VAR	In detent

Input Coupling

CH 1	1 MΩ DC

Horizontal

POSITION	Midrange
X10 MAG	Off
A SEC/DIV	100 ns (knob in)
SEC/DIV VAR	In detent

Trigger

HOLDOFF	MIN (Fully CCW)
LEVEL	Midrange
A/B TRIG	A
SLOPE	+
MODE	AUTO LVL
SOURCE	VERT
COUPLING	DC

b. Connect a fast-rise, positive-going square-wave output via a precision 50-ohm cable, a 50-ohm 10X attenuator, and a 50-ohm termination to the CH 1 input connector.

c. Set the generator to produce a 100-kHz, 5-division display.

d. ADJUST—Coil L644 for as flat a response as possible.

e. Disconnect the test equipment from the instrument.

DC BALANCE, AND X-Y PHASE DIFFERENTIAL ADJUSTMENTS

Equipment Required (see Table 4-1)
Primary Leveled-Sine wave Generator (Item 2)
Calibration Generator (Item 3)
50 Ω BNC Cable (Item 10)
5X Attenuator (Item 17)
Alignment Tool (Item 20)

See **ADJUSTMENT LOCATIONS 1** *and* **ADJUSTMENT LOCATIONS 4** *at the back of this manual for test point and adjustment locations.*

Initial Control Settings.

Control settings not listed do not affect the procedure.

VERTICAL VOLTS/DIV

CH 1 and CH 2	10 mV
CH 1 VAR	CCW (out of detent)
CH 2 VAR	In detent

Input Coupling

CH 1 and CH 2	50 Ω DC

VERTICAL Mode

CH 1	On
CH 2, CH 3, CH 4	Off
ADD, INVERT, and BW LIMIT	Off
ALT/CHOP	ALT

VERTICAL POSITION

CH 1	Midrange

Horizontal

A SEC/DIV	1 ms
SEC/DIV VAR	In detent
POSITION	Midrange

TRIGGER

MODE	AUTO LVL
SOURCE	VERT
COUPLING	DC
SLOPE	+ (plus)
LEVEL	Midrange
HOLDOFF	Minimum

Delta

ΔV	On (RATIO readout)
TRACKING/INDEP	INDEP
Δ REF OR DLY POS and Δ	Cursors near the 3rd line above and 3rd line below graticule center (6 division spacing)
INTENSITY	Left of center
READOUT INTENSITY	Right of center
SCALE ILLUM	Fully CCW
FOCUS	Best focused display

1. Check/Adjust Readout Jitter (R805 and R618).

a. Rotate the Δ REF OR DLY POS control CCW until the RATIO readout is constant.

b. Rotate the Δ control until the readout display indicates 130.0%.

c. CHECK—One cursor is near the bottom horizontal graticule line and the other is near dotted graticule line marked 100(%).

d. Rotate the Δ REF OR DLY POS control until the readout displays exactly 100.0%. The cursors should now be on or near the dotted graticule lines marked 0% and 100(%).

e. Set the CH 1 VOLTS/DIV VAR to the detent position.

NOTE

Care must be taken not to disturb the position of the controls adjusted in parts b through e during the balance of this procedure. If they are accidentally moved, repeat the procedure from the beginning.

f. Connect a 1 kHz, fast-rise signal from the Calibration Generator to the CH 1 OR X input connector via a 50 Ω BNC cable and 5X attenuator.

g. Set the generator output level for an 8-division display.

h. Use the CH1 Vertical and the Horizontal POSITION controls to center the CH 1 display on the graticule.

i. ADJUST—Vertical Readout Jitter (R618) for minimum vertical jitter of the readout characters and cursors.

j. ADJUST—Gain (R638) and Centering (R639) to align cursors with the 0 and 100% graticule markings.

k. Disconnect the 1 kHz signal.

l. Press the Δt button to obtain a Δt cursor display.

m. Using the ΔREF OR DLY POS and Δ controls, position the cursors to the 2nd and 10th graticule lines.

n. X10 MAG on.

o. ADJUST—Horizontal Readout Jitter (R805) for minimum horizontal jitter of the readout characters and cursors.

p. Set X10 MAG off.

2. Set CH 1 and CH 2 DC Balance.

NOTE

The instrument must have had a 20-minute warmup prior to performing the next step to ensure accuracy.

a. Press and hold momentarily and release the CH 1 and CH 2 upper Input Coupling switches simultaneously.

b. CHECK—The display reads DC BALANCE IN PROGRESS for approximately 10 seconds, then the display returns to normal.

c. CHECK—For less than 0.2-division + 0.5 mV vertical trace shift when the CH 1 VOLTS/DIV switch is rotated through all of its settings.

d. Set the VERTICAL MODE switches to disable CH 1 and display CH 2.

e. CHECK—For less than 0.2-division + 0.5 mV vertical trace shift when the CH 2 VOLTS/DIV switch is rotated through all of its settings.

3. Adjust X-Y Phasing (C118).

a. Set:

CH 1 VOLTS/DIV	50 mV
Input Coupling	50 Ω DC
A SEC/DIV	X-Y
CH 1 VERTICAL MODE	On
CH 2, CH 3, CH 4 VERTICAL MODE	Off

b. Connect the Primary Leveled Sine-Wave Generator to the CH 1 OR X input connector via a 50 Ω BNC cable.

c. Set the generator frequency to 1 MHz and adjust the amplitude for a 6-division vertical signal display.

d. Use the CH 1 POSITION control to vertically center the display on the graticule.

e. ADJUST—X-Y Phasing (C118) for no opening in the ellipse.

f. Set the generator frequency to 2 MHz and adjust the amplitude for a 6-division vertical signal display.

g. CHECK—Horizontal opening in the ellipse is 0.3 division or less, measured at the center horizontal graticule line.

h. Disconnect the test setup.

MAINTENANCE

This section of the manual contains information for conducting preventive maintenance, troubleshooting, and corrective maintenance on the instruments.

STATIC-SENSITIVE COMPONENTS

The following precautions are applicable when performing any maintenance involving internal access to the instrument.

CAUTION

Static discharge can damage any semiconductor component in this instrument.

This instrument contains electrical components that are susceptible to damage from static discharge. Table 6-1 lists the relative susceptibility of various classes of semiconductors. Static voltages of 1 kV to 30 kV are common in unprotected environments.

When performing maintenance, observe the following precautions to avoid component damage:

1. Minimize handling of static-sensitive components.
2. Transport and store static-sensitive components or assemblies in their original containers or on a metal rail. Label any package that contains static-sensitive components or assemblies.
3. Discharge the static voltage from your body by wearing a grounded antistatic wrist strap while handling these components. Servicing static-sensitive components or assemblies should be performed only at a static-free work station by qualified service personnel.
4. Nothing capable of generating or holding a static charge should be allowed on the work station surface.
5. Keep the component leads shorted together whenever possible.
6. Pick up components by their bodies, never by their leads.

Table 6-1
Susceptibility to Static Discharge Damage

Semiconductor Classes	Relative Susceptibility Levels[a]
MOS or CMOS microcircuits or discretes, or linear microcircuits with MOS inputs. (Most Sensitive)	1
ECL	2
Schottky signal diodes	3
Schottky TTL	4
High-frequency bipolar transistors	5
JFETs	6
Linear microcircuits	7
Low-power Schottky TTL	8
TTL (Least Sensitive)	9

[a]Voltage equivalent for levels: (Voltage discharged from a 100 pF capacitor through a resistance of 100 Ω).

1 = 100 to 500 V
2 = 200 to 500 V
3 = 250 V
4 = 500 V
5 = 400 to 600 V
6 = 600 to 800 V
7 = 400 to 1000 V (est.)
8 = 900 V
9 = 1200 V

7. Do not slide the components over any surface.
8. Avoid handling components in areas that have a floor or work-surface covering capable of generating a static charge.

9. Use a soldering iron that is connected to earth ground.

10. Use only approved antistatic, vacuum-type desoldering tools for component removal.

PREVENTIVE MAINTENANCE

INTRODUCTION

Preventive maintenance consists of cleaning, visual inspection, and checking instrument performance. When accomplished regularly, it may prevent instrument malfunction and enhance instrument reliability. The severity of the environment in which the instrument is used determines the required frequency of maintenance. An appropriate time to accomplish preventive maintenance is just before instrument adjustment.

GENERAL CARE

The cabinet minimizes accumulation of dust inside the instrument and should normally be in place when operating the instrument. The front cover supplied with the instrument provides both dust and damage protection for the front panel and CRT, and it should be on whenever the instrument is stored or is being transported.

INSPECTION AND CLEANING

The instrument should be visually inspected and cleaned as often as operating conditions require. Accumulation of dirt in the instrument can cause overheating and component breakdown. Dirt on components acts as an insulating blanket, preventing efficient heat dissipation. It also provides an electrical conduction path that could result in instrument failure, especially under high-humidity conditions.

CAUTION

Avoid the use of chemical cleaning agents which might damage the plastics used in this instrument. Use a nonresidue-type cleaner, preferably isopropyl alcohol or a solution of 1% mild detergent with 99% water. Before using any other type of cleaner, consult your Tektronix Service Center or representative.

Exterior

INSPECTION. Inspect the external portions of the instrument for damage, wear, and missing parts; use Table 6-2 as a guide. Instruments that appear to have been dropped or otherwise abused should be checked thoroughly to verify correct operation and performance. Deficiencies found that could cause personal injury or could lead to further damage to the instrument should be repaired immediately.

Table 6-2
External Inspection Check List

Item	Inspect For	Repair Action
Cabinet, Lid, Front Panel	Cracks, scratches, deformations, damaged hardware or gaskets.	Touch up paint scratches and replace defective components.
Front-Panel Controls	Missing, damaged, or loose knobs, buttons, and controls.	Repair or replace missing or defective items.
Connectors	Broken shells, cracked insulation, and deformed contacts. Dirt in connectors.	Replace defective parts, Clear or wash out dirt.
Carrying Handle	Correct operation.	Replace defective parts.
Accessories	Missing items or parts of items, bent pins, broken or frayed cables, and damaged connectors.	Replace damaged or missing items, frayed cables, and defective parts.

CAUTION

To prevent getting moisture inside the instrument during external cleaning, use only enough liquid to dampen the cloth or applicator.

CLEANING. Loose dust on the outside of the instrument can be removed with a soft cloth or small soft-bristle brush. The brush is particularly useful for dislodging dirt on and around the controls and connectors. Dirt that remains can be removed with a soft cloth dampened in a mild detergent- and-water solution. Do not use abrasive cleaners.

Two plastic light filters, one blue and one clear, are provided with the oscilloscope. Clean the light filters and the CRT face with a soft lint-free cloth dampened with either isopropyl alcohol or a mild detergent-and-water solution.

Interior

To gain access to internal portions of the instrument for inspection and cleaning, refer to the "Removal and Replacement Instructions" in the "Corrective Maintenance" part of this section.

INSPECTION. Inspect the internal portions of the instrument for damage and wear, using Table 6-3 as a guide. Deficiencies found should be repaired immediately. The corrective procedure for most visible defects is obvious; however, particular care must be taken if heat-damaged components are found. Overheating usually indicates other trouble in the instrument, therefore it is important that the cause of overheating be corrected to prevent recurrence of the damage.

If any electrical component is replaced, conduct a Performance Check for the affected circuit and for other closely related circuits (see Section 4). If repair or replacement work is done on any of the power supplies, conduct a complete Performance Check and, if so indicated, an instrument readjustment (see Sections 4 and 5).

Table 6-3
Internal Inspection Check List

Item	Inspect For	Repair Action
Circuit Boards	Loose, broken, or corroded solder connections. Burned circuit boards. Burned, broken, or cracked circuit-run plating.	Clean solder corrosion with an eraser and flush with isopropyl alcohol. Resolder defective connections. Determine cause of burned items and repair. Repair defective circuit runs.
Resistors	Burned, cracked, broken, blistered.	Replace defective resistors. Check for cause of burned component and repair as necessary.
Solder Connections	Cold solder or rosen joints.	Resolder joint and clean with isopropyl alcohol.
Capacitors	Damaged or leaking cases. Corroded solder on leads or terminals.	Replace defective capacitors. Clean solder connections and flush with isopropyl alcohol.
Semiconductors	Loosely inserted in sockets. Distorted pins.	Firmly seat loose semiconductors. Remove devices having distorted pins. Carefully straighten pins (as required to fit the socket), using long-nose pliers, and reinsert firmly. Ensure that straightening action does not crack pins, causing them to break off.
Wiring and Cables	Loose plugs or connectors. Burned, broken, or frayed wiring.	Firmly seat connectors. Repair or replace defective wires or cables.
Chassis	Dents, deformations, and damaged hardware.	Straighten, repair, or replace defective hardware.

CAUTION

To prevent damage from electrical arcing, ensure that circuit boards and components are dry before applying power to the instrument.

CLEANING. To clean the interior, blow off dust with dry, low-pressure air (approximately 9 psi). Remove any remaining dust with a soft brush or a cloth dampened with a solution of mild detergent and water. A cotton-tipped applicator is useful for cleaning in narrow spaces and on circuit boards.

If these methods do not remove all the dust or dirt, the instrument may be spray washed using a solution of 5% mild detergent and 95% water as follows:

CAUTION

Exceptions to the following procedure are the Attenuator assemblies. Clean these assemblies only with isopropyl alcohol as described in step 4.

1. Gain access to the parts to be cleaned by removing easily accessible shields and panels.

2. Spray-wash dirty parts with the detergent-and-water solution, then use clean water to thoroughly rinse them.

3. Dry all parts with low-pressure air.

NOTE

Most of the switches used in the instrument are sealed and the contacts are inaccessible. If cleaning is deemed necessary, use only isopropyl alcohol.

4. Clean switches with isopropyl alcohol and wait 60 seconds for the majority of the alcohol to evaporate. Then complete drying with low-pressure air.

5. Dry all components and assemblies in an oven or drying compartment using low-temperature (125°F to 150°F) circulating air.

LUBRICATION

There is no periodic lubrication required for this instrument.

SEMICONDUCTOR CHECKS

Periodic checks of the transistors and other semiconductors in the oscilloscope are not recommended. The best check of semiconductor performance is actual operation in the instrument.

PERIODIC READJUSTMENT

To ensure accurate measurements, check the performance of this instrument every 2000 hours of operation, or if used infrequently, once each year. In addition, replacement of components may necessitate readjustment of the affected circuits.

Complete Performance Check and Adjustment instructions are given in Sections 4 and 5. The Performance Check Procedure can also be helpful in localizing certain troubles in the instrument.

TROUBLESHOOTING

INTRODUCTION

Preventive maintenance performed on a regular basis should reveal most potential problems before an instrument malfunctions. However, should troubleshooting be required, the following information is provided to facilitate location of a fault. In addition, the material presented in the "Theory of Operation" and "Diagrams" sections of this manual may be helpful while troubleshooting.

TROUBLESHOOTING AIDS

Diagnostic Firmware

The operating firmware in this instrument contains diagnostic routines that aid in locating malfunctions. When instrument power is applied, power-up tests are performed to verify proper operation of much of the instrument's circuitry. If a failure is detected, this information is passed on to the operator in the form of either a CRT readout or illuminated LED indicators. The failure information directs the operator to the failing block of circuitry. If the failure is such that the processor can still execute the diagnostic routines, the user can call up specific tests to further check the failing circuitry. The specific diagnostic routines are explained later in this section.

Schematic Diagrams

Complete schematic diagrams are located on tabbed foldout pages in the "Diagrams" section. Portions of circuitry mounted on each circuit board are enclosed by heavy black lines. The assembly number and name of the circuit are shown near either the top or the bottom edge of the diagram.

Functional blocks on schematic diagrams are outlined with a wide grey line. Components within the outlined area perform the function designated by the block label. The "Theory of Operation" uses these functional block names when describing circuit operation as an aid in cross-referencing between the theory and the schematic diagrams.

Component numbers and electrical values of components in this instrument are shown on the schematic diagrams. Refer to the first page of the "Diagrams" section for the reference designators and symbols used to identify components. Important voltages and waveform reference numbers (enclosed in hexagonal-shaped boxes) are also shown on each diagram. Waveform illustrations are located adjacent to their respective schematic diagram.

Circuit Board Illustrations

Circuit board illustrations showing the physical location of each component are provided for use in conjunction with each schematic diagram. Each board illustration is found in the "Diagrams" section on the back of a foldout page, preceding the first schematic diagram(s) to which it relates.

The locations of waveform test points are marked on the circuit board illustrations with hexagonal outlined numbers corresponding to the waveform numbers on both the schematic diagram and the waveform illustrations.

Circuit Board Locations

The placement in the instrument of each circuit board is shown in a board locator illustration. This illustration is located on the foldout page along with the circuit board illustration.

Power Distribution Diagrams

Power Distribution diagrams (diagrams 11 and 12) are provided in the "Diagrams" section to aid in troubleshooting power-supply problems.

Circuit Board Interconnection Diagram

A circuit board interconnection diagram (diagram 13) and tables listing the interconnecting pins and signals carried are provided in the "Diagrams" section following the Power Distribution diagrams.

Grid Coordinate System

Each schematic diagram and circuit board illustration has a grid border along its left and top edges. A table located adjacent to each diagram lists the grid coordinates of each component shown on that diagram. To aid in physically locating components on the circuit board, this table also lists the grid coordinates of each component on the circuit board illustration.

Near each circuit board illustration is an alphanumeric listing of all components mounted on that board. The second column in each listing identifies the schematic diagram on which each component can be found. These component-locator tables are especially useful when more than one schematic diagram is associated with a particular circuit board.

Troubleshooting Charts

The troubleshooting charts contained in the "Diagrams" section are to be used as an aid in locating malfunctioning circuitry. To use the charts, begin with the Preliminary Tests flowchart. This chart will help identify problem areas and will direct you to other appropriate charts for further troubleshooting.

Some malfunctions, especially those involving multiple simultaneous failures, may require more elaborate troubleshooting approaches with references to circuit descriptions in the "Theory of Operation" section of this manual.

Component Color Coding

Information regarding color codes and markings of resistors and capacitors is located on the color-coding illustration (Figure 9-1) at the beginning of the "Diagrams" section.

RESISTOR COLOR CODE. Resistors used in this instrument are carbon-film, composition, or precision metal-film types. They are usually color coded with the EIA color code; however, some metal-film type resistors may have the value printed on the body. The color code is interpreted starting with the stripe nearest to one end of the resistor. Composition resistors have four stripes; these represent two significant digits, a multiplier, and a tolerance value. Metal-film resistors have five stripes representing three significant digits, a multiplier, and a tolerance value.

CAPACITOR MARKINGS. Capacitance values of common disc capacitors and small electrolytics are marked on the side of the capacitor body. White ceramic capacitors are color coded in picofarads, using a modified EIA code.

Dipped tantalum capacitors are color coded in microfarads. The color dot indicates both the positive lead and the voltage rating. Since these capacitors are easily destroyed by reversed or excessive voltage, be careful to observe the polarity and voltage rating when replacing them.

DIODE COLOR CODE. The cathode end of each glass-encased diode is indicated by either a stripe, a series of stripes or a dot. For most diodes marked with a series of stripes, the color combination of the stripes identifies three digits of the Tektronix Part Number, using the resistor color-code system. The cathode and anode ends of a metal-encased diode may be identified by the diode symbol marked on its body.

Semiconductor Lead Configurations

Figure 9-2 in the "Diagrams" section shows the lead configurations for semiconductor devices used in the instrument. These lead configurations and case styles are typical of those used at completion of the instrument design. Vendor changes and performance improvement changes may result in changes of case styles or lead configurations. If the device in question does not appear to match the configuration shown in Figure 9-2, examine the associated circuitry or consult a manufacturer's data sheet.

Multipin Connectors

Multipin connector orientation is indexed by a triangle on the cable connector and a 1 or triangle on the circuit board. Slot numbers may be molded into the connector. When a connection is made to circuit board pins or header, ensure that the index on the connector is aligned with the index on the circuit board (see Figure 6-1). Cable connectors can be removed by inserting a screw driver into the center slot of its header.

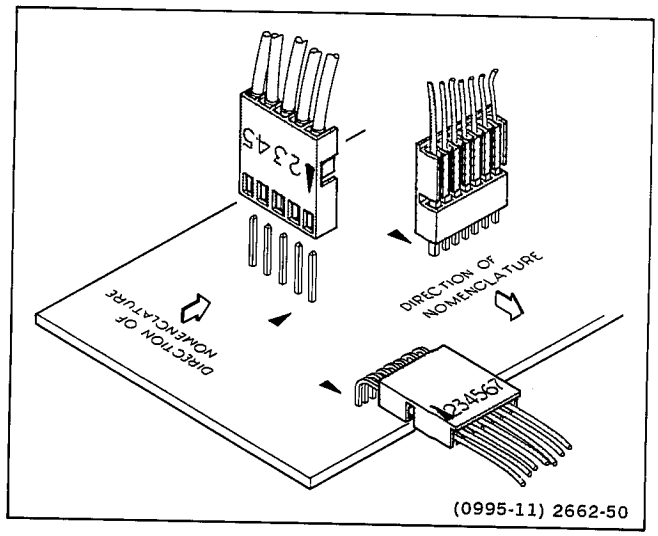


Figure 6-1. Multipin connector orientation.

TROUBLESHOOTING EQUIPMENT

The equipment listed in Table 4-1 of this manual, or equivalent equipment, may be useful when troubleshooting this instrument.

TROUBLESHOOTING TECHNIQUES

The following procedure is arranged in an order that enables checking simple trouble possibilities before requiring more extensive troubleshooting. The first two steps

use diagnostic aids inherent in the instrument's operating firmware and will locate many circuit faults. The next four procedures are check steps that ensure proper control settings, connections, operation, and adjustment. If the trouble is not located by these checks, the remaining steps will aid in locating the defective component. When the defective component is located, replace it using the appropriate replacement procedure given under "Corrective Maintenance" in this section.

CAUTION

Before using any test equipment to make measurements on static-sensitive, current-sensitive, or voltage-sensitive components or assemblies, ensure that any voltage or current supplied by the test equipment does not exceed the limits of the component to be tested.

1. Power-up Tests

The instrument performs automatic verification of much of the instrument's circuitry when power is first applied. The Kernel tests verify proper operation of the Microprocessor, the ROM, and the RAM. If all Kernel tests pass, a second level of checks, the Confidence tests, are performed. The Confidence tests, when passed, give the user a high degree of assurance that the instrument is functioning properly.

If a Kernel test or Confidence test fails, the area of failure is identified either by a message on the CRT (if the instrument is able to produce a display) or by an error code displayed on the front-panel LED indicators. If a failure occurs, refer to the "Diagnostic Routines" discussion later in this section for definitions of error messages and LED error codes.

Once a problem area has been identified, the associated troubleshooting procedure should be performed to further isolate the problem. The troubleshooting procedures are located on tabbed-foldout pages in the "Diagrams" section at the rear of this manual.

2. Diagnostic Test and Exerciser Routines

Each of the tests automatically performed at power-up, along with several other circuit exercising routines, may be individually selected by the user to further clarify the nature of a suspected failure. The desired test or exerciser is selected by scrolling through a menu of the available routines when under control of the Diagnostic Monitor. Entry into the Diagnostic Monitor and its uses are explained in the "Diagnostic Routines" discussion later in this section.

3. Check Control Settings

Incorrect control settings can give a false indication of instrument malfunction. If there is any question about the correct function or operation of any control, refer to the Operators Manual.

4. Check Associated Equipment

Before proceeding, ensure that any equipment used with the instrument is operating correctly. Verify that input signals are properly connected and that the interconnecting cables are not defective. Check that the ac-power-source voltage to all equipment is correct.

5. Visual Check

Perform a visual inspection. This check may reveal broken connections or wires, damaged components, semiconductors not firmly mounted, damaged circuit boards, or other clues to the cause of an instrument malfunction.

6. Check Instrument Performance and Adjustment.

Check the performance of either those circuits where trouble appears to exist or the entire instrument. The apparent trouble may be the result of misadjustment. Complete performance check and adjustment instructions are given in Sections 4 and 5 of this manual.

7. Isolate Trouble to a Circuit

To isolate problems to a particular area, use any symptoms noticed to help locate the trouble. Refer to the trouble-shooting charts in the "Diagrams" section as an aid in locating a faulty circuit.

When trouble symptoms appear in more than one circuit, first check the power supplies, then check the affected circuits by taking voltage and waveform readings. Check first for the correct output voltage of each individual supply. These voltages are measured between the power supply test points and ground. (See schematic diagrams 8, 9, 10, and associated circuit board illustrations in the "Diagrams" section.) If the power-supply voltages and ripple are within the listed ranges, the supply can be assumed to be working correctly. If they are outside the range, the supply may be either misadjusted or operating incorrectly.

The Low Voltage Power Supply levels are interdependent. All the low voltage supplies use the +10 V reference for their reference levels. If more than one of the low voltage supplies appears defective, repair them in the following order: +10 V REF, +5 V Digital, +87 V, +42 V, +15 V, +5 V Analog, −15 V, −8 V, and −5 V.

A defective component elsewhere in the instrument can create the appearance of a power-supply problem and may also affect the operation of other circuits. Use the power supply troubleshooting charts to aid in locating the problem.

8. Check Circuit Board Interconnections

After the trouble has been isolated to a particular circuit, again check for loose or broken connections, improperly seated semiconductors, and heat-damaged components.

9. Check Voltages and Waveforms

Often the defective component can be located by checking circuit voltages or waveforms. Typical voltages are listed on the schematic diagrams. Waveforms indicated on the schematic diagrams by hexagonal-outlined numbers are shown adjacent to the diagrams. Waveform test points are shown on the circuit board illustrations.

NOTE

Voltages and waveforms indicated on the schematic diagrams are not absolute and may vary slightly between instruments. To establish operating conditions similar to those used to obtain these readings, see the voltage and waveform setup conditions preceding the waveform illustrations.

Note the recommended test equipment, front-panel control settings, voltage and waveform conditions, and cable-connection instructions. Any special control settings required to obtain a given waveform are noted under the waveform illustration. Changes to the control settings from the initial setup, other than those noted, are not required.

10. Check Individual Components

The following procedures describe methods of checking individual components. Two-lead components that are soldered in place are most accurately checked by first disconnecting one end from the circuit board. This isolates the measurement from the effects of the surrounding circuitry. See Figure 9-1 for component value identification and Figure 9-2 for semiconductor lead configurations.

WARNING

To avoid electric shock, always disconnect the instrument from the ac power source before removing or replacing components.

CAUTION

When checking semiconductors, observe the static-sensitivity precautions located at the beginning of this section.

TRANSISTORS. A good check of a transistor is actual performance under operating conditions. A transistor can most effectively be checked by substituting a known-good component. However, be sure that circuit conditions are not such that a replacement transistor might also be damaged. If substitute transistors are not available, use a dynamic-type transistor checker for testing. Static-type transistor checkers are not recommended, since they do not check operation under simulated operating conditions.

When troubleshooting transistors in the circuit with a voltmeter, measure both the emitter-to-base and emitter-to-collector voltages to determine whether they are consistent with normal circuit voltages. Voltages across a transistor may vary with the type of device and its circuit function.

Some of these voltages are predictable. The emitter-to-base voltage for a conducting silicon transistor will normally range from 0.6 V to 0.8 V. The emitter-to-collector voltage for a saturated transistor is about 0.2 V. Because these values are small, the best way to check them is by connecting a sensitive voltmeter across the junction rather than comparing two voltages taken with respect to ground. If the former method is used, both leads of the voltmeter must be isolated from ground.

If voltage values measured are less that those just given, either the device is shorted or no current is flowing in the external circuit. If values exceed the emitter-to-base values given, either the junction is reverse biased or the device is defective. Voltages exceeding those given for typical emitter-to-collector values could indicate either a nonsaturated device operating normally or a defective (open-circuited) transistor. If the device is conducting, voltage will be developed across the resistors in series with it; if open, no voltage will be developed across the resistors unless current is being supplied by a parallel path.

CAUTION

When checking emitter-to-base junctions, do not use an ohmmeter range that has a high internal current. High current may damage the transistor. Reverse biasing the emitter-to-base junction with a high current may degrade the current-transfer ratio (Beta) of the transistor.

A transistor emitter-to-base junction also can be checked for an open or shorted condition by measuring the resistance between terminals with an ohmmeter set to a range having a low internal source current, such as the R × 1 kΩ range. The junction resistance should be very high in one direction and much lower when the meter leads are reversed.

When troubleshooting a field-effect transistor (FET), the voltage across its elements can be checked in the same manner as previously described for other transistors. However, remember that in the normal depletion mode of operation, the gate-to-source junction is reverse biased; in the enhanced mode, the junction is forward biased.

INTEGRATED CIRCUITS. An integrated circuit (IC) can be checked with a voltmeter, test oscilloscope, or by direct substitution. A good understanding of circuit operation is essential when troubleshooting a circuit having IC components. Use care when checking voltages and waveforms around the IC so that adjacent leads are not shorted together. An IC test clip provides a convenient means of clipping a test probe to an IC.

HYBRIDS. Hybrid components can best be checked by observing voltages and waveforms on the circuit board. Measurements should not be made on any hybrid component while out of the circuit as they may easily be damaged. Direct substitution is the best troubleshooting method when a hybrid failure is suspected. The CH 1 and CH 2 hybrids are matched, and should be replaced as a matched pair.

CAUTION

When checking a diode, do not use an ohmmeter scale that has a high internal current. High current may damage a diode. Checks on diodes can be performed in much the same manner as those on transistor emitter-to-base junctions. Do not check tunnel diodes or back diodes with an ohmmeter; use a dynamic tester, such as the TEKTRONIX 576 Curve Tracer.

DIODES. A diode can be checked for either an open or a shorted condition by measuring the resistance between terminals with an ohmmeter set to a range having a low internal source current, such as the R × 1 kΩ range. The diode resistance should be very high in one direction and much lower when the meter leads are reversed.

Silicon diodes should have 0.6 to 0.8 V across their junctions when conducting. Higher readings indicate that they are either reverse biased or defective, depending on polarity.

Light Emitting Diodes (LEDs) should have 1.5 to 2.2 V, depending on their current and color, across their junctions when conducting. Higher readings usually indicate the diodes are open, especially if they are not illuminated (ON).

RESISTORS. Check resistors with an ohmmeter. Refer to the "Replaceable Electrical Parts" list for the tolerances of resistors used in this instrument. A resistor normally does not require replacement unless its measured value varies widely from its specified value and tolerance.

INDUCTORS. Check for open inductors by checking continuity with an ohmmeter. Shorted or partially shorted inductors can usually be found by checking the waveform response when high-frequency signals are passed through the circuit.

CAPACITORS. A leaky or shorted capacitor can best be detected by checking resistance with an ohmmeter set to one of the highest ranges. Do not exceed the voltage rating of the capacitor. The resistance reading should be high after the capacitor is charged to the output voltage of the ohmmeter. An open capacitor can be detected with a capacitance meter or by checking whether the capacitor passes ac signals.

ATTENUATORS. The Attenuators are built as complete assemblies and should not be taken apart. If an Attenuator is suspected as having failed, direct substitution is the recommended troubleshooting method.

11. Repair and Adjust the Circuit.

If any defective parts are located, follow the replacement procedures given under "Corrective Maintenance" in this section. After any electrical component has been replaced, the performance of that circuit and any other closely related circuit should be checked. Since the power supplies affect all circuits, performance of the entire instrument should be checked if work has been done on the power supplies or if the power transformer has been replaced. Readjustment of the affected circuitry may be necessary. Refer to the "Performance Check" and "Adjustment Procedure", Sections 4 and 5 of this manual.

DIAGNOSTIC ROUTINES

The diagnostic routines contained in the instrument operating firmware consist of the various power-up tests, automatically performed when power is first applied, and several circuit exerciser routines. The test or exerciser routines are selected by scrolling through a menu of available routines when the firmware is under control of the Diagnostic Monitor. Monitor control is indicated by the message "DIAGNSTIC. PUSH A/B TRIG TO EXIT" displayed in the top CRT graticule division.

Entry into the monitor is automatic if a power-up test fails. The user may also force entry into the Diagnostic Monitor from the normal operating mode by holding in the front-panel ΔV and Δt push buttons and then pressing the front-panel SLOPE push button. Exiting the monitor is accomplished by pressing in the A/B TRIG push button, as instructed by the CRT readout display.

Depending on how the Diagnostic Monitor was entered (from normal mode or as a result of a power-up test failure), the first menu item displayed may vary; entry into the monitor from the normal mode begins at ALL TESTS while entry from power-up starts at the first failed test. Since, in a failure mode, the CRT readout may not be able to display the selected menu item, the VERT TRIGGER SOURCE indicator illuminates as a reference when ALL TESTS is selected. With the VERT TRIGGER SOURCE indicator illuminated, the user may scroll to the desired test or exerciser routine using the test order called out in Table 6-4 or Table 6-5 respectively. Whether the menu is displayed or not, scrolling is accomplished by pressing either the front-panel upper TRIGGER MODE switch to increment or the lower TRIGGER MODE switch to decrement the menu position by one.

Table 6-4
Sequence of Diagnostic Tests

Routine Type	Type Number	Lit LED	Routine Name	Error Code	Error Code Meaning
All Tests[a]	00	VERT	All	ZZ	The left digit is the option number and the right digit is the test number of the first failing test of the last ALL TESTS run. When looping, it shows the last failing test.
Test	00		Kernel Test	ZZ	Left digit is option number and right digit is device number. See Table 6-6 for main box kernel test failure codes.[d]
Test	01	CH 1	Interrupt Request	01	Interrupt request is missing or has wrong period.
Test	02	CH 2	Switch Stuck	01	Trigger COUPLING lower.
				02	Trigger COUPLING upper.
				03	MEASURE/HELP
				04	CH 1 Coupling lower.
				05	CH 1 Coupling upper.
				11	CH 4 VOLTS/DIV
				12	CH 3 VOLTS/DIV
				13	INIT@50%
				14	CH 2 Coupling lower.
				15	CH 2 Coupling upper.
				25	CH 2 INVERT
				31	CH 1 VERTICAL MODE
				32	CH 2 VERTICAL MODE
				33	ADD VERTICAL MODE
				34	CH 3 VERTICAL MODE
				35	CH 4 VERTICAL MODE
				41	STEP/AUTO
				42	SAVE HELP
				43	RECALL HELP
				44	CHOP/ALT VERTICAL MODE
				45	20 MHz BW LIMIT
				51	X10 MAG
				52	TRACK/INDEP
				53	Δt (delta time).
				54	ΔV (delta volts).
				55	Trigger SLOPE
				61	Trigger SOURCE lower.
				62	Trigger SOURCE upper.
				63	Trigger MODE lower.
				64	Trigger MODE upper.
				65[b]	A/B TRIG select.

[a]VERT TRIG SOURCE indicator lights when in ALL TESTS as a visual reference in the event a CRT display can not be produced.

[b]If the A/B TRIG switch is stuck during power-up, the oscilloscope will branch to "normal" operation after a short delay. The associated error message will only be visible momentarily if the CRT is warmed-up.

[c]Readout Board error codes are also displayed on the + and − Trigger SLOPE LEDs.

[d]This test is not user-selectable but is run automatically during cycle mode.

Table 6-4 (cont)

Routine Type	Type Number	Lit LED	Routine Name	Error Code	Error Code Meaning
Test	03	CH 3	Readout Board	01	Shift register failure.[c](– Trigger LED).
				02	Readout RAM failure[c] (+ Trigger LED).
Test	04	CH 4	Calibration Data	X1	Parity error on read (bit 0 set).
				X2	Out of limits (bit 1 set).
				1X	Bad checksum (bit 4 set).
Test	05	ADD	Main Board	01	AUTO LVL failed to trigger.
				X2	Negative level not negative enough.
				X4	Negative level too negative.
				2X	Positive level not positive enough.
				4X	Positive level too positive.
Test	06	INVERT	RAM Battery	01	Battery voltage too low.
				02	Battery voltage too high.

[a]VERT TRIG SOURCE indicator lights when in ALL TESTS as a visual reference in the event a CRT display can not be produced.

[b]If the A/B TRIG switch is stuck during power-up, the oscilloscope will branch to "normal" operation after a short delay. The associated error message will only be visible momentarily if the CRT is warmed-up.

[c]Readout Board error codes are also displayed on the + and – Trigger

[d]This test is not user-selectable but is run automatically during cycle mode.

Table 6-5
Sequence of Exerciser Routines[a]

Routine Type	Type Number	ON LED	Routine Function
Exerciser	01	CH 1	Display Pots and Switches.
Exerciser	02	CH 2	Examine Calibration Data in RAM.
Exerciser	03	CH 3	Clears Cycle Errors.
Exerciser	04	CH 4	Display ROM Headers.
Exerciser	05	ADD	Display Operating Time and Power Cycle Count.
Exerciser	06	INVERT	Select Setup to Use at Power-Up.
Exerciser	07	CHOP	Enable/Disable Setup SAVE and Sequence Definition.
Exerciser	08	BW LIMIT	Initialize Setups.

[a]Additional Diagnostic Exercisers for extended functions are in Appendix A of the Operators Manual.

Routine Control

When the desired Test or Exerciser has been selected, the operator has two types of control that may be exercised over the routine: START/STOP and LOOP.

Starting or stopping the execution of the selected routine is controlled by the front-panel TRIGGER COUPLING switches. Pressing the upper switch starts the routine; pressing the lower switch stops it.

All of the test routines may be set to LOOP mode (continuously repeated) by pressing the front-panel upper TRIGGER SOURCE switch while the routine is selected but not executing. The LOOP feature will cause the routine to be continuously repeated once started until stopped when the operator presses the lower TRIGGER COUPLING switch. Once the routine is stopped, the LOOP feature may be disabled by pressing the lower TRIGGER SOURCE switch.

While a Test or Exerciser routine is executing, the Diagnostic Monitor Control message on the top line of the CRT display will be cleared as an indication that a routine is running. When test routines are looping, the message "LOOP" is displayed in the bottom division of the CRT graticule.

Display Format

The Tests and Exercisers routines display information about the routine type and number, as well as any test results, at the bottom of the CRT display. The readout line is formatted as follows:

OD TYPE XY STATUS ZZ LOOP OD<ABCC>

The information is defined as follows:

"OD" is a two-character option designator identifying the option that this particular line of diagnostic information refers to (see Options manual for details). For the basic instrument, the OD location is blank.

"TYPE" refers to routine type: All Tests (ALL), Test (TEST), Exerciser (EXER), or Calibration (CAL).

"X" indicates which bit of the "Option Select Register" is set to turn on the option called out by "OD" (see Options manual for description of Options Select Register). This bit is zero for the basic instrument.

"Y" is the TYPE number of the routine (see the "Type Number" column of Table 6-4).

"STATUS" shows the results of the last time a selected test routine ran: either PASS or FAIL. This space is blank for exerciser and calibration routines. When the diagnostics are called up from normal operating mode, the space will be blank until the selected test is executed.

"ZZ" is a two-digit error code identifying the nature of the failure in a failed test (see the "Error Code" column of Table 6-4).

"LOOP" indicates when a selected test is set to the LOOP mode.

"OD<ABCC>" is the CYCLE mode failure indicator. CYCLE mode, when entered by removing the NO CAL/CAL jumper (P501) before turning the instrument on, causes the instrument to continuously LOOP through the Power Up Diagnostic Tests. If a failure occurs, the cycle-failure data, identifying the first failure encountered, is written to RAM. Thereafter, at each power-up, the Diagnostic Monitor is automatically entered, and the failure data is displayed. The failure data must be cleared from the RAM location to eliminate the CYCLE mode failure display (see CYCLE ERROR CLEAR Exerciser 03). The information displayed is an abbreviated version of the previous items:

"OD" is a two-character option designator showing which option failed first while in the CYCLE mode (the same codes as for "OD" at the start of the readout line).

"A" identifies the option-select bit for the failing option (the same code as for "X").

"B" is the test Type Number where the failure occurred (the same codes as for "Y").

"CC" is the error code for the test (the same codes as for "ZZ").

Kernel Tests

The Kernel tests are those tests which, when failed, are considered "fatal" to the operation of the Microprocessor. Failure of a Kernel test will cause the front-panel TRIG'D indicator to flash, and certain of the other front-panel indicators will be illuminated with an error code. The code points to the area of failure as indicated in Table 6-6. Tables 6-7 and 6-8 are used to determine the option and device numbers used in Table 6-6. Only the basic instrument codes are given in Table 6-6. Option codes are defined in the "Options Service Manual."

Table 6-6
Kernel Test Failure Codes

Failure Codes		
Option	**Device**	**Failing Device**
0	0	Control Board RAM
0	1	ROM U2160
0	2	ROM U2260
0	3	Reset Control Circuitry
F	1	Buffer ROM U2160

Table 6-7
Front-Panel LED Option Codes

Option Code					
CH 1 LED (bit 3)	**CH 2 LED (bit 2)**	**CH 3 LED (bit 1)**	**CH 4 LED (bit 0)**	**Option Number (in Hex)**	**Option Name**
OFF	OFF	OFF	OFF	0	Basic Instrument
ON	ON	ON	ON	F	Options Buffer Board

Table 6-8
Front-Panel LED Device Codes

READY LED (bit 2)	Device Code		Device Number
	+ LED (bit 1)	– LED (bit 0)	
OFF	OFF	OFF	0
OFF	OFF	ON	1
OFF	ON	OFF	2
OFF	ON	ON	3
ON	OFF	OFF	4
ON	OFF	ON	5
ON	ON	OFF	6
ON	ON	ON	7

Even if a Kernel test fails, the operator may try to go to normal oscilloscope operation by pressing the A/B TRIG select push button. Depending on the exact nature of the failure, the instrument may or may not be functional.

Kernel tests are automatically executed at power-up. The Kernel tests are divided into RAM tests and ROM tests as follows:

RAM TEST. This test is done with a complementary data pattern starting at the highest RAM address available and continuing to the lowest. The process reads and saves the original data, and then writes a pattern of 01010101's (55 Hex) at the highest RAM memory address. The data is then read back to see if it is still 55 (Hex). Next a complementary pattern of 10101010 (AA Hex) is written to the same address. Then the address content is read back and tested to see if it is still AA (Hex). After the memory is checked, the original data is written back into the memory address. RAM TEST then checks the next lower address. The testing continues until all of RAM is checked.

Test checks: RAM address decoding, RAM address lines, RAM data lines, RAM memory, and Data Bus Buffers.

ROM TEST. The ROM test performs three checks on each of the system read-only memories.

Data Bus Drive—Two locations containing complementary data patterns are read.

Test checks: Data bus lines and the Data Bus Driver.

Correct Part—A byte in the ROM being checked is compared to the most-significant byte of the addressed ROM block (starting address of where the ROM should be installed).

Test checks: ROM address decoding and proper installation of ROM components.

Checksum—A sixteen bit, spiral-add checksum is calculated and compared to a two-byte value stored in ROM being checked.

Test checks: ROM contents, ROM addressing, ROM data lines, and the Data Bus Driver.

Confidence Tests

The Confidence tests provide checks for much of the remaining circuitry to ensure that instrument operation is correct. Confidence tests are performed automatically at power-up after the Kernel is determined to be functional or initiated by the operator from the Diagnostic Monitor.

A failure of any Confidence test during power-up will pass control to the Diagnostic Monitor; this permits the test results to be examined. Descriptions of the Confidence tests follow.

INTERRUPT REQUEST (Test 01). Ten consecutive interrupt cycles are checked to ensure that succeeding interrupts occur not more than 4.5 ms apart (5600 "E" cycles).

Test checks: Interrupt Timer circuitry.

SWITCH STUCK (Test 02). The front-panel, momentary-contact switches are scanned, checking for a closed switch. At power-up, the test runs immediately.

By holding one of the momentary switches in a closed position when power is first applied, this test will fail, and the Diagnostic Monitor will be entered. When the test is started from the Diagnostic Monitor, a one-half second delay is incorporated to allow the COUPLING (test start) switch to return to its normal (open) position. Table 6-4 defines the error codes that may be encountered when a switch is detected as closed.

NOTE

When pressing the lower TRIGGER COUPLING switch to stop this test, an error code may be generated. This is normal and does not indicate an actual failure.

Test checks: Momentary switches, row scanning circuitry, and column scanning circuitry.

READOUT BOARD (Test 03). This two-part test checks the interface to the Readout Board from the Microprocessor and the character RAM circuits.

Processor Interface Test—The Microprocessor loads the three, eight-bit shift registers with an alternating bit pattern that is then shifted back to the processor for comparison.

Test checks: Data Registers, data strobes (clocks), and the data input and output lines.

RAM Test—A "1" is rotated through each byte of the Readout RAM, one bit at a time. Each time an additional bit is rotated into the byte, the byte is loaded into the processor interface and clocked back to the processor for comparison. The byte is then restored to its original content, and each successive byte is tested in the same manner.

Test checks: Readout RAM addressing, Readout RAM data lines, and RAM read/write capability.

CALIBRATION DATA (Test 04). Three checks are performed on the RAM to verify its contents.

Checksum Test—The contents of locations containing calibration constants are checksummed using a spiral-add technique. The result is compared to the stored checksum generated at the time of calibration.

Test checks: RAM addressing and RAM contents.

Parity Test—As each of the calibration constants is read for the Checksum test above, the parity of each 14-bit word is checked.

Test checks: CALIBRATION DATA integrity and RAM CALIBRATION DATA retention.

Limit Test—Checks for valid calibration data.

Test checks: The contents of locations containing calibration data are compared to their stored limits.

MAIN BOARD (Test 05). The AUTO LVL triggering feature (a routine stored in firmware) is operated to detect the peaks of a Line Trigger signal. Detected peaks are compared to expected values to verify operation (and calibration) of interrelated signal processing circuits.

Test checks: Line Trigger source, the A-Trigger generation circuitry, and Control DAC U2101 (located on the Control board, diagram 2).

BATTERY VOLTS (Test 06). The battery voltage is read and compared to stored constants. If the voltage is above or below the stored limits the appropriate error code is displayed.

Test checks: Battery voltage, voltage follower operational amplifier U2620C, and CR2770.

Exerciser Routines

The Exerciser routines (see Table 6-5) allow the operator to set and examine various bytes of control data used in determining instrument function.

POTS AND SWITCHES (Exerciser 01). This routine displays the values that the Microprocessor detects as the various digitized pots and switches are activated. The left half of the top line of the display appears after turning a pot. The right half of the top line of the CRT display appears after pressing a switch. The top line of the CRT display has the following format:

AA BB CC DEEE FF GG HI JJ KL

The format is defined as follows:

"AA" is the code of the most-recently-activated potentiometer (see Table 6-9 for definition of pot codes).

"BB" is the current value (in hexadecimal) of pot AA. See Table 6-9 for the approximate range of codes for the CCW (counter clockwise) and CW (clockwise) potentiometer rotations.

"CC" is the previous value (in hexadecimal) of pot AA.

"D" is the DAC Multiplexer code used to select pot AA (see Table 6-9).

"EEE" is the 12-bit DAC value (in hexadecimal) associated with pot AA. See Table 6-9 for the approximate range of codes for the CCW (counter clockwise) and CW (clockwise) potentiometer rotations.

"FF" is the code of the previously-activated potentiometer (see Table 6-9).

"GG" is the row code of the most-recently-activated switch (see Table 6-10 for definition of row codes).

"H" is the switch-position code: 0 for open; C for closed.

"I" is the column code of the most-recently-activated switch (see Table 6-10).

"JJ" is the row code for the previously-activated switch.

"K" is the switch-position code: 0 for open; C for closed.

"L" is the column code for the previously-activated switch.

NOTE

For all momentary switches (except A/B TRIG) only the closed position will be shown in the switch-position code locations (H and K). The A/B TRIG switch has both the open and the closed positions shown.

Table 6-9
Potentiometer Codes and Values (Exerciser 01)

Rotation Values				Potentiometer	
CCW		CW		AA	Name
BB	DEEE	BB	DEEE	Code	
FF	6FFF	00	6000	01	HOLDOFF
FF	3FFF	00	3000	02	Trigger LEVEL
00	1000	FF	1FFF	03	SEC/DIV VAR
FF	5FFF	00	5000	04	Horizontal POSITION
00	0000	FF	3FFF	05	Δ (A section)[a]
00	0000	FF	3FFF	06	Δ (B section)[a]
00	0000	FF	3FFF	07	Δ REF OR DLY POS (A section)[a]
00	0000	FF	3FFF	08	Δ REF OR DLY POS (B section)[a]
FF	07FF	00	0000	09	CH 1 VOLTS/DIV VAR
FF	27FF	00	2000	0A	CH 2 VOLTS/DIV VAR
FF	0FFF	00	0000	11	CH 1 Vertical POSITION
FF	1FFF	00	1000	12	CH 2 Vertical POSITION
FF	27FF	00	2000	13	CH 3 Vertical POSITION
FF	37FF	00	3000	14	CH 4 Vertical POSITION
FF	4FFF	00	4800	15	TRACE SEP
FF	5FFF[b]	00	5FFF[b]	16	READOUT INTENSITY
80	6800	FF	6FFF	17	Trace INTENSITY

[a]The Δ REF OR DLY POS and Δ controls are both 180° offset pairs that continuously rotate. Displayed BB values jump and the AA code changes when instrument software switches between the A and B sections. The D code position shows the two most-significant bits of the 14-bit DAC output (in hexadecimal), effectively generating 5.5 turn potentiometer values.

[b]The potentiometer midpoint value is 5800, and the intensity is off (MIN). Maximum intensity is at both the CCW and CW positions.

Table 6-10
Pots and Switches Column
and Row Code Definitions (Exerciser 01)

Row Code (GG)	Column Code (I)	Definition	Row Code (GG)	Column Code (I)	Definition
0	0	Trig COUPLING Down	5	0	READOUT Scale Factors
0	1	Trig COUPLING Up	5	1	Unused
0	2	MEASURE/HELP	5	2	Unused
0	3	CH 1 Coupling Down	5	3	Unused
0	4	CH 1 Coupling Up	5	4	Unused
1	0	CH 4 VOLTS/DIV	6	0	CH 1 VERT MODE
1	1	CH 3 VOLTS/DIV	6	1	CH 2 VERT MODE
1	2	INIT @ 50%	6	2	ADD VERT MODE
1	3	CH 2 Coupling Down	6	3	CH 3 VERT MODE
1	4	CH 2 Coupling Up	6	4	CH 4 VERT MODE
2	0	CH 1 VOLTS/DIV LSB	7	0	STEP/AUTO
2	1	CH 1 VOLTS/DIV Bit 2	7	1	SAVE HELP
2	2	CH 1 VOLTS/DIV Bit 3	7	2	RECALL HELP
2	3	CH 1 VOLTS DIV MSB	7	3	CHOP/ALT
2	4	CH 2 INVERT	7	4	BW LIMIT
3	0	CH 2 VOLTS/DIV LSB	8	0	X10 MAG
3	1	CH 2 VOLTS/DIV Bit 2	8	1	TRACKING/INDEP
3	2	CH 2 VOLTS/DIV Bit 3	8	2	Δt
3	3	CH 2 VOLTS/DIV MSB	8	3	ΔV
3	4	B ENDS A	8	4	Trig SLOPE
4	0	SEC/DIV LSB	9	0	Trig SOURCE Down
4	1	SEC/DIV Bit 2	9	1	Trig SOURCE Up
4	2	SEC/DIV Bit 3	9	2	Trig MODE Down
4	3	SEC/DIV MSB	9	3	Trig MODE Up
4	4	A/B SWP Select	9	4	A/B TRIG Select

CALIBRATION RAM EXAMINE (Exerciser 02). This routine allows the operator to examine the contents of 256 decimal locations, 00 (Hex) through FF (Hex), in RAM. When entered, the Exerciser displays the contents of RAM location 00 (Hex) on the top line of the CRT display. One hundred and seventy calibration constants reside between addresses 01 (Hex) and AA (Hex). Calibration constants residing between 01 (Hex) and 6E (Hex) should have odd parity as explained below. The remaining locations may be of either parity. The readout display line has the following format:

AA DDDD P

The format is defined as follows:

"AA" is the eight-bit address in hexadecimal notation.

"DDDD" is the 14-bit word stored at that location (13 bits of data and one parity bit).

"P" is a parity indicator for the data word: X indicates even parity; blank is odd parity.

Pushing the upper or lower TRIGGER MODE switch will increment or decrement the RAM address by 16 (10 Hex) respectively. Similarly, pushing the upper or lower TRIGGER SOURCE switch will increment or decrement the address by 1 respectively.

CYCLE ERROR CLEAR (Exerciser 03). This routine provides a way to clear the cycle-failure data written to the RAM when a CYCLE mode failure occurs. Interpretation of the cycle failure data is explained in the "Display Format" description provided earlier in this section. Until the data is cleared, each time the instrument is powered up, the Diagnostic Monitor is entered.

Clearing the RAM location (and the CYCLE ERROR message) is done by scrolling to EXER 03 (CLEAR CYCLE ERROR) and pressing the following switches in sequence:

TRIGGER COUPLING upper (starts exerciser),
TRIGGER SOURCE lower,
TRIGGER MODE lower, then
TRIGGER COUPLING lower (exits the exerciser).

When the CYCLE ERROR CLEAR routine is successfully executed, the cycle failure data will disappear from the display.

DISPLAY ROM HEADERS (Exerciser 04). This routine displays the Standard Tektronix ROM Header of each system ROM on the top line of the CRT display. The readout line has the following format:

CCCC PPPP SS AAAA OD

The definition of the format is as follows:

"CCCC" is a two-byte hexadecimal checksum.

"PPPP" is the four middle digits of the ROM part number.

"SS" is the suffix of the ROM part number (version number).

"AAAA" is the starting address of the ROM (address where the ROM should be installed).

"OD" is a two-character option designator identifying the option that this particular line of diagnostic information refers to (see Options manual for details). For the basic instrument, the OD location is blank.

Pressing the upper TRIGGER COUPLING switch increments the routine to the next ROM Header; pressing the lower TRIGGER COUPLING switch exits the routine.

HRS ON and OFF/ON CYCLES (Exerciser 05). This routine displays the Operating Time and Power Cycle Count (see Operators Manual).

POWER-UP SETUP (Exerciser 06). This routine selects the setup to use at power-up (see Operators Manual).

SAVE ENABLE (Exerciser 07). This routine will Enable/Disable setup SAVE and sequence definition (see Operators Manual).

SETUP INIT (Exerciser 08). This routine destroys all saved setups (see Operators Manual).

CONTROLLER LATCHES EXERCISER. This routine is not user selectable, but runs automatically when the Diagnostic Monitor is waiting for a key activation.

The routine first sets latches U2301 and U2201 (diagram 2). It then pulses the B SWP CLK line (pin 13 of U2660, diagram 1), as a scope trigger, and rotates a "0" through 15 of the 16 latched bits. Bit 16 is not set since it would reset Interrupt Timer U2640 (diagram 1) and upset processor interrupt timing. By externally triggering a test oscilloscope on the B SWP CLK signal line and observing the shifted timing relationships of the latched signals, proper operation of the DAC latches may be verified.

NOP KERNEL EXERCISER. This exerciser is not a firmware routine, but rather a forced hardware condition. It is best suited for troubleshooting an inoperative Control Board, as it exercises only the Microprocessor address bus (see Table 6-11) and the associated Address Decode circuitry. By moving Jumper P503 (diagram 1) to the Diagnostic position, Data Bus Buffers U2350 and U2450 are disabled, and the Microprocessor is forced into a NOP (no operation) loop. This causes the address on the address bus to be continuously incremented for exercising the Address Decode circuitry. Troubleshooting of kernel addressing with an oscilloscope or logic analyzer is then possible.

Table 6-11
NOP Test Data

U2140 Pin #	Signal Name	1 CYCLE Time	Frequency
9	A0	3.199 μs	312.5 kHz
10	A1	6.39 μs	156.3 kHz
11	A2	12.79 μs	78.15 kHz
12	A3	25.59 μs	39.075 kHz
13	A4	51.18 μs	19.53 kHz
14	A5	102.4 μs	9.769 kHz
15	A6	204.7 μs	4.88 kHz
16	A7	409.4 μs	2.44 kHz
17	A8	818.9 μs	1.22 kHz
18	A9	1638 μs	610.6 Hz
19	A10	3275 μs	305.3 Hz
20	A11	6.55 ms	152.6 Hz
22	A12	13.1 ms	76.3 Hz
23	A13	26.2 ms	38.16 Hz
24	A14	52.4 ms	19.08 Hz
25	A15	104.8 ms	9.54 Hz

CORRECTIVE MAINTENANCE

INTRODUCTION

Corrective maintenance consists of component replacement and instrument repair. This part of the manual describes special techniques and procedures required to replace components in this instrument. If it is necessary to ship your instrument to a Tektronix Service Center for repair or service, refer to the "Instrument Repackaging Instructions" in Section 2.

MAINTENANCE PRECAUTIONS

To reduce the possibility of personal injury or instrument damage, observe the following precautions.

1. Disconnect the instrument from the ac power source before removing or installing components.

2. Verify that the line-rectifier filter capacitors are discharged prior to performing any servicing.

3. Use care not to interconnect instrument grounds which may be at different potentials (cross grounding).

4. When soldering on circuit boards or small insulated wires, use only a 15-watt, pencil-type soldering iron.

OBTAINING REPLACEMENT PARTS

Most electrical and mechanical parts can be obtained through your local Tektronix Field Office or representative. However, many of the standard electronic components can usually be obtained from a local commercial source. Before purchasing or ordering a part from a source other than Tektronix, Inc., please check the "Replaceable Electrical Parts" list for the proper value, rating, tolerance, and description.

NOTE

Physical size and shape of a component may affect instrument performance, particularly at high frequencies. Always use direct-replacement components, unless it is known that a substitute will not degrade instrument performance.

Special Parts

In addition to the standard electronic components, some special parts are used in the instrument. These components are manufactured or selected by Tektronix, Inc. to meet specific performance requirements, or are manufactured for Tektronix, Inc. in accordance with our specifications. The various manufacturers can be identified by referring to the "Cross Index-Manufacturer's Code Number to Manufacturer" at the beginning of the "Replaceable Electrical Parts" list. Many of the mechanical parts used in this instrument were manufactured by Tektronix, Inc. Order all special parts directly from your local Tektronix Field Office or representative.

Ordering Parts

When ordering replacement parts from Tektronix, Inc., be sure to include all of the following information:

1. Instrument type (include modification or option numbers).

2. Instrument serial number.

3. A description of the part (if electrical, include its full circuit component number).

4. Tektronix part number.

MAINTENANCE AIDS

The maintenance aids listed in Table 6-12 include items required for performing most of the maintenance procedures in this instrument. Equivalent products may be substituted for the examples given, provided their characteristics are similar.

Table 6-12
Maintenance Aids

Description	Specification	Usage	Example
1. Soldering Iron	15 to 25 W.	General soldering and unsoldering.	Antex Precision Model C.
2. Flat-bit Screwdriver	3-inch shaft, 3/32 inch bit.	Assembly and disassembly.	Xcelite Model R3323.
3. Torx Screwdriver	Tip sizes: #T9, #T10, #T15, #T20. Handles	Assembly and disassembly.	Tektronix Part Numbers #T9 003-0965-00 #T10 003-0815-00 #T15 003-0966-00 #T20 003-0866-00 8 1/2 in. 003-0293-00 3 1/2 in. 003-0445-00.
4. Nutdrivers	3/16 inch,1/4 inch, and 5/16 inch	Assembly and disassembly.	Xcelite #6, #8 and #10.
5. Open-end Wrenches	1/4 inch, 5/16 inch, 7/16 inch.	Assembly and disassembly.	
6. Allen Wrenches	0.050 inch, 1/16 inch.	Assembly and disassembly.	
7. Long-nose Pliers		Component removal and replacement.	Diamolloy Model LN55-3.
8. Diagonal Cutters		Component removal and replacement.	Diamalloy Model M554-3.
9. Vacuum Solder Extractor	No static charge retention.	Unsoldering static sensitive devices and components on multilayer boards.	Pace Model PC-10.
10. Spray Cleaner	No-Noise	Switch and Pot cleaning.	Tektronix Part Number 006-0442-02.
11. Pin-replacement kit		Replace circuit board connector pins.	Tektronix Part Number 040-0542-00.
12. IC-Removal Tool		Removing DIP IC packages.	Augat T114-1.
13. Isopropyl Alcohol	Reagent grade.	Cleaning attenuator and front panel assemblies.	2-Isopropanol.

INTERCONNECTIONS

Interconnections in this instrument are made with pins soldered onto the circuit boards. Several types of mating connectors are used for the interconnecting pins. The following information provides the replacement procedures for the various type connectors.

End-Lead Pin Connectors

Pin connectors used to connect the wires to the interconnect pins are factory assembled. They consist of machine-inserted pin connectors mounted in plastic holders. If the connectors are faulty, the entire wire assembly should be replaced.

Multipin Connectors

When pin connectors are grouped together and mounted in a plastic holder, they are removed, reinstalled, or replaced as a unit. If any individual wire or connector in the assembly is faulty, the entire cable assembly should be replaced. Multipin connector orientation is indexed by a triangle on the cable connector and a 1 or triangle on the circuit board. Slot numbers may be molded into the connector. Be sure these index marks are aligned with each other when the multipin connector is reinstalled.

TRANSISTORS, INTEGRATED CIRCUITS, AND HYBRID CIRCUITS

Transistors, integrated circuits, and hybrid circuits should not be replaced unless they are actually defective. If removed from their sockets or unsoldered from the circuit board during routine maintenance, return them to their original board locations. Unnecessary replacement or transposing of semiconductor devices may affect the adjustment of the instrument. When a semiconductor is replaced, check the performance of any circuit that may be affected.

Any replacement component should be of the original type or a direct replacement. Bend transistor leads to fit their circuit board holes, and cut the leads to the same length as the original component. See Figure 9-2 in the "Diagrams" section for lead-configuration illustrations.

The heat-sink-mounted power supply transistors are insulated from the heat sink with a heat-transferring insulator pad. Reinstall the insulator pads and bushings when replacing these transistors. Do not use any type of heat-transferring compound on the insulator pads.

NOTE

After replacing a power transistor, check that the collector is not shorted to the heat sink before applying power to the instrument.

To remove socketed dual-in-line packaged (DIP) integrated circuits, pull slowly and evenly on both ends of the device. Avoid disengaging one end of the integrated circuit from the socket before the other since this may damage the pins.

To remove a soldered DIP IC when it is going to be replaced, clip all the leads of the device and remove the leads from the circuit board one at a time. If the device must be removed intact for possible reinstallation, do not heat adjacent conductors consecutively. Apply heat to pins at alternate sides and ends of the IC as solder is removed. Allow a moment for the circuit board to cool before proceeding to the next pin.

Hybrid circuits and heatsinks are removed as a unit by removing the mounting nuts at the four corners of the heatsink/housing. A firm downward pressure at the center of the heatsink will aid in installation/removal of the nuts. The hybrid circuit substrate is bonded to the heatsink/housing casting. Attempting to separate the hybrid device from its heatsink will damage the device.

SOLDERING TECHNIQUES

The reliability and accuracy of this instrument can be maintained only if proper soldering techniques are used to remove or replace parts. General soldering techniques, which apply to maintenance of any precision electronic equipment, should be used when working on this instrument.

WARNING

To avoid an electric-shock hazard, observe the following precautions before attempting any soldering: turn the instrument off, disconnect it from the ac power source, and verify that the line-rectifier filter capacitors have discharged (see label on the primary power shield). If, due to a component failure, the capacitors are not discharging, it may be necessary to discharge them. Use a 1 kΩ 5-watt resistor and discharge the capacitors from point to point through the access holes in the primary power shield.

Use rosin-core wire solder containing 63% tin and 37% lead. Contact your local Tektronix Field Office or representative to obtain the names of approved solder types.

When soldering on circuits boards or small insulated wires, use only a 15-watt, pencil-type soldering iron. A higher wattage soldering iron may cause etched circuit conductors to separate from the board base material and melt the insulation on small wires. Always keep the soldering-iron tip properly tinned to ensure best heat transfer from the iron tip to the solder joint. Apply only

enough solder to make a firm joint. After soldering, clean the area around the solder connection with an approved flux-removing solvent (such as isopropyl alcohol) and allow it to air dry.

Circuit boards in this instrument may have as many as four conductive layers. Conductive paths between the top and bottom board layers may connect to one or more inner layers. If any inner-layer conductive path becomes broken due to poor soldering practices, the board becomes unusable and must be replaced. Damage of this nature can void the instrument warranty.

CAUTION

Only an experienced maintenance person proficient in the use of vacuum-type desoldering equipment should attempt repair of any circuit board in this instrument.

Desoldering parts from multilayer circuit boards is especially critical. Many integrated circuits are static sensitive and may be damaged by solder extractors that generate static charges. Perform work involving static-sensitive devices only at a static-free work station while wearing a grounded antistatic wrist strap. Use only an antistatic vacuum-type solder extractor approved by a Tektronix Service Center.

CAUTION

Attempts to unsolder, remove, and resolder leads from the component side of a circuit board may cause damage to the reverse side of the circuit board.

The following techniques should be used to replace a component on a circuit board:

1. Touch the vacuum desoldering tool to the lead at the solder connection. Never place the iron directly on the board; doing so may damage the board.

NOTE

Some components are difficult to remove from the circuit board due to a bend placed in the component leads during machine insertion. To make removal of machine-inserted components easier, straighten the component leads on the reverse side of the circuit board.

2. When removing a multipin component, especially an IC, do not heat adjacent pins consecutively. Apply heat to the pins at alternate sides and ends of the IC as solder is removed. Allow a moment for the circuit board to cool before proceeding to the next pin.

CAUTION

Excessive heat can cause the etched circuit conductors to separate from the circuit board. Never allow the solder extractor tip to remain at one place on the board for more than three seconds. Solder wick, spring-actuated or squeeze-bulb solder suckers, and heat blocks (for desoldering multipin components) must not be used. Damage caused by poor soldering techniques can void the instrument warranty.

3. Bend the leads of the replacement component to fit the holes in the circuit board. If the component is replaced while the board is installed in the instrument, cut the leads so they protrude only a small amount through the reverse side of the circuit board. Excess lead length may cause shorting to other conductive parts.

4. Insert the leads into the holes of the board so that the replacement component is positioned the same as the original component. Most components should be firmly seated against the circuit board.

5. Touch the soldering iron to the connection and apply enough solder to make a firm solder joint. Do not move the component while the solder hardens.

6. Cut off any excess lead protruding through the circuit board (if not clipped to the correct length in step 3).

7. Clean the area around the solder connection with an approved flux-removing solvent. Be careful not to remove any of the printed information from the circuit board.

8. When soldering to the ceramic CRT-termination network, a slightly larger soldering iron can be used. It is recommended that a solder containing about 3% silver be used when soldering to the ceramic material to avoid destroying the bond. The bond can be broken by repeated use of ordinary tin-lead solder or by the application of too much heat; however, occasional use of ordinary solder will not break the bond, provided excessive heat is not applied when making the connection.

REMOVAL AND REPLACEMENT INSTRUCTIONS

WARNING

To avoid electric shock, disconnect the instrument from the ac power source before removing or replacing any component or assembly.

WARNING

Removal of the cabinet and other external panels leaves the CRT exposed for possible damage. All procedures in these instructions require careful attention to avoid damage to the CRT which could cause it to implode. An implosion creates high speed glass fragments. Wear protective clothing and use safety shields as required. See "WARNING" in "CRT REMOVAL".

The exploded view drawing in the "Replaceable Mechanical Parts" list at the rear of this manual may be helpful during the removal and reinstallation of individual components or subassemblies. Circuit board and component locations are illustrated in the "Diagrams" section of this manual.

Cabinet Removal

Removal of the instrument wrap-around cabinet is accomplished by the following steps:

1. Unplug the power cord from the ac power source.

2. Unplug the power cord from the rear-panel connector.

3. Install the front cover, place the cabinet carrying handle against the bottom of the cabinet, and set the instrument face down on a flat surface.

4. Unwrap the power cord and remove it.

5. Remove the four screws in the rear feet.

6. Remove the two screws from the top-center and bottom-center of the rear cover.

7. Lift the rear cover and power cord away from the instrument, leaving the rear feet attached.

WARNING

Dangerous potentials exist at several points throughout this instrument. If it is operated with the cabinet removed, do not touch exposed connections or components. Some transistors may have elevated case voltages. Disconnect the ac power source from the instrument and verify that the line-rectifier filter capacitors have discharged before cleaning the instrument or replacing parts (see label on the primary power shield).

8. Slide the cabinet off the instrument.

To reinstall the wrap-around cabinet, perform the reverse of the preceding instructions. Ensure that the cabinet fits properly into the EMI gasket grooves in the front frame and rear panel.

WARNING

The line-rectifier filter capacitors normally retain a charge for a short period (approximately 15 to 20 seconds) after the instrument is turned off and can remain charged for a longer period if a bleeder-resistor or power-supply problem occurs. Before beginning any cleaning or work on the internal circuitry of the oscilloscope, disconnect the ac power source from the instrument and verify that the capacitors have discharged to 24 V or less. Measurement is made at the three points indicated on the plastic primary input shield at the rear of the instrument (after the Top-Cover Plate is removed). If the capacitors retain charges of greater than 24 V for more than 20 seconds, discharge them using a 1 kΩ, 5-watt resistor connected point-to-point across the capacitors through the access holes. Ensure that the capacitors are discharged before starting to troubleshoot.

Vertical Bracket (Top-Cover Plate) Removal

To remove the Vertical Bracket from instruments that do not have the DMM option installed, perform the following steps:

1. Remove the instrument Cabinet as described in that procedure.

2. Set the instrument, bottom down, on a flat surface.

3. Remove two top securing screws at the front edge of the Vertical Bracket.

4. (SN B050000 and above.) Remove the screw in the right-center of the Vertical Bracket.

4. (SN B049999 and below.) Remove the two screws in the right-center of the Vertical Bracket.

5. Remove the top securing screw at the left-rear of the Vertical Bracket.

6. Remove the securing screw from the chassis rear plate.

7. Remove the securing screw from the left side of the chassis.

8. Lift the Vertical Bracket up and away from the instrument.

(SN B050000 and above.) To reinstall the Vertical Bracket, perform the reverse of the preceding instructions. Align the black plastic pin on the power supply assembly with its mating hole before installing and tightening screws.

(SN B049999 and below.) To reinstall the Vertical Bracket, perform the reverse of the preceding instructions. Be certain to align the circuit board at the right rear with the two black grommets installed in the Vertical Bracket. Align the two black plastic pins on the power supply assembly with their mating holes before installing and tightening screws.

A5—Control Board Removal

Removal of the Control Board is accomplished by the following steps:

1. Remove the instrument wrap-around cabinet as described in that procedure.

2. Place the instrument on its left side on a flat surface.

3. Disconnect the two ribbon-cable and one flex-circuit connectors (P251, P651, and P652) from the Control board (see Figure 6-2).

4. (SN B050000 and above.) Disconnect the three ribbon-cable connectors (P411, P511, and P512) from the Main Board.

4. (SN B049999 and below.) Disconnect the two ribbon-cable connectors (P511 and P512) from the Main Board.

5. Remove the five mounting screws securing the Control board to the chassis, one at each corner of the board and one at the center.

6. Lift the Control board away from the chassis.

To reinstall the Control board, perform the reverse of the preceding instructions.

FAN REMOVAL. (If your instrument has the DMM option installed *and* has a serial number of SN B049999 or below, use the "Fan Removal" procedure in your options Service Manual.) For instruments with SN B050000 and above, removal of the fan is accomplished by the following steps:

1. Desolder the wires from the feed-through capacitor (C10) and ground lug, noting color code for reassembly.

2. Remove the Fan retainer screw, located above the Fan.

To reinstall the Fan, perform the reverse of the above instructions. Align the holes in the Fan flange with the pins on the rear plate before tightening the screw.

A2/A2A1 and A3—Power Supply Assembly Removal

Removal of the Power Supply assembly from instruments that do not contain options is accomplished by the following steps:

1. Remove the instrument Cabinet as described in that procedure.

2. Remove the Vertical Bracket as described in that procedure.

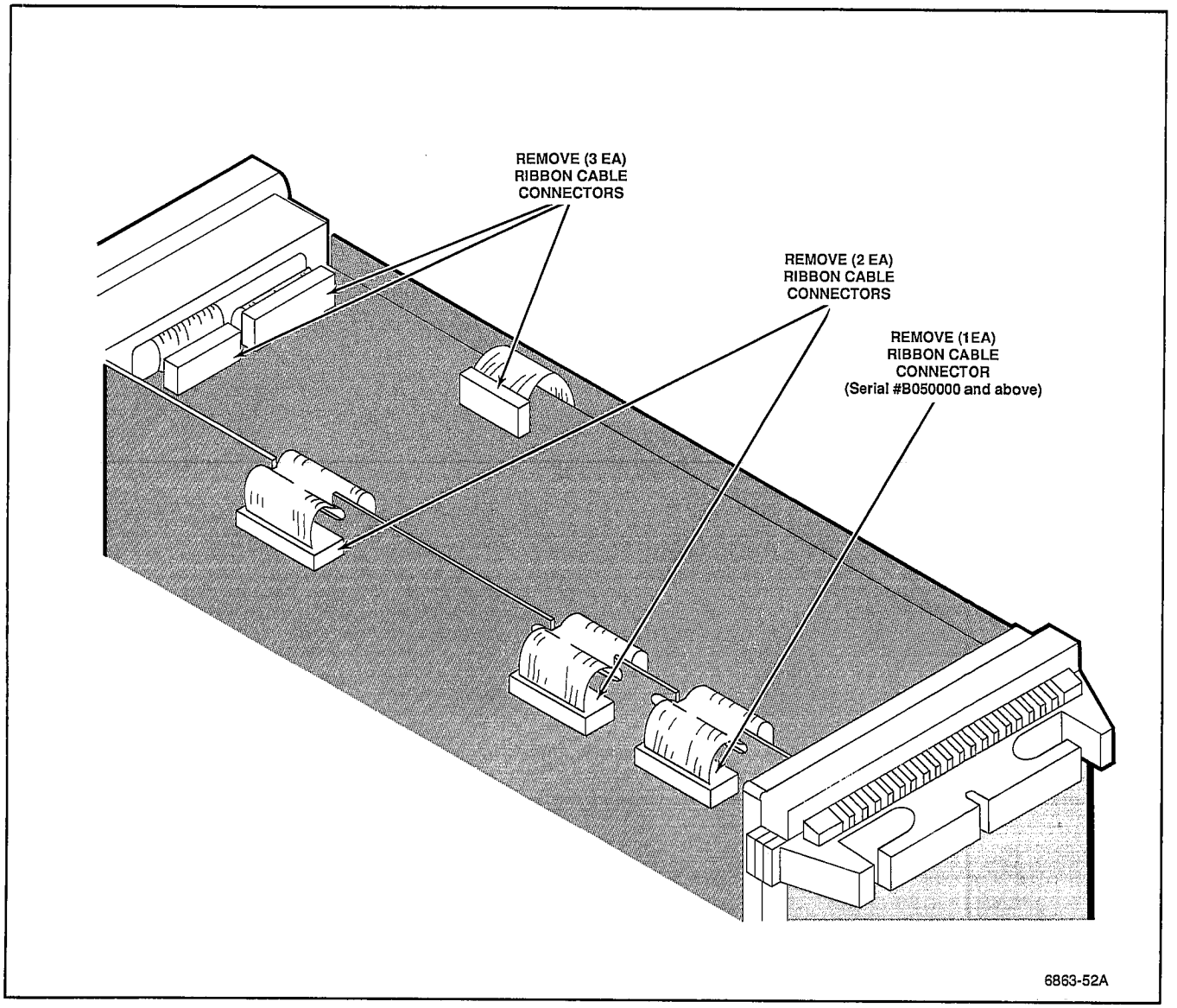


Figure 6-2. Ribbon cable removal.

3. Remove the Fan as described in that procedure.

4. Desolder the Fan power cable connecting the power supply to the feed-through capacitor (C10) on the inside of the rear plate.

5. Remove the two screws in the rear plate holding the black plastic primary circuit shield (located inside the chassis) and remove the shield.

6. Remove the two screws holding the rear of the Power Supply assembly to the rear plate.

7. Remove the three screws securing the power-transistor heatsink to the chassis.

8. Disconnect the power supply ribbon-cable connector (P251) from the Control board and feed the cable through the notch in the Control board and slot in the chassis.

9. Disconnect the two cables (P121 and P122) connecting the Main board to the Power Supply from the side of the Power Supply assembly.

10. Disconnect the four primary power connectors (P204, P205, P206, and P207) at the rear of the Power Supply assembly. Note their orientation for reinstallation.

11. If the Probe Power option is installed, disconnect the Probe Power connectors (P201 and P202) from the Power Supply assembly.

12. Lift the Power Supply assembly from the instrument.

To reinstall the Power Supply assembly, perform the reverse of the preceding instructions.

The following procedures describe the further disassembly of the Power Supply assembly circuit boards once the assembly is removed from the instrument.

INVERTER BOARD AND REGULATOR BOARD SEPARATION. To separate the Inverter and Regulator boards, perform the following steps:

1. Remove the rear-corner securing screw from the Regulator board and the two screws at the front edge of this board.

2. Unplug the four pin disconnect terminals (J231, J232, J233, and J234) while disabling the locking leg on the connector retainer.

3. (SN B050000 and above.) Separate the two circuit boards by removing the three black and one white plastic spacers from the top and bottom edges of the assembly. Note the location of the white spacer for reassembly.

3. (SN B049999 and below.) Separate the two circuit boards by removing the four black plastic spacers from the top and bottom edges of the assembly.

To rejoin the Inverter and Regulator boards, perform the reverse of the preceding steps.

A9—High-Voltage Board Removal

Removal of the High-Voltage board is accomplished by the following steps:

1. Remove the instrument Cabinet as described in that procedure.

2. Remove the Vertical Bracket as described in that procedure.

WARNING

The CRT anode lead may retain a high-voltage charge after the instrument is turned off. To avoid electrical shock, ground the CRT anode lead to the chassis after disconnecting the plug. Reconnect and disconnect the anode-lead plug several times, grounding the anode lead to chassis ground each time it is disconnected to fully dissipate the charge.

3. Unplug the CRT anode lead and discharge it to chassis ground.

4. Remove the high-voltage lead from the retainer cap.

5. Unplug the two leads connecting the CRT to the ceramic CRT terminator. Use long-nose pliers to pull the connectors straight away from the CRT neck pins. Avoid putting pressure on the metal-to-glass seal at the base of the pins.

6. Disconnect the single conductor connector from the ceramic CRT terminator.

7. Remove the two nuts retaining the ceramic CRT terminator to the chassis and remove the terminator.

8. Remove the nut retaining the high-voltage lead clamp to the chassis and remove the clamp.

9. Remove three screws on the rear CRT cover. Remove the cover.

10. Remove the five screws securing the High-Voltage Shield and remove the shield. If optional assembly cables are mounted in the shield's groove, it will be necessary to loosen these cables from the option board enough to slip the cover out underneath them.

11. Remove the high-voltage lead from the u-shaped grommet in the rear plate.

12. Unplug the CRT socket by gently prying evenly on both sides of the socket until the socket can be disengaged from the CRT pins. Do not apply side pressure on the socket.

13. Disconnect the connectors (P901, P902, P903, and P904) from the High-Voltage board. Note connector orientation for reinstallation.

14. Remove the four spacer posts securing the High-Voltage Board to the chassis.

15. Carefully tilt the top of the High Voltage board out far enough to clear the chassis side flange while pulling the board up gently to disengage the High-Voltage board pin connectors from the Main board.

16. Lift the board from the chassis while carefully feeding the CRT socket, cabling, and high-voltage lead through the rear plate slot.

To reinstall the High-Voltage Board, perform the reverse of the preceding instructions.

A4—Readout Board Removal (SN B049999 and Below)

Removal of the Readout Board is accomplished by the following steps:

1. Remove the instrument Cabinet as described in that procedure.

2. Remove the Vertical bracket as described in that procedure.

3. Place the instrument, left side down, on a flat surface.

4. Disconnect the Readout board ribbon-cable connector (P411) from the Main board.

5. With the instrument still on its side, pull the Readout board out of its plastic board mounts. Remove it from the instrument while guiding the ribbon cable and connector through the slots in the Main board and chassis.

To reinstall the Readout board, perform the reverse of the preceding steps.

A6—Front-Panel Circuit Board Assembly Removal

Removal of the Front-Panel circuit board assembly is accomplished by the following steps:

1. Remove the instrument Cabinet as described in that procedure.

2. Set the instrument back into its rear cover with the CRT facing up. Using a small-bladed screwdriver, gently pry up on the top cover trim strip to release it from the top edge of the front decorative trim ring.

3. Remove the four screws from the top edge of the front decorative trim ring.

4. Remove the four screws and the two plastic feet from the bottom edge of the front decorative trim ring.

5. Using firm outward pressure, pull the knobs from the four controls directly below the CRT (INTENSITY, FOCUS, READOUT INTENSITY, and SCALE ILLUM).

6. Slide off the front decorative trim ring. The clear implosion shield is retained by the trim ring. Use care to avoid dislodging the shield accidentally from its recess in the CRT frame.

7. Disconnect the ribbon-cable connector (P652) and the flex-circuit connector (P651) from the front of the Control Board. Feed the flex-circuit connector through the slot carefully while sliding the front panel gently outward.

8. Pull out the Front-Panel Circuit Board Assembly.

The following steps describe the further disassembly of the Front-Panel Assembly once it is removed from the instrument.

ASSEMBLY SEPARATION. Separation of the pot holder module from the Front-Panel Board is accomplished by the following steps:

1. Using a 1/16-inch Allen wrench, loosen the set screws in the CH 1 VOLTS/DIV VAR, CH 2 VOLTS/DIV VAR, and A and B SEC/DIV VAR knobs and remove these three knobs from their control shafts.

2. Using a 1/16-inch Allen wrench, loosen the six set screws in the CH 1 and CH 2 VOLTS/DIV knobs, and the SEC/DIV knob. Remove these three knobs from their control shafts.

3. Using firm outward pressure, pull off the remaining knobs. Note the locations of the knobs with indicator bars for reference during reinstallation.

4. On the rear of the assembly, remove the four screws securing the black variable resistor holder assembly.

5. Separate and slide out the above assembly with attached variable-control shafts. Avoid stressing the shafts to the side while sliding the assembly out.

FRONT-PANEL REMOVAL. Use the following procedure to further disassemble the Front-Panel circuit board assembly.

1. Separate the Front-Panel and variable resistor holder assembly as described above (if not already done).

2. Lift up the circuit board carefully to avoid dislodging any of the square push buttons from their switches.

3. Lift off the black plastic switch guide and mounting ring.

To reassemble and reinstall the Front-Panel assembly, perform the reverse of the preceding instructions. When reinstalling the circuit board, align all push buttons and LEDs with the black plastic switch guides before installing and tightening the screws.

A1A11 and A1A12—Channel 1 and Channel 2 Attenuator Assembly Removal

Removal of either the Channel 1 or Channel 2 Attenuator assembly is accomplished by the following steps:

1. Remove the instrument Cabinet as described in that procedure.

2. Remove the Front-Panel assembly as described in that procedure.

3. Remove the two screws holding the Attenuator support bar and remove the bar.

4. For each attenuator, remove the two screws holding the Attenuator to the front subpanel and the two screws holding it to the Main board (through access holes in the front panel compartment of the chassis).

5. Disconnect the associated multipin connector (either P10 for Channel 1 or P11 for Channel 2) from the Main board.

6. Remove the two screws holding the preamplifier shield and ground clip and remove them.

7. Desolder the two Attenuator output leads and the compensation capacitor lead.

8. Unplug the Attenuator by gently pulling the assembly straight up and away from the Main Board.

To reinstall a removed Attenuator assembly, perform the reverse of the preceding steps.

A1—Main Board Removal

Removal of the Main Board is accomplished by the following steps:

1. Remove the instrument Cabinet as described in that procedure.

2. Remove the Vertical Bracket as described in that procedure.

3. Remove the Front-Panel circuit board assembly as described in that procedure.

4. Disconnect the two power-supply multipin connectors (P121 and P122) from the side of the Power Supply assembly.

5. Disconnect the three ribbon-cable connectors (P411, P511, and P512) from the bottom of the Main board.

NOTE

See "Warning" under CRT Removal before proceeding.

6. Disconnect the vertical and horizontal deflection leads from the neck pins of the CRT. Access is via holes in the Main board. Use long-nose pliers to disconnect the pins by gently pulling straight up on the connectors. Avoid putting side pressure on the metal-to-glass seal of the CRT neck pins.

7. Desolder the rear-panel BNC connector leads from the BNCs. Unplug the CH 2 OUT cable (P105) from the Main board, and remove its cable retaining clamp.

8. Disconnect the flex-circuit connector (P120) for the CRT controls from the Main board.

9. Disconnect the two-conductor connector (P181) for the Scale Illumination board near the ASTIG and the SCALE ILLUM controls.

10. Remove the STEP/AUTO jack (J12) retaining nut from the rear plate after desoldering its wire from the Main board using correct vacuum desoldering techniques. Remove the jack.

11. Turn the long extension shaft (see Figure 6-3) CCW and unsnap it from the pivot bracket at the rear middle of the Main board, sliding it out of the bracket sideways.

CAUTION

Do not pull on the power switch push button or it will be damaged.

12. Remove the power switch push button mounting screw (item A) shown in Figure 6-3. Separate the long extension shaft from the short extension shaft at point B by inserting a small screwdriver tip in the slot while pulling out on the bracket at point C. Remove the screw (item D) and slide the long extension shaft out the rear of the front frame.

13. Remove the two screws holding the Attenuator support bar and remove the bar.

14. Remove the six screws holding the Attenuator assemblies and the CH 3 and CH 4 input connectors to the front subpanel.

15. Remove the Main board mounting screws (ten screws total) securing the Main board to the chassis.

16. Lift the rear of the Main board away from the chassis to unplug J191 and separate the Main board from the High Voltage board. When the plug pins are completely disengaged and the rear of the board clears the rear frame, slide the Main board rearward out of the front subpanel. Lift the Main board (with attached Delay Line) clear of the instrument while working the power supply cables through the slot in the chassis.

To reinstall the Main board, perform the reverse of the preceding instructions.

A8—Scale Illumination Circuit Board Removal

NOTE

See "Warning" under CRT removal instructions before proceeding.

Removal of the Scale-Illumination Circuit Board is accomplished by the following steps:

1. Remove the instrument Cabinet as described in that procedure.

2. Remove the front decorative trim ring as described in the A6—Front Panel board removal procedure.

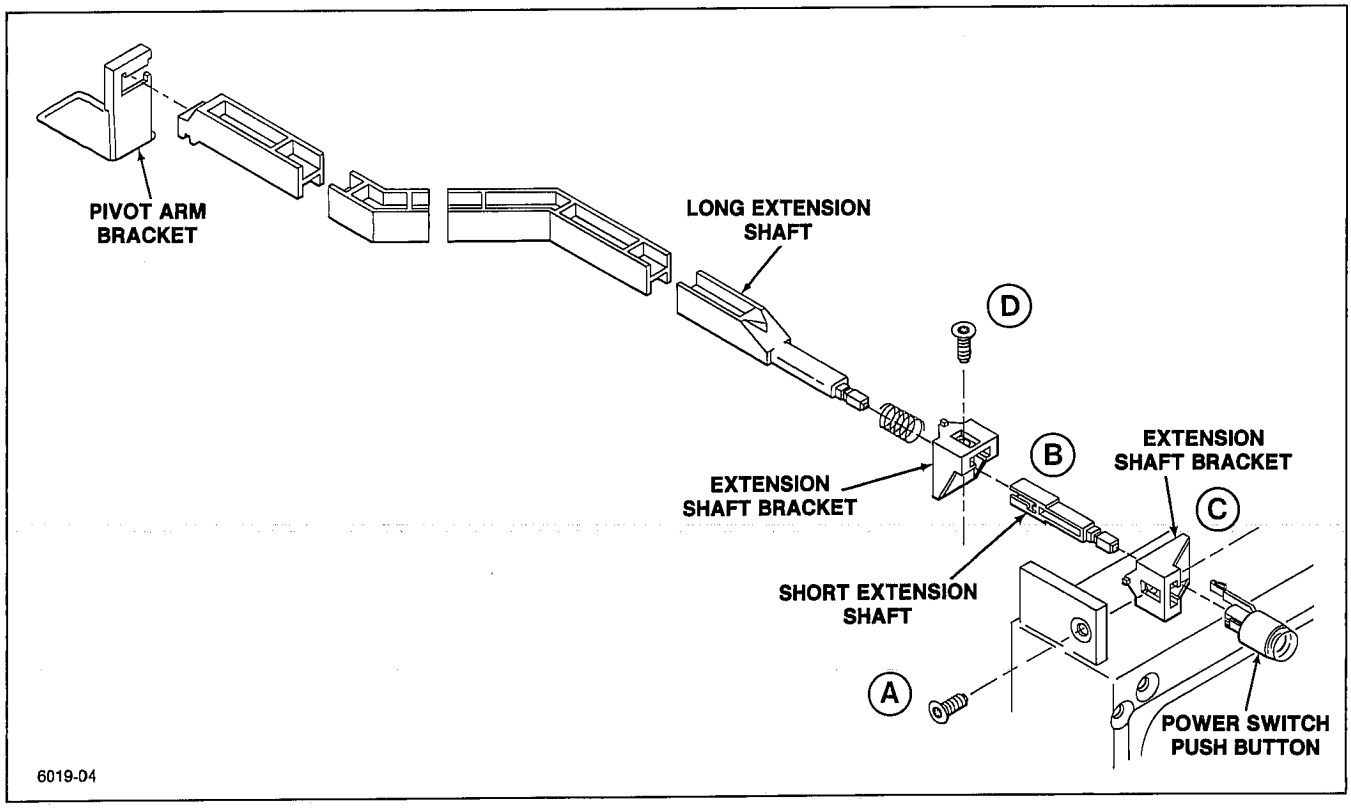


Figure 6-3. Power Switch Push Button Disassembly.

3. Remove the eight screws in the CRT frame. Remove frame and black plastic gasket. Note the difference in length of the screws for reinstallation.

4. Remove the clear plastic light reflector from the Scale-Illumination circuit board and the black plastic mounting spacer.

5. Disconnect the scale-illumination multipin connector (P181) from the Main board.

6. Remove the Scale-Illumination circuit board by lifting it away from the front subpanel while working the wires and connector through the slot in the subpanel.

To reinstall the Scale-Illumination circuit board, perform the reverse of the preceding instructions.

CRT Removal

WARNING

Use care when handling a CRT. Breakage of the CRT may cause high-speed scattering of glass fragments (implosion). Protective clothing and safety glasses (preferably a full-face shield) should be worn. Avoid striking the CRT on any object which may cause it to crack or implode. When storing a CRT, place it in a protective carton or set it face down on a smooth surface in a protected location. When stored face down, it should be placed on a soft, nonabrasive surface to prevent the CRT face plate from being scratched.

1. Remove the instrument Cabinet as described in that procedure.

2. Remove the Vertical Bracket as described in that procedure.

3. Remove three screws on the rear CRT cover. Remove the cover.

4. Unplug the CRT socket by gently prying the socket evenly on both sides until the pins can be disengaged. Do not apply side pressure on the socket.

WARNING

The CRT anode lead and the output terminal of the High-Voltage Multiplier can retain a high-voltage charge after the instrument is turned off. To avoid electrical shock, ground both the CRT anode lead and the high-voltage lead to the main instrument chassis. Repeat the grounding process several times to fully dissipate the charge.

5. Disconnect the CRT anode lead connector and discharge it to chassis ground.

6. Using long-nosed pliers, disconnect the horizontal and vertical deflection leads from the bottom of the CRT. Pull straight out on these connectors to prevent strain on the metal-to-glass seal. (Access to the connectors is through holes in the Main board).

7. Using long-nosed pliers, disconnect the vertical termination leads from the top of the CRT. Also disconnect the CRT shield ground lead from the top of the CRT.

8. Remove the five screws securing the High-Voltage Shield and remove the shield. If optional assembly cables are mounted in the shield's groove, it will be necessary to loosen these cables from the option board enough to slip the cover out underneath them.

9. Disconnect the connector (P903) from the front of the High-Voltage board. Note connector orientation for reinstallation.

10. Remove the front decorative trim ring as described in the A6—Front-Panel circuit board assembly removal instructions.

11. Remove the eight retaining screws from the CRT-mounting bezel at the front of the CRT. Note the difference in length of the screws for reinstallation. Push in on the four longer (outer) screws to disengage the CRT retainers.

12. Remove the CRT frame and black plastic gasket from the front of the instrument, working the frame gently from side to side to free it from the CRT (if required).

13. Slide the CRT out of the instrument while feeding the CRT leads through their respective holes in the CRT shield and front subpanel.

NOTE

Once the CRT is removed, it should be stored in such a manner as to protect it from impact. If stored face down, it should be placed on a soft, nonabrasive surface to prevent the CRT face plate from being scratched. To reinstall the CRT, perform the reverse of the preceding instructions. Be certain the two pins on the lower edge of the CRT frame align with the hole and slot in the front subpanel of the chassis. Tighten the shorter screws to 10 in-lb of torque before tightening any of the longer screws. Then tighten the longer screws in sequence:

2 1

4 3

Screw number one aligns the CRT. On the third time through the sequence, tighten each screw to 10 in-lb of torque.

OPTIONS

INTRODUCTION

This section contains a general description of instrument options available at the time of publication of this manual. Additional information about instrument options and option availability can be obtained either by consulting the current Tektronix Product Catalog or by contacting your local Tektronix Field Office or representative.

POWER CORD OPTIONS

Instruments are shipped with the detachable power-cord configuration ordered by the customer. Descriptive information about the international power-cord options is provided in Section 2, "Preparation for Use." The following list identifies the Tektronix part numbers for the optional power cords and associated fuses.

Universal Euro

Power cord (2.5 m)	Option A1
Fuse (1.6 A, 250 V, 5 x 20 mm, Quick-acting)	159-0098-00

UK

Power cord (2.5 m)	Option A2
Fuse (1.6 A, 250 V, 5 x 20 mm, Quick-acting)	159-0098-00

Australian

Power Cord (2.5 m)	Option A3
Fuse (1.6 A, 250V, 5 x 20 mm, Quick-acting)	159-0098-00

North American

Power Cord (2.5 m)	Option A4
Fuse (2 A, 250 V, AGC/3AG, Fast-blow)	159-0021-00

Switzerland

Power Cord (2.5 m)	Option A5
Fuse (1.6 A, 250 V, 5 x 20 mm, Quick-acting)	159-0098-00

OPTION 1R

When the oscilloscope is ordered with Option 1R, it is shipped in a configuration that permits easy installation into a 19-inch-wide electronic-equipment rack.

An optional rear-support kit is also available for use when rackmounting the instrument. Using this optional rear-support kit enables the rackmounted instrument to meet appropriate electrical and environmental specifications.

Connector-mounting holes are provided in the front panel of the rackmounted instrument. These enable convenient accessing of the four bnc connectors (CH 2 SIGNAL OUT, A GATE OUT, B GATE OUT, and EXT Z AXIS IN) and the two PROBE POWER connectors located on the rear panel. Additional cabling and connectors required to implement any front-panel access to the rear-panel connectors are supplied by the user; however, these items can be separately ordered from Tektronix.

Complete rackmounting instructions are provided in a separate document shipped with Option 1R. These instructions also contain appropriate procedures to convert a standard instrument into the Option 1R configuration by using the rackmounting conversion kit.

OPTION 01

Option 01 (DMM) adds a 4-1/2 digit, fully autoranging digital multimeter which measures dc and ac voltage and current, resistance, dBV, dBm, continuity, and temperature. Option 1B is the same as Option 01 except that the temperature probe is not included. Measurement results and DMM messages are displayed on the top line of the oscilloscope CRT readout.

OPTION 05

Option 05 (TV) simplifies triggering and viewing of television signals. The option adds TV (back-porch) clamp circuitry to the Channel 2 input and TV trigger coupling modes, allowing selection of either horizontal or vertical sync pulses to obtain horizontal-line-sync or field-sync pulse triggering. This option permits triggering on a specific line number within a TV field and provides sync polarity switching for either sync-negative or sync-positive composite video signals.

OPTIONS 06 AND 09

Options 06 (Counter/Timer/Trigger) and 09 (Counter/Timer/Trigger with Word Recognizer) allow precision time-interval measurement, event and frequency counting, delay-by-events triggering, and logic triggering. The 17-bit Word Recognizer probe of Option 09 extends the capabilities of these functions.

OPTION 10

Option 10 allows the instrument to be remotely controlled and queried using a standard interface system. The interface implemented conforms to the specifications contained in *IEEE Standard Digital Interface for Programmable Instrumentation (ANSI/IEEE Std 488-1978)*, commonly referred to as the General Purpose Interface Bus (GPIB). It also complies with a Tektronix Standard relating to GPIB Codes, Formats, Conventions and Features.

OPTION 11

Option 11 provides two probe-power connectors on the rear panel of the instrument. Voltages supplied at these connectors meet the power requirements of standard Tektronix active oscilloscope probes.

REPLACEABLE ELECTRICAL PARTS

PARTS ORDERING INFORMATION

Replacement parts are available from or through your local Tektronix, Inc. Field Office or representative.

Changes to Tektronix instruments are sometimes made to accommodate improved components as they become available, and to give you the benefit of the latest circuit improvements developed in our engineering department. It is therefore important, when ordering parts, to include the following information in your order: Part number, instrument type or number, serial number, and modification number if applicable.

If a part you have ordered has been replaced with a new or improved part, your local Tektronix, Inc. Field Office or representative will contact you concerning any change in part number.

Change information, if any, is located at the rear of this manual.

LIST OF ASSEMBLIES

A list of assemblies can be found at the beginning of the Electrical Parts List. The assemblies are listed in numerical order. When the complete component number of a part is known, this list will identify the assembly in which the part is located.

CROSS INDEX-MFR. CODE NUMBER TO MANUFACTURER

The Mfr. Code Number to Manufacturer index for the Electrical Parts List is located immediately after this page. The Cross Index provides codes, names and addresses of manufacturers of components listed in the Electrical Parts List.

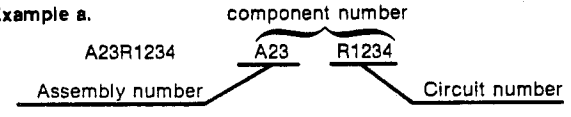
ABBREVIATIONS

Abbreviations conform to American National Standard Y1.1.

COMPONENT NUMBER (column one of the Electrical Parts List)

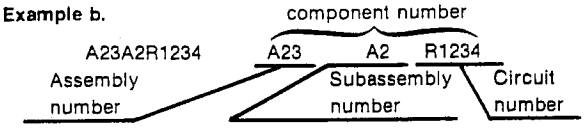
A numbering method has been used to identify assemblies, subassemblies and parts. Examples of this numbering method and typical expansions are illustrated by the following:

Example a.



Read: Resistor 1234 of Assembly 23

Example b.



Read: Resistor 1234 of Subassembly 2 of Assembly 23

Only the circuit number will appear on the diagrams and circuit board illustrations. Each diagram and circuit board illustration is clearly marked with the assembly number. Assembly numbers are also marked on the mechanical exploded views located in the Mechanical Parts List. The component number is obtained by adding the assembly number prefix to the circuit number.

The Electrical Parts List is divided and arranged by assemblies in numerical sequence (e.g., assembly A1 with its subassemblies and parts, precedes assembly A2 with its subassemblies and parts).

Chassis-mounted parts have no assembly number prefix and are located at the end of the Electrical Parts List.

TEKTRONIX PART NO. (column two of the Electrical Parts List)

Indicates part number to be used when ordering replacement part from Tektronix.

SERIAL/MODEL NO. (columns three and four of the Electrical Parts List)

Column three (3) indicates the serial number at which the part was first used. Column four (4) indicates the serial number at which the part was removed. No serial number entered indicates part is good for all serial numbers.

NAME & DESCRIPTION (column five of the Electrical Parts List)

In the Parts List, an Item Name is separated from the description by a colon (:). Because of space limitations, an Item Name may sometimes appear as incomplete. For further Item Name identification, the U.S. Federal Cataloging Handbook H6-1 can be utilized where possible.

MFR. CODE (column six of the Electrical Parts List)

Indicates the code number of the actual manufacturer of the part. (Code to name and address cross reference can be found immediately after this page.)

MFR. PART NUMBER (column seven of the Electrical Parts List)

Indicates actual manufacturers part number.

CROSS INDEX - MFR. CODE NUMBER TO MANUFACTURER

Mfr. Code	Manufacturer	Address	City, State, Zip Code
00213	NYTRONICS COMPONENTS GROUP INC SUBSIDIARY OF NYTRONICS INC	ORANGE ST	DARLINGTON SC 29532
00779	AMP INC	2800 FULLING MILL PO BOX 3608	HARRISBURG PA 17105
01121	ALLEN-BRADLEY CO	1201 S 2ND ST	MILWAUKEE WI 53204-2410
01295	TEXAS INSTRUMENTS INC SEMICONDUCTOR GROUP	13500 N CENTRAL EXPY PO BOX 655012	DALLAS TX 75265
02113	COILCRAFT INC	1102 SILVER LAKE RD	CARY IL 60013-1658
02735	RCA CORP SOLID STATE DIVISION	ROUTE 202	SOMERVILLE NJ 08876
03508	GENERAL ELECTRIC CO SEMI-CONDUCTOR PRODUCTS DEPT	W GENESEE ST	AUBURN NY 13021
04222	AVX CERAMICS DIV OF AVX CORP	19TH AVE SOUTH P O BOX 867	MYRTLE BEACH SC 29577
04713	MOTOROLA INC SEMICONDUCTOR PRODUCTS SECTOR	5005 E MCDOWELL RD	PHOENIX AZ 85008-4229
05292	ITT COMPONENTS DIV		CLIFTON NJ
05397	UNION CARBIDE CORP MATERIALS SYSTEMS DIV	11901 MADISON AVE	CLEVELAND OH 44101
05828	GENERAL INSTRUMENT CORP GOVERNMENT SYSTEMS DIV	600 W JOHN ST	HICKSVILLE NY 11802
06665	PRECISION MONOLITHICS INC SUB OF BOURNS INC	1500 SPACE PARK DR	SANTA CLARA CA 95050
07263	FAIRCHILD SEMICONDUCTOR CORP NORTH AMERICAN SALES SUB OF SCHLUMBERGER LTD MS 118	10400 RIDGEVIEW CT	CUPERTINO CA 95014
07716	TRW INC TRW IRC FIXED RESISTORS/BURLINGTON	2850 MT PLEASANT AVE	BURLINGTON IA 52601
09019	GENERAL ELECTRIC CO POWER ELECTRONICS SYSTEMS DEPT	ELECTRONICS PARK BLDG 7	SYRACUSE NY 13221
09353	C AND K COMPONENTS INC	15 RIVERDALE AVE	NEWTON MA 02158-1057
09922	BURNDY CORP	RICHARDS AVE	NORWALK CT 06852
11236	CTS CORP BERNE DIV THICK FILM PRODUCTS GROUP	406 PARR ROAD	BERNE IN 46711-9506
12954	MICROSEMI CORP - SCOTTSDALE	8700 E THOMAS RD P O BOX 1390	SCOTTSDALE AZ 85252
12969	UNITRODE CORP	5 FORBES RD	LEXINGTON MA 02173-7305
14301	ANDERSON ELECTRONICS INC	310 PENN ST PO BOX 89	HOLLIDAYSBURG PA 16648-2009
14433	ITT SEMICONDUCTORS DIV		WEST PALM BEACH FL
14552	MICROSEMI CORP	2830 S FAIRVIEW ST	SANTA ANA CA 92704-5948
14674	CORNING GLASS WORKS	HOUGHTON PK	CORNING NY 14830
14752	ELECTRO CUBE INC	1710 S DEL MAR AVE	SAN GABRIEL CA 91776-3825
15454	KETMA RODAN DIVISION	2900 BLUE STAR STREET	ANAHEIM CA 92806-2591
18324	SIGNETICS CORP MILITARY PRODUCTS DIV	4130 S MARKET COURT	SACRAMENTO CA 95834-1222
19701	MEPCO/CENTRALAB A NORTH AMERICAN PHILIPS CO MINERAL WELLS AIRPORT	PO BOX 760	MINERAL WELLS TX 76067-0760
20462	PREM MAGNETICS INC	3519 N CHAPEL HILL	MCHENRY IL 60050-2504
20932	KYOCERA INTERNATIONAL INC	11620 SORRENTO VALLEY RD PO BOX 81543 PLANT NO 1	SAN DIEGO CA 92121
22526	DU PONT E I DE NEMOURS AND CO INC DU PONT CONNECTOR SYSTEMS DIV MILITARY PRODUCTS GROUP	515 FISHING CREEK RD	NEW CUMBERLAND PA 17070-3007
24226	GOWANDA ELECTRONICS CORP	NO 1 INDUSTRIAL PL	GOWANDA NY 14070-1409
24546	CORNING GLASS WORKS	550 HIGH ST	BRADFORD PA 16701-3737
25088	SIEMENS CORP	186 WOOD AVE S	ISELIN NJ 08830-2704
27264	MOLEX INC	2222 WELLINGTON COURT	LISLE IL 60532-1613
31471	AMERICAN MICRO SYSTEMS INC	3800 HOMESTEAD RD	SANTA CLARA CA 95051-4542
31918	ITT SCHADOW INC	8081 WALLACE RD	EDEN PRAIRIE MN 55344-2224
32159	WEST-CAP ARIZONA SUB OF SFE TECHNOLOGIES	2201 E ELVIRA ROAD	TUCSON AZ 85706-7026

CROSS INDEX - MFR. CODE NUMBER TO MANUFACTURER

Mfr. Code	Manufacturer	Address	City, State, Zip Code
32997	BOURNS INC TRIMPOT DIV	1200 COLUMBIA AVE	RIVERSIDE CA 92507-2114
34335	ADVANCED MICRO DEVICES	901 THOMPSON PL	SUNNYVALE CA 94086-4518
34479	RENCO CORP	26 COROMAR DRIVE	GOLETA CA 93117-3024
34899	FAIR-RITE PRODUCTS CORP	1 COMMERCIAL ROW	WALLKILL NY 12589
50434	HEWLETT-PACKARD CO OPTOELECTRONICS DIV	370 W TRIMBLE RD	SAN JOSE CA 95131
51406	MURATA ERIE NORTH AMERICA INC HEADQUARTERS AND GEORGIA OPERATIONS	2200 LAKE PARK DR	SMYRNA GA 30080
52769	SPRAGUE-GOODMAN ELECTRONICS INC	134 FULTON AVE	GARDEN CITY PARK NY 11040-5352
53387	MINNESOTA MINING MFG CO 3M ELECTRONIC PRODUCTS DIV	3M CENTER	ST PAUL MN 55101-1428
54473	MATSUSHITA ELECTRIC CORP OF AMERICA	ONE PANASONIC WAY PO BOX 1501	SECAUCUS NJ 07094-2917
54583	TDK ELECTRONICS CORP	12 HARBOR PARK DR	PORT WASHINGTON NY 11550
54937	DEYOUNG MANUFACTURING INC	12920 NE 125TH WAY	KIRKLAND WA 98034-7716
55112	WESTLAKE CAPACITORS INC	5334 STERLING CENTER DRIVE	WESTLAKE VILLAGE CA 91361
55680	NICHICON /AMERICA/ CORP	927 E STATE PKY	SCHAUMBURG IL 60195-4526
56289	SPRAGUE ELECTRIC CO WORLD HEADQUARTERS	92 HAYDEN AVE	LEXINGTON MA 02173-7929
56845	DALE ELECTRONICS INC	2300 RIVERSIDE BLVD PO BOX 74	NORFOLK NE 68701-2242
57668	ROHM CORP	8 WHATNEY PO BOX 19515	IRVINE CA 92713
58224	XENELL CORP	11 DUNBARTON RD PO BOX 4401	CHERRY HILL NJ 08003-2107
59660	TUSONIX INC	7741 N BUSINESS PARK DR PO BOX 37144	TUCSON AZ 85740-7144
59821	MEPCO/CENTRALAB A NORTH AMERICAN PHILIPS CO	7158 MERCHANT AVE	EL PASO TX 79915-1207
61271	FUJITSU MICROELECTRONICS INC	2985 KIFER RD	SANTA CLARA CA 95051-0802
62786	HITACHI AMERICA LTD	1800 BERING DRIVE	SAN JOSE CA 95122
65786	CYPRESS SEMICONDUCTOR CORP	3901 N 1ST ST	SAN JOSE CA 95134-1506
71400	BUSSMANN DIV OF COOPER INDUSTRIES INC	114 OLD STATE RD PO BOX 14460	ST LOUIS MO 63178
71744	GENERAL INSTRUMENT CORP LAMP DIV/WORLD WIDE/	4433 N RAVENSWOOD AVE	CHICAGO IL 60640-5802
72982	ERIE SPECIALTY PRODUCTS INC	645 W 11TH ST	ERIE PA 16512
73138	BECKMAN INDUSTRIAL CORP BECKMAN ELECTRONIC TECHNOLOGIES	4141 PALM ST	FULLERTON CA 92635
75042	SUB OF EMERSON ELECTRIC IRC ELECTRONIC COMPONENTS PHILADELPHIA DIV TRW FIXED RESISTORS	401 N BROAD ST	PHILADELPHIA PA 19108-1001
80009	TEKTRONIX INC	14150 SW KARL BRAUN DR PO BOX 500	BEAVERTON OR 97077-0001
81483	INTERNATIONAL RECTIFIER	9220 SUNSET BLVD	LOS ANGELES CA 90069-3501
81855	EAGLE-PICHER INDUSTRIES INC ELECTRONICS DIV	COUPLES DEPT C AND PORTER STS PO BOX 47	JOPLIN MO 64801
91637	DALE ELECTRONICS INC	2064 12TH AVE PO BOX 609	COLUMBUS NE 68601-3632
93410	ESSEX GROUP ING CONTROLS DIV LEXINGTON PLANT	45-55 PLYMOUTH ST P O BOX 1007	LEXINGTON OH 44904
S4431	MURATA MFG CO LTD	16 KAIDEN NISHIJM CHO NAGAOKAKY-CITY	KYOTO JAPAN
TK0515	ERICSSON COMPONENTS INC	403 INTERNATIONAL PKY PO BOX 853904	RICHARDSON TX 75085-3904
TK0935	MARQUARDT SWITCHES INC	67 ALBANY ST PO BOX 465	CAZENOVIA NY 13035-1219
TK0946	SAN-O INDUSTRIAL CORP	170 WILBUR PL	BAHEMIA LONG ISLAND NY 11716
TK0961	NEC ELECTRONICS USA INC ELECTRON DIV	401 ELLIS ST PO BOX 7241	MOUNTAIN VIEW CA 94039
TK1345	ZMAN AND ASSOCIATES	7633 S 180TH	KENT WA 98032
TK1450	TOKYO COSMOS ELECTRIC CO LTD	2-268 SOBUDAI ZAWA	KANAGAWA 228 JAPAN

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Mfr. Code	Manufacturer	Address	City, State, Zip Code
TK1492	COFER COMPONENT PROCESSING	3270 KELLER ST UNIT 11	SANTA CLARA CA 95050
TK1544	COMPUTER CONNECTIONS	30608 SAN ANTONIO ST	HAYWARD CA 94544
TK1573	WILHELM WESTERMAN	PO BOX 2345 AUGUSTA-ANLAGE 56	6800 MANNHEIM 1 WEST GERMANY
TK1727	PHILIPS NEDERLAND BV AFD ELONCO	POSTBUS 90050	5600 PB EINDHOVEN THE NETHERLANDS
TK1899	MINNESOTA MINING AND MFG CO	5400 RT B PO BOX 1228	COLUMBIA MO 65205
TK2042	ZMAN & ASSOCIATES	7633 S 180TH	KENT WA 98032
TK2282	KYOCERA AMERICA INC	5701 E FOURTH PLAIN BLVD	VANCOUVER WA 98661

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A1	671-0720-00	B010100	B010513	CIRCUIT BD ASSY:MAIN	80009	671-0720-00
A1	671-0720-05	B010514	B051039	CIRCUIT BD ASSY:MAIN	80009	671-0720-05
A1	671-0720-06	B051040		CIRCUIT BD ASSY:MAIN	80009	671-0720-06
A2	672-1037-12			CIRCUIT BD ASSY:LV PWR SPLY MODULE	80009	672-1037-12
A2A1	----- -----			CIRCUIT BD ASSY:REGULATOR (AVAILABLE AT THE 672-1037-XX LEVEL ONLY)		
A3	----- -----			CIRCUIT BD ASSY:INVERTER (AVAILABLE AT THE 672-1037-XX LEVEL ONLY)		
A4	670-9493-02	B010100	B049999	CIRCUIT BD ASSY:READOUT	80009	670-9493-02
A5	670-9052-02	B010100	B049999	CIRCUIT BD ASSY:DIGITAL CONTROL	80009	670-9052-02
A5	671-0965-00	B050000		CIRCUIT BD ASSY:CONTROL/READOUT/BUFFER	80009	671-0965-00
A6	614-0825-00			FRONT PNL ASSY:STANDARD,2445B/55B/65B & 67B (STANDARD)	80009	614-0825-00
A6	614-0826-00			FRONT PNL ASSY:TV OPTION,2445B/55B/65B/67B (OPTION 05)	80009	614-0826-00
A6A1	----- -----			CIRCUIT BD ASSY:FRONT PANEL (REPLACEABLE AT A6 LEVEL ONLY)		
A8	670-7280-00			CIRCUIT BD ASSY:SCALE ILLUM	80009	670-7280-00
A9	670-7277-09			CIRCUIT BD ASSY:HIGH VOLTAGE	80009	670-7277-09
A14	670-8000-00			CIRCUIT BD ASSY:DYNAMIC CENTERING	80009	670-8000-00

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A1	671-0720-00	B010100	B010513	CIRCUIT BD ASSY:MAIN	80009	671-0720-00
A1	671-0720-05	B010514	B051039	CIRCUIT BD ASSY:MAIN	80009	671-0720-05
A1	671-0720-06	B051040		CIRCUIT BD ASSY:MAIN	80009	671-0720-06
A1A11	119-2342-05	B010100	B010689	ATTENUATOR,VAR:PROGRAMMABLE 1X-100X	80009	119-2342-05
A1A11	119-2342-07	B010690	B051100	ATTENUATOR,VAR:1X-100X,CHANNEL 1	80009	119-2342-07
A1A11	119-2342-09	B051101		ATTENUATOR,VAR:1X-100X,CHANNEL 1	80009	119-2342-09
A1A12	119-2342-06	B010100	B010689	ATTENUATOR,VAR:PROGRAMMABLE 1X-100X	80009	119-2342-06
A1A12	119-2342-08	B010690	B051100	ATTENUATOR,VAR:1X-100X,CHANNEL 2	80009	119-2342-08
A1A12	119-2342-10	B051101		ATTENUATOR,VAR:1X-100X,CHANNEL 2	80009	119-2342-10
A1C100	283-0000-00			CAP,FXD,CER DI:0.001UF,+100-0%,500V (COMBO WITH A1C200)	59660	831-610-Y5U0102P
A1C102	290-0973-00			CAP,FXD,ELCTLT:100UF,20%,25VDC	55680	ULB1E101MPA
A1C103	281-0812-00			CAP,FXD,CER DI:1000PF,10%,100V	04222	MA101C102KAA
A1C105	281-0064-00			CAP,VAR,PLASTIC:0.25-1.5PF,600V	52769	ER-530-013
A1C106	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C107	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C108	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C109	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C110	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C113	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C114	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C115	281-0761-00			CAP,FXD,CER DI:27PF,5%,100V	04222	MA101A270JAA
A1C116	281-0814-00			CAP,FXD,CER DI:100 PF,10%,100V	04222	MA101A101KAA
A1C117	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C118	281-0205-00			CAP,VAR,PLASTIC:5.5-65 PF,100V	TK1727	2222-808-32659
A1C119	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C120	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C121	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C125	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C130	290-0776-01			CAP,FXD,ELCTLT:22UF,20%,10WVDC	55680	ULB1A220MAA1TD
A1C152	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C154	281-0812-00			CAP,FXD,CER DI:1000PF,10%,100V	04222	MA101C102KAA
A1C171	281-0851-00			CAP,FXD,CER DI:180PF,5%,100VDC	04222	MA101A181JAA
A1C175	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A1C176	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A1C177	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A1C179	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A1C180	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A1C181	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A1C182	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A1C183	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A1C184	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C185	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C200	283-0000-00			CAP,FXD,CER DI:0.001UF,+100-0%,500V (COMBO WITH A1C100)	59660	831-610-Y5U0102P
A1C202	281-0812-00			CAP,FXD,CER DI:1000PF,10%,100V	04222	MA101C102KAA
A1C203	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A1C205	281-0064-00			CAP,VAR,PLASTIC:0.25-1.5PF,600V	52769	ER-530-013
A1C207	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C209	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C210	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C211	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C217	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C218	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C219	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C220	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C221	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C223	281-0812-00			CAP,FXD,CER DI:1000PF,10%,100V	04222	MA101C102KAA

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A1C225	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C301	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C302	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C307	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C310	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C311	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C325	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C329	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A1C332	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A1C336	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C351	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C402	281-0762-00			CAP,FXD,CER DI:27PF,20%,100V	04222	MA101A270MAA
A1C403	281-0218-00			CAP,VAR,CER DI:1-5PF,+2 -2.5%,100V	59660	513-011A1-5
A1C404	281-0218-00			CAP,VAR,CER DI:1-5PF,+2 -2.5%,100V	59660	513-011A1-5
A1C412	281-0762-00			CAP,FXD,CER DI:27PF,20%,100V	04222	MA101A270MAA
A1C415	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C458	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C460	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C464	281-0763-00			CAP,FXD,CER DI:47PF,10%,100V	04222	MA101A470KAA
A1C466	281-0763-00			CAP,FXD,CER DI:47PF,10%,100V	04222	MA101A470KAA
A1C478	281-0759-00			CAP,FXD,CER DI:22PF,10%,100V	04222	MA101A220KAA
A1C480	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C487	281-0823-00			CAP,FXD,CER DI:470PF,10%,50V	04222	MA105A471KAA
A1C488	281-0814-00			CAP,FXD,CER DI:100 PF,10%,100V	04222	MA101A101KAA
A1C500	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C501	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C512	290-0246-00			CAP,FXD,ELCTLT:3.3UF,10%,15V	12954	D3R3EA15K1
A1C513	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A1C520	281-0814-00	B010100	B011521	CAP,FXD,CER DI:100 PF,10%,100V	04222	MA101A101KAA
A1C520	281-0777-00	B011522		CAP,FXD,CER DI:51PF,5%,100V	04222	MA101A510JAA
A1C520	281-0814-00	B050000	B051076	CAP,FXD,CER DI:100 PF,10%,100V	04222	MA101A101KAA
A1C520	281-0777-00	B051077		CAP,FXD,CER DI:51PF,5%,100V	04222	MA101A510JAA
A1C521	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C528	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C536	290-0246-00			CAP,FXD,ELCTLT:3.3UF,10%,15V	12954	D3R3EA15K1
A1C537	281-0812-00			CAP,FXD,CER DI:1000PF,10%,100V	04222	MA101C102KAA
A1C544	281-0814-00	B010100	B011521	CAP,FXD,CER DI:100 PF,10%,100V	04222	MA101A101KAA
A1C544	281-0777-00	B011522		CAP,FXD,CER DI:51PF,5%,100V	04222	MA101A510JAA
A1C544	281-0814-00	B050000	B051076	CAP,FXD,CER DI:100 PF,10%,100V	04222	MA101A101KAA
A1C544	281-0777-00	B051077		CAP,FXD,CER DI:51PF,5%,100V	04222	MA101A510JAA
A1C601	281-0270-00			CAP,VAR,CER DI:9-90PF,50V	51406	TZ03R900E
A1C617	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A1C625	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C645	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A1C650	281-0823-00			CAP,FXD,CER DI:470PF,10%,50V	04222	MA105A471KAA
A1C653	281-0819-00			CAP,FXD,CER DI:33 PF,5%,50V	04222	GC105A330J
A1C660	281-0851-00			CAP,FXD,CER DI:180PF,5%,100VDC	04222	MA101A181JAA
A1C669	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C675	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C707	281-0808-00			CAP,FXD,CER DI:7 PF,20%,100V	04222	MA101A7R04AA
A1C708	285-0676-01			CAP,FXD,PLASTIC:0.1UF,3.5%,35V	80009	285-0676-01
A1C709	285-1060-00			CAP,FXD,PLASTIC:10UF,3%,25V	80009	285-1060-00
A1C710	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C712	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A1C722	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C723	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C730	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C731	290-0944-01			CAP,FXD,ELCTLT:220UF,20%,10V	55680	UVX1C221MPA1TA

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A1C732	290-0944-01			CAP,FXD,ELCTLT:220UF,20%,10V	55680	UVX1C221MPA1TA
A1C733	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C735	281-0823-00			CAP,FXD,CER DI:470PF,10%,50V	04222	MA105A471KAA
A1C738	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C740	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C742	281-0812-00			CAP,FXD,CER DI:1000PF,10%,100V	04222	MA101C102KAA
A1C744	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C755	281-0759-00			CAP,FXD,CER DI:22PF,10%,100V	04222	MA101A220KAA
A1C803	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C804	281-0797-00			CAP,FXD,CER DI:15PF,10%,100V	04222	SA106A150KAA
A1C805	281-0823-00			CAP,FXD,CER DI:470PF,10%,50V	04222	MA105A471KAA
A1C806	283-0156-00			CAP,FXD,CER DI:1000PF,+80-20%,200V	04222	SR152E102ZAA
A1C808	281-0757-00			CAP,FXD,CER DI:10PF,20%,100V TUBULAR,MI	04222	MA101A100MAA
A1C809	281-0819-00			CAP,FXD,CER DI:33 PF,5%,50V	04222	GC105A330J
A1C810	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C811	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C817	281-0812-00			CAP,FXD,CER DI:1000PF,10%,100V	04222	MA101C102KAA
A1C819	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C822	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C823	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C830	281-0812-00			CAP,FXD,CER DI:1000PF,10%,100V	04222	MA101C102KAA
A1C848	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C849	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C850	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C851	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A1C852	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A1C853	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A1C854	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A1C900	281-0763-00			CAP,FXD,CER DI:47PF,10%,100V	04222	MA101A470KAA
A1C903	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C907	281-0808-00			CAP,FXD,CER DI:7 PF,20%,100V	04222	MA101A7R04AA
A1C908	285-0752-03			CAP,FXD,PLASTIC:1UF,3%,50V	80009	285-0752-03
A1C912	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C933	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C938	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C940	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C941	281-0763-00			CAP,FXD,CER DI:47PF,10%,100V	04222	MA101A470KAA
A1C943	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C947	281-0759-00			CAP,FXD,CER DI:22PF,10%,100V	04222	MA101A220KAA
A1C957	290-0804-00			CAP,FXD,ELCTLT:10UF,+50-20%,25V	55680	ULB1E100TAAANA
A1C958	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C966	281-0783-00			CAP,FXD,CER DI:0.1 UF 20%,100V	04222	MA401C104MAA
A1C967	281-0783-00			CAP,FXD,CER DI:0.1 UF 20%,100V	04222	MA401C104MAA
A1C972	281-0756-00			CAP,FXD,CER DI:2.2PF,+/-0.5PF,200V	04222	SA102A2R2DAA
A1C973	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C975	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C976	283-1001-00			CAP,FXD,CER DI:0.03UF,50VDC	80009	283-1001-00
A1C977	290-0246-00			CAP,FXD,ELCTLT:3.3UF,10%,15V	12954	D3R3EA15K1
A1C980	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C981	283-1000-00			CAP,FXD,CER DI:0.02UF,50VDC	80009	283-1000-00
A1C982	281-0759-00			CAP,FXD,CER DI:22PF,10%,100V	04222	MA101A220KAA
A1C985	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C988	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C990	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C995	281-0810-00			CAP,FXD,CER DI:5.6PF,+/-0.5PF,100V	04222	MA101A5R6DAA
A1CR100	152-0323-01			SEMICOND DVC,DI:SW,SI,50V,25PA AT 20V,20PF	14552	MT5127
A1CR101	152-0323-01			SEMICOND DVC,DI:SW,SI,50V,25PA AT 20V,20PF	14552	MT5127
A1CR107	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020

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A1CR130	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR131	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR140	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR141	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR142	152-0141-02			SEMICOND DVC,DI:SW,SI;30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR143	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR144	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR145	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR146	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR147	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR148	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR149	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR150	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR151	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR152	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR153	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR154	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR155	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR161	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR162	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR163	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR180	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR181	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR200	152-0323-01			SEMICOND DVC,DI:SW,SI,50V,25PA AT 20V,20PF	14552	MT5127
A1CR201	152-0323-01			SEMICOND DVC,DI:SW,SI,50V,25PA AT 20V,20PF	14552	MT5127
A1CR354	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR360	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR460	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR461	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR476	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR484	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR485	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR495	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR503	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR538	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR539	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR600	152-0141-02			SEMICOND DVC,DI:SW,SI;30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR601	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR616	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR619	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR620	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR621	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR652	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR653	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR707	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR741	152-0951-00			SEMICOND DVC,DI:SCHOTTKY,SI,60V,2.25PF	80009	152-0951-00
A1CR742	152-0951-00			SEMICOND DVC,DI:SCHOTTKY,SI,60V,2.25PF	80009	152-0951-00
A1CR746	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR747	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR752	152-0075-00			SEMICOND DVC,DI:SW,GE,22V,80MW,DO-7	80009	152-0075-00
A1CR753	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR807	152-0574-00			SEMICOND DVC,DI:SW,SI,120V,0.150MA,4NS,DO35	12969	NDP566
A1CR811	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR850	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR941	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR942	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR950	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR951	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A1CR956	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR966	152-0574-00			SEMICOND DVC,DI:SW,SI,120V,0.150MA,4NS,DO35	12969	NDP566
A1CR972	152-0574-00			SEMICOND DVC,DI:SW,SI,120V,0.150MA,4NS,DO35	12969	NDP566
A1CR982	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR987	152-0574-00			SEMICOND DVC,DI:SW,SI,120V,0.150MA,4NS,DO35	12969	NDP566
A1CR995	152-0061-00			SEMICOND DVC,DI:SW,SI,175V,0.1A,DO-35	07263	FDH2161
A1DL100	119-1490-04			DELAY LINE,ELEC:73NS,150 OHM	80009	119-1490-04
A1E900	276-0712-00			CORE,EM:BALUN,FERRITE	34899	2843002402
A1J1	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 3)	22526	48283-036
A1J9	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL	22526	48283-036
A1J11	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 3)	22526	48283-036
A1J100	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 2)	22526	48283-036
A1J101	131-3520-00			CONN,RCPT,ELEC:HEADER,10 CONT,STR SLDR PIN	53387	3591-6002
A1J102	131-3520-00			CONN,RCPT,ELEC:HEADER,10 CONT,STR SLDR PIN	53387	3591-6002
A1J103	131-2921-00			CONN,RCPT,ELEC:HEADER,1 X 2,0.1 SPACING	00779	1-86479-3
A1J104	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 4)	22526	48283-036
A1J105	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 2)	22526	48283-036
A1J109	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 2)	22526	48283-036
A1J120	131-3152-00			CONN,RCPT,ELEC:HEADER,2 X 8 0.1 SPACING	22526	66506-043
A1J181	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 2)	22526	48283-036
A1J411	131-3362-00			CONN,RCPT,ELEC:HEADER,STR,26 PIN	53387	3593-6002
A1J511	131-3362-00			CONN,RCPT,ELEC:HEADER,STR,26 PIN	53387	3593-6002
A1J512	131-3364-00			CONN,RCPT,ELEC:HEADER,STRAIGHT,34 PIN	53387	3594-6002
A1L101	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L107	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L113	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L115	108-0317-00			COIL,RF:FIXED,15 UH	32159	71501M+10PERCENT
A1L120	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L180	108-0602-00			COIL,RF:FIXED,45NH	TK2042	ORDER BY DESCR
A1L200	108-0509-00			COIL,RF:FIXED,2.45UH	TK2042	ORDER BY DESCR
A1L219	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L220	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L280	108-0602-00			COIL,RF:FIXED,45NH	TK2042	ORDER BY DESCR
A1L307	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L325	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L336	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L403	108-0552-00			COIL,RF:FIXED,80NH	TK1345	108-0552-00
A1L521	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L605	108-0170-01			COIL,RF:FIXED,360NH	TK2042	ORDER BY DESCR
A1L606	108-0736-00			COIL,RF:FIXED,828NH	TK2042	ORDER BY DESCR
A1L607	108-0736-00			COIL,RF:FIXED,828NH	TK2042	ORDER BY DESCR
A1L608	108-0170-01			COIL,RF:FIXED,360NH	TK2042	ORDER BY DESCR
A1L609	108-0509-00			COIL,RF:FIXED,2.45UH	TK2042	ORDER BY DESCR
A1L610	108-0509-00			COIL,RF:FIXED,2.45UH	TK2042	ORDER BY DESCR
A1L619	108-0736-00			COIL,RF:FIXED,828NH	TK2042	ORDER BY DESCR
A1L628	108-0327-00			COIL,RF:FIXED,48NH	TK2042	ORDER BY DESCR
A1L633	108-0327-00			COIL,RF:FIXED,48NH	TK2042	ORDER BY DESCR
A1L644	114-0353-00			COIL,RF:VARIABLE,0.6-1.0UH	24226	ORDER BY DESCR
A1L733	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L738	108-0317-00			COIL,RF:FIXED,15 UH	32159	71501M+10PERCENT
A1L740	108-0317-00			COIL,RF:FIXED,15 UH	32159	71501M+10PERCENT
A1L743	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A1L938	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L973	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L980	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1LR101	108-0325-00			COIL,RF:FIXED,489NH	TK2042	ORDER BY DESCR
A1LR107	108-0325-00			COIL,RF:FIXED,489NH	TK2042	ORDER BY DESCR
A1LR201	108-0325-00			COIL,RF:FIXED,489NH	TK2042	ORDER BY DESCR
A1LR218	108-0330-00			COIL,RF:FIXED,403NH	TK2042	ORDER BY DESCR
A1LR219	108-0330-00			COIL,RF:FIXED,403NH	TK2042	ORDER BY DESCR
A1Q130	151-0622-00			TRANSISTOR:PNP,SI,40V,1A,TO-226AE/237	04713	SPS8956(MPSW51A)
A1Q131	151-0622-00			TRANSISTOR:PNP,SI,40V,1A,TO-226AE/237	04713	SPS8956(MPSW51A)
A1Q154	151-0188-00			TRANSISTOR:PNP,SI,TO-92	80009	151-0188-00
A1Q155	151-0188-00			TRANSISTOR:PNP,SI,TO-92	80009	151-0188-00
A1Q190	151-0190-00			TRANSISTOR:NPN,SI,TO-92	80009	151-0190-00
A1Q460	151-0198-01			TRANSISTOR:NPN,SI,TO-92 PLSTC (QUANTITY OF 2,LOCATIONS A & B)	80009	151-0198-01
A1Q550	151-0190-00			TRANSISTOR:NPN,SI,TO-92	80009	151-0190-00
A1Q600	151-0190-00			TRANSISTOR:NPN,SI,TO-92	80009	151-0190-00
A1Q623	151-0190-00			TRANSISTOR:NPN,SI,TO-92	80009	151-0190-00
A1Q624	151-1025-00			TRANSISTOR:FET,N-CHAN,SI,TO-92	04713	SPF3036
A1Q645	151-0188-00			TRANSISTOR:PNP,SI,TO-92	80009	151-0188-00
A1Q700	151-0190-00			TRANSISTOR:NPN,SI,TO-92	80009	151-0190-00
A1Q709	151-0736-00			TRANSISTOR:NPN,SI,TO-92	80009	151-0736-00
A1Q710	151-0736-00			TRANSISTOR:NPN,SI,TO-92	80009	151-0736-00
A1Q740	151-0223-00	B010514	B041039	TRANSISTOR:NPN,SI,625MW,TO-92	80009	151-0223-00
A1Q741	151-0190-00			TRANSISTOR:NPN,SI,TO-92	80009	151-0190-00
A1Q742	151-0190-00			TRANSISTOR:NPN,SI,TO-92	80009	151-0190-00
A1Q743	151-0188-00			TRANSISTOR:PNP,SI,TO-92	80009	151-0188-00
A1Q745	151-0188-00			TRANSISTOR:PNP,SI,TO-92	80009	151-0188-00
A1Q941	151-0188-00			TRANSISTOR:PNP,SI,TO-92	80009	151-0188-00
A1Q942	151-0188-00			TRANSISTOR:PNP,SI,TO-92	80009	151-0188-00
A1R100	315-0474-00			RES,FXD,FILM:470K OHM,5%,0.25W	19701	5043CX470K0J92U
A1R101	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A1R102	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A1R112	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R114	321-0130-03			RES,FXD,FILM:221 OHM,0.25%,0.125W,TC=T2 MI	91637	MFF1816D221R0C
A1R115	321-0146-00			RES,FXD,FILM:324 OHM,1%,0.125W,TC=T0	07716	CEAD324R0F
A1R117	321-0320-00			RES,FXD,FILM:21.0K OHM,1%,0.125W,TC=T0	19701	5033ED21K00F
A1R118	321-0212-00			RES,FXD,FILM:1.58K OHM,1%,0.125W,TC=T0	19701	5033ED1K58F
A1R121	313-1121-00			RES,FXD,FILM:120 OHM,5%,0.2W	80009	313-1121-00
A1R123	313-1622-00			RES,FXD,FILM:6.2K OHM,5%,0.2W	57668	TR20JE 06K2
A1R125	301-0361-00			RES,FXD,FILM:360 OHM,5%,0.5W	19701	5053CX360R0J
A1R129	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R130	313-1561-00			RES,FXD,FILM:560 OHM,5%,0.2W	57668	TR20JE 560E
A1R131	313-1561-00			RES,FXD,FILM:560 OHM,5%,0.2W	57668	TR20JE 560E
A1R133	322-3201-00			RES,FXD,FILM:1.21K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K21
A1R135	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R136	313-1622-00			RES,FXD,FILM:6.2K OHM,5%,0.2W	57668	TR20JE 06K2
A1R140	313-1471-00			RES,FXD,FILM:470 OHM,5%,0.2W	57668	TR20JE 470E
A1R141	313-1471-00			RES,FXD,FILM:470 OHM,5%,0.2W	57668	TR20JE 470E
A1R142	313-1391-00			RES,FXD,FILM:390 OHM,5%,0.2W	57668	TR20JE 390E
A1R143	313-1391-00			RES,FXD,FILM:390 OHM,5%,0.2W	57668	TR20JE 390E
A1R144	307-0108-00			RES,FXD,CMPSN:6.8 OHM,5%,0.25W	01121	CB68G5
A1R149	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R150	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R152	322-3265-00			RES,FXD,FILM:5.62K OHM,1%,0.2W,TC=T0	80009	322-3265-00
A1R153	322-3277-00			RES,FXD,FILM:7.5K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 7K50
A1R154	322-3210-00			RES,FXD,FILM:1.5K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K50
A1R155	321-0206-00			RES,FXD,FILM:1.37K OHM,1%,0.125W,TC=T0	07716	CEAD13700F

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A1R156	321-0255-00			RES,FXD,FILM:4.42K OHM,1%,0.125W,TC=T0	19701	5033ED4K420F
A1R159	322-3242-00			RES,FXD,FILM:3.24K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 3K24
A1R161	322-3293-00			RES,FXD,FILM:11K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 11K0
A1R162	322-3293-00			RES,FXD,FILM:11K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 11K0
A1R163	322-3242-00			RES,FXD,FILM:3.24K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 3K24
A1R165	313-1822-00			RES,FXD,FILM:8.2K,OHM,5%,0.2W	57668	TR20JE 08K2
A1R180	322-3242-00			RES,FXD,FILM:3.24K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 3K24
A1R181	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R182	322-3242-00			RES,FXD,FILM:3.24K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 3K24
A1R183	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R190	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R191	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R192	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R193	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R194	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R195	322-3143-00			RES,FXD,FILM:301 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 301E
A1R196	322-3277-00			RES,FXD,FILM:7.5K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 7K50
A1R197	322-3265-00			RES,FXD,FILM:5.62K OHM,1%,0.2W,TC=T0	80009	322-3265-00
A1R198	321-1700-04			RES,FXD,FILM:10.44K OHM,0.1%,0.125W,TC=T2	19701	5033RC10K440B
A1R199	321-1700-04			RES,FXD,FILM:10.44K OHM,0.1%,0.125W,TC=T2	19701	5033RC10K440B
A1R200	315-0474-00			RES,FXD,FILM:470K OHM,5%,0.25W	19701	5043CX470K0J92U
A1R201	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A1R202	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A1R216	313-1121-00			RES,FXD,FILM:120 OHM,5%,0.2W	80009	313-1121-00
A1R217	321-0320-00			RES,FXD,FILM:21.0K OHM,1%,0.125W,TC=T0	19701	5033ED21K00F
A1R218	321-0212-00			RES,FXD,FILM:1.58K OHM,1%,0.125W,TC=T0	19701	5033ED1K58F
A1R225	301-0361-00			RES,FXD,FILM:360 OHM,5%,0.5W	19701	5053CX360R0J
A1R230	322-3226-00			RES,FXD,FILM:2.21K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K21
A1R231	322-3226-00			RES,FXD,FILM:2.21K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K21
A1R232	322-3226-00			RES,FXD,FILM:2.21K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K21
A1R301	315-0180-00			RES,FXD,FILM:18 OHM,5%,0.25W	19701	5043CX18R00J
A1R302	315-0180-00			RES,FXD,FILM:18 OHM,5%,0.25W	19701	5043CX18R00J
A1R303	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R304	315-0101-00			RES,FXD,FILM:100 OHM,5%,0.25W	57668	NTR25J-E 100E
A1R311	315-0101-00			RES,FXD,FILM:100 OHM,5%,0.25W	57668	NTR25J-E 100E
A1R312	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R329	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R332	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R353	322-3239-00			RES,FXD,FILM:3.01K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 3K01
A1R361	322-3265-00			RES,FXD,FILM:5.62K OHM,1%,0.2W,TC=T0	80009	322-3265-00
A1R401	322-3202-00			RES,FXD,FILM:1.24K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K24
A1R402	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R403	311-0607-00			RES,VAR,NONWW:TRMR,10K OHM,0.5W	73138	82-25-2
A1R404	313-1200-00			RES,FXD,FILM:20 OHM,5%,0.2W	57668	TR20JE20E
A1R405	313-1200-00			RES,FXD,FILM:20 OHM,5%,0.2W	57668	TR20JE20E
A1R411	311-0635-00			RES,VAR,NONWW:TRMR,1K OHM,0.5W	32997	3329H-L58-102
A1R412	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R416	313-1432-00			RES,FXD,FILM:4.3K OHM,5%,0.2W	57668	TR20JE 04K3
A1R430	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R450	321-0310-00			RES,FXD,FILM:16.5K OHM,1%,0.125W,TC=T0	19701	5033ED16K50F
A1R451	321-0275-00			RES,FXD,FILM:7.15K OHM,1%,0.125W,TC=T0	07716	CEAD71500F
A1R452	321-0310-00			RES,FXD,FILM:16.5K OHM,1%,0.125W,TC=T0	19701	5033ED16K50F
A1R453	321-0275-00			RES,FXD,FILM:7.15K OHM,1%,0.125W,TC=T0	07716	CEAD71500F
A1R454	321-0310-00			RES,FXD,FILM:16.5K OHM,1%,0.125W,TC=T0	19701	5033ED16K50F
A1R455	321-0310-00			RES,FXD,FILM:16.5K OHM,1%,0.125W,TC=T0	19701	5033ED16K50F
A1R456	321-0333-00			RES,FXD,FILM:28.7K OHM,1%,0.125W,TC=T0	19701	5043ED28K70F
A1R457	321-0275-00			RES,FXD,FILM:7.15K OHM,1%,0.125W,TC=T0	07716	CEAD71500F
A1R458	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0

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A1R459	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R460	321-0062-00			RES,FXD,FILM:43.2 OHM,0.5%,0.125W,TC=T0	57668	CRB14 FXE 43.2
A1R461	322-3139-00			RES,FXD,FILM:274 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 274E
A1R462	322-3201-00			RES,FXD,FILM:1.21K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K21
A1R463	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R464	321-0063-00			RES,FXD,FILM:44.2 OHM,0.5%,0.125W,TC=T0	91637	CMF55116G44R20F
A1R465	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R468	321-0287-00			RES,FXD,FILM:9.53K OHM,1%,0.125W,TC=T0	19701	5033ED9K530F
A1R469	313-1200-00			RES,FXD,FILM:20 OHM,5%,0.2W	57668	TR20JE20E
A1R470	322-3322-00			RES,FXD,FILM:22.1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 22K1
A1R471	322-3322-00			RES,FXD,FILM:22.1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 22K1
A1R473	313-1471-00			RES,FXD,FILM:470 OHM,5%,0.2W	57668	TR20JE 470E
A1R476	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R477	322-3258-00			RES,FXD,FILM:4.75K OHM,1%,0.2W,TC=T0	56845	ORDER BY DESCR
A1R478	321-0193-03			RES,FXD,FILM:1K OHM,0.25%,0.125W,TC=T2	07716	CEAC10000C
A1R479	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R480	321-0375-00			RES,FXD,FILM:78.7K OHM,1%,0.125W,TC=T0	07716	CEAD78701F
A1R481	321-0347-00			RES,FXD,FILM:40.2K OHM,1%,0.125W,TC=T0	91637	CMF55116G40201F
A1R482	313-1471-00			RES,FXD,FILM:470 OHM,5%,0.2W	57668	TR20JE 470E
A1R483	321-0347-00			RES,FXD,FILM:40.2K OHM,1%,0.125W,TC=T0	91637	CMF55116G40201F
A1R484	322-3222-00			RES,FXD,FILM:2K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K00
A1R485	322-3222-00			RES,FXD,FILM:2K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K00
A1R486	321-0347-00			RES,FXD,FILM:40.2K OHM,1%,0.125W,TC=T0	91637	CMF55116G40201F
A1R487	321-0130-03			RES,FXD,FILM:221 OHM,0.25%,0.125W,TC=T2 MI	91637	MFF1816D221R0C
A1R488	321-1216-03			RES,FXD,FILM:1.76K OHM,0.25%,0.125W,TC=T2	24546	NC55C1761C
A1R489	321-1216-03			RES,FXD,FILM:1.76K OHM,0.25%,0.125W,TC=T2	24546	NC55C1761C
A1R490	321-0375-00			RES,FXD,FILM:78.7K OHM,1%,0.125W,TC=T0	07716	CEAD78701F
A1R491	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R492	321-0193-03			RES,FXD,FILM:1K OHM,0.25%,0.125W,TC=T2	07716	CEAC10000C
A1R493	322-3258-00			RES,FXD,FILM:4.75K OHM,1%,0.2W,TC=T0	56845	ORDER BY DESCR
A1R494	313-1201-00			RES,FXD,FILM:200 OHM,5%,0.2W	57668	TR20JE200E
A1R496	322-3293-00			RES,FXD,FILM:11K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 11K0
A1R497	313-1821-00			RES,FXD,FILM:820 OHM,5%,0.2W	57668	TR20JE 820E
A1R498	313-1821-00			RES,FXD,FILM:820 OHM,5%,0.2W	57668	TR20JE 820E
A1R501	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R502	313-1622-00			RES,FXD,FILM:6.2K OHM,5%,0.2W	57668	TR20JE 06K2
A1R503	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R504	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R511	321-0320-00			RES,FXD,FILM:21.0K OHM,1%,0.125W,TC=T0	19701	5033ED21K00F
A1R512	322-3293-00			RES,FXD,FILM:11K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 11K0
A1R513	313-1470-00			RES,FXD,FILM:47 OHM,5%,0.2W	57668	TR20JE 47E
A1R518	313-1680-00			RES,FXD,FILM:68 OHM,0.2W,5%	57668	TR20JT68 68E
A1R519	313-1621-00			RES,FXD,FILM:620 OHM,5%,0.2W	57668	TR20JE 620E
A1R520	313-1393-00			RES,FXD,FILM:39K OHM,5%,0.2W	57668	TR20JE 39K
A1R521	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R527	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R529	313-1561-00			RES,FXD,FILM:560 OHM,5%,0.2W	57668	TR20JE 560E
A1R537	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R538	313-1621-00			RES,FXD,FILM:620 OHM,5%,0.2W	57668	TR20JE 620E
A1R542	313-1680-00			RES,FXD,FILM:68 OHM,0.2W,5%	57668	TR20JT68 68E
A1R543	313-1621-00			RES,FXD,FILM:620 OHM,5%,0.2W	57668	TR20JE 620E
A1R544	313-1393-00			RES,FXD,FILM:39K OHM,5%,0.2W	57668	TR20JE 39K
A1R545	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R550	313-1471-00			RES,FXD,FILM:470 OHM,5%,0.2W	57668	TR20JE 470E
A1R551	321-1682-07			RES,FXD,FILM:5.7K OHM,0.1%,0.125W,TC=T9	19701	5033RE5K701B
A1R552	321-0641-07			RES,FXD,FILM:1.8K OHM,0.1,0.125W,TC=T9	07716	CEAE 18000B
A1R553	322-3210-00			RES,FXD,FILM:1.5K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K50
A1R554	322-3213-00			RES,FXD,FILM:1.62K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K62

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A1R555	321-0294-00			RES,FXD,FILM:11.3K OHM,1%,0.125W,TC=T0	19701	5043ED11K30F
A1R556	322-3282-00			RES,FXD,FILM:8.45K OHM,1%,0.2W,TC=T0	80009	322-3282-00
A1R557	321-0808-07			RES,FXD,FILM:300 OHM,0.1%,0.125W,TC=T9	24546	NE55E3000B
A1R558	321-0657-07			RES,FXD,FILM:60 OHM,0.1%,0.125W,TC=T9	57668	RB14BZE 60E
A1R560	313-1621-00			RES,FXD,FILM:620 OHM,5%,0.2W	57668	TR20JE 620E
A1R600	313-1270-00			RES,FXD,FILM:27 OHM 5%,0.2W	57668	TR20JT68 27E
A1R601	313-1750-00			RES,FXD,FILM:75 OHM,5%,0.2W	57668	TR20JE 75E
A1R602	313-1470-00			RES,FXD,FILM:47 OHM,5%,0.2W	57668	TR20JE 47E
A1R605	311-2227-00			RES,VAR,NONWW:TRMR,100 OHM,20%,0.5W LINEAR	TK1450	GF06UT 100
A1R606	313-1100-00			RES,FXD,FILM:10 OHM,5%,0.2W	57668	TR20JE10E0
A1R607	313-1100-00			RES,FXD,FILM:10 OHM,5%,0.2W	57668	TR20JE10E0
A1R614	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R615	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R617	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R618	311-2234-00			RES,VAR,NONWW:TRMR,5K OHM,20%,0.5W LINEAR	TK1450	GF06UT 5K
A1R619	315-0510-00			RES,FXD,FILM:51 OHM,5%,0.25W	19701	5043CX51R00J
A1R620	322-3258-00			RES,FXD,FILM:4.75K OHM,1%,0.2W,TC=T0	56845	ORDER BY DESCR
A1R622	322-3226-00			RES,FXD,FILM:2.21K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K21
A1R623	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R624	313-1100-00			RES,FXD,FILM:10 OHM,5%,0.2W	57668	TR20JE10E0
A1R637	322-3222-00			RES,FXD,FILM:2K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K00
A1R638	311-2234-00			RES,VAR,NONWW:TRMR,5K OHM,20%,0.5W LINEAR	TK1450	GF06UT 5K
A1R639	311-2230-00			RES,VAR,NONWW:TRMR,500 OHM,20%,0.50 LINEAR	TK1450	GF06UT 500
A1R642	313-1432-00			RES,FXD,FILM:4.3K OHM,5%,0.2W	57668	TR20JE 04K3
A1R643	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R644	322-3258-00			RES,FXD,FILM:4.75K OHM,1%,0.2W,TC=T0	56845	ORDER BY DESCR
A1R645	321-0265-00			RES,FXD,FILM:5.62K OHM,1%,0.125W,TC=T0	19701	5043ED5K620F
A1R646	321-0252-00			RES,FXD,FILM:4.12K OHM,1%,0.125W,TC=T0	07716	CEAD41200F
A1R650	322-3318-00			RES,FXD,FILM:20K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 20K0
A1R651	322-3189-00			RES,FXD,FILM:909 OHM,1%,0.2W,TC=T0	57668	CRB 20 FXE 909E
A1R652	315-0274-00			RES,FXD,FILM:270K OHM,5%,0.25W	57668	NTR25J-E270K
A1R653	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R655	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R659	321-0147-00			RES,FXD,FILM:332 OHM,1%,0.125W,TC=T0	07716	CEAD332R0F
A1R669	321-0995-00			RES,FXD,FILM:549K OHM,1%,0.125W,TC=T0	24546	NA55D5493F
A1R670	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R671	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R678	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R700	313-1221-00			RES,FXD,FILM:220 OHM,5%,0.2W	57668	TR20JE220E
A1R701	322-3223-00			RES,FXD,FILM:2.05K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K05
A1R702	321-0252-00			RES,FXD,FILM:4.12K OHM,1%,0.125W,TC=T0	07716	CEAD41200F
A1R707	322-3201-00			RES,FXD,FILM:1.21K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K21
A1R708	313-1242-00			RES,FXD,FILM:2.4K OHM,5%,0.2W	57668	TR20JE 02K4
A1R709	322-3258-00			RES,FXD,FILM:4.75K OHM,1%,0.2W,TC=T0	56845	ORDER BY DESCR
A1R710	315-0396-00			RES,FXD,FILM:39M OHM,5%,0.25W	01121	CB3965
A1R713	313-1822-00			RES,FXD,FILM:8.2K,OHM,5%,0.2W	57668	TR20JE 08K2
A1R723	321-0240-00			RES,FXD,FILM:3.09K OHM,1%,0.125W,TC=T0	07716	CEAD30900F
A1R724	321-0680-00			RES,FXD,FILM:35.3K OHM,0.5%,0.125W,TC=T2	19701	5033RC35K30D
A1R731	322-3306-00			RES,FXD,FILM:15K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 15K0
A1R732	322-3273-00			RES,FXD,FILM:6.81K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 6K81
A1R733	322-3218-00			RES,FXD,FILM:1.82K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K82
A1R734	313-1221-00			RES,FXD,FILM:220 OHM,5%,0.2W	57668	TR20JE220E
A1R735	313-1273-00			RES,FXD,FILM:27K OHM,5%,0.2W	57668	TR20JE 27K
A1R736	321-0217-00			RES,FXD,FILM:1.78K OHM,1%,0.125W,TC=T0	19701	5043D1K780F
A1R737	322-3263-00			RES,FXD,FILM:5.36K OHM,1%,0.2W,TC=T0	56845	ORDER BY DESCR
A1R738	322-3273-00			RES,FXD,FILM:6.81K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 6K81
A1R742	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A1R743	313-1331-00			RES,FXD,FILM:330 OHM,5%,0.2W	57668	TR20JE 330E

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A1R744	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R745	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A1R746	301-0470-00			RES,FXD,FILM:47 OHM,5%,0.5W	19701	5053CX47R00J
A1R747	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R748	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R749	313-1333-00			RES,FXD,FILM:33K OHM,5%,0.2W	57668	TR20JE 33K
A1R750	313-1151-00			RES,FXD,FILM:150 OHM,5%,0.2W	57668	TR20JE150E
A1R753	313-1242-00			RES,FXD,FILM:2.4K OHM,5%,0.2W	57668	TR20JE 02K4
A1R754	313-1333-00			RES,FXD,FILM:33K OHM,5%,0.2W	57668	TR20JE 33K
A1R755	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R757	313-1151-00			RES,FXD,FILM:150 OHM,5%,0.2W	57668	TR20JE150E
A1R800	321-0147-00			RES,FXD,FILM:332 OHM,1%,0.125W,TC=T0	07716	CEAD332R0F
A1R801	311-2230-00			RES,VAR,NONWW:TRMR,500 OHM,20%,0.50 LINEAR	TK1450	GF06UT 500
A1R802	313-1222-00			RES,FXD,FILM:2.2K OHM,5%,0.2W	57668	TR20JE 02K2
A1R803	313-1821-00			RES,FXD,FILM:820 OHM,5%,0.2W	57668	TR20JE 820E
A1R804	313-1151-00			RES,FXD,FILM:150 OHM,5%,0.2W	57668	TR20JE150E
A1R805	311-1242-00			RES,VAR,NONWW:TRMR,200K OHM,0.5W	32997	3386X-T07-204
A1R806	322-3414-00			RES,FXD,FILM:200K OHM,1%,0.2W,TC=T0	91637	CCF50G20002F
A1R809	313-1151-00			RES,FXD,FILM:150 OHM,5%,0.2W	57668	TR20JE150E
A1R811	301-0331-00			RES,FXD,FILM:330 OHM,5%,0.5W	19701	5053CX330R0J
A1R817	313-1221-00			RES,FXD,FILM:220 OHM,5%,0.2W	57668	TR20JE220E
A1R820	321-0337-00			RES,FXD,FILM:31.6K OHM,1%,0.125W,TC=T0	07716	CEAD31601F
A1R821	321-0330-00			RES,FXD,FILM:26.7K OHM,1%,0.125W,TC=T0	07716	CEAD26701F
A1R822	322-3139-00			RES,FXD,FILM:274 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 274E
A1R823	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R849	313-1333-00			RES,FXD,FILM:33K OHM,5%,0.2W	57668	TR20JE 33K
A1R850	311-2234-00			RES,VAR,NONWW:TRMR,5K OHM,20%,0.5W LINEAR	TK1450	GF06UT 5K
A1R852	313-1240-00			RES,FXD,FILM:24 OHM,5%,0.2W	57668	TR20JT6824E0
A1R853	313-1240-00			RES,FXD,FILM:24 OHM,5%,0.2W	57668	TR20JT6824E0
A1R855	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R856	322-3210-00			RES,FXD,FILM:1.5K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K50
A1R858	322-3239-00			RES,FXD,FILM:3.01K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 3K01
A1R860	311-2234-00			RES,VAR,NONWW:TRMR,5K OHM,20%,0.5W LINEAR	TK1450	GF06UT 5K
A1R900	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R901	322-3197-00			RES,FXD,FILM:1.1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K10
A1R903	322-3258-00			RES,FXD,FILM:4.75K OHM,1%,0.2W,TC=T0	56845	ORDER BY DESCR
A1R904	313-1124-00			RES,FXD,FILM:120K OHM,5%,0.2W	57668	TR20JE120K
A1R907	313-1471-00			RES,FXD,FILM:470 OHM,5%,0.2W	57668	TR20JE 470E
A1R910	315-0396-00			RES,FXD,FILM:39M OHM,5%,0.25W	01121	CB3965
A1R912	313-1822-00			RES,FXD,FILM:8.2K,OHM,5%,0.2W	57668	TR20JE 08K2
A1R924	322-3325-00			RES,FXD,FILM:23.7K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 23K7
A1R936	322-3225-00			RES,FXD,FILM:2.15K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K15
A1R937	322-3268-00			RES,FXD,FILM:6.04K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 6K04
A1R939	322-3243-00			RES,FXD,FILM:3.32K OHM,1%,0.2W,TC=T0	80009	322-3243-00
A1R941	313-1151-00			RES,FXD,FILM:150 OHM,5%,0.2W	57668	TR20JE150E
A1R942	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A1R943	313-1151-00			RES,FXD,FILM:150 OHM,5%,0.2W	57668	TR20JE150E
A1R944	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R945	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A1R946	313-1221-00	B010100	B010513	RES,FXD,FILM:220 OHM,5%,0.2W	57668	TR20JE220E
A1R946	322-3193-00	B010514	B051039	RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R946	313-1221-00	B051039		RES,FXD,FILM:220 OHM,5%,0.2W	57668	TR20JE220E
A1R947	322-3117-00			RES,FXD,FILM:162 OHM,1%,0.2W,TC=T0	57668	CRB 20 FXE 162E
A1R950	301-0470-00			RES,FXD,FILM:47 OHM,5%,0.5W	19701	5053CX47R00J
A1R951	308-0555-00			RES,FXD,WW:5 OHM,5%,3W	00213	1200S-5.0-5
A1R952	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R956	322-3239-00			RES,FXD,FILM:3.01K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 3K01
A1R957	321-0291-00			RES,FXD,FILM:10.5K OHM,1%,0.125W,TC=T0	19701	5033ED10K50F

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A1R972	313-1510-00			RES,FXD,FILM:51 OHM,5%,0.2W	80009	313-1510-00
A1R973	313-1513-00			RES,FXD,CMPSN:51K OHM,5%,0.2W	57668	TR20JE 51K
A1R975	322-3097-00	B010514	B051039	RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R981	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R982	321-0103-00			RES,FXD,FILM:115 OHM,1%,0.125W,TC=T0	01121	RNK1150F
A1R985	313-1332-00			RES,FXD,FILM:3.3K OHM,5%,0.2W	57668	TR20JE 03K3
A1R986	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R995	313-1512-00			RES,FXD,FILM:5.1K OHM,5%,0.2W	57668	TR20JE 5K1
A1S615	260-1421-00			SWITCH,PUSH:1 BTN,2 POLE,INSTRUMENT ID	59821	ORDER BY DESCR
A1TP800	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL	22526	48283-036
A1U100	153-2235-03 #195			MICROCKT,LINEAR:LOW NOISE VERT PREAMP (MATCHED WITH A1U200)	80009	153-2235-03
A1U110	156-1245-00			MICROCKT,LINEAR:7 XSTR,NPN,SI,HV/HIGH CUR	01295	ULN2003AN-P3
A1U120	156-1245-00			MICROCKT,LINEAR:7 XSTR,NPN,SI,HV/HIGH CUR	01295	ULN2003AN-P3
A1U130	156-1245-00			MICROCKT,LINEAR:7 XSTR,NPN,SI,HV/HIGH CUR	01295	ULN2003AN-P3
A1U140	156-0651-00			MICROCKT,DGTL:8-BIT PRL-OUT SER SHF RGTR	80009	156-0651-00
A1U150	156-0651-00			MICROCKT,DGTL:8-BIT PRL-OUT SER SHF RGTR	80009	156-0651-00
A1U160	156-1200-01			MICROCKT,LINEAR:BIFET,QUAD OPNL AMPL,SCRN	80009	156-1200-01
A1U165	156-2854-00			MICROCKT,LINEAR:OPNL AMPL QUAD	80009	156-2854-00
A1U170	156-0513-03			MICROCKT,LINEAR:CMOS,8 CHAN ANALOG MUX	04713	MC14051BCL
A1U180	156-1191-01			MICROCKT,LINEAR:BIFET,DUAL OPNL AMPL,SCRN	80009	156-1191-01
A1U200	153-2235-03			MICROCKT,LINEAR:LOW NOISE VERT PREAMP (MATCHED WITH A1U100)	80009	153-2235-03
A1U300	155-0238-00			MICROCKT,LINEAR:TRIGGER PREAMP	80009	155-0238-00
A1U350	156-1191-01			MICROCKT,LINEAR:BIFET,DUAL OPNL AMPL,SCRN	80009	156-1191-01
A1U400	155-0236-00			MICROCKT,LINEAR:VERTICAL CHANNEL SWITCH	80009	155-0236-00
A1U450	156-0158-07			MICROCKT,LINEAR:DUAL OPNL AMPL,SCREENED	01295	MC1458JG4
A1U475	156-0048-00			MICROCKT,LINEAR:5 XSTR ARRAY	02735	CA3046
A1U485	156-0048-00			MICROCKT,LINEAR:5 XSTR ARRAY	02735	CA3046
A1U500	155-0239-02			MICROCKT,LINEAR:TRIGGER	80009	155-0239-02
A1U550	156-0048-00			MICROCKT,LINEAR:5 XSTR ARRAY	02735	CA3046
A1U600	155-0237-00	B010100	B011576	MICROCKT,LINEAR:VERTICAL OUTPUT	80009	155-0237-00
A1U600	165-2393-00	B011577	B049999	MICROCKT,LINEAR:VERT OUTPUT TESTED,H2393	80009	165-2393-00
A1U600	155-0237-00	B050000	B051629	MICROCKT,LINEAR:VERTICAL OUTPUT	80009	155-0237-00
A1U600	165-2393-00	B051630		MICROCKT,LINEAR:VERT OUTPUT TESTED,H2393	80009	165-2393-00
A1U650	155-0244-01			MICROCKT,DGTL:SYSTEM LOGIC INTERFACE	31471	M 217
A1U700	155-0240-00			MICROCKT,LINEAR:SWEEP	80009	155-0240-00
A1U735	156-0048-00			MICROCKT,LINEAR:5 XSTR ARRAY	02735	CA3046
A1U800	155-0241-02			MICROCKT,DGTL:HORIZONTAL AMP SYS	80009	155-0241-02
A1U850	156-0515-00			MICROCKT,DGTL:CMOS,TRIPLE 2-CHAN MUX	02735	CD4053BF
A1U860	156-0515-00			MICROCKT,DGTL:CMOS,TRIPLE 2-CHAN MUX	02735	CD4053BF
A1U900	155-0240-00			MICROCKT,LINEAR:SWEEP	80009	155-0240-00
A1U910	156-1191-01			MICROCKT,LINEAR:BIFET,DUAL OPNL AMPL,SCRN	80009	156-1191-01
A1U950	155-0242-01			MICROCKT,LINEAR:Z AXIS AUTOFOCUS	80009	155-0242-01
A1U975	160-5062-00	B010100	B010513	MICROCKT,DGTL:STTL,DECA 20 INP AND/OR PLD	80009	160-5062-00
A1U975	160-5062-01	B010514	B051039	MICROCKT,DGTL:STTL,DECA 20 INP,PRGM	80009	160-5062-01
A1U975	160-5062-00	B051040		MICROCKT,DGTL:STTL,DECA 20 INP AND/OR PLD	80009	160-5062-00
A1U980	156-1611-01			MICROCKT,DGTL:ASTTL,DUAL D-TYPE FF	80009	156-1611-01
A1U985	156-0341-00			MICROCKT,DGTL:DUAL 2-INP OR DRIVER	07263	75453BTC
A1VR112	152-0166-00			SEMICOND DVC,DI:ZEN,SI,6.2V,5%,400MW,DO-7	04713	SZ11738RL
A1VR125	152-0166-00			SEMICOND DVC,DI:ZEN,SI,6.2V,5%,400MW,DO-7	04713	SZ11738RL
A1VR225	152-0166-00			SEMICOND DVC,DI:ZEN,SI,6.2V,5%,400MW,DO-7	04713	SZ11738RL
A1VR550	152-0195-00			SEMICOND DVC,DI:ZEN,SI,5.1V,5%,0.4W,DO-7	04713	SZ11755RL
A1W101	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A1W102	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A1W103	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A1W104	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A1W105	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A1W106	195-6500-02			LEAD,ELECTRICAL:22 AWG,1.75 L,9-N	TK1544	195-6500-02

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A1W107	195-6500-02			LEAD,ELECTRICAL:22 AWG,1.75 L,9-N	TK1544	195-6500-02
A1W108	195-6500-02			LEAD,ELECTRICAL:22 AWG,1.75 L,9-N	TK1544	195-6500-02
A1W109	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A1W120	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A1W121	175-4594-01			CA ASSY,SP,ELEC:6,22 AWG,5.25 L	80009	175-4594-01
A1W122	175-4598-00			CA ASSY,SP,ELEC:8,26 AWG,7.0 L,RIBBON	80009	175-4598-00
A1W141	174-0385-00			CA ASSY,SP,ELEC:5,22 AWG,2.6 L,9-N	80009	174-0385-00
A1W151	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A1W500	131-0566-00	B010100	B051039	BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A1W610	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A1W850	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A1W918	195-3991-01			LEAD,ELECTRICAL:22 AWG,3.5 L,0-N	80009	195-3991-01
A1W919	195-3991-01			LEAD,ELECTRICAL:22 AWG,3.5 L,0-N	80009	195-3991-01
A1XU100	136-0763-00			SKT,PL-IN ELEK:26 LINE CONT IMPD HYBRID	00779	ORDER BY DESCR
A1XU119	136-0728-00			SKT,PL-IN ELEK:MICROCKT,14 CONTACT	09922	DILB14P-108
A1XU191	136-0263-07			SOCKET,PIN TERM:U/W 0.025 SQ PIN (QUANTITY OF 16)	22526	ORDER BY DESCR
A1XU200	136-0763-00			SKT,PL-IN ELEK:26 LINE CONT IMPD HYBRID	00779	ORDER BY DESCR
A1XU300	136-0764-00			SKT,PL-IN ELEK:48 LINE CONT IMPD HYBRID	00779	ORDER BY DESCR
A1XU400	136-0763-00			SKT,PL-IN ELEK:26 LINE CONT IMPD HYBRID	00779	ORDER BY DESCR
A1XU500	136-0764-00			SKT,PL-IN ELEK:48 LINE CONT IMPD HYBRID	00779	ORDER BY DESCR
A1XU600	136-0764-00			SKT,PL-IN ELEK:48 LINE CONT IMPD HYBRID	00779	ORDER BY DESCR
A1XU650	136-0757-00			SKT,PL-IN ELEK:MICROCIRCUIT,40 DIP	09922	DILB40P-108
A1XU700	136-0764-00			SKT,PL-IN ELEK:48 LINE CONT IMPD HYBRID	00779	ORDER BY DESCR
A1XU900	136-0764-00			SKT,PL-IN ELEK:48 LINE CONT IMPD HYBRID	00779	ORDER BY DESCR
A1XU950	136-0764-00			SKT,PL-IN ELEK:48 LINE CONT IMPD HYBRID	00779	ORDER BY DESCR

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A2	672-1037-12			CIRCUIT BD ASSY:LV PWR SPLY MODULE	80009	672-1037-12

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A2A1	----- -----			CIRCUIT BD ASSY:REGULATOR (AVAILABLE AT THE 672-1037-XX LEVEL ONLY)		
A2A1C1016	285-1222-00			CAP,FXD,PLASTIC:0.068UF,20%,250V	55112	158/.068/M/250/H
A2A1C1018	285-1222-00			CAP,FXD,PLASTIC:0.068UF,20%,250V	55112	158/.068/M/250/H
A2A1C1208	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A2A1C1220	290-0939-00			CAP,FXD,ELCTLT:10UF,+100-10%,100V	56289	672D106H100CG2C
A2A1C1222	281-0783-00			CAP,FXD,CER DI:0.1 UF 20%,100V	04222	MA401C104MAA
A2A1C1226	281-0791-00			CAP,FXD,CER DI:270PF,10%,100V	04222	MA101C271KAA
A2A1C1240	290-0939-00			CAP,FXD,ELCTLT:10UF,+100-10%,100V	56289	672D106H100CG2C
A2A1C1245	281-0783-00			CAP,FXD,CER DI:0.1 UF 20%,100V	04222	MA401C104MAA
A2A1C1246	281-0791-00			CAP,FXD,CER DI:270PF,10%,100V	04222	MA101C271KAA
A2A1C1260	290-0942-00			CAP,FXD,ELCTLT:100UF,+100-10%,25V	55680	UPA1E101MAH
A2A1C1261	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A2A1C1270	281-0791-00			CAP,FXD,CER DI:270PF,10%,100V	04222	MA101C271KAA
A2A1C1272	281-0774-00			CAP,FXD,CER DI:0.022MFD,20%,100V	04222	MA201E223MAA
A2A1C1280	290-0942-00			CAP,FXD,ELCTLT:100UF,+100-10%,25V	55680	UPA1E101MAH
A2A1C1290	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A2A1C1291	290-0778-00			CAP,FXD,ELCTLT:1UF,20%,50V,NPLZD	54473	ECE-A50N1
A2A1C1292	290-0778-01			CAP,FXD,ELCTLT:1UF,+20%,50V	55680	UEB1H010MAAITD
A2A1C1300	290-0942-00			CAP,FXD,ELCTLT:100UF,+100-10%,25V	55680	UPA1E101MAH
A2A1C1330	290-0942-00			CAP,FXD,ELCTLT:100UF,+100-10%,25V	55680	UPA1E101MAH
A2A1C1331	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A2A1C1350	290-0942-00			CAP,FXD,ELCTLT:100UF,+100-10%,25V	55680	UPA1E101MAH
A2A1C1357	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A2A1C1374	281-0791-00			CAP,FXD,CER DI:270PF,10%,100V	04222	MA101C271KAA
A2A1C1400	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A2A1C1402	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A2A1C1974	290-0778-00			CAP,FXD,ELCTLT:1UF,20%,50V,NPLZD	54473	ECE-A50N1
A2A1CR1011	152-0750-00			SEMICOND DVC,DI:RECT,BRIDGE,SI,600V,3A	05828	RKBPC606-12
A2A1CR1220	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1221	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1241	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1242	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1243	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1244	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1260	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1261	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1262	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1CR1263	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1CR1264	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1CR1281	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1CR1282	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1CR1283	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1290	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1CR1294	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1CR1295	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1CR1300	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1CR1301	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1CR1302	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1CR1303	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1330	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1331	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1332	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1334	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1351	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1376	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1E1001	119-0181-00			ARSR,ELEC SURGE:230,GAS FILLED	25088	B1-A230
A2A1E1002	119-0181-00			ARSR,ELEC SURGE:230,GAS FILLED	25088	B1-A230

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A2A1F1330	159-0295-00			FUSE,CARTRIDGE:5 X 20MM,125V,1AMP	TK0946	TSC-1
A2A1J121	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 6)	22526	48283-036
A2A1J122	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 7)	22526	48283-036
A2A1J201	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 4)	22526	48283-036
A2A1J202	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 4)	22526	48283-036
A2A1J203	131-2925-00			CONN,RCPT,ELEC:CKT BD,1 X 6,0.2 SPACING	27264	10-10-1064
A2A1J204	131-1048-00			TERM,QIK DISC.:CKT BD MT,0.11 X 0.02 BL	00779	61134-1
A2A1J205	131-1048-00			TERM,QIK DISC.:CKT BD MT,0.11 X 0.02 BL	00779	61134-1
A2A1J206	131-1048-00			TERM,QIK DISC.:CKT BD MT,0.11 X 0.02 BL	00779	61134-1
A2A1J207	131-1048-00			TERM,QIK DISC.:CKT BD MT,0.11 X 0.02 BL	00779	61134-1
A2A1J208	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL	22526	48283-036
A2A1L1011	108-0473-00			COIL,RF:FIXED,174UH	TK2042	ORDER BY DESCR
A2A1L1012	108-0473-00			COIL,RF:FIXED,174UH	TK2042	ORDER BY DESCR
A2A1L1402	108-0443-00			COIL,RF:FIXED,23.5UH	80009	108-0443-00
A2A1P208	131-3957-00			BUS,CONDUCTOR:SHUNT ASSEMBLY,BLACK	80009	131-3957-00
A2A1Q1220	151-0497-00			TRANSISTOR:NPN,SI,TO-220	80009	151-0497-00
A2A1Q1221	151-0347-00			TRANSISTOR:NPN,SI,TO-92	04713	SPS7951
A2A1Q1222	151-0347-00			TRANSISTOR:NPN,SI,TO-92	04713	SPS7951
A2A1Q1223	151-0347-00			TRANSISTOR:NPN,SI,TO-92	04713	SPS7951
A2A1Q1240	151-0464-00			TRANSISTOR:NPN,SI,TO-220	80009	151-0464-00
A2A1Q1241	151-0347-00			TRANSISTOR:NPN,SI,TO-92	04713	SPS7951
A2A1Q1243	151-0347-00			TRANSISTOR:NPN,SI,TO-92	04713	SPS7951
A2A1Q1245	151-0347-00			TRANSISTOR:NPN,SI,TO-92	04713	SPS7951
A2A1Q1280	151-0476-00			TRANSISTOR:NPN,SI,TO-220	80009	151-0476-00
A2A1Q1281	151-0347-00			TRANSISTOR:NPN,SI,TO-92	04713	SPS7951
A2A1Q1290	151-1059-00			TRANSISTOR:FET,N-CHAN,30MW,TO-92 CASE	04713	ORDER BY DESCR
A2A1Q1300	151-0482-00			TRANSISTOR:PNP,SI,TO-220	04713	SJE1977
A2A1Q1301	151-0342-00			TRANSISTOR:PNP,SI,TO-92	07263	S035928
A2A1Q1351	151-0429-00			TRANSISTOR:DARLINGTON,PNP,SI,TO-126	80009	151-0429-00
A2A1Q1354	151-0342-00			TRANSISTOR:PNP,SI,TO-92	07263	S035928
A2A1Q1370	151-0341-00			TRANSISTOR:NPN,SI,TO-106	04713	SPS6919
A2A1Q1376	151-0341-00			TRANSISTOR:NPN,SI,TO-106	04713	SPS6919
A2A1R1010	301-0150-00			RES,FXD,FILM:15 OHM,5%,0.5W	19701	5053CX15R00J
A2A1R1011	315-0560-00			RES,FXD,FILM:56 OHM,5%,0.25W	57668	NTR25J-E56E0
A2A1R1012	315-0560-00			RES,FXD,FILM:56 OHM,5%,0.25W	57668	NTR25J-E56E0
A2A1R1013	315-0683-00			RES,FXD,FILM:68K OHM,5%,0.25W	57668	NTR25J-E68K0
A2A1R1014	313-1363-00			RES,FXD,FILM:36K OHM,5%,0.2W	57668	TR20JE 36K
A2A1R1015	313-1363-00			RES,FXD,FILM:36K OHM,5%,0.2W	57668	TR20JE 36K
A2A1R1016	301-0680-00			RES,FXD,FILM:68 OHM,5%,0.5W	19701	5053CX68R00J
A2A1R1017	315-0474-00			RES,FXD,FILM:470K OHM,5%,0.25W	19701	5043CX470K0J92U
A2A1R1018	301-0300-00			RES,FXD,FILM:30 OHM,5%,0.5W	19701	5053CX30R00J
A2A1R1019	301-0150-00			RES,FXD,FILM:15 OHM,5%,0.5W	19701	5053CX15R00J
A2A1R1204	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A2A1R1208	313-1201-00			RES,FXD,FILM:200 OHM,5%,0.2W	57668	TR20JE200E
A2A1R1212	313-1393-00			RES,FXD,FILM:39K OHM,5%,0.2W	57668	TR20JE 39K
A2A1R1220	304-0822-00			RES,FXD,CMPSN:8.2K OHM,10%,1W	01121	GB8221
A2A1R1221	315-0100-02			RES,FXD,CMPSN:10 OHM,5%,0.25W	01121	CB1005
A2A1R1222	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A2A1R1223	313-1823-00			RES,FXD,FILM:82K OHM,5%,0.2W	57668	TR20JE 82K
A2A1R1226	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A2A1R1227	321-0634-00			RES,FXD,FILM:84.65K OHM,0.25%,0.125W,TC=T2	19701	5033RC84K65C
A2A1R1228	321-0293-03			RES,FXD,FILM:11.0K OHM,0.25%,0.125W,TC=T2	24546	NC55C1102C
A2A1R1229	313-1683-00			RES,FXD,FILM:68K OHM,5%,0.2W	57668	TR20JE 68K
A2A1R1240	303-0202-00			RES,FXD,CMPSN:2K OHM,5%,1W	01121	GB 2025

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A2A1R1241	307-0105-00			RES,FXD,CMPSN:3.9 OHM,5%,0.25W	01121	CB 39G5
A2A1R1242	313-1152-00			RES,FXD,FILM:1.5K OHM,5%,0.2W	57668	TR20JE01K5
A2A1R1243	313-1393-00			RES,FXD,FILM:39K OHM,5%,0.2W	57668	TR20JE 39K
A2A1R1244	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A2A1R1246	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A2A1R1247	321-0368-00			RES,FXD,FILM:66.5K OHM,1%,0.125W,TC=T0	07716	CEAD66501F
A2A1R1248	321-0319-00			RES,FXD,FILM:20.5K OHM,1%,0.125W,TC=T0	19701	5033ED20K50F
A2A1R1249	313-1473-00			RES,FXD,FILM:47K OHM,5%,0.2W	57668	TR20JE 47K
A2A1R1261	321-0289-00			RES,FXD,FILM:10.0K OHM,1%,0.125W,TC=T0	19701	5033ED10K0F
A2A1R1262	321-0318-00			RES,FXD,FILM:20.0K OHM,1%,0.125W,TC=T0	19701	5033ED20K00F
A2A1R1264	313-1473-00			RES,FXD,FILM:47K OHM,5%,0.2W	57668	TR20JE 47K
A2A1R1270	313-1432-00			RES,FXD,FILM:4.3K OHM,5%,0.2W	57668	TR20JE 04K3
A2A1R1273	313-1473-00			RES,FXD,FILM:47K OHM,5%,0.2W	57668	TR20JE 47K
A2A1R1274	313-1683-00			RES,FXD,FILM:68K OHM,5%,0.2W	57668	TR20JE 68K
A2A1R1280	303-0470-00			RES,FXD,CMPSN:47 OHM,5%,1W	01121	GB4705
A2A1R1281	308-0839-00			RES,FXD,WW:0.1 OHM,5%,1.0W	75042	BW-20-R1000J
A2A1R1282	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A2A1R1283	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A2A1R1284	321-0318-00			RES,FXD,FILM:20.0K OHM,1%,0.125W,TC=T0	19701	5033ED20K00F
A2A1R1285	321-0318-00			RES,FXD,FILM:20.0K OHM,1%,0.125W,TC=T0	19701	5033ED20K00F
A2A1R1286	313-1243-00			RES,FXD,FILM:24K OHM,5%,0.2W	80009	313-1243-00
A2A1R1287	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A2A1R1291	321-0334-00			RES,FXD,FILM:29.4K OHM,1%,0.125W,TC=T0	07716	CEAD29401F
A2A1R1292	311-2258-00			RES,VAR,NONWW:TRMR,1K OHM,20%,0.5W	TK1450	GF06VT 1 K OHM
A2A1R1293	321-0639-00			RES,FXD,FILM:9.6K OHM,1%,0.125W,TC=T0	19701	5043ED9K600F
A2A1R1294	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A2A1R1295	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A2A1R1296	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A2A1R1297	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A2A1R1298	322-3275-00			RES,FXD,FILM:7.15K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 7K15
A2A1R1299	313-1224-00			RES,FXD,FILM:220K,5%,0.2W	57668	TR20JE 220K
A2A1R1300	303-0470-00			RES,FXD,CMPSN:47 OHM,5%,1W	01121	GB4705
A2A1R1301	308-0839-00			RES,FXD,WW:0.1 OHM,5%,1.0W	75042	BW-20-R1000J
A2A1R1302	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A2A1R1304	313-1243-00			RES,FXD,FILM:24K OHM,5%,0.2W	80009	313-1243-00
A2A1R1305	321-0289-06			RES,FXD,FILM:10.0K OHM,0.25%,0.125W,TC=T9	19701	5033RE10K00C
A2A1R1306	321-0318-03			RES,FXD,FILM:20.0K OHM,0.125%,0.125W,TC=T2	19701	5033RC20K00C
A2A1R1307	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A2A1R1309	313-1222-00			RES,FXD,FILM:2.2K OHM,5%,0.2W	57668	TR20JE 02K2
A2A1R1331	321-0685-00			RES,FXD,FILM:30K OHM,0.5%,0.125W,TC=T2	19701	5033RC30K00D
A2A1R1332	321-0318-03			RES,FXD,FILM:20.0K OHM,0.125%,0.125W,TC=T2	19701	5033RC20K00C
A2A1R1333	313-1751-00			RES,FXD,FILM:750 OHM,5%,0.2W	57668	TR20JE 750E
A2A1R1334	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A2A1R1351	313-1202-00			RES,FXD,FILM:2K OHM,5%,0.2W	57668	TR20JE02K0
A2A1R1352	301-0150-00			RES,FXD,FILM:15 OHM,5%,0.5W	19701	5053CX15R00J
A2A1R1353	301-0150-00			RES,FXD,FILM:15 OHM,5%,0.5W	19701	5053CX15R00J
A2A1R1354	313-1222-00			RES,FXD,FILM:2.2K OHM,5%,0.2W	57668	TR20JE 02K2
A2A1R1355	313-1682-00			RES,FXD,FILM:6.8K OHM,5%,0.2W	57668	TR20JE 06K8
A2A1R1356	313-1512-00			RES,FXD,FILM:5.1K OHM,5%,0.2W	57668	TR20JE 5K1
A2A1R1357	321-0318-03			RES,FXD,FILM:20.0K OHM,0.125%,0.125W,TC=T2	19701	5033RC20K00C
A2A1R1358	321-0689-00			RES,FXD,FILM:24.9K OHM,0.5%,0.125W,TC=T0	19701	5033RD24K90D
A2A1R1359	313-1682-00			RES,FXD,FILM:6.8K OHM,5%,0.2W	57668	TR20JE 06K8
A2A1R1370	321-0363-00			RES,FXD,FILM:59.0K OHM,1%,0.125W,TC=T0	07716	CEAD59001F
A2A1R1372	321-0299-00			RES,FXD,FILM:12.7K OHM,1%,0.125W,TC=T0	19701	5033ED12K70F
A2A1R1374	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A2A1R1376	321-0318-03			RES,FXD,FILM:20.0K OHM,0.125%,0.125W,TC=T2	19701	5033RC20K00C
A2A1R1378	313-1202-00			RES,FXD,FILM:2K OHM,5%,0.2W	57668	TR20JE02K0
A2A1R1400	315-0101-03			RES,FXD,CMPSN:100 OHM,5%,0.25W	01121	CB1015

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A2A1R1402	315-0101-03			RES,FXD,CMPSN:100 OHM,5%,0.25W	01121	CB1015
A2A1RT1010	307-0350-00			RES,THERMAL:7.5 OHM,10%,3.9%/DEG C	80009	307-0350-00
A2A1RT1016	307-0746-00			RES,THERMAL:5 OHM,10%,7A/DEG C	15454	SG200-S
A2A1S350	260-1849-00			SWITCH,PUSH:DPDT,4A,250VAC	31918	NE15/F2U103EE
A2A1T1229	120-1401-00			XFMR,TRIGGER:LINE,1:1 TURNS RATIO	54937	DMI 500-2044
A2A1U1260	156-1161-00			MICROCKT,LINEAR:VOLTAGE REGULATOR,POS,ADJ	12969	UC317T
A2A1U1270	156-0495-00			MICROCKT,LINEAR:OPNL AMPL	01295	LM324N
A2A1U1281	156-0158-07			MICROCKT,LINEAR:DUAL OPNL AMPL,SCREENED	01295	MC1458JG4
A2A1U1290	156-1173-00			MICROCKT,LINEAR:VOLTAGE REFERENCE	04713	MC1403UDS
A2A1U1300	156-0495-00			MICROCKT,LINEAR:OPNL AMPL	01295	LM324N
A2A1U1330	156-0872-00			MICROCKT,LINEAR:VOLTAGE REGULATOR	04713	MC7912CT
A2A1U1371	156-0495-00			MICROCKT,LINEAR:OPNL AMPL	01295	LM324N
A2A1VR1293	152-0055-00			SEMICOND DVC,DI:ZEN,SI,11V,5%,0.4W,DO-7	14433	Z5407
A2A1W251	175-4585-00			CA ASSY,SP,ELEC:20,28 AWG,13.0 L	80009	175-4585-00

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A3	----- -----			CIRCUIT BD ASSY:INVERTER (AVAILABLE AT THE 672-1037-XX LEVEL ONLY)		
A3C1020	285-1192-00			CAP,FXD,PPR DI:0.0022 UF,20%,250VAC	TK0515	PME271Y510
A3C1021	290-0971-00			CAP,FXD,ELCTLT:290UF +50-10%,200V	56289	39DX1314
A3C1022	290-0971-00			CAP,FXD,ELCTLT:290UF +50-10%,200V	56289	39DX1314
A3C1023	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A3C1025	290-0942-00			CAP,FXD,ELCTLT:100UF,+100-10%,25V	55680	UPA1E101MAH
A3C1029	281-0850-00			CAP,FXD,CER DI:820PF,5%,50VDC	04222	SA101A821JAA
A3C1032	281-0812-00			CAP,FXD,CER DI:1000PF,10%,100V	04222	MA101C102KAA
A3C1033	281-0772-00			CAP,FXD,CER DI:4700PF,10%,100V	04222	MA201C472KAA
A3C1034	290-0524-00			CAP,FXD,ELCTLT:4.7UF,20%,10V	05397	T368A475M010AZ
A3C1035	281-0772-00			CAP,FXD,CER DI:4700PF,10%,100V	04222	MA201C472KAA
A3C1040	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A3C1042	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A3C1048	281-0826-00			CAP,FXD,CER DI:2200PF,10%,100V	20932	401EM100AD222K
A3C1050	285-1254-00			CAP,FXD,PLASTIC:0.22UF,10%,400WVDC	56289	730P0167
A3C1051	285-1192-00			CAP,FXD,PPR DI:0.0022 UF,20%,250VAC	TK0515	PME271Y510
A3C1052	285-1196-00			CAP,FXD,PPR DI:0.01UF,20%,250V	TK0515	PME 265 MB 510
A3C1062	281-0850-00			CAP,FXD,CER DI:820PF,5%,50VDC	04222	SA101A821JAA
A3C1065	285-1190-00			CAP,FXD,MTLZD:0.056 UF,5%,250 V	05292	PMT3R ADVISE
A3C1066	290-0782-01			CAP,FXD,ELCTLT:4.7UF,20%,35VDC	55680	UVX1V4R7MAA1TD
A3C1067	281-0850-00			CAP,FXD,CER DI:820PF,5%,50VDC	04222	SA101A821JAA
A3C1071	281-0772-00			CAP,FXD,CER DI:4700PF,10%,100V	04222	MA201C472KAA
A3C1072	290-0806-00			CAP,FXD,ELCTLT:3.3UF,+75-10%,350VDC	55680	UHU2V3R3TPA
A3C1075	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A3C1101	290-0942-00			CAP,FXD,ELCTLT:100UF,+100-10%,25V	55680	UPA1E101MAH
A3C1102	290-0942-00			CAP,FXD,ELCTLT:100UF,+100-10%,25V	55680	UPA1E101MAH
A3C1110	290-0800-00			CAP,FXD,ELCTLT:250UF,+100-10%,20V	56289	672D257H020DM5C
A3C1111	290-0800-00			CAP,FXD,ELCTLT:250UF,+100-10%,20V	56289	672D257H020DM5C
A3C1112	290-0782-01			CAP,FXD,ELCTLT:4.7UF,20%,35VDC	55680	UVX1V4R7MAA1TD
A3C1113	290-0798-00			CAP,FXD,ELCTLT:180UF,+100-10%,40V	56289	672D187H040DM5C
A3C1114	290-0800-00			CAP,FXD,ELCTLT:250UF,+100-10%,20V	56289	672D257H020DM5C
A3C1115	290-0800-00			CAP,FXD,ELCTLT:250UF,+100-10%,20V	56289	672D257H020DM5C
A3C1116	290-0798-00			CAP,FXD,ELCTLT:180UF,+100-10%,40V	56289	672D187H040DM5C
A3C1120	290-0939-00			CAP,FXD,ELCTLT:10UF,+100-10%,100V	56289	672D106H100CG2C
A3C1130	290-0939-00			CAP,FXD,ELCTLT:10UF,+100-10%,100V	56289	672D106H100CG2C
A3C1132	290-0880-00			CAP,FXD,ELCTLT:10UF,+50-10%,160V	54473	ECE-A160V10U
A3CR1022	152-0333-00			SEMICOND DVC,DI:SW,SI,55V,200MA,DO-35	07263	FDH-6012
A3CR1023	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A3CR1028	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A3CR1030	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A3CR1034	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A3CR1035	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A3CR1040	152-0075-00			SEMICOND DVC,DI:SW,GE,22V,80MW,DO-7	80009	152-0075-00
A3CR1050	152-0661-01			SEMICOND DVC,DI:RECT,SI,600V,3A	04713	S.R.3523-1RL
A3CR1060	152-0040-00			SEMICOND DVC,DI:RECT,SI,600V,1A,DO-41	80009	152-0040-00
A3CR1062	152-0333-00			SEMICOND DVC,DI:SW,SI,55V,200MA,DO-35	07263	FDH-6012
A3CR1063	152-0333-00			SEMICOND DVC,DI:SW,SI,55V,200MA,DO-35	07263	FDH-6012
A3CR1064	152-0333-00			SEMICOND DVC,DI:SW,SI,55V,200MA,DO-35	07263	FDH-6012
A3CR1065	152-0333-00			SEMICOND DVC,DI:SW,SI,55V,200MA,DO-35	07263	FDH-6012
A3CR1070	152-0040-00			SEMICOND DVC,DI:RECT,SI,600V,1A,DO-41	80009	152-0040-00
A3CR1072	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A3CR1101	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3CR1102	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3CR1103	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3CR1104	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3CR1105	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3CR1106	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL

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A3CR1110	152-0794-00			SEMICOND DVC,DI:RECT,SI,10A,30V,TO-220	81483	95-4269
A3CR1113	152-0946-00			SEMICOND DVC,DI:RECT,SI,40V,3.0A	80009	152-0946-00
A3CR1114	152-0946-00			SEMICOND DVC,DI:RECT,SI,40V,3.0A	80009	152-0946-00
A3CR1115	152-0946-00			SEMICOND DVC,DI:RECT,SI,40V,3.0A	80009	152-0946-00
A3CR1116	152-0946-00			SEMICOND DVC,DI:RECT,SI,40V,3.0A	80009	152-0946-00
A3CR1121	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3CR1122	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3CR1123	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3CR1124	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3CR1131	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3CR1132	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3F1101	159-0255-00			FUSE,CARTRIDGE:FAST BLOW,4A,125V	80009	159-0255-00
A3F1102	159-0059-00			FUSE,WIRE LEAD:5A,125V	71400	A5
A3J301	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 3)	22526	48283-036
A3J302	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 3)	22526	48283-036
A3J303	131-2926-00			CONN,RCPT,ELEC:CKT BD,1 X 2,0.2 SPACING	27264	10-10-1024
A3L1110	108-0554-00			COIL,RF:FIXED,5UH,+/-20%	TK1345	108-0554-00
A3L1113	108-1144-00			COIL,RF:FIXED,27 UH,20%	34479	RL1284
A3L1114	108-1144-00			COIL,RF:FIXED,27 UH,20%	34479	RL1284
A3L1115	108-1144-00			COIL,RF:FIXED,27 UH,20%	34479	RL1284
A3L1116	108-1144-00			COIL,RF:FIXED,27 UH,20%	34479	RL1284
A3Q1021	151-0301-00			TRANSISTOR:PNP,SI,TO-18	80009	151-0301-00
A3Q1022	151-0192-00			TRANSISTOR:NPN,SI,TO-92	04713	SPS8801
A3Q1029	151-0254-00			TRANSISTOR:DARLINGTON,NPN,SI,625MW,TO-92	03508	X38L3118
A3Q1030	151-0301-00			TRANSISTOR:PNP,SI,TO-18	80009	151-0301-00
A3Q1040	151-0302-00			TRANSISTOR:NPN,SI,TO-18	04713	ST899
A3Q1050	151-1152-00			TRANSISTOR:MOSFE,N-CHANNEL,SI,TO-220	04713	IRF820
A3Q1060	151-1152-00			TRANSISTOR:MOSFE,N-CHANNEL,SI,TO-220	04713	IRF820
A3Q1062	151-0302-00			TRANSISTOR:NPN,SI,TO-18	04713	ST899
A3Q1070	151-1152-00			TRANSISTOR:MOSFE,N-CHANNEL,SI,TO-220	04713	IRF820
A3Q1110	151-0188-00			TRANSISTOR:PNP,SI,TO-92	80009	151-0188-00
A3R1018	313-1394-00			RES,FXD,FILM:390K,5%,0.2W	57668	TR20JE 390K
A3R1019	313-1394-00			RES,FXD,FILM:390K,5%,0.2W	57668	TR20JE 390K
A3R1020	301-0274-00			RES,FXD,FILM:270K OHM,5%,0.5W	19701	5053CX270K0J
A3R1022	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A3R1023	313-1122-00			RES,FXD,FILM:1.2K OHM,5%,0.2W	57668	TR20JE01K2
A3R1024	313-1473-00			RES,FXD,FILM:47K OHM,5%,0.2W	57668	TR20JE 47K
A3R1025	313-1302-00			RES,FXD,FILM:3K OHM,5%,0.2W	57668	TR20JE 03K0
A3R1027	321-0431-00			RES,FXD,FILM:301K OHM,1%,0.125W,TC=T0	07716	CEAD30102F
A3R1028	321-0481-04			RES,FXD,FILM:1M OHM,0.1%,0.125W,TC=T2	91637	CMF55116D10003B
A3R1029	313-1152-00			RES,FXD,FILM:1.5K OHM,5%,0.2W	57668	TR20JE01K5
A3R1030	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A3R1031	313-1334-00			RES,FXD,FILM:330K OHM,5%,0.2W	80009	313-1334-00
A3R1032	321-0335-00			RES,FXD,FILM:30.1K OHM,1%,0.125W,TC=T0	57668	RB14FXE30K1
A3R1033	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A3R1034	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A3R1035	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A3R1036	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A3R1037	313-1272-00			RES,FXD,FILM:2.7K OHM,5%,0.2W	57668	TR20JE 02K7
A3R1040	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A3R1041	313-1471-00			RES,FXD,FILM:470 OHM,5%,0.2W	57668	TR20JE 470E
A3R1042	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A3R1044	321-0334-00			RES,FXD,FILM:29.4K OHM,1%,0.125W,TC=T0	07716	CEAD29401F
A3R1045	321-0289-00			RES,FXD,FILM:10.0K OHM,1%,0.125W,TC=T0	19701	5033ED10K0F
A3R1046	321-0422-00			RES,FXD,FILM:243K OHM,1%,0.125W,TC=T0	07716	CEAD24302F
A3R1050	308-0843-00			RES,FXD,WW:0.2 OHM,5%,1/0W	91637	RS1A-90-R2J

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A3R1052	313-1470-00			RES,FXD,FILM:47 OHM,5%,0.2W	57668	TR20JE 47E
A3R1060	313-1470-00			RES,FXD,FILM:47 OHM,5%,0.2W	57668	TR20JE 47E
A3R1061	313-1202-00			RES,FXD,FILM:2K OHM,5%,0.2W	57668	TR20JE02K0
A3R1062	313-1682-00			RES,FXD,FILM:6.8K OHM,5%,0.2W	57668	TR20JE 06K8
A3R1063	313-1202-00			RES,FXD,FILM:2K OHM,5%,0.2W	57668	TR20JE02K0
A3R1064	313-1202-00			RES,FXD,FILM:2K OHM,5%,0.2W	57668	TR20JE02K0
A3R1065	315-0154-00			RES,FXD,FILM:150K OHM,5%,0.25W	57668	NTR25J-E150K
A3R1066	313-1202-00			RES,FXD,FILM:2K OHM,5%,0.2W	57668	TR20JE02K0
A3R1067	313-1682-00			RES,FXD,FILM:6.8K OHM,5%,0.2W	57668	TR20JE 06K8
A3R1068	313-1202-00			RES,FXD,FILM:2K OHM,5%,0.2W	57668	TR20JE02K0
A3R1069	303-0363-00			RES,FXD,CMPSN:36K OHM,5%,1W	01121	GB3635
A3R1070	313-1470-00			RES,FXD,FILM:47 OHM,5%,0.2W	57668	TR20JE 47E
A3R1071	315-0431-00			RES,FXD,FILM:430 OHM,5%,0.25W	19701	5043CX430R0J
A3R1072	321-0318-03			RES,FXD,FILM:20.0K OHM,0.125%,0.125W,TC=T2	19701	5033RC20K00C
A3R1075	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A3R1110	321-0219-00			RES,FXD,FILM:1.87K OHM,1%,0.125W,TC=T0	07716	CEAD18700F
A3R1111	315-0510-00			RES,FXD,FILM:51 OHM,5%,0.25W	19701	5043CX51R00J
A3R1112	321-0271-00			RES,FXD,FILM:6.49K OHM,1%,0.125W,TC=T0	07716	CEAD64900F
A3R1113	321-0271-00			RES,FXD,FILM:6.49K OHM,1%,0.125W,TC=T0	07716	CEAD64900F
A3R1114	321-0297-00			RES,FXD,FILM:12.1K OHM,1%,0.125W,TC=T0	07716	CEAD12101F
A3R1115	301-0301-00			RES,FXD,FILM:300 OHM,5%,0.5W	19701	5053CX300R0J
A3R1129	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A3R1130	313-1273-00			RES,FXD,FILM:27K OHM,5%,0.2W	57668	TR20JE 27K
A3RL1060	108-0329-00			COIL,RF:FIXED,2.4UH	TK2042	ORDER BY DESCR
A3RT1110	307-0124-00			RES,THERMAL:5K OHM,10%,NTC	15454	1DC502K-220-EC
A3S1020	260-0907-01			SWITCH,THRMSTC:NC,OPEN 97.8,CL 75.6,10A	93410	430-1537
A3T1020	120-1244-00			TRANSFORMER,RF:COMMON MODE,13MH,0.5A	20462	4096
A3T1060	120-1437-00			XFMR,PWR,STPDN:	02113	C1310
A3U1029	156-0885-05			CPLR,OPTOELECTR:LED,5KV,ISOLATION	09019	H11AX1139R
A3U1030	156-1627-00			MICROCKT,LINEAR:BIPOLAR,PWM PWR SPLY CONT	12969	UC494ACN
A3U1040	156-0885-05			CPLR,OPTOELECTR:LED,5KV,ISOLATION	09019	H11AX1139R
A3U1064	156-0366-00			MICROCKT,DGTL:DUAL D FLIP-FLOP	02735	CD4013BF
A3U1066	156-0328-00			MICROCKT,DGTL:DUAL MOS CLOCK DRIVER	04713	MMH0026CP1D
A3U1110	156-1161-00			MICROCKT,LINEAR:VOLTAGE REGULATOR,POS,ADJ	12969	UC317T
A3VR1020	152-0166-00			SEMICOND DVC,DI:ZEN,SI,6.2V,5%,400MW,DO-7	04713	SZ11738RL
A3VR1062	152-0168-00			SEMICOND DVC,DI:ZEN,SI,12V,5%,0.4W,DO-763B	14552	TD331689
A3W1021	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A3W1022	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A3W1050	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A3W1060	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A3W1102	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A3Y1050	120-1417-00			TRANSFORMER,RF:POWER HIGH FREQUENCY	54937	500-2311

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A4	670-9493-02	B010100	B049999	CIRCUIT BD ASSY:READOUT	80009	670-9493-02
A4C2830	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2835	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2851	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2855	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2860	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2885	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2901	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2911	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A4C2912	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2913	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2926	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2940	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2950	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2960	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2970	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2980	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2990	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4R2805	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A4R2830	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A4R2841	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A4R2842	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A4R2843	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A4R2844	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A4R2850	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A4R2901	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A4R2902	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A4R2903	321-1296-03			RES,FXD,FILM:12.0K OHM,0.25%,0.125W,TC=T2	07716	CEAC12001C
A4R2905	321-0816-03			RES,FXD,FILM:5K OHM,0.25%,0.125W,TC=T2	19701	5033RC5K000C
A4R2910	321-0685-00			RES,FXD,FILM:30K OHM,0.5%,0.125W,TC=T2	19701	5033RC30K00D
A4R2911	321-0685-00			RES,FXD,FILM:30K OHM,0.5%,0.125W,TC=T2	19701	5033RC30K00D
A4R2912	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A4R2913	321-0198-00			RES,FXD,FILM:1.13K OHM,1%,0.125W,TC=T0	07716	CEAD11300F
A4R2914	322-3306-00			RES,FXD,FILM:15K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 15K0
A4R2915	313-1202-00			RES,FXD,FILM:2K OHM,5%,0.2W	57668	TR20JE02K0
A4R2916	322-3414-00			RES,FXD,FILM:200K OHM,1%,0.2W,TC=T0	91637	CCF50G20002F
A4R2917	322-3385-00			RES,FXD,FILM:100K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100K
A4R2918	311-2270-00			RES,VAR,NONWW:TRMR,10K OHM,20%,0.5W	TK1450	GF06VT 10 K OHM
A4R2919	321-0756-00			RES,FXD,FILM:50K OHM,1%,0.125W,TC=T0	24546	NA55D5002F
A4R2920	313-1334-00			RES,FXD,FILM:330K OHM,5%,0.2W	80009	313-1334-00
A4R2921	322-3297-00			RES,FXD,FILM:12.1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 12K1
A4R2922	321-0756-00			RES,FXD,FILM:50K OHM,1%,0.125W,TC=T0	24546	NA55D5002F
A4R2923	321-0385-00			RES,FXD,FILM:100K OHM,1%,0.125W,TC=T0	19701	5033ED100K0F
A4R2924	322-3414-00			RES,FXD,FILM:200K OHM,1%,0.2W,TC=T0	91637	CCF50G20002F
A4R2925	321-0235-02			RES,FXD,FILM:2.74K OHM,0.5%,0.125W,TC=T2	24546	NC55C2741D
A4R2926	322-3222-00	B010100	B010548	RES,FXD,FILM:2K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K00
A4R2926	322-3210-00	B010549		RES,FXD,FILM:1.5K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K50
A4R2927	322-3318-00			RES,FXD,FILM:20K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 20K0
A4R2928	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A4R2929	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A4R2930	313-1152-00			RES,FXD,FILM:1.5K OHM,5%,0.2W	57668	TR20JE01K5
A4R2931	311-2258-00	B010100	B010548	RES,VAR,NONWW:TRMR,1K OHM,20%,0.5W	TK1450	GF06VT 1 K OHM
A4R2931	311-2270-00	B010549		RES,VAR,NONWW:TRMR,10K OHM,20%,0.5W	TK1450	GF06VT 10 K OHM
A4R2932	322-3414-00			RES,FXD,FILM:200K OHM,1%,0.2W,TC=T0	91637	CCF50G20002F
A4R2933	322-3385-00			RES,FXD,FILM:100K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100K
A4R2934	322-3297-00			RES,FXD,FILM:12.1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 12K1
A4R2940	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A4R2945	313-1471-00			RES,FXD,FILM:470 OHM,5%,0.2W	57668	TR20JE 470E

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A4R2975	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A4R2985	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A4U2800	156-0514-00			MICROCKT,DGTL:CMOS,DIFF 4-CHANNEL MUX	02735	CD4052BF-98
A4U2805	156-0514-00			MICROCKT,DGTL:CMOS,DIFF 4-CHANNEL MUX	02735	CD4052BF-98
A4U2810	156-0382-00			MICROCKT,DGTL:QUAD 2-INP NAND GATE	01295	SN74LS00(N OR J)
A4U2820	156-1191-01			MICROCKT,LINEAR:BIFET,DUAL OPNL AMPL,SCRN	80009	156-1191-01
A4U2830	156-1172-00			MICROCKT,DGTL:DUAL 4 BIT BIN CNTR	80009	156-1172-00
A4U2835	156-0479-00			MICROCKT,DGTL:QUAD 2-INP OR GATE	80009	156-0479-00
A4U2850	156-0388-00			MICROCKT,DGTL:DUAL D FLIP-FLOP	01295	SN74LS74 N OR J
A4U2855	156-0383-00			MICROCKT,DGTL:QUAD 2-INP NOR GATE	01295	SN74LS02 N OR J
A4U2860	156-0975-00			MICROCKT,DGTL:UNIV SHIFT/STORAGE REGISTER	34335	SN74LS299N
A4U2865	156-0796-00			MICROCKT,DGTL:8 STG SHF & STORE BUS RGTR	02735	CD4094BF
A4U2870	156-1172-00			MICROCKT,DGTL:DUAL 4 BIT BIN CNTR	80009	156-1172-00
A4U2880	156-0388-00			MICROCKT,DGTL:DUAL D FLIP-FLOP	01295	SN74LS74 N OR J
A4U2885	156-0386-00			MICROCKT,DGTL:TRIPLE 3-INP NAND GATE	01295	SN74LS10(N OR J)
A4U2890	156-0382-00			MICROCKT,DGTL:QUAD 2-INP NAND GATE	01295	SN74LS00(N OR J)
A4U2900	156-0386-00			MICROCKT,DGTL:TRIPLE 3-INP NAND GATE	01295	SN74LS10(N OR J)
A4U2905	156-1702-00			MICROCKT,DGTL:STTL,10 BIT REGISTER	34335	AM29821DCB
A4U2910	156-1555-00			MICROCKT,LINEAR:D/A CONVERTER	34335	AM6080PC
A4U2920	156-1594-00			IC,MEMORY:NMOS,SRAM:2K X 8,150NS;,DIP24.6	65786	CY6116-55PC
A4U2930	160-1631-02			MICROCKT,DGTL:4096 X 8 EPROM,PRGM	80009	160-1631-02
A4U2935	156-0956-00			MICROCKT,DGTL:OCTAL BFR W/3 STATE OUT	18324	N74LS244(N OR F)
A4U2940	156-1172-00			MICROCKT,DGTL:DUAL 4 BIT BIN CNTR	80009	156-1172-00
A4U2950	156-0388-00			MICROCKT,DGTL:DUAL D FLIP-FLOP	01295	SN74LS74 N OR J
A4U2960	156-0796-00			MICROCKT,DGTL:8 STG SHF & STORE BUS RGTR	02735	CD4094BF
A4U2965	156-0382-00			MICROCKT,DGTL:QUAD 2-INP NAND GATE	01295	SN74LS00(N OR J)
A4U2970	156-0480-00			MICROCKT,DGTL:TTL,QUAD 2-INP AND GATE	80009	156-0480-00
A4U2980	156-0382-00			MICROCKT,DGTL:QUAD 2-INP NAND GATE	01295	SN74LS00(N OR J)
A4U2985	156-0768-00			MICROCKT,DGTL:BIDIRECT UNIV SR	01295	SN74LS194AN
A4U2990	156-0381-00			MICROCKT,DGTL:QUAD 2-INP ECXL OR GATE	01295	SN74LS86 N OR J
A4U2995	156-0651-00			MICROCKT,DGTL:8-BIT PRL-OUT SER SHF RGTR	80009	156-0651-00
A4VR2805	152-0217-00			SEMICOND DVC,DI:ZEN,SI,8.2V,5%,0.4W,DO-7	04713	SZG20
A4VR2925	152-0662-00			SEMICOND DVC,DI:ZEN,SI,5V,1%,400MW,DO-7	04713	SZG195RL
A4W411	175-4581-01			CA ASSY,SP,ELEC:26,28 AWG,2.25 L,RIBBON	22526	ORDER BY DESCR
A4W2851	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A4W2913	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07

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A5	670-9052-02	B010100	B049999	CIRCUIT BD ASSY:DIGITAL CONTROL	80009	670-9052-02
A5BT2570	146-0049-00			BATTERY,STORAGE:3.5V,750MAH	81855	LTC-7P
A5C2010	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A5C2011	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A5C2101	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2110	281-0814-00			CAP,FXD,CER DI:100 PF,10%,100V	04222	MA101A101KAA
A5C2111	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2112	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2113	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A5C2160	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2220	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2221	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2230	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2240	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2320	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2321	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2322	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A5C2330	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2331	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A5C2332	285-1300-01			CAP,FXD,MTLZD:0.1UF,10%,63V	55112	185/0.1/K/63/ABA
A5C2333	285-1300-01			CAP,FXD,MTLZD:0.1UF,10%,63V	55112	185/0.1/K/63/ABA
A5C2340	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2350	290-0527-00			CAP,FXD,ELCTLT:15UF,20%,20V	05397	T368B156M020AS
A5C2351	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2360	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2420	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A5C2421	285-1300-01			CAP,FXD,MTLZD:0.1UF,10%,63V	55112	185/0.1/K/63/ABA
A5C2422	281-0791-00			CAP,FXD,CER DI:270PF,10%,100V	04222	MA101C271KAA
A5C2430	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2431	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A5C2432	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A5C2450	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2470	290-0527-00			CAP,FXD,ELCTLT:15UF,20%,20V	05397	T368B156M020AS
A5C2501	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2510	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2511	281-0791-00			CAP,FXD,CER DI:270PF,10%,100V	04222	MA101C271KAA
A5C2520	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2521	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2530	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2550	281-0819-00			CAP,FXD,CER DI:33 PF,5%,50V	04222	GC105A330J
A5C2551	281-0816-00			CAP,FXD,CER DI:82 PF,5%,100V	04222	MA106A820JAA
A5C2552	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2601	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2610	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2620	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2621	285-1300-01			CAP,FXD,MTLZD:0.1UF,10%,63V	55112	185/0.1/K/63/ABA
A5C2622	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A5C2630	285-1300-01			CAP,FXD,MTLZD:0.1UF,10%,63V	55112	185/0.1/K/63/ABA
A5C2631	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A5C2632	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2640	285-1300-01			CAP,FXD,MTLZD:0.1UF,10%,63V	55112	185/0.1/K/63/ABA
A5C2650	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2660	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2720	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2721	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A5C2730	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A5C2731	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2732	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A5C2733	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2740	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2800	281-0812-00			CAP,FXD,CER DI:1000PF,10%,100V	04222	MA101C102KAA
A5CR2070	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2071	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2170	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2230	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2231	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2232	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2233	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2370	152-0951-00			SEMICOND DVC,DI:SCHOTTKY,SI,60V,2.25PF	80009	152-0951-00
A5CR2371	152-0951-00			SEMICOND DVC,DI:SCHOTTKY,SI,60V,2.25PF	80009	152-0951-00
A5CR2420	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2610	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2620	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2621	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2622	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2630	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2631	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2640	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2770	152-0951-00			SEMICOND DVC,DI:SCHOTTKY,SI,60V,2.25PF	80009	152-0951-00
A5J251	131-3360-00			CONN,RCPT,ELEC:HEADER,STR,20 PIN	53387	3592-6002
A5J500	131-3364-00			CONN,RCPT,ELEC:HEADER,STRAIGHT,34 PIN	53387	3594-6002
A5J501	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 3)	22526	48283-036
A5J503	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 3)	22526	48283-036
A5J651	131-3360-00			CONN,RCPT,ELEC:HEADER,STR,20 PIN	53387	3592-6002
A5J652	131-3360-00			CONN,RCPT,ELEC:HEADER,STR,20 PIN	53387	3592-6002
A5Q2070	151-0342-00			TRANSISTOR:PNP,SI,TO-92	07263	S035928
A5Q2270	151-0342-00			TRANSISTOR:PNP,SI,TO-92	07263	S035928
A5Q2320	151-0341-00			TRANSISTOR:NPN,SI,TO-106	04713	SPS6919
A5R2001	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2002	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2004	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2005	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2006	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2007	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2010	311-2234-00			RES,VAR,NONWW:TRMR,5K OHM,20%,0.5W LINEAR	TK1450	GF06UT 5K
A5R2011	322-3431-00			RES,FXD,FILM:301K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 301K
A5R2012	322-3289-02			RES,FXD,FILM:10K OHM,0.5%,0.2W,TC=T2	57668	CRB 20 DYE 10K0
A5R2013	322-3289-02			RES,FXD,FILM:10K OHM,0.5%,0.2W,TC=T2	57668	CRB 20 DYE 10K0
A5R2101	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2102	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2103	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2104	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2110	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2170	313-1512-00			RES,FXD,FILM:5.1K OHM,5%,0.2W	57668	TR20JE 5K1
A5R2170	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A5R2171	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2172	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2201	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2202	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2203	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2204	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2205	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2206	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2220	313-1681-00			RES,FXD,FILM:680 OHM,5%,0.2W	57668	TR20JE 680E

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A5R2230	322-3482-02			RES,FXD,FILM:14.2K OHM,0.5%,0.2W,TC=T0	57668	CRB 20 DYE 14K2
A5R2231	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2232	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2241	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2242	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2244	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2250	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2251	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2301	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2302	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2303	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2304	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2305	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2306	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2320	313-1203-00			RES,FXD,FILM:20K OHM,5%,0.2W	57668	TR20JE20K
A5R2330	322-3360-02			RES,FXD,FILM:54.9K OHM,0.5%,0.2W,TC=T2	57668	CRB20 DYE 54K9
A5R2331	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A5R2332	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A5R2333	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A5R2334	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A5R2340	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2341	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2342	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2343	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2344	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2345	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2346	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2370	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2401	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2402	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2403	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2404	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2405	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2406	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2407	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2408	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2409	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2410	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2411	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2412	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2413	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2414	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2415	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2416	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2417	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2420	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2421	322-3300-02			RES,FXD,FILM:13K OHM,0.5%,0.2W,TC=T2	57668	CRB20 DYE 13K0
A5R2422	322-3482-02			RES,FXD,FILM:14.2K OHM,0.5%,0.2W,TC=T0	57668	CRB 20 DYE 14K2
A5R2430	322-3289-02			RES,FXD,FILM:10K OHM,0.5%,0.2W,TC=T2	57668	CRB 20 DYE 10K0
A5R2431	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2432	322-3325-00			RES,FXD,FILM:23.7K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 23K7
A5R2433	322-3289-02			RES,FXD,FILM:10K OHM,0.5%,0.2W,TC=T2	57668	CRB 20 DYE 10K0
A5R2434	322-3289-02			RES,FXD,FILM:10K OHM,0.5%,0.2W,TC=T2	57668	CRB 20 DYE 10K0
A5R2440	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2441	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2442	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2443	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2444	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A5R2470	313-1681-00			RES,FXD,FILM:680 OHM,5%,0.2W	57668	TR20JE 680E
A5R2471	313-1681-00			RES,FXD,FILM:680 OHM,5%,0.2W	57668	TR20JE 680E
A5R2500	313-1331-00			RES,FXD,FILM:330 OHM,5%,0.2W	57668	TR20JE 330E
A5R2501	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2502	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2503	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2504	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2505	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2506	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A5R2510	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2511	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2512	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2513	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2520	322-3177-02			RES,FXD,FILM:681 OHM,0.5%,0.2W,TC=T2	57668	CRB 20 DYE 681E
A5R2521	322-3177-02			RES,FXD,FILM:681 OHM,0.5%,0.2W,TC=T2	57668	CRB 20 DYE 681E
A5R2522	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2523	313-1683-00			RES,FXD,FILM:68K OHM,5%,0.2W	57668	TR20JE 68K
A5R2524	313-1683-00			RES,FXD,FILM:68K OHM,5%,0.2W	57668	TR20JE 68K
A5R2530	315-0106-00			RES,FXD,FILM:10M OHM,5%,0.25W	01121	CB1065
A5R2531	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2532	313-1683-00			RES,FXD,FILM:68K OHM,5%,0.2W	57668	TR20JE 68K
A5R2533	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A5R2534	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A5R2535	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A5R2536	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2537	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2540	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2541	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2542	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2543	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2544	313-1681-00			RES,FXD,FILM:680 OHM,5%,0.2W	57668	TR20JE 680E
A5R2545	313-1331-00			RES,FXD,FILM:330 OHM,5%,0.2W	57668	TR20JE 330E
A5R2560	313-1222-00			RES,FXD,FILM:2.2K OHM,5%,0.2W	57668	TR20JE 02K2
A5R2601	313-1331-00			RES,FXD,FILM:330 OHM,5%,0.2W	57668	TR20JE 330E
A5R2602	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2603	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2604	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A5R2610	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2611	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2612	313-1512-00			RES,FXD,FILM:5.1K OHM,5%,0.2W	57668	TR20JE 5K1
A5R2613	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2620	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2621	313-1222-00			RES,FXD,FILM:2.2K OHM,5%,0.2W	57668	TR20JE 02K2
A5R2622	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2623	313-1222-00			RES,FXD,FILM:2.2K OHM,5%,0.2W	57668	TR20JE 02K2
A5R2624	313-1512-00			RES,FXD,FILM:5.1K OHM,5%,0.2W	57668	TR20JE 5K1
A5R2630	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A5R2631	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A5R2632	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A5R2640	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2641	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2642	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2643	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2644	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2645	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2660	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2661	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2701	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A5R2702	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2703	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2704	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2705	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2706	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2707	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2708	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2709	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2710	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2711	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2712	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2720	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2721	313-1203-00			RES,FXD,FILM:20K OHM,5%,0.2W	57668	TR20JE20K
A5R2730	313-1203-00			RES,FXD,FILM:20K OHM,5%,0.2W	57668	TR20JE20K
A5R2731	315-0107-00			RES,FXD,FILM:100M OHM,5%,0.25W	01121	CB1075
A5R2732	315-0107-00			RES,FXD,FILM:100M OHM,5%,0.25W	01121	CB1075
A5R2733	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A5R2734	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2735	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2740	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A5R2741	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2742	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2770	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5TP2070	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL	22526	48283-036
A5TP2420	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL	22526	48283-036
A5TP2421	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL	22526	48283-036
A5TP2701	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL	22526	48283-036
A5U2101	156-1589-00			MICROCKT,LINEAR:D/A CONVERTER,12 BIT,HIGH SPEED,MONOLITHIC	06665	DAC312FR
A5U2140	156-1342-01			MICROCKT,DGTL:NMOS,MPU,8-BIT W/CLK	04713	SC67127P
A5U2160	160-5370-04	B010100	B011267	MICROCKT,DGTL:65536 X 8 EPROM,PRGM	80009	160-5370-04
A5U2160	160-5370-08	B011268	B011462	MICROCKT,DGTL:65536 X 8 EPROM,PRGM	80009	160-5370-08
A5U2160	160-5370-09	B011463		MICROCKT,DGTL:65536 X 8 EPROM,PRGM	80009	160-5370-09
A5U2201	156-0865-00			MICROCKT,DGTL:OCTAL D FF W/CLR	80009	156-0865-00
A5U2210	156-0391-00			MICROCKT,DGTL:LSTTL,HEX D TYPE FF W/CLEAR	04713	74LS174(N OR J)
A5U2220	156-0956-00			MICROCKT,DGTL:OCTAL BFR W/3 STATE OUT	18324	N74LS244(N OR F)
A5U2240	156-2396-00			MICROCKT,LINEAR:BIPOLAR,MPU RESET GENERATOR	01295	TL7705 ACP
A5U2250	160-5061-00			MICROCKT,DGTL:PROGRAMMABLE LOGIC DEVICE	80009	160-5061-00
A5U2260	160-5371-04	B010100	B011267	MICROCKT,DGTL:65536 X 8 EPROM,PRGM	80009	160-5371-04
A5U2260	160-5371-08	B011268	B011462	MICROCKT,DGTL:65536 X 8 EPROM,PRGM	80009	160-5371-08
A5U2260	160-5371-09	B011463		MICROCKT,DGTL:65536 X 8 EPROM,PRGM	80009	160-5371-09
A5U2301	156-0865-00			MICROCKT,DGTL:OCTAL D FF W/CLR	80009	156-0865-00
A5U2310	156-0865-00			MICROCKT,DGTL:OCTAL D FF W/CLR	80009	156-0865-00
A5U2350	156-0956-00			MICROCKT,DGTL:OCTAL BFR W/3 STATE OUT	18324	N74LS244(N OR F)
A5U2401	156-0513-03			MICROCKT,LINEAR:CMOS,8 CHAN ANALOG MUX	04713	MC14051BCL
A5U2410	156-1486-00			MICROCKT,DGTL:CMOS,8 CHANNEL DATA SEL	02735	CD4512BFX
A5U2420	156-1200-01			MICROCKT,LINEAR:BIFET,QUAD OPNL AMPL,SCRN	80009	156-1200-01
A5U2430	156-1200-01			MICROCKT,LINEAR:BIFET,QUAD OPNL AMPL,SCRN	80009	156-1200-01
A5U2440	156-0388-00			MICROCKT,DGTL:DUAL D FLIP-FLOP	01295	SN74LS74 N OR J
A5U2450	156-1065-00			MICROCKT,DGTL:OCTAL D TYPE TRANS LATCHES	01295	SN74LS373N
A5U2460	156-2473-00			IC,MEMORY:CMOS,SRAM;8K X 8,200NS,10UA	TK0961	uPD4464C-20
A5U2501	156-0513-03			MICROCKT,LINEAR:CMOS,8 CHAN ANALOG MUX	04713	MC14051BCL
A5U2510	156-1126-01			MICROCKT,LINEAR:VOLTAGE COMPARATOR,SELECTED	01295	LM311JG4
A5U2520	156-1191-01			MICROCKT,LINEAR:BIFET,DUAL OPNL AMPL,SCRN	80009	156-1191-01
A5U2521	156-0513-03			MICROCKT,LINEAR:CMOS,8 CHAN ANALOG MUX	04713	MC14051BCL
A5U2530	156-0513-03			MICROCKT,LINEAR:CMOS,8 CHAN ANALOG MUX	04713	MC14051BCL
A5U2540	156-1722-00			MICROCKT,DGTL:FTTL,HEX INVERTER	04713	MC74F04ND
A5U2550	156-0469-00			MICROCKT,DGTL:3-LINE TO 8-LINE DECODER	01295	SN74LS138N

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A5U2601	156-0513-03			MICROCKT,LINEAR:CMOS,8 CHAN ANALOG MUX	04713	MC14051BCL
A5U2620	156-1200-01			MICROCKT,LINEAR:BIFET,QUAD OPNL AMPL,SCRN	80009	156-1200-01
A5U2630	156-1200-01			MICROCKT,LINEAR:BIFET,QUAD OPNL AMPL,SCRN	80009	156-1200-01
A5U2640	156-0895-00			MICROCKT,DGTL:14-BIT BINARY COUNTER	04713	MC14020BCL
A5U2650	156-0804-00			MICROCKT,DGTL:QUADRUPLE S-R LATCH	04713	74LS279(N OR J)
A5U2660	156-1026-00			MICROCKT,DGTL:4 LINE TO 1 LINE DECODER	18324	74LS154N
A5VR2420	152-0278-00			SEMICOND DVC,DI:ZEN,SI,3V,5%,0.4W,DO-7	80009	152-0278-00
A5W511	174-0002-00			CA ASSY,SP,ELEC:26,28 AWG,2.0 L	80009	174-0002-00
A5W512	174-0001-00			CA ASSY,SP,ELEC:34,28 AWG,2.0 L	80009	174-0001-00
A5W2070	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A5W2540	131-1817-01			BUS,CONDUCTOR:22 AWG,2.0 TO 2.125 SPACING	TK1492	ORDER BY DESCR
A5W2610	131-1817-01			BUS,CONDUCTOR:22 AWG,2.0 TO 2.125 SPACING	TK1492	ORDER BY DESCR
A5W2701	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A5Y2540	158-0248-01			XTAL UNIT,QTZ:10.000MHZ,0.01% SER RESONANT	14301	011-669-02923

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A5	671-0965-00	B050000		CIRCUIT BD ASSY:CONTROL/READOUT/BUFFER	80009	671-0965-00
A5C2010	290-5009-00			CAP,FXD,ELCTLT:15UF,25V	56289	293D156X0025D2T
A5C2011	290-5009-00			CAP,FXD,ELCTLT:15UF,25V	56289	293D156X0025D2T
A5C2101	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2110	283-5188-00			CAP,FXD,CER DI:100PF,5%,100V	04222	12061A101J1T050R
A5C2111	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2113	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A5C2160	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2220	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2221	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2222	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2230	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2240	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2241	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2250	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2321	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2322	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2323	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2324	283-5003-00			CAP,FXD,CER DI:0.01UF,10%,50V	14674	12065C103KAT060R
A5C2325	283-5003-00			CAP,FXD,CER DI:0.01UF,10%,50V	14674	12065C103KAT060R
A5C2330	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2331	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A5C2332	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2333	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2350	290-5009-00			CAP,FXD,ELCTLT:15UF,25V	56289	293D156X0025D2T
A5C2352	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2360	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2415	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2420	290-5009-00			CAP,FXD,ELCTLT:15UF,25V	56289	293D156X0025D2T
A5C2421	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2422	283-5197-00			CAP,FXD,CER DI:330PF,5%,100V	TK2282	W1206C331J3B05
A5C2425	283-5003-00			CAP,FXD,CER DI:0.01UF,10%,50V	14674	12065C103KAT060R
A5C2430	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2431	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2432	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2433	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2434	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2440	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2450	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2451	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2452	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2460	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2465	283-5188-00			CAP,FXD,CER DI:100PF,5%,100V	04222	12061A101J1T050R
A5C2501	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2510	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2511	283-5197-00			CAP,FXD,CER DI:330PF,5%,100V	TK2282	W1206C331J3B05
A5C2520	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2521	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2530	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2540	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2542	283-5114-00	B050000	B051130	CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2542	283-5098-00	B051131		CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2550	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2610	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2621	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2622	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2623	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2629	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A5C2630	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2631	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2632	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2633	283-5003-00			CAP,FXD,CER DI:0.01UF,10%,50V	14674	12065C103KAT060R
A5C2634	283-5003-00			CAP,FXD,CER DI:0.01UF,10%,50V	14674	12065C103KAT060R
A5C2640	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2641	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2650	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2720	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2721	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2722	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2730	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2731	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2732	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2733	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2734	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2820	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2821	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2830	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2831	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2835	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2836	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2850	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2851	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2855	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2860	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2861	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2870	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2875	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2885	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2890	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2901	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2905	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2911	283-5003-00			CAP,FXD,CER DI:0.01UF,10%,50V	14674	12065C103KAT060R
A5C2913	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2926	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2940	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2950	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2960	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2965	290-5009-00			CAP,FXD,ELCTLT:15UF,25V	56289	293D156X0025D2T
A5C2970	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2980	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2981	290-5009-00			CAP,FXD,ELCTLT:15UF,25V	56289	293D156X0025D2T
A5C2990	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2995	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5CR2230	152-5004-00			SEMICOND DVC,DI:SI,SW,SER PR,70V	04713	BAV99T1
A5CR2332	152-5004-00			SEMICOND DVC,DI:SI,SW,SER PR,70V	04713	BAV99T1
A5CR2420	152-5004-00			SEMICOND DVC,DI:SI,SW,SER PR,70V	04713	BAV99T1
A5CR2421	152-5004-00			SEMICOND DVC,DI:SI,SW,SER PR,70V	04713	BAV99T1
A5CR2422	152-5004-00			SEMICOND DVC,DI:SI,SW,SER PR,70V	04713	BAV99T1
A5CR2423	152-5004-00			SEMICOND DVC,DI:SI,SW,SER PR,70V	04713	BAV99T1
A5CR2610	152-5005-00			SEMICOND DVC,DI:DUAL,COMMON ANODE,70V,BAW56	04713	MBAW56TI
A5CR2620	152-5005-00			SEMICOND DVC,DI:DUAL,COMMON ANODE,70V,BAW56	04713	MBAW56TI
A5CR2621	152-5005-00			SEMICOND DVC,DI:DUAL,COMMON ANODE,70V,BAW56	04713	MBAW56TI
A5CR2640	152-5005-00			SEMICOND DVC,DI:DUAL,COMMON ANODE,70V,BAW56	04713	MBAW56TI
A5J251	131-3360-00			CONN,RCPT,ELEC:HEADER,STR,20 PIN	53387	3592-6002
A5J411	131-3362-00			CONN,RCPT,ELEC:HEADER,STR,26 PIN	53387	3593-6002
A5J501	131-4671-00			CONN,RCPT,ELEC:1 X 3,0.1 SPACING	80009	131-4671-00

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A5J503	131-4671-00			CONN,RCPT,ELEC:1 X 3,0.1 SPACING	80009	131-4671-00
A5J504	131-4671-00			CONN,RCPT,ELEC:1 X 3,0.1 SPACING	80009	131-4671-00
A5J511	131-3362-00			CONN,RCPT,ELEC:HEADER,STR,26 PIN	53387	3593-6002
A5J512	131-3364-00			CONN,RCPT,ELEC:HEADER,STRAIGHT,34 PIN	53387	3594-6002
A5J651	131-3360-00			CONN,RCPT,ELEC:HEADER,STR,20 PIN	53387	3592-6002
A5J652	131-3360-00			CONN,RCPT,ELEC:HEADER,STR,20 PIN	53387	3592-6002
A5J4241	131-3323-00			CONN,RCPT,ELEC:HEADER,STR,2 X 20,0.1 CTR	22526	66506-025
A5J4330	131-3152-00			CONN,RCPT,ELEC:HEADER,2 X 8 0.1 SPACING	22526	66506-043
A5P501	131-0993-00			BUS,CONDUCTOR:SHUNT ASSEMBLY,BLACK	22526	65474-005
A5P503	131-0993-00			BUS,CONDUCTOR:SHUNT ASSEMBLY,BLACK	22526	65474-005
A5P504	131-0993-00			BUS,CONDUCTOR:SHUNT ASSEMBLY,BLACK	22526	65474-005
A5Q2320	151-5001-00			TRANSISTOR:NPN,SI,SOT-23	80009	151-5001-00
A5Q2805	151-5001-00			TRANSISTOR:NPN,SI,SOT-23	80009	151-5001-00
A5R2001	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2002	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2004	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2005	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2006	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2007	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2010	311-5038-00			RES,VAR,NONWW:TRMR,20K OHM,25%,0.1W	32997	3314A-1-203E
A5R2011	321-5026-00			RES,FXD,FILM:4.75K,1%,0.125W	01121	BCK4751FT
A5R2012	321-5165-00			RES,FXD,FILM:10K OHM,0.1%,0.125W,TC=T9	80009	321-5165-00
A5R2013	321-5165-00			RES,FXD,FILM:10K OHM,0.1%,0.125W,TC=T9	80009	321-5165-00
A5R2014	321-5167-00			RES,FXD,FILM:221K OHM,1%,0.125W	80009	321-5167-00
A5R2015	321-5041-00			RES,FXD,FILM:82.5K,1%,0.125W	01121	BCK8252FT
A5R2016	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2101	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2102	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2103	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2104	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2201	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2202	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2203	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2204	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2205	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2210	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2211	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2212	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2213	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2214	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2215	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2220	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2230	321-5165-00			RES,FXD,FILM:10K OHM,0.1%,0.125W,TC=T9	80009	321-5165-00
A5R2231	321-5022-00			RES,FXD,FILM:2.21K,1%,0.125W	01121	BCK2211FT
A5R2232	321-5022-00			RES,FXD,FILM:2.21K,1%,0.125W	01121	BCK2211FT
A5R2241	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2242	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2244	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2251	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2301	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2302	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2303	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2304	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2305	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2320	321-5034-00			RES,FXD,FILM:22.1K,1%,0.125W	01121	BCK2212FT
A5R2321	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2322	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2323	321-5032-00			RES,FXD,FILM:15.0K,1%,0.125W	01121	BCK1502FT

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A5R2329	321-5036-00			RES,FXD,FILM:33.2K,1%,0.125W	01121	BCK3322FT
A5R2330	321-5027-00			RES,FXD,FILM:5.62K,1%,0.125W	01121	BCK5621FT
A5R2331	321-5023-00			RES,FXD,FILM:2.74K,1%,0.125W	01121	BCK2741FT
A5R2332	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2333	321-5023-00			RES,FXD,FILM:2.74K,1%,0.125W	01121	BCK2741FT
A5R2334	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2340	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2341	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2342	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2343	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2344	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2345	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2346	321-5022-00			RES,FXD,FILM:2.21K,1%,0.125W	01121	BCK2211FT
A5R2401	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2402	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2403	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2404	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2405	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2406	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2407	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2408	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2409	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2410	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2411	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2412	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2413	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2414	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2415	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2416	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2417	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2420	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2421	321-5165-00			RES,FXD,FILM:10K OHM,0.1%,0.125W,TC=T9	80009	321-5165-00
A5R2422	321-5165-00			RES,FXD,FILM:10K OHM,0.1%,0.125W,TC=T9	80009	321-5165-00
A5R2423	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2424	321-5031-00			RES,FXD,FILM:12.1K,1%,0.125W	01121	BCK1212FT
A5R2430	321-5165-00			RES,FXD,FILM:10K OHM,0.1%,0.125W,TC=T9	80009	321-5165-00
A5R2431	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2432	321-5036-00			RES,FXD,FILM:33.2K,1%,0.125W	01121	BCK3322FT
A5R2433	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2434	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2435	321-5041-00			RES,FXD,FILM:82.5K,1%,0.125W	01121	BCK8252FT
A5R2440	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2441	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2442	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2443	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2444	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2461	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2465	321-5016-00			RES,FXD,FILM:681 OHM,1%,0.125W	01121	BCK6810FT
A5R2501	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2502	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2503	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2504	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2505	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2511	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2512	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2513	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2520	321-5164-00			RES,FXD,FILM:681 OHM,0.1%,0.125W,TC=T9	80009	321-5164-00
A5R2521	321-5164-00			RES,FXD,FILM:681 OHM,0.1%,0.125W,TC=T9	80009	321-5164-00

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A5R2522	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2523	321-5040-00			RES,FXD,FILM:68.1K,1%,0.125W	01121	BCK6812FT
A5R2524	321-5040-00			RES,FXD,FILM:68.1K,1%,0.125W	01121	BCK6812FT
A5R2531	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2532	321-5040-00			RES,FXD,FILM:68.1K,1%,0.125W	01121	BCK6812FT
A5R2533	321-5023-00			RES,FXD,FILM:2.74K,1%,0.125W	01121	BCK2741FT
A5R2534	321-5023-00			RES,FXD,FILM:2.74K,1%,0.125W	01121	BCK2741FT
A5R2535	321-5023-00			RES,FXD,FILM:2.74K,1%,0.125W	01121	BCK2741FT
A5R2536	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2537	321-5022-00			RES,FXD,FILM:2.21K,1%,0.125W	01121	BCK2211FT
A5R2540	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2560	321-5022-00			RES,FXD,FILM:2.21K,1%,0.125W	01121	BCK2211FT
A5R2601	321-5012-00			RES,FXD,FILM:332 OHM,1%,0.125W	01121	BCK3320FT
A5R2602	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2603	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2611	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2612	321-5026-00			RES,FXD,FILM:4.75K,1%,0.125W	01121	BCK4751FT
A5R2613	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2620	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2621	321-5022-00			RES,FXD,FILM:2.21K,1%,0.125W	01121	BCK2211FT
A5R2622	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2623	321-5022-00			RES,FXD,FILM:2.21K,1%,0.125W	01121	BCK2211FT
A5R2624	321-5026-00			RES,FXD,FILM:4.75K,1%,0.125W	01121	BCK4751FT
A5R2625	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2626	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2630	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2631	321-5023-00			RES,FXD,FILM:2.74K,1%,0.125W	01121	BCK2741FT
A5R2632	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2640	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2643	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2644	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2645	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2646	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2647	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2648	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2649	321-5012-00			RES,FXD,FILM:332 OHM,1%,0.125W	01121	BCK3320FT
A5R2701	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2702	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2703	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2704	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2705	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2706	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2707	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2708	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2709	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2710	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2711	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2712	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2720	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2721	321-5034-00			RES,FXD,FILM:22.1K,1%,0.125W	01121	BCK2212FT
A5R2730	321-5034-00			RES,FXD,FILM:22.1K,1%,0.125W	01121	BCK2212FT
A5R2731	321-5199-00			RES,FXD,FILM:100M OHM,10%,0.0625 W	80009	321-5199-00
A5R2732	321-5199-00			RES,FXD,FILM:100M OHM,10%,0.0625 W	80009	321-5199-00
A5R2733	321-5023-00			RES,FXD,FILM:2.74K,1%,0.125W	01121	BCK2741FT
A5R2734	321-5022-00			RES,FXD,FILM:2.21K,1%,0.125W	01121	BCK2211FT
A5R2735	321-5022-00			RES,FXD,FILM:2.21K,1%,0.125W	01121	BCK2211FT
A5R2740	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2741	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A5R2830	321-5012-00	B050000	B050689	RES,FXD,FILM:332 OHM,1%,0.125W	01121	BCK3320FT
A5R2830	321-5051-00	B050690		RES,FXD,FILM:0 OHM,1%,0.125W	80009	321-5051-00
A5R2865	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2866	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2885	321-5018-00	B050000	B050352	RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2885	321-5012-00	B050353		RES,FXD,FILM:332 OHM,1%,0.125W	01121	BCK3320FT
A5R2890	321-5018-00	B050000	B050352	RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2890	321-5012-00	B050353		RES,FXD,FILM:332 OHM,1%,0.125W	01121	BCK3320FT
A5R2902	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2903	321-5165-00			RES,FXD,FILM:10K OHM,0.1%,0.125W,TC=T9	80009	321-5165-00
A5R2904	321-5051-00	B050000	B050689	RES,FXD,FILM:0 OHM,1%,0.125W	80009	321-5051-00
A5R2904	321-0512-00	B050690		RES,FXD,FILM:2.1M OHM,1%,0.125W,TC=T0	80009	321-0512-00
A5R2905	321-5028-00			RES,FXD,FILM:6.81K,1%,0.125W	01121	BCK6811FT
A5R2906	321-5165-00			RES,FXD,FILM:10K OHM,0.1%,0.125W,TC=T9	80009	321-5165-00
A5R2907	321-5033-00			RES,FXD,FILM:18.2K,1%,0.125W	01121	BCK1822FT
A5R2908	321-5032-00			RES,FXD,FILM:15.0K,1%,0.125W	01121	BCK1502FT
A5R2909	321-5032-00			RES,FXD,FILM:15.0K,1%,0.125W	01121	BCK1502FT
A5R2910	321-5032-00			RES,FXD,FILM:15.0K,1%,0.125W	01121	BCK1502FT
A5R2911	321-5032-00			RES,FXD,FILM:15.0K,1%,0.125W	01121	BCK1502FT
A5R2912	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2913	321-5015-00			RES,FXD,FILM:562 OHM,1%,0.125W	01121	BCK5620FT
A5R2914	321-5032-00			RES,FXD,FILM:15.0K,1%,0.125W	01121	BCK1502FT
A5R2915	321-5015-00			RES,FXD,FILM:562 OHM,1%,0.125W	01121	BCK5620FT
A5R2916	321-5064-00			RES,FXD,FILM:200K,1%,0.125W,1206,8MM	80009	321-5064-00
A5R2917	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2918	311-5038-00			RES,VAR,NONWW:TRMR,20K OHM,25%,0.1W	32997	3314A-1-203E
A5R2919	321-5038-00			RES,FXD,FILM:47.5K,1%,0.125W	01121	BCK4752FT
A5R2920	321-5064-00			RES,FXD,FILM:200K,1%,0.125W,1206,8MM	80009	321-5064-00
A5R2921	321-5031-00			RES,FXD,FILM:12.1K,1%,0.125W	01121	BCK1212FT
A5R2922	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2923	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2924	321-5064-00			RES,FXD,FILM:200K,1%,0.125W,1206,8MM	80009	321-5064-00
A5R2925	321-5023-00			RES,FXD,FILM:2.74K,1%,0.125W	01121	BCK2741FT
A5R2926	321-5020-00			RES,FXD,FILM:1.50K,1%,0.125W	01121	BCK1501FT
A5R2927	321-5026-00			RES,FXD,FILM:4.75K,1%,0.125W	01121	BCK4751FT
A5R2928	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2929	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2930	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2931	311-5040-00	B050000	B051665	RES,VAR,NONWW:TRMR,10K OHM,25%,0.1W	32997	3314J-1-103E
A5R2931	311-5034-00	B051666		RES,VAR,NONWW:TRMR,2K OHM,25%,0.1W	51406	RVG4E-202VM-TA
A5R2932	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2933	321-5064-00			RES,FXD,FILM:200K,1%,0.125W,1206,8MM	80009	321-5064-00
A5R2934	321-5064-00			RES,FXD,FILM:200K,1%,0.125W,1206,8MM	80009	321-5064-00
A5R2935	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2960	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2961	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2995	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5U2101	156-5157-01			MICROCKT,INTFC:DAC,BIPOLAR,12 BIT	80009	156-5157-01
A5U2140	156-1342-01			MICROCKT,DGTL:NMOS,MPU,8-BIT W/CLK	04713	SC67127P
A5U2160	160-5876-00	B050000	B050958	MICROCKT,DGTL:8K X 8 EPROM,PRGM	80009	160-5876-00
A5U2160	160-5876-01	B050959		MICROCKT,DGTL:8K X 8 EPROM,PRGM	80009	160-5876-01
A5U2201	156-5147-01			MICROCKT,DGTL:CMOS,OCTAL D TYPE FF W/RESET	80009	156-5147-01
A5U2210	156-5147-01			MICROCKT,DGTL:CMOS,OCTAL D TYPE FF W/RESET	80009	156-5147-01
A5U2220	156-5071-01			MICROCKT,DGTL:CMOS,OCTAL BUS TRANS	80009	156-5071-01
A5U2240	156-5489-01			MICROCKT,LINEAR:MPU RESET GEN FOR 5V SYS	80009	156-5489-01
A5U2250	160-5874-00			MICROCKT,DGTL:LOGIC DEVICE,PRGM	80009	160-5874-00
A5U2301	156-5147-01			MICROCKT,DGTL:CMOS,OCTAL D TYPE FF W/RESET	80009	156-5147-01

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A5U2310	156-5147-01			MICROCKT,DGTL:CMOS,OCTAL D TYPE FF W/RESET	80009	156-5147-01
A5U2350	156-5071-01			MICROCKT,DGTL:CMOS,OCTAL BUS TRANS	80009	156-5071-01
A5U2360	160-5877-00	B050000	B050581	MICROCKT,DGTL:16K X 8 X 8 EPROM,PRGM	80009	160-5877-00
A5U2360	160-5877-01	B050582		MICROCKT,DGTL:16K X 8 X 8 EPROM,PRGM	80009	160-5877-01
A5U2401	156-5050-01			MICROCKT,DGTL:HCMOS,ANALOG SW,8 CHAN	80009	156-5050-01
A5U2405	156-5409-01			MICROCKT,DGTL:HCMOS,OCTAL D-TYPE TRANS	80009	156-5409-01
A5U2410	156-5459-01			MICROCKT,DGTL:CMOS,OCTAL BUS TRANSCEIVER,	80009	156-5459-01
A5U2415	156-5409-01			MICROCKT,DGTL:HCMOS,OCTAL D-TYPE TRANS	80009	156-5409-01
A5U2420	156-2051-01			MICROCKT,LINEAR:OPNL AMPL,QUAD,JET INPUT	80009	156-2051-01
A5U2425	156-5409-01			MICROCKT,DGTL:HCMOS,OCTAL D-TYPE TRANS	80009	156-5409-01
A5U2430	156-2051-01			MICROCKT,LINEAR:OPNL AMPL,QUAD,JET INPUT	80009	156-2051-01
A5U2440	156-5145-01			MICROCKT,DGTL:HCMOS,DUAL D-TYPE FF	80009	156-5145-01
A5U2450	156-5409-01			MICROCKT,DGTL:HCMOS,OCTAL D-TYPE TRANS	80009	156-5409-01
A5U2460	156-2991-00			IC,MEMORY:CMOS,NVRAM;8K X 8,200NS,SRAM	80009	156-2991-00
A5U2501	156-5050-01			MICROCKT,DGTL:HCMOS,ANALOG SW,8 CHAN	80009	156-5050-01
A5U2510	156-5000-01			MICROCKT,LINEAR:VOLTAGE COMPARATOR	80009	156-5000-01
A5U2520	156-5138-01			MICROCKT,LINEAR:OP AMP,BIFET,DUAL	80009	156-5138-01
A5U2521	156-5050-01			MICROCKT,DGTL:HCMOS,ANALOG SW,8 CHAN	80009	156-5050-01
A5U2530	156-5050-01			MICROCKT,DGTL:HCMOS,ANALOG SW,8 CHAN	80009	156-5050-01
A5U2540	156-5081-01			MICROCKT,DGTL:HCMOS,HEX INVERTER	80009	156-5081-01
A5U2550	156-5088-01			MICROCKT,DGTL:CMOS,3 TO 8 DECODER/	80009	156-5088-01
A5U2560	156-5145-01			MICROCKT,DGTL:HCMOS,DUAL D-TYPE FF	80009	156-5145-01
A5U2570	156-5145-01			MICROCKT,DGTL:HCMOS,DUAL D-TYPE FF	80009	156-5145-01
A5U2601	156-5050-01			MICROCKT,DGTL:HCMOS,ANALOG SW,8 CHAN	80009	156-5050-01
A5U2620	156-2051-01			MICROCKT,LINEAR:OPNL AMPL,QUAD,JET INPUT	80009	156-2051-01
A5U2630	156-2051-01			MICROCKT,LINEAR:OPNL AMPL,QUAD,JET INPUT	80009	156-2051-01
A5U2640	156-5567-01			MICROCKT,DGTL:CMOS,14 STAGES BIN CNTR	80009	156-5567-01
A5U2650	156-5088-01			MICROCKT,DGTL:CMOS,3 TO 8 DECODER/	80009	156-5088-01
A5U2660	156-5088-01			MICROCKT,DGTL:CMOS,3 TO 8 DECODER/	80009	156-5088-01
A5U2800	156-5120-01			MICROCKT,DGTL:CMOS,DUAL 4 CHAN ANALOG MUX	80009	156-5120-01
A5U2805	156-5120-01			MICROCKT,DGTL:CMOS,DUAL 4 CHAN ANALOG MUX	80009	156-5120-01
A5U2810	156-5098-01			MICROCKT,DGTL:HCMOS,QUAD 2-INPUT NAND GATE	80009	156-5098-01
A5U2820	156-2051-01			MICROCKT,LINEAR:OPNL AMPL,QUAD,JET INPUT	80009	156-2051-01
A5U2830	156-5306-01			MICROCKT,DGTL:CMOS,DUAL 4 BIT	80009	156-5306-01
A5U2835	156-5085-01			MICROCKT,DGTL:CMOS,QUAD 2-INPUT OR GATE	80009	156-5085-01
A5U2850	156-5145-01			MICROCKT,DGTL:HCMOS,DUAL D-TYPE FF	80009	156-5145-01
A5U2855	156-5106-01			MICROCKT,DGTL:CMOS,QUAD 2 INPUT N OR GATE	80009	156-5106-01
A5U2860	156-5569-01			MICROCKT,DGTL:CMOS,8-BIT UNIVERSIAL SHIFT	80009	156-5569-01
A5U2865	156-5021-01			MICROCKT,DGTL:CMOS,8 STATE SHIFT ANS STOR	80009	156-5021-01
A5U2870	156-5306-01			MICROCKT,DGTL:CMOS,DUAL 4 BIT	80009	156-5306-01
A5U2875	156-5145-01			MICROCKT,DGTL:HCMOS,DUAL D-TYPE FF	80009	156-5145-01
A5U2880	156-5145-01			MICROCKT,DGTL:HCMOS,DUAL D-TYPE FF	80009	156-5145-01
A5U2885	156-5130-01			MICROCKT,DGTL:CMOS,TRIPLE 3-INPUT N AND D	80009	156-5130-01
A5U2890	156-5098-01			MICROCKT,DGTL:HCMOS,QUAD 2-INPUT NAND GATE	80009	156-5098-01
A5U2900	156-5130-01			MICROCKT,DGTL:CMOS,TRIPLE 3-INPUT N AND D	80009	156-5130-01
A5U2905	156-5147-01			MICROCKT,DGTL:CMOS,OCTAL D TYPE FF W/RESET	80009	156-5147-01
A5U2910	156-1555-00			MICROCKT,LINEAR:D/A CONVERTER	34335	AM6080PC
A5U2920	156-5011-00	B050000	B051130	IC,MEMORY:CMOS,SRAM;8K X 8,150NS	62786	HM6264LFP-15
A5U2920	156-5011-01	B051131		IC,MEMORY:CMOS,SRAM;8K X 8,150NS	80009	156-5011-01
A5U2930	160-5875-00			MICROCKT,DGTL:8K X 8 EPROM,PRGM	80009	160-5875-00
A5U2935	156-5071-01			MICROCKT,DGTL:CMOS,OCTAL BUS TRANS	80009	156-5071-01
A5U2940	156-5306-01			MICROCKT,DGTL:CMOS,DUAL 4 BIT	80009	156-5306-01
A5U2950	156-5145-01			MICROCKT,DGTL:HCMOS,DUAL D-TYPE FF	80009	156-5145-01
A5U2960	156-5021-01			MICROCKT,DGTL:CMOS,8 STATE SHIFT ANS STOR	80009	156-5021-01
A5U2965	156-5098-01			MICROCKT,DGTL:HCMOS,QUAD 2-INPUT NAND GATE	80009	156-5098-01
A5U2970	156-5098-01			MICROCKT,DGTL:HCMOS,QUAD 2-INPUT NAND GATE	80009	156-5098-01
A5U2975	156-5098-01			MICROCKT,DGTL:HCMOS,QUAD 2-INPUT NAND GATE	80009	156-5098-01

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A5U2980	156-5098-01			MICROCKT,DGTL:HCMOS,QUAD 2-INPUT NAND GATE	80009	156-5098-01
A5U2985	156-5568-01			MICROCKT,DGTL:HCMOS,4-BIT BIDIRECTIONAL	80009	156-5568-01
A5U2990	156-5198-01			MICROCKT,DGTL:CMOS,QUAD 2-INPUT X OR GATE	80009	156-5198-01
A5U2995	156-5135-01			MICROCKT,DGTL:CMOS,8 BIT SER/PAR SHIFT	80009	156-5135-01
A5W411	174-1366-00			CA ASSY,SP,ELEC:26,28 AWG,3.0 L	TK1899	ORDER BY DESCR
A5W511	174-1501-00			CA ASSY,SP,ELEC:26,28 AWG,2.0 L,RIBBON	80009	174-1501-00
A5W512	174-1502-00			CA ASSY,SP,ELEC:34,28 AWG,2.0 L,RIBBON	80009	174-1502-00
A5XU2360	136-0755-00			SKT,PL-IN ELEK:MICROCIRCUIT,28 DIP	09922	DILB28P-108
A5Y2540	158-5005-00			OSC,XTAL CLOCK:10MHZ	80009	158-5005-00

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A6	614-0825-00			FRONT PNL ASSY:STANDARD,2445B/55B/65B & 67B (STANDARD)	80009	614-0825-00
A6	614-0826-00			FRONT PNL ASSY:TV OPTION,2445B/55B/65B/67B (OPTION 05)	80009	614-0826-00
A6P3001	131-3478-01			CONN,RCPT,ELEC:VERT,2 X 10,0.1 SPACING	80009	131-3478-01
A6R3007	311-2318-00			RES,VAR,NONWW:5K OHM,30%,0.5W	32997	ORDER BY DESCR
A6R3008	311-2316-00			RES,VAR,NONWW:2K OHM,20%,0.5W	32997	ORDER BY DESCR
A6R3009	311-2317-00			RES,VAR,NONWW:5K OHM,30%,0.25W	32997	ORDER BY DESCR
A6R3010	311-2318-00			RES,VAR,NONWW:5K OHM,30%,0.5W	32997	ORDER BY DESCR
A6R3011	311-2316-00			RES,VAR,NONWW:2K OHM,20%,0.5W	32997	ORDER BY DESCR
A6R3012	311-2317-00			RES,VAR,NONWW:5K OHM,30%,0.25W	32997	ORDER BY DESCR
A6R3013	311-2316-00			RES,VAR,NONWW:2K OHM,20%,0.5W	32997	ORDER BY DESCR
A6R3014	311-2318-00			RES,VAR,NONWW:5K OHM,30%,0.5W	32997	ORDER BY DESCR
A6R3015	311-2316-00			RES,VAR,NONWW:2K OHM,20%,0.5W	32997	ORDER BY DESCR
A6R3016	311-2316-00			RES,VAR,NONWW:2K OHM,20%,0.5W	32997	ORDER BY DESCR
A6R3017	311-2316-00			RES,VAR,NONWW:2K OHM,20%,0.5W	32997	ORDER BY DESCR
A6R3018	311-2318-00			RES,VAR,NONWW:5K OHM,30%,0.5W	32997	ORDER BY DESCR
A6R3019	311-2316-00			RES,VAR,NONWW:2K OHM,20%,0.5W	32997	ORDER BY DESCR

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A6A1	----- -----			CIRCUIT BD ASSY:FRONT PANEL (REPLACEABLE AT A6 LEVEL ONLY)		
A6A1C3001	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A6A1C3002	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A6A1C3019	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A6A1CR3001	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3002	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3003	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3004	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3005	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3006	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3007	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3008	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3009	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3010	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3011	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3012	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3013	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3014	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3015	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3016	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3017	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3018	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3019	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3020	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3021	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3022	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3023	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3024	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3025	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3026	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3027	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3028	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3029	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3030	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3031	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3032	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3033	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3034	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3035	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3036	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3037	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3038	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3039	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3040	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3041	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3042	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3043	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3044	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1DS3001	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3002	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3003	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3004	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3005	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3006	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3007	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3008	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3009	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A6A1DS3010	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3011	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3012	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3013	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3014	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3015	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3016	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3017	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3018	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3019	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3020	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3021	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3022	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3023	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3024	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3025	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3026	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3027	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3028	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3029	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3030	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3031	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3032	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3033	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3034	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3035	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3036	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3037	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3038	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3039	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3040	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3041	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3042	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3043	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3044	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3045	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3046	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3047	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3048	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3049	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1R3001	307-0486-00			RES NTWK,FXD,FI:100 OHM,20%,1.125W	11236	750-101-R100 OHM
A6A1R3002	307-0695-00			RES NTWK,FXD,FI:9,150 OHM,2%,0.2W EA	11236	750-101-R150 OHM
A6A1R3003	307-0486-00			RES NTWK,FXD,FI:100 OHM,20%,1.125W	11236	750-101-R100 OHM
A6A1R3004	313-1151-00			RES,FXD,FILM:150 OHM,5%,0.2W	57668	TR20JE150E
A6A1R3005	313-1151-00			RES,FXD,FILM:150 OHM,5%,0.2W	57668	TR20JE150E
A6A1R3006	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A6A1S3001	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3002	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3003	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3004	260-2283-00			SWITCH,ROTARY:VOLTS/DIV	80009	260-2283-00
A6A1S3005	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3006	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3007	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3008	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3009	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3010	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3011	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3012	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A6A1S3013	260-2283-00			SWITCH,ROTARY:VOLTS/DIV	80009	260-2283-00
A6A1S3014	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3015	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3016	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3017	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3018	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3019	260-2283-00			SWITCH,ROTARY:VOLTS/DIV	80009	260-2283-00
A6A1S3020	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3021	260-2164-01			SWITCH,SLIDE:SPDT,4A,20VAC	09353	1101 AV2 BE2
A6A1S3022	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3023	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3024	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3025	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3026	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3027	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3028	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3029	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3030	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3031	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3032	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3033	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3034	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3035	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1U3001	156-2120-00			MICROCKT,DGTL:SER-IN PRL-OUT SHIFT RGTR	80009	156-2120-00
A6A1U3002	156-2120-00			MICROCKT,DGTL:SER-IN PRL-OUT SHIFT RGTR	80009	156-2120-00
A6A1U3003	156-2120-00			MICROCKT,DGTL:SER-IN PRL-OUT SHIFT RGTR	80009	156-2120-00
A6A1U3004	156-2120-00			MICROCKT,DGTL:SER-IN PRL-OUT SHIFT RGTR	80009	156-2120-00
A6A1U3005	156-2120-00			MICROCKT,DGTL:SER-IN PRL-OUT SHIFT RGTR	80009	156-2120-00
A6A1U3006	156-2120-00			MICROCKT,DGTL:SER-IN PRL-OUT SHIFT RGTR	80009	156-2120-00

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A8	670-7280-00			CIRCUIT BD ASSY:SCALE ILLUM	80009	670-7280-00
A8DS100	150-0057-01			LAMP,INCAND:5V,0.115A,WIRE LD,AGED & SEL	71744	7153 AS 15
A8DS101	150-0057-01			LAMP,INCAND:5V,0.115A,WIRE LD,AGED & SEL	71744	7153 AS 15
A8DS102	150-0057-01			LAMP,INCAND:5V,0.115A,WIRE LD,AGED & SEL	71744	7153 AS 15

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A9	670-7277-09			CIRCUIT BD ASSY:HIGH VOLTAGE	80009	670-7277-09
A9C91	283-0084-00			CAP,FXD,CER DI:270PF,5%,1000V	59660	838533X5F02715
A9C1812	285-1430-00			CAP,FXD,PLASTIC:0.047UF,10%,400V	80009	285-1430-00
A9C1813	285-1430-00			CAP,FXD,PLASTIC:0.047UF,10%,400V	80009	285-1430-00
A9C1814	285-1430-00			CAP,FXD,PLASTIC:0.047UF,10%,400V	80009	285-1430-00
A9C1815	285-1430-00			CAP,FXD,PLASTIC:0.047UF,10%,400V	80009	285-1430-00
A9C1870	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A9C1885	285-1430-00			CAP,FXD,PLASTIC:0.047UF,10%,400V	80009	285-1430-00
A9C1886	285-1430-00			CAP,FXD,PLASTIC:0.047UF,10%,400V	80009	285-1430-00
A9C1888	285-1430-00			CAP,FXD,PLASTIC:0.047UF,10%,400V	80009	285-1430-00
A9C1889	285-1430-00			CAP,FXD,PLASTIC:0.047UF,10%,400V	80009	285-1430-00
A9C1890	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A9C1912	281-0798-00			CAP,FXD,CER DI:51PF,1%,100V	04222	MA101A510GAA
A9C1915	281-0783-00			CAP,FXD,CER DI:0.1 UF 20%,100V	04222	MA401C104MAA
A9C1932	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A9C1951	290-0269-01			CAP,FXD,ELCTLT:0.22UF,5%,35V,1KHZ,TANTULUM	56289	173D224X5035U
A9C1971	285-1430-00			CAP,FXD,PLASTIC:0.047UF,10%,400V	80009	285-1430-00
A9C1972	290-0747-00			CAP,FXD,ELCTLT:100UF,+50-20%,25WVDC	54473	ECE-B25V100L
A9C1973	281-0826-00			CAP,FXD,CER DI:2200PF,10%,100V	20932	401EM100AD222K
A9C1980	281-0771-00			CAP,FXD,CER DI:2200PF,20%,200V	04222	SA106E222MAA
A9C1990	285-1096-00			CAP,FXD,PLASTIC:1UF,10%,50V	14752	230B1A105K
A9C1991	281-0826-00			CAP,FXD,CER DI:2200PF,10%,100V	20932	401EM100AD222K
A9CR1894	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A9CR1895	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A9CR1915	152-0061-00			SEMICOND DVC,DI:SW,SI,175V,0.1A,DO-35	07263	FDH2161
A9CR1930	152-0061-00			SEMICOND DVC,DI:SW,SI,175V,0.1A,DO-35	07263	FDH2161
A9CR1950	152-0061-00			SEMICOND DVC,DI:SW,SI,175V,0.1A,DO-35	07263	FDH2161
A9CR1953	152-0061-00			SEMICOND DVC,DI:SW,SI,175V,0.1A,DO-35	07263	FDH2161
A9CR1990	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A9DS90	150-0030-00			LAMP,GLOW:60-90V MAX,0.6MA,A28-T,WIRE LEADS	58224	A2B-T
A9DS91	150-0030-00			LAMP,GLOW:60-90V MAX,0.6MA,A28-T,WIRE LEADS	58224	A2B-T
A9F1900	159-0185-00			FUSE,CARTRIDGE:5.2 X 20MM,0.75A,125V	TK0946	TSC-750MA
A9J901	131-0589-00			TERMINAL,PIN:0.46 L X 0.025 SQ,PH BRZ (QUANTITY OF 9)	22526	48283-029
A9J902	131-0589-00			TERMINAL,PIN:0.46 L X 0.025 SQ,PH BRZ (QUANTITY OF 2)	22526	48283-029
A9J903	131-0589-00			TERMINAL,PIN:0.46 L X 0.025 SQ,PH BRZ (QUANTITY OF 2)	22526	48283-029
A9J904	131-0589-00			TERMINAL,PIN:0.46 L X 0.025 SQ,PH BRZ	22526	48283-029
A9L1921	108-0262-00			COIL,RF:FIXED,505NH	80009	108-0262-00
A9L1974	108-0318-00			COIL,RF:FIXED,100UH	32159	81000M
A9P191	131-3461-00			HEADER,MICROCKT:14 PIN,0.5 L,GOLD PL (SUBPART OF A9W900)	80009	131-3461-00
A9P900	----- -----					
A9Q1851	151-0443-00			TRANSISTOR:PNP,SI,TO-92	04713	SPS7950
A9Q1852	151-0443-00			TRANSISTOR:PNP,SI,TO-92	04713	SPS7950
A9Q1890	151-0443-00			TRANSISTOR:PNP,SI,TO-92	04713	SPS7950
A9Q1891	151-0745-00			TRANSISTOR:PNP,SI,TO-220	61271	2SA1077G
A9Q1980	151-0444-00			TRANSISTOR:NPN,SI,TO-92	04713	SPS797
A9R1812	315-0100-02			RES,FXD,CMPSN:10 OHM,5%,0.25W	01121	CB1005
A9R1813	315-0100-02			RES,FXD,CMPSN:10 OHM,5%,0.25W	01121	CB1005
A9R1814	315-0103-00			RES,FXD,FILM:10K OHM,5%,0.25W	19701	5043CX10K00J
A9R1815	315-0103-00			RES,FXD,FILM:10K OHM,5%,0.25W	19701	5043CX10K00J
A9R1833	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A9R1834	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A9R1842	311-2234-00			RES,VAR,NONWW:TRMR,5K OHM,20%,0.5W LINEAR	TK1450	GF06UT 5K
A9R1848	311-2234-00			RES,VAR,NONWW:TRMR,5K OHM,20%,0.5W LINEAR	TK1450	GF06UT 5K
A9R1853	321-0447-00			RES,FXD,FILM:442K OHM,1%,0.125W,TC=T0	24546	NA55D4423F
A9R1854	321-0435-00			RES,FXD,FILM:332K OHM,1%,0.125W,TC=T0	07716	CEAD33202F

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A9R1855	321-0407-00			RES,FXD,FILM:169K OHM,1%,0.125W,TC=T0	07716	CEAD16902F
A9R1856	321-0367-00			RES,FXD,FILM:64.9K OHM,1%,0.125W,TC=T0	07716	CEAD64901F
A9R1857	321-0364-00			RES,FXD,FILM:60.4K OHM,1%,0.125W,TC=T0	19701	5043ED60K40F
A9R1858	313-1105-00			RES,FXD,FILM:1M OHM,5%,0.2W	57668	TR20JE1M
A9R1864	311-2236-00			RES,VAR,NONWW:TRMR,20K OHM,20%,0.5W LINEAR	TK1450	GF06UT 20K
A9R1870	311-2239-00			RES,VAR,NONWW:TRMR,100K OHM,20%,0.5W LINEAR	TK1450	GF06UT 100K
A9R1871	315-0154-00			RES,FXD,FILM:150K OHM,5%,0.25W	57668	NTR25J-E150K
A9R1872	315-0184-00			RES,FXD,FILM:180K OHM,5%,0.25W	19701	5043CX180K0J
A9R1873	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A9R1878	311-2239-00			RES,VAR,NONWW:TRMR,100K OHM,20%,0.5W LINEAR	TK1450	GF06UT 100K
A9R1880	315-0434-00			RES,FXD,FILM:430K OHM,5%,0.25W	57668	NTR25J-E430K
A9R1881	321-0385-00			RES,FXD,FILM:100K OHM,1%,0.125W,TC=T0	19701	5033ED100K0F
A9R1885	315-0103-00			RES,FXD,FILM:10K OHM,5%,0.25W	19701	5043CX10K00J
A9R1888	315-0100-02			RES,FXD,CMPSN:10 OHM,5%,0.25W	01121	CB1005
A9R1890	313-1473-00			RES,FXD,FILM:47K OHM,5%,0.2W	57668	TR20JE 47K
A9R1891	321-0481-04			RES,FXD,FILM:1M OHM,0.1%,0.125W,TC=T2	91637	CMF55116D10003B
A9R1892	321-0639-00			RES,FXD,FILM:9.6K OHM,1%,0.125W,TC=T0	19701	5043ED9K600F
A9R1892	321-0693-00			RES,FXD,FILM:68.1K OHM,0.5%,0.125W,TC=T0	19701	5033RD6812DB2980
A9R1893	321-0481-04			RES,FXD,FILM:1M OHM,0.1%,0.125W,TC=T2	91637	CMF55116D10003B
A9R1895	313-1302-00			RES,FXD,FILM:3K OHM,5%,0.2W	57668	TR20JE 03K0
A9R1896	315-0100-02			RES,FXD,CMPSN:10 OHM,5%,0.25W	01121	CB1005
A9R1897	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A9R1898	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A9R1901	315-0101-03			RES,FXD,CMPSN:100 OHM,5%,0.25W	01121	CB1015
A9R1910	321-0271-00			RES,FXD,FILM:6.49K OHM,1%,0.125W,TC=T0	07716	CEAD64900F
A9R1911	321-0245-00			RES,FXD,FILM:3.48K OHM,1%,0.125W,TC=T0	19701	5033ED3K48F
A9R1913	315-0101-03			RES,FXD,CMPSN:100 OHM,5%,0.25W	01121	CB1015
A9R1920	315-0152-00			RES,FXD,FILM:1.5K OHM,5%,0.25W	57668	NTR25J-E01K5
A9R1922	315-0331-03			RES,FXD,CMPSN:330 OHM 5%,0.25W	01121	CB3315
A9R1941	313-1201-00			RES,FXD,FILM:200 OHM,5%,0.2W	57668	TR20JE200E
A9R1944	321-0306-00			RES,FXD,FILM:15.0K OHM,1%,0.125W,TC=T0	19701	5033ED15J00F
A9R1945	321-0963-07			RES,FXD,FILM:98.73K OHM,0.1%,0.125W,TC=T9	07716	CEA 98.73KOHM 1%
A9R1950	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A9R1951	313-1220-00			RES,FXD,FILM:22 OHM,5%,0.2W	57668	TR20JE22E
A9R1952	313-1202-00			RES,FXD,FILM:2K OHM,5%,0.2W	57668	TR20JE02K0
A9R1953	313-1393-00			RES,FXD,FILM:39K OHM,5%,0.2W	57668	TR20JE 39K
A9R1971	313-1202-00			RES,FXD,FILM:2K OHM,5%,0.2W	57668	TR20JE02K0
A9R1972	313-1224-00			RES,FXD,FILM:220K,5%,0.2W	57668	TR20JE 220K
A9R1973	313-1124-00			RES,FXD,FILM:120K OHM,5%,0.2W	57668	TR20JE120K
A9R1990	321-0693-00			RES,FXD,FILM:68.1K OHM,0.5%,0.125W,TC=T0	19701	5033RD6812DB2980
A9R1991	315-0107-00			RES,FXD,FILM:100M OHM,5%,0.25W	01121	CB1075
A9R1992	313-1394-00			RES,FXD,FILM:390K,5%,0.2W	57668	TR20JE 390K
A9R1994	321-0402-00			RES,FXD,FILM:150K OHM,1%,0.125W,TC=T0	19701	5033ED150K0F
A9T1970	120-1418-01			XFMR,PWR,SDN&SU:HIGH VOLTAGE	80009	120-1418-01
A9U1062	156-0411-02			MICROCKT,LINEAR:QUAD COMPARATOR,SCREENED	04713	LM339JDS
A9U1830	152-0805-00			SEMICOND DVC,DI:HV MULTR,4.67KV INPUT,+14KV	S4431	MSR8506
A9U1890	156-1191-01			MICROCKT,LINEAR:BIFET,DUAL OPNL AMPL,SCRN	80009	156-1191-01
A9U1956	156-0158-07			MICROCKT,LINEAR:DUAL OPNL AMPL,SCREENED	01295	MC1458JG4
A9VR1891	152-0282-00			SEMICOND DVC,DI:ZEN,SI,30V,2%,400MW,DO-35	14552	1N972B
A9W900	198-4603-01			WIRE SET,ELEC:W/CRT SOCKET	80009	198-4603-01
A9W1909	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A14	670-8000-00			CIRCUIT BD ASSY:DYNAMIC CENTERING	80009	670-8000-00
A14C3401	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A14J141	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 5)	22526	48283-036
A14R3401	311-2234-00			RES,VAR,NONWW:TRMR,5K OHM,20%,0.5W LINEAR	TK1450	GF06UT 5K
A14R3402	313-1222-00			RES,FXD,FILM:2.2K OHM,5%,0.2W	57668	TR20JE 02K2
A14R3403	313-1750-00			RES,FXD,FILM:75 OHM,5%,0.2W	57668	TR20JE 75E
A14R3404	321-0284-00			RES,FXD,FILM:8.87K OHM,1%,0.125W,TC=T0	19701	5043ED8K870F
A14R3405	313-1750-00			RES,FXD,FILM:75 OHM,5%,0.2W	57668	TR20JE 75E
A14R3407	311-2234-00			RES,VAR,NONWW:TRMR,5K OHM,20%,0.5W LINEAR	TK1450	GF06UT 5K
A14R3408	321-0284-00			RES,FXD,FILM:8.87K OHM,1%,0.125W,TC=T0	19701	5043ED8K870F
A14R3409	313-1222-00			RES,FXD,FILM:2.2K OHM,5%,0.2W	57668	TR20JE 02K2
A14R3410	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A14R3411	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A14U3401	156-0130-00			MICROCKT,LINEAR:MODULATOR/DEMODULATOR	80009	156-0130-00
A14U3402	156-0130-00			MICROCKT,LINEAR:MODULATOR/DEMODULATOR	80009	156-0130-00
A14VR3401	152-0227-00			SEMICOND DVC,DI:ZEN,SI,6.2V,5%,0.4W,DO-7	04713	SZ13903

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
C10	281-0697-00			CAP,FXD,CER DI:5000PF,+100-0%,100V	72982	2425-003W5W0502Z
F90	159-0021-00			FUSE,CARTRIDGE:3AG,2A,250V,FAST BLOW	71400	AGC-CW-2
L90	119-1478-01			COIL,TUBE DEFL:FXD,TRACE ROTATION	80009	119-1478-01
R134	311-2312-01			RES,VAR,NONWW:PNL,5K OHM,20%,0.5W	80009	311-2312-01
R351	311-2312-01			RES,VAR,NONWW:PNL,5K OHM,20%,0.5W	80009	311-2312-01
R352	311-2312-01			RES,VAR,NONWW:PNL,5K OHM,20%,0.5W	80009	311-2312-01
R975	311-2312-01			RES,VAR,NONWW:PNL,5K OHM,20%,0.5W	80009	311-2312-01
R976	311-2312-01			RES,VAR,NONWW:PNL,5K OHM,20%,0.5W	80009	311-2312-01
R977	311-2312-01			RES,VAR,NONWW:PNL,5K OHM,20%,0.5W	80009	311-2312-01
S90	260-1967-00			SWITCH,SLIDE:DPDT 5A/250V 10A/125V MKD	TK0935	4021.0512
V900	154-0850-01			CRT ASSEMBLY:FINISHED 2445	80009	154-0850-01

REPLACEABLE ELECTRICAL PARTS

PARTS ORDERING INFORMATION

Replacement parts are available from or through your local Tektronix, Inc. Field Office or representative.

Changes to Tektronix instruments are sometimes made to accommodate improved components as they become available, and to give you the benefit of the latest circuit improvements developed in our engineering department. It is therefore important, when ordering parts, to include the following information in your order: Part number, instrument type or number, serial number, and modification number if applicable.

If a part you have ordered has been replaced with a new or improved part, your local Tektronix, Inc. Field Office or representative will contact you concerning any change in part number.

Change information, if any, is located at the rear of this manual.

LIST OF ASSEMBLIES

A list of assemblies can be found at the beginning of the Electrical Parts List. The assemblies are listed in numerical order. When the complete component number of a part is known, this list will identify the assembly in which the part is located.

CROSS INDEX-MFR. CODE NUMBER TO MANUFACTURER

The Mfr. Code Number to Manufacturer index for the Electrical Parts List is located immediately after this page. The Cross Index provides codes, names and addresses of manufacturers of components listed in the Electrical Parts List.

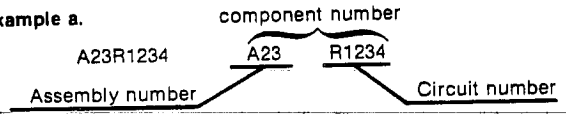
ABBREVIATIONS

Abbreviations conform to American National Standard Y1.1.

COMPONENT NUMBER (column one of the Electrical Parts List)

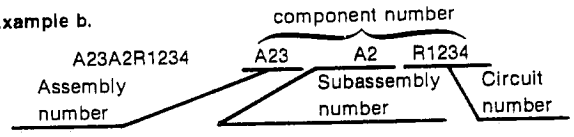
A numbering method has been used to identify assemblies, subassemblies and parts. Examples of this numbering method and typical expansions are illustrated by the following:

Example a.



Read: Resistor 1234 of Assembly 23

Example b.



Read: Resistor 1234 of Subassembly 2 of Assembly 23

Only the circuit number will appear on the diagrams and circuit board illustrations. Each diagram and circuit board illustration is clearly marked with the assembly number. Assembly numbers are also marked on the mechanical exploded views located in the Mechanical Parts List. The component number is obtained by adding the assembly number prefix to the circuit number.

The Electrical Parts List is divided and arranged by assemblies in numerical sequence (e.g., assembly A1 with its subassemblies and parts, precedes assembly A2 with its subassemblies and parts).

Chassis-mounted parts have no assembly number prefix and are located at the end of the Electrical Parts List.

TEKTRONIX PART NO. (column two of the Electrical Parts List)

Indicates part number to be used when ordering replacement part from Tektronix.

SERIAL/MODEL NO. (columns three and four of the Electrical Parts List)

Column three (3) indicates the serial number at which the part was first used. Column four (4) indicates the serial number at which the part was removed. No serial number entered indicates part is good for all serial numbers.

NAME & DESCRIPTION (column five of the Electrical Parts List)

In the Parts List, an Item Name is separated from the description by a colon (:). Because of space limitations, an Item Name may sometimes appear as incomplete. For further Item Name identification, the U.S. Federal Cataloging Handbook H6-1 can be utilized where possible.

MFR. CODE (column six of the Electrical Parts List)

Indicates the code number of the actual manufacturer of the part. (Code to name and address cross reference can be found immediately after this page.)

MFR. PART NUMBER (column seven of the Electrical Parts List)

Indicates actual manufacturers part number.

CROSS INDEX - MFR. CODE NUMBER TO MANUFACTURER

Mfr. Code	Manufacturer	Address	City, State, Zip Code
00213	NYTRONICS COMPONENTS GROUP INC SUBSIDIARY OF NYTRONICS INC	ORANGE ST	DARLINGTON SC 29532
00779	AMP INC	2800 FULLING MILL PO BOX 3608	HARRISBURG PA 17105
01121	ALLEN-BRADLEY CO	1201 S 2ND ST	MILWAUKEE WI 53204-2410
01295	TEXAS INSTRUMENTS INC SEMICONDUCTOR GROUP	13500 N CENTRAL EXPY PO BOX 655012	DALLAS TX 75265
02113	COILCRAFT INC	1102 SILVER LAKE RD	CARY IL 60013-1658
02735	RCA CORP SOLID STATE DIVISION	ROUTE 202	SOMERVILLE NJ 08876
03508	GENERAL ELECTRIC CO SEMI-CONDUCTOR PRODUCTS DEPT	W GENESEE ST	AUBURN NY 13021
04222	AVX CERAMICS DIV OF AVX CORP	19TH AVE SOUTH P O BOX 867	MYRTLE BEACH SC 29577
04713	MOTOROLA INC SEMICONDUCTOR PRODUCTS SECTOR	5005 E MCDOWELL RD	PHOENIX AZ 85008-4229
05292	ITT COMPONENTS DIV		CLIFTON NJ
05397	UNION CARBIDE CORP MATERIALS SYSTEMS DIV	11901 MADISON AVE	CLEVELAND OH 44101
05828	GENERAL INSTRUMENT CORP GOVERNMENT SYSTEMS DIV	600 W JOHN ST	HICKSVILLE NY 11802
06665	PRECISION MONOLITHICS INC SUB OF BOURNS INC	1500 SPACE PARK DR	SANTA CLARA CA 95050
07263	FAIRCHILD SEMICONDUCTOR CORP NORTH AMERICAN SALES SUB OF SCHLUMBERGER LTD MS 118	10400 RIDGEVIEW CT	CUPERTINO CA 95014
07716	TRW INC TRW IRC FIXED RESISTORS/BURLINGTON	2850 MT PLEASANT AVE	BURLINGTON IA 52601
09019	GENERAL ELECTRIC CO POWER ELECTRONICS SYSTEMS DEPT	ELECTRONICS PARK BLDG 7	SYRACUSE NY 13221
09353	C AND K COMPONENTS INC	15 RIVERDALE AVE	NEWTON MA 02158-1057
09922	BURNDY CORP	RICHARDS AVE	NORWALK CT 06852
11236	CTS CORP BERNE DIV THICK FILM PRODUCTS GROUP	406 PARR ROAD	BERNE IN 46711-9506
12954	MICROSEMI CORP - SCOTTSDALE	8700 E THOMAS RD P O BOX 1390	SCOTTSDALE AZ 85252
12969	UNITRODE CORP	5 FORBES RD	LEXINGTON MA 02173-7305
14301	ANDERSON ELECTRONICS INC	310 PENN ST PO BOX 89	HOLLIDAYSBURG PA 16648-2009
14433	ITT SEMICONDUCTORS DIV		WEST PALM BEACH FL
14552	MICROSEMI CORP	2830 S FAIRVIEW ST	SANTA ANA CA 92704-5948
14674	CORNING GLASS WORKS	HOUGHTON PK	CORNING NY 14830
14752	ELECTRO CUBE INC	1710 S DEL MAR AVE	SAN GABRIEL CA 91776-3825
15454	KETMA RODAN DIVISION	2900 BLUE STAR STREET	ANAHEIM CA 92806-2591
18324	SIGNETICS CORP MILITARY PRODUCTS DIV	4130 S MARKET COURT	SACRAMENTO CA 95834-1222
19613	MINNESOTA MINING AND MFG CO TEXTOOL PRODUCTS DEPT ELECTRONIC PRODUCT DIV	1410 E PIONEER DR	IRVING TX 75061-7847
19701	MEPCO/CENTRALAB A NORTH AMERICAN PHILIPS CO MINERAL WELLS AIRPORT	PO BOX 760	MINERAL WELLS TX 76067-0760
20462	PREM MAGNETICS INC	3519 N CHAPEL HILL	MCHENRY IL 60050-2504
20932	KYOCERA INTERNATIONAL INC	11620 SORRENTO VALLEY RD PO BOX 81543 PLANT NO 1	SAN DIEGO CA 92121
22526	DU PONT E I DE NEMOURS AND CO INC DU PONT CONNECTOR SYSTEMS DIV MILITARY PRODUCTS GROUP	515 FISHING CREEK RD	NEW CUMBERLAND PA 17070-3007
24226	GOWANDA ELECTRONICS CORP	NO 1 INDUSTRIAL PL	GOWANDA NY 14070-1409
24546	CORNING GLASS WORKS	550 HIGH ST	BRADFORD PA 16701-3737
25088	SIEMENS CORP	186 WOOD AVE S	ISELIN NJ 08830-2704
27264	MOLEX INC	2222 WELLINGTON COURT	LISLE IL 60532-1613
31471	AMERICAN MICRO SYSTEMS INC	3800 HOMESTEAD RD	SANTA CLARA CA 95051-4542
31918	ITT SCHADOW INC	8081 WALLACE RD	EDEN PRAIRIE MN 55344-2224

CROSS INDEX - MFR. CODE NUMBER TO MANUFACTURER

Mfr. Code	Manufacturer	Address	City, State, Zip Code
32159	WEST-CAP ARIZONA SUB OF SFE TECHNOLOGIES	2201 E ELVIRA ROAD	TUCSON AZ 85706-7026
32997	BOURNS INC TRIMPOT DIV	1200 COLUMBIA AVE	RIVERSIDE CA 92507-2114
34335	ADVANCED MICRO DEVICES	901 THOMPSON PL	SUNNYVALE CA 94086-4518
34479	RENCO CORP	26 COROMAR DRIVE	GOLETA CA 93117-3024
34899	FAIR-RITE PRODUCTS CORP	1 COMMERCIAL ROW	WALLKILL NY 12589
50434	HEWLETT-PACKARD CO OPTOELECTRONICS DIV	370 W TRIMBLE RD	SAN JOSE CA 95131
51406	MURATA ERIE NORTH AMERICA INC HEADQUARTERS AND GEORGIA OPERATIONS	2200 LAKE PARK DR	SMYRNA GA 30080
52769	SPRAGUE-GOODMAN ELECTRONICS INC	134 FULTON AVE	GARDEN CITY PARK NY 11040-5352
53387	MINNESOTA MINING MFG CO 3M ELECTRONIC PRODUCTS DIV	3M CENTER	ST PAUL MN 55101-1428
54473	MATSUSHITA ELECTRIC CORP OF AMERICA	ONE PANASONIC WAY PO BOX 1501	SECAUCUS NJ 07094-2917
54583	TDK ELECTRONICS CORP	12 HARBOR PARK DR	PORT WASHINGTON NY 11550
54937	DEYOUNG MANUFACTURING INC	12920 NE 125TH WAY	KIRKLAND WA 98034-7716
55112	WESTLAKE CAPACITORS INC	5334 STERLING CENTER DRIVE	WESTLAKE VILLAGE CA 91361
55680	NICHICON /AMERICA/ CORP	927 E STATE PKY	SCHAUMBURG IL 60195-4526
56289	SPRAGUE ELECTRIC CO WORLD HEADQUARTERS	92 HAYDEN AVE	LEXINGTON MA 02173-7929
56845	DALE ELECTRONICS INC	2300 RIVERSIDE BLVD PO BOX 74	NORFOLK NE 68701-2242
57668	ROHM CORP	8 WHATNEY	IRVINE CA 92713
58224	XENELL CORP	PO BOX 19515 11 DUNBARTON RD	CHERRY HILL NJ 08003-2107
59660	TUSONIX INC	PO BOX 4401 7741 N BUSINESS PARK DR	TUCSON AZ 85740-7144
59821	MEPCO/CENTRALAB A NORTH AMERICAN PHILIPS CO	PO BOX 37144 7158 MERCHANT AVE	EL PASO TX 79915-1207
61271	FUJITSU MICROELECTRONICS INC	2985 KIFER RD	SANTA CLARA CA 95051-0802
62786	HITACHI AMERICA LTD	1800 BERING DRIVE	SAN JOSE CA 95122
65786	CYPRESS SEMICONDUCTOR CORP	3901 N 1ST ST	SAN JOSE CA 95134-1506
71400	BUSSMANN DIV OF COOPER INDUSTRIES INC	114 OLD STATE RD PO BOX 14460	ST LOUIS MO 63178
71744	GENERAL INSTRUMENT CORP LAMP DIV/WORLD WIDE/	4433 N RAVENSWOOD AVE	CHICAGO IL 60640-5802
72982	ERIE SPECIALTY PRODUCTS INC	645 W 11TH ST	ERIE PA 16512
73138	BECKMAN INDUSTRIAL CORP BECKMAN ELECTRONIC TECHNOLOGIES SUB OF EMERSON ELECTRIC	4141 PALM ST	FULLERTON CA 92635
75042	IRC ELECTRONIC COMPONENTS PHILADELPHIA DIV TRW FIXED RESISTORS	401 N BROAD ST	PHILADELPHIA PA 19108-1001
80009	TEKTRONIX INC	14150 SW KARL BRAUN DR PO BOX 500	BEAVERTON OR 97077-0001
81483	INTERNATIONAL RECTIFIER	9220 SUNSET BLVD	LOS ANGELES CA 90069-3501
81855	EAGLE-PICHER INDUSTRIES INC ELECTRONICS DIV	COUPLES DEPT C AND PORTER STS PO BOX 47	JOPLIN MO 64801
91637	DALE ELECTRONICS INC	2064 12TH AVE PO BOX 609	COLUMBUS NE 68601-3632
93410	ESSEX GROUP ING CONTROLS DIV LEXINGTON PLANT	45-55 PLYMOUTH ST P O BOX 1007	LEXINGTON OH 44904
S4431	MURATA MFG CO LTD	16 KAIDEN NISHIJM CHO NAGAOKAKY-CITY	KYOTO JAPAN
TK0515	ERICSSON COMPONENTS INC	403 INTERNATIONAL PKY PO BOX 853904	RICHARDSON TX 75085-3904
TK0935	MARQUARDT SWITCHES INC	67 ALBANY ST PO BOX 465	CAZENOVIA NY 13035-1219
TK0946	SAN-O INDUSTRIAL CORP	170 WILBUR PL	BAHEMIA LONG ISLAND NY 11716
TK0961	NEC ELECTRONICS USA INC ELECTRON DIV	401 ELLIS ST PO BOX 7241	MOUNTAIN VIEW CA 94039
TK1345	ZMAN AND ASSOCIATES	7633 S 180TH	KENT WA 98032

CROSS INDEX - MFR. CODE NUMBER TO MANUFACTURER

Mfr. Code	Manufacturer	Address	City, State, Zip Code
TK1450	TOKYO COSMOS ELECTRIC CO LTD	2-268 SOBUDAI ZAWA	KANAGAWA 228 JAPAN
TK1492	COFER COMPONENT PROCESSING	3270 KELLER ST UNIT 11	SANTA CLARA CA 95050
TK1544	COMPUTER CONNECTIONS	30608 SAN ANTONIO ST	HAYWARD CA 94544
TK1573	WILHELM WESTERMAN	PO BOX 2345 AUGUSTA-ANLAGE 56	6800 MANNHEIM 1 WEST GERMANY
TK1727	PHILIPS NEDERLAND BV AFD ELONCO	POSTBUS 90050	5600 PB EINDHOVEN THE NETHERLANDS
TK1899	MINNESOTA MINING AND MFG CO	5400 RT B PO BOX 1228	COLUMBIA MO 65205
TK2042	ZMAN & ASSOCIATES	7633 S 180TH	KENT WA 98032
TK2282	KYOCERA AMERICA INC	5701 E FOURTH PLAIN BLVD	VANCOUVER WA 98661

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A1	671-0721-00			CIRCUIT BD ASSY:MAIN	80009	671-0721-00
A2	672-1037-12			CIRCUIT BD ASSY:LV PWR SPLY MODULE	80009	672-1037-12
A2A1	----- -----			CIRCUIT BD ASSY:REGULATOR (AVAILABLE AT THE 672-1037-XX LEVEL ONLY)		
A3	----- -----			CIRCUIT BD ASSY:INVERTER (AVAILABLE AT THE 672-1037-XX LEVEL ONLY)		
A4	670-9493-02			CIRCUIT BD ASSY:READOUT	80009	670-9493-02
A5	670-9052-02	121001	152999	CIRCUIT BD ASSY:DIGITAL CONTROL	80009	670-9052-02
A5	671-0965-00	152001		CIRCUIT BD ASSY:CONTROL/READOUT/BUFFER	80009	671-0965-00
A6	614-0825-00			FRONT PNL ASSY:STANDARD,2445B/55B/65B & 67B (STANDARD)	80009	614-0825-00
A6	614-0826-00			FRONT PNL ASSY:TV OPTION,2445B/55B/65B/67B (OPTION 05)	80009	614-0826-00
A6A1	----- -----			CIRCUIT BD ASSY:FRONT PANEL (REPLACEABLE AT A6 LEVEL ONLY)		
A8	670-7280-00			CIRCUIT BD ASSY:SCALE ILLUM	80009	670-7280-00
A9	670-7277-09			CIRCUIT BD ASSY:HIGH VOLTAGE	80009	670-7277-09
A14	670-8000-00			CIRCUIT BD ASSY:DYNAMIC CENTERING	80009	670-8000-00

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A1	671-0721-00			CIRCUIT BD ASSY:MAIN	80009	671-0721-00
A1A11	119-2342-05			ATTENUATOR,VAR:PROGRAMMABLE 1X-100X	80009	119-2342-05
A1A12	119-2342-06			ATTENUATOR,VAR:PROGRAMMABLE 1X-100X	80009	119-2342-06
A1C100	283-0000-00			CAP,FXD,CER DI:0.001UF,+100-0%,500V	59660	831-610-Y5U0102P
A1C102	290-0973-00			CAP,FXD,ELCTLT:100UF,20%,25VDC	55680	ULB1E101MPA
A1C103	281-0812-00			CAP,FXD,CER DI:1000PF,10%,100V	04222	MA101C102KAA
A1C105	281-0064-00			CAP,VAR,PLASTIC:0.25-1.5PF,600V	52769	ER-530-013
A1C106	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C107	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C108	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C109	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C110	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C113	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C114	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C115	281-0761-00			CAP,FXD,CER DI:27PF,5%,100V	04222	MA101A270JAA
A1C116	281-0814-00			CAP,FXD,CER DI:100 PF,10%,100V	04222	MA101A101KAA
A1C117	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C118	281-0205-00			CAP,VAR,PLASTIC:5.5-65 PF,100V	TK1727	2222-808-32659
A1C119	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C120	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C121	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C125	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C130	290-0776-01			CAP,FXD,ELCTLT:22UF,20%,10WVDC	55680	ULB1A220MAA1TD
A1C152	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C154	281-0812-00			CAP,FXD,CER DI:1000PF,10%,100V	04222	MA101C102KAA
A1C171	281-0851-00			CAP,FXD,CER DI:180PF,5%,100VDC	04222	MA101A181JAA
A1C175	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A1C176	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A1C177	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A1C179	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A1C180	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A1C181	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A1C182	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A1C183	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A1C184	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C185	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C200	283-0000-00			CAP,FXD,CER DI:0.001UF,+100-0%,500V	59660	831-610-Y5U0102P
A1C202	281-0812-00			CAP,FXD,CER DI:1000PF,10%,100V	04222	MA101C102KAA
A1C203	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A1C205	281-0064-00			CAP,VAR,PLASTIC:0.25-1.5PF,600V	52769	ER-530-013
A1C207	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C209	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C210	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C211	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C217	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C218	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C219	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C220	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C221	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C223	281-0812-00			CAP,FXD,CER DI:1000PF,10%,100V	04222	MA101C102KAA
A1C225	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C301	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C302	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C307	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C310	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C311	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C325	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C329	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA

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A1C332	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A1C336	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C351	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C402	281-0762-00			CAP,FXD,CER DI:27PF,20%,100V	04222	MA101A270MAA
A1C403	281-0218-00			CAP,VAR,CER DI:1-5PF,+2 -2.5%,100V	59660	513-011A1-5
A1C404	281-0218-00			CAP,VAR,CER DI:1-5PF,+2 -2.5%,100V	59660	513-011A1-5
A1C412	281-0762-00			CAP,FXD,CER DI:27PF,20%,100V	04222	MA101A270MAA
A1C415	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C458	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C460	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C464	281-0763-00			CAP,FXD,CER DI:47PF,10%,100V	04222	MA101A470KAA
A1C466	281-0763-00			CAP,FXD,CER DI:47PF,10%,100V	04222	MA101A470KAA
A1C478	281-0759-00			CAP,FXD,CER DI:22PF,10%,100V	04222	MA101A220KAA
A1C480	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C487	281-0823-00			CAP,FXD,CER DI:470PF,10%,50V	04222	MA105A471KAA
A1C488	281-0814-00			CAP,FXD,CER DI:100 PF,10%,100V	04222	MA101A101KAA
A1C500	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C501	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C512	290-0246-00			CAP,FXD,ELCTLT:3.3UF,10%,15V	12954	D3R3EA15K1
A1C513	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A1C520	281-0814-00			CAP,FXD,CER DI:100 PF,10%,100V	04222	MA101A101KAA
A1C521	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C528	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C536	290-0246-00			CAP,FXD,ELCTLT:3.3UF,10%,15V	12954	D3R3EA15K1
A1C537	281-0812-00			CAP,FXD,CER DI:1000PF,10%,100V	04222	MA101C102KAA
A1C544	281-0814-00			CAP,FXD,CER DI:100 PF,10%,100V	04222	MA101A101KAA
A1C601	281-0270-00			CAP,VAR,CER DI:9-90PF,50V	51406	TZ03R900E
A1C617	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A1C625	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C645	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A1C650	281-0823-00			CAP,FXD,CER DI:470PF,10%,50V	04222	MA105A471KAA
A1C653	281-0811-00			CAP,FXD,CER DI:10PF,10%,100V	04222	MA101A100KAA
A1C669	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C675	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C707	281-0808-00			CAP,FXD,CER DI:7 PF,20%,100V	04222	MA101A7R04AA
A1C708	285-0676-01			CAP,FXD,PLASTIC:0.1UF,3.5%,35V	80009	285-0676-01
A1C709	285-1060-00			CAP,FXD,PLASTIC:10UF,3%,25V	80009	285-1060-00
A1C710	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C712	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A1C722	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C723	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C730	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C731	290-0944-01			CAP,FXD,ELCTLT:220UF,20%,10V	55680	UVX1C221MPA1TA
A1C732	290-0944-01			CAP,FXD,ELCTLT:220UF,20%,10V	55680	UVX1C221MPA1TA
A1C733	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C735	281-0823-00			CAP,FXD,CER DI:470PF,10%,50V	04222	MA105A471KAA
A1C738	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C740	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A1C742	281-0812-00			CAP,FXD,CER DI:1000PF,10%,100V	04222	MA101C102KAA
A1C744	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C755	281-0759-00			CAP,FXD,CER DI:22PF,10%,100V	04222	MA101A220KAA
A1C803	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C805	281-0823-00			CAP,FXD,CER DI:470PF,10%,50V	04222	MA105A471KAA
A1C806	283-0156-00			CAP,FXD,CER DI:1000PF,+80-20%,200V	04222	SR152E102ZAA
A1C808	281-0757-00			CAP,FXD,CER DI:10PF,20%,100V TUBULAR,MI	04222	MA101A100MAA
A1C809	281-0819-00			CAP,FXD,CER DI:33 PF,5%,50V	04222	GC105A330J
A1C810	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C811	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T

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A1C817	281-0812-00			CAP,FXD,CER DI:1000PF,10%,100V	04222	MA101C102KAA
A1C819	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C822	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C823	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C830	281-0814-00			CAP,FXD,CER DI:100 PF,10%,100V	04222	MA101A101KAA
A1C848	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C849	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C850	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C851	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A1C852	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A1C853	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A1C854	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A1C900	281-0763-00			CAP,FXD,CER DI:47PF,10%,100V	04222	MA101A470KAA
A1C903	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C907	281-0808-00			CAP,FXD,CER DI:7 PF,20%,100V	04222	MA101A7R04AA
A1C908	285-0752-03			CAP,FXD,PLASTIC:1UF,3%,50V	80009	285-0752-03
A1C912	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C933	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C938	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C940	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C943	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C947	281-0759-00			CAP,FXD,CER DI:22PF,10%,100V	04222	MA101A220KAA
A1C957	290-0804-00			CAP,FXD,ELCTLT:10UF,+50-20%,25V	55680	ULB1E100TAAANA
A1C958	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C966	281-0783-00			CAP,FXD,CER DI:0.1 UF 20%,100V	04222	MA401C104MAA
A1C967	281-0783-00			CAP,FXD,CER DI:0.1 UF 20%,100V	04222	MA401C104MAA
A1C972	281-0756-00			CAP,FXD,CER DI:2.2PF,+/-0.5PF,200V	04222	SA102A2R2DAA
A1C973	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C975	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C976	283-1001-00			CAP,FXD,CER DI:0.03UF,50VDC	80009	283-1001-00
A1C977	290-0246-00			CAP,FXD,ELCTLT:3.3UF,10%,15V	12954	D3R3EA15K1
A1C980	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C981	283-1000-00			CAP,FXD,CER DI:0.02UF,50VDC	80009	283-1000-00
A1C982	281-0759-00			CAP,FXD,CER DI:22PF,10%,100V	04222	MA101A220KAA
A1C985	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A1C988	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C990	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A1C995	281-0810-00			CAP,FXD,CER DI:5.6PF,+/-0.5PF,100V	04222	MA101A5R6DAA
A1CR100	152-0323-01			SEMICOND DVC,DI:SW,SI,50V,25PA AT 20V,20PF	14552	MT5127
A1CR101	152-0323-01			SEMICOND DVC,DI:SW,SI,50V,25PA AT 20V,20PF	14552	MT5127
A1CR107	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A1CR130	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR131	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR140	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR141	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR142	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR143	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR144	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR145	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR146	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR147	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR148	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR149	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR150	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR151	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR152	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR153	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR154	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)

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A1CR155	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR161	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR162	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR163	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR180	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR181	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR200	152-0323-01			SEMICOND DVC,DI:SW,SI,50V,25PA AT 20V,20PF	14552	MT5127
A1CR201	152-0323-01			SEMICOND DVC,DI:SW,SI,50V,25PA AT 20V,20PF	14552	MT5127
A1CR354	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR360	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR460	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR461	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR476	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR484	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR485	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR495	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR503	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR538	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR539	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR600	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR601	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR616	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR619	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR620	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR621	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR652	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR653	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR707	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR741	152-0951-00			SEMICOND DVC,DI:SCHOTTKY,SI,60V,2.25PF	80009	152-0951-00
A1CR742	152-0951-00			SEMICOND DVC,DI:SCHOTTKY,SI,60V,2.25PF	80009	152-0951-00
A1CR746	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR747	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR752	152-0075-00			SEMICOND DVC,DI:SW,GE,22V,80MW,DO-7	80009	152-0075-00
A1CR753	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR807	152-0574-00			SEMICOND DVC,DI:SW,SI,120V,0.150MA,4NS,DO35	12969	NDP566
A1CR811	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR850	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR941	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR942	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR950	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR951	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR956	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A1CR966	152-0574-00			SEMICOND DVC,DI:SW,SI,120V,0.150MA,4NS,DO35	12969	NDP566
A1CR972	152-0574-00			SEMICOND DVC,DI:SW,SI,120V,0.150MA,4NS,DO35	12969	NDP566
A1CR987	152-0574-00			SEMICOND DVC,DI:SW,SI,120V,0.150MA,4NS,DO35	12969	NDP566
A1CR995	152-0061-00			SEMICOND DVC,DI:SW,SI,175V,0.1A,DO-35	07263	FDH2161
A1DL100	119-1490-03			DELAY LINE,ELEC:73NS,150 OHM	80009	119-1490-03
A1E900	276-0712-00			CORE,EM:BALUN,FERRITE	34899	2843002402
A1J1	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 3)	22526	48283-036
A1J9	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL	22526	48283-036
A1J11	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 3)	22526	48283-036
A1J100	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 2)	22526	48283-036
A1J101	131-3520-00			CONN,RCPT,ELEC:HEADER,10 CONT,STR SLDR PIN	53387	3591-6002
A1J102	131-3520-00			CONN,RCPT,ELEC:HEADER,10 CONT,STR SLDR PIN	53387	3591-6002
A1J103	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL	22526	48283-036

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
				(QUANTITY OF 2)		
A1J104	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL	22526	48283-036
				(QUANTITY OF 4)		
A1J105	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL	22526	48283-036
				(QUANTITY OF 2)		
A1J109	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL	22526	48283-036
				(QUANTITY OF 2)		
A1J120	131-3152-00			CONN,RCPT,ELEC:HEADER,2 X 8 0.1 SPACING	22526	66506-043
A1J181	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL	22526	48283-036
				(QUANTITY OF 2)		
A1J411	131-3362-00			CONN,RCPT,ELEC:HEADER,STR,26 PIN	53387	3593-6002
A1J511	131-3362-00			CONN,RCPT,ELEC:HEADER,STR,26 PIN	53387	3593-6002
A1J512	131-3364-00			CONN,RCPT,ELEC:HEADER,STRAIGHT,34 PIN	53387	3594-6002
A1J949	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL	22526	48283-036
				(QUANTITY OF 2)		
A1L101	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L107	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L113	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L115	108-0317-00			COIL,RF:FIXED,15 UH	32159	71501M+10PERCENT
A1L120	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L200	108-0509-00			COIL,RF:FIXED,2.45UH	TK2042	ORDER BY DESCR
A1L219	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L220	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L307	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L325	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L336	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L403	108-0552-00			COIL,RF:FIXED,80NH	TK1345	108-0552-00
A1L521	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L605	108-0170-01			COIL,RF:FIXED,360NH	TK2042	ORDER BY DESCR
A1L606	108-0736-00			COIL,RF:FIXED,828NH	TK2042	ORDER BY DESCR
A1L607	108-0736-00			COIL,RF:FIXED,828NH	TK2042	ORDER BY DESCR
A1L608	108-0170-01			COIL,RF:FIXED,360NH	TK2042	ORDER BY DESCR
A1L609	108-0509-00			COIL,RF:FIXED,2.45UH	TK2042	ORDER BY DESCR
A1L610	108-0509-00			COIL,RF:FIXED,2.45UH	TK2042	ORDER BY DESCR
A1L619	108-0736-00			COIL,RF:FIXED,828NH	TK2042	ORDER BY DESCR
A1L628	108-0327-00			COIL,RF:FIXED,48NH	TK2042	ORDER BY DESCR
A1L633	108-0327-00			COIL,RF:FIXED,48NH	TK2042	ORDER BY DESCR
A1L644	114-0353-00			COIL,RF:VARIABLE,0.6-1.0UH	24226	ORDER BY DESCR
A1L733	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L738	108-0317-00			COIL,RF:FIXED,15 UH	32159	71501M+10PERCENT
A1L740	108-0317-00			COIL,RF:FIXED,15 UH	32159	71501M+10PERCENT
A1L743	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L938	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L973	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1L980	108-1251-00			COIL,RF:FXD,2.7UH,10%	54583	SPT 0406-2R7K-6
A1LR101	108-0325-00			COIL,RF:FIXED,489NH	TK2042	ORDER BY DESCR
A1LR107	108-0325-00			COIL,RF:FIXED,489NH	TK2042	ORDER BY DESCR
A1LR180	108-0602-00			COIL,RF:FIXED,45NH	TK2042	ORDER BY DESCR
A1LR201	108-0325-00			COIL,RF:FIXED,489NH	TK2042	ORDER BY DESCR
A1LR218	108-0330-00			COIL,RF:FIXED,403NH	TK2042	ORDER BY DESCR
A1LR219	108-0330-00			COIL,RF:FIXED,403NH	TK2042	ORDER BY DESCR
A1LR280	108-0602-00			COIL,RF:FIXED,45NH	TK2042	ORDER BY DESCR
A1Q130	151-0622-00			TRANSISTOR:PNP,SI,40V,1A,TO-226AE/237	04713	SPS8956(MPSW51A)
A1Q131	151-0622-00			TRANSISTOR:PNP,SI,40V,1A,TO-226AE/237	04713	SPS8956(MPSW51A)
A1Q154	151-0188-00			TRANSISTOR:PNP,SI,TO-92	80009	151-0188-00
A1Q155	151-0188-00			TRANSISTOR:PNP,SI,TO-92	80009	151-0188-00
A1Q190	151-0190-00			TRANSISTOR:NPN,SI,TO-92	80009	151-0190-00
A1Q460	151-0198-01			TRANSISTOR:NPN,SI,TO-92 PLSTC	80009	151-0198-01
				(QUANTITY OF A & B)		

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A1Q550	151-0190-00			TRANSISTOR:NPN,SI,TO-92	80009	151-0190-00
A1Q600	151-0190-00			TRANSISTOR:NPN,SI,TO-92	80009	151-0190-00
A1Q623	151-0190-00			TRANSISTOR:NPN,SI,TO-92	80009	151-0190-00
A1Q624	151-1025-00			TRANSISTOR:FET,N-CHAN,SI,TO-92	04713	SPF3036
A1Q645	151-0188-00			TRANSISTOR:PNP,SI,TO-92	80009	151-0188-00
A1Q700	151-0190-00			TRANSISTOR:NPN,SI,TO-92	80009	151-0190-00
A1Q709	151-0736-00			TRANSISTOR:NPN,SI,TO-92	80009	151-0736-00
A1Q710	151-0736-00			TRANSISTOR:NPN,SI,TO-92	80009	151-0736-00
A1Q741	151-0190-00			TRANSISTOR:NPN,SI,TO-92	80009	151-0190-00
A1Q742	151-0190-00			TRANSISTOR:NPN,SI,TO-92	80009	151-0190-00
A1Q743	151-0188-00			TRANSISTOR:PNP,SI,TO-92	80009	151-0188-00
A1Q745	151-0188-00			TRANSISTOR:PNP,SI,TO-92	80009	151-0188-00
A1Q941	151-0188-00			TRANSISTOR:PNP,SI,TO-92	80009	151-0188-00
A1Q942	151-0188-00			TRANSISTOR:PNP,SI,TO-92	80009	151-0188-00
A1R100	315-0474-00			RES,FXD,FILM:470K OHM,5%,0.25W	19701	5043CX470K0J92U
A1R101	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A1R102	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A1R112	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R114	321-0130-03			RES,FXD,FILM:221 OHM,0.25%,0.125W,TC=T2 MI	91637	MFF1816D221R0C
A1R115	321-0146-00			RES,FXD,FILM:324 OHM,1%,0.125W,TC=T0	07716	CEAD324R0F
A1R117	321-0320-00			RES,FXD,FILM:21.0K OHM,1%,0.125W,TC=T0	19701	5033ED21K00F
A1R118	321-0212-00			RES,FXD,FILM:1.58K OHM,1%,0.125W,TC=T0	19701	5033ED1K58F
A1R121	313-1121-00			RES,FXD,FILM:120 OHM,5%,0.2W	80009	313-1121-00
A1R123	313-1622-00			RES,FXD,FILM:6.2K OHM,5%,0.2W	57668	TR20JE 06K2
A1R125	301-0361-00			RES,FXD,FILM:360 OHM,5%,0.5W	19701	5053CX360R0J
A1R129	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R130	313-1561-00			RES,FXD,FILM:560 OHM,5%,0.2W	57668	TR20JE 560E
A1R131	313-1561-00			RES,FXD,FILM:560 OHM,5%,0.2W	57668	TR20JE 560E
A1R133	322-3201-00			RES,FXD,FILM:1.21K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K21
A1R135	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R136	313-1622-00			RES,FXD,FILM:6.2K OHM,5%,0.2W	57668	TR20JE 06K2
A1R140	313-1471-00			RES,FXD,FILM:470 OHM,5%,0.2W	57668	TR20JE 470E
A1R141	313-1471-00			RES,FXD,FILM:470 OHM,5%,0.2W	57668	TR20JE 470E
A1R142	313-1391-00			RES,FXD,FILM:390 OHM,5%,0.2W	57668	TR20JE 390E
A1R143	313-1391-00			RES,FXD,FILM:390 OHM,5%,0.2W	57668	TR20JE 390E
A1R144	307-0108-00			RES,FXD,CMPSN:6.8 OHM,5%,0.25W	01121	CB68G5
A1R149	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R150	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R152	313-1242-00			RES,FXD,FILM:2.4K OHM,5%,0.2W	57668	TR20JE 02K4
A1R153	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R154	322-3242-00			RES,FXD,FILM:3.24K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 3K24
A1R155	321-0250-00			RES,FXD,FILM:3.92K OHM,1%,0.125W,TC=T0	07716	CEAD39200F
A1R156	322-3242-00			RES,FXD,FILM:3.24K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 3K24
A1R159	322-3242-00			RES,FXD,FILM:3.24K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 3K24
A1R161	322-3293-00			RES,FXD,FILM:11K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 11K0
A1R162	322-3293-00			RES,FXD,FILM:11K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 11K0
A1R163	322-3242-00			RES,FXD,FILM:3.24K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 3K24
A1R165	313-1822-00			RES,FXD,FILM:8.2K,OHM,5%,0.2W	57668	TR20JE 08K2
A1R180	322-3242-00			RES,FXD,FILM:3.24K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 3K24
A1R181	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R182	322-3242-00			RES,FXD,FILM:3.24K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 3K24
A1R183	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R190	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R191	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R192	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R193	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R194	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R195	322-3143-00			RES,FXD,FILM:301 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 301E

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A1R196	322-3277-00			RES,FXD,FILM:7.5K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 7K50
A1R197	322-3265-00			RES,FXD,FILM:5.62K OHM,1%,0.2W,TC=T0	80009	322-3265-00
A1R198	321-1700-04			RES,FXD,FILM:10.44K OHM,0.1%,0.125W,TC=T2	19701	5033RC10K440B
A1R199	321-1700-04			RES,FXD,FILM:10.44K OHM,0.1%,0.125W,TC=T2	19701	5033RC10K440B
A1R200	315-0474-00			RES,FXD,FILM:470K OHM,5%,0.25W	19701	5043CX470K0J92U
A1R201	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A1R202	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A1R216	313-1121-00			RES,FXD,FILM:120 OHM,5%,0.2W	80009	313-1121-00
A1R217	321-0320-00			RES,FXD,FILM:21.0K OHM,1%,0.125W,TC=T0	19701	5033ED21K00F
A1R218	321-0212-00			RES,FXD,FILM:1.58K OHM,1%,0.125W,TC=T0	19701	5033ED1K58F
A1R225	301-0361-00			RES,FXD,FILM:360 OHM,5%,0.5W	19701	5053CX360R0J
A1R230	322-3226-00			RES,FXD,FILM:2.21K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K21
A1R231	322-3226-00			RES,FXD,FILM:2.21K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K21
A1R232	322-3226-00			RES,FXD,FILM:2.21K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K21
A1R301	315-0180-00			RES,FXD,FILM:18 OHM,5%,0.25W	19701	5043CX18R00J
A1R302	315-0180-00			RES,FXD,FILM:18 OHM,5%,0.25W	19701	5043CX18R00J
A1R303	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R304	315-0101-00			RES,FXD,FILM:100 OHM,5%,0.25W	57668	NTR25J-E 100E
A1R311	315-0101-00			RES,FXD,FILM:100 OHM,5%,0.25W	57668	NTR25J-E 100E
A1R312	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R329	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R332	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R353	322-3239-00			RES,FXD,FILM:3.01K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 3K01
A1R361	322-3265-00			RES,FXD,FILM:5.62K OHM,1%,0.2W,TC=T0	80009	322-3265-00
A1R401	322-3202-00			RES,FXD,FILM:1.24K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K24
A1R402	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R403	311-0607-00			RES,VAR,NONWW:TRMR,10K OHM,0.5W	73138	82-25-2
A1R404	313-1200-00			RES,FXD,FILM:20 OHM,5%,0.2W	57668	TR20JE20E
A1R405	313-1200-00			RES,FXD,FILM:20 OHM,5%,0.2W	57668	TR20JE20E
A1R411	311-0635-00			RES,VAR,NONWW:TRMR,1K OHM,0.5W	32997	3329H-L58-102
A1R412	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R416	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R417	311-2234-00			RES,VAR,NONWW:TRMR,5K OHM,20%,0.5W LINEAR	TK1450	GF06UT 5K
A1R430	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R450	321-0310-00			RES,FXD,FILM:16.5K OHM,1%,0.125W,TC=T0	19701	5033ED16K50F
A1R451	321-0275-00			RES,FXD,FILM:7.15K OHM,1%,0.125W,TC=T0	07716	CEAD71500F
A1R452	321-0310-00			RES,FXD,FILM:16.5K OHM,1%,0.125W,TC=T0	19701	5033ED16K50F
A1R453	321-0275-00			RES,FXD,FILM:7.15K OHM,1%,0.125W,TC=T0	07716	CEAD71500F
A1R454	321-0310-00			RES,FXD,FILM:16.5K OHM,1%,0.125W,TC=T0	19701	5033ED16K50F
A1R455	321-0310-00			RES,FXD,FILM:16.5K OHM,1%,0.125W,TC=T0	19701	5033ED16K50F
A1R456	321-0333-00			RES,FXD,FILM:28.7K OHM,1%,0.125W,TC=T0	19701	5043ED28K70F
A1R457	321-0275-00			RES,FXD,FILM:7.15K OHM,1%,0.125W,TC=T0	07716	CEAD71500F
A1R458	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R459	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R460	321-0062-00			RES,FXD,FILM:43.2 OHM,0.5%,0.125W,TC=T0	57668	CRB14 FXE 43.2
A1R461	322-3139-00			RES,FXD,FILM:274 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 274E
A1R462	322-3201-00			RES,FXD,FILM:1.21K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K21
A1R463	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R464	321-0063-00			RES,FXD,FILM:44.2 OHM,0.5%,0.125W,TC=T0	91637	CMF55116G44R20F
A1R465	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R468	321-0287-00			RES,FXD,FILM:9.53K OHM,1%,0.125W,TC=T0	19701	5033ED9K530F
A1R469	313-1200-00			RES,FXD,FILM:20 OHM,5%,0.2W	57668	TR20JE20E
A1R470	322-3322-00			RES,FXD,FILM:22.1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 22K1
A1R471	322-3322-00			RES,FXD,FILM:22.1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 22K1
A1R473	313-1471-00			RES,FXD,FILM:470 OHM,5%,0.2W	57668	TR20JE 470E
A1R476	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R477	322-3258-00			RES,FXD,FILM:4.75K OHM,1%,0.2W,TC=T0	56845	ORDER BY DESCR
A1R478	321-0193-03			RES,FXD,FILM:1K OHM,0.25%,0.125W,TC=T2	07716	CEAC10000C

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A1R479	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R480	321-0375-00			RES,FXD,FILM:78.7K OHM,1%,0.125W,TC=T0	07716	CEAD78701F
A1R481	321-0347-00			RES,FXD,FILM:40.2K OHM,1%,0.125W,TC=T0	91637	CMF55116G40201F
A1R482	313-1471-00			RES,FXD,FILM:470 OHM,5%,0.2W	57668	TR20JE 470E
A1R483	321-0347-00			RES,FXD,FILM:40.2K OHM,1%,0.125W,TC=T0	91637	CMF55116G40201F
A1R484	322-3222-00			RES,FXD,FILM:2K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K00
A1R485	322-3222-00			RES,FXD,FILM:2K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K00
A1R486	321-0347-00			RES,FXD,FILM:40.2K OHM,1%,0.125W,TC=T0	91637	CMF55116G40201F
A1R487	321-0130-03			RES,FXD,FILM:221 OHM,0.25%,0.125W,TC=T2 MI	91637	MFF1816D221R0C
A1R488	321-1216-03			RES,FXD,FILM:1.76K OHM,0.25%,0.125W,TC=T2	24546	NC55C1761C
A1R489	321-1216-03			RES,FXD,FILM:1.76K OHM,0.25%,0.125W,TC=T2	24546	NC55C1761C
A1R490	321-0375-00			RES,FXD,FILM:78.7K OHM,1%,0.125W,TC=T0	07716	CEAD78701F
A1R491	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R492	321-0193-03			RES,FXD,FILM:1K OHM,0.25%,0.125W,TC=T2	07716	CEAC10000C
A1R493	322-3258-00			RES,FXD,FILM:4.75K OHM,1%,0.2W,TC=T0	56845	ORDER BY DESCR
A1R494	313-1201-00			RES,FXD,FILM:200 OHM,5%,0.2W	57668	TR20JE200E
A1R495	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R496	322-3293-00			RES,FXD,FILM:11K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 11K0
A1R497	313-1821-00			RES,FXD,FILM:820 OHM,5%,0.2W	57668	TR20JE 820E
A1R498	313-1821-00			RES,FXD,FILM:820 OHM,5%,0.2W	57668	TR20JE 820E
A1R501	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R502	313-1622-00			RES,FXD,FILM:6.2K OHM,5%,0.2W	57668	TR20JE 06K2
A1R503	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R504	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R511	321-0320-00			RES,FXD,FILM:21.0K OHM,1%,0.125W,TC=T0	19701	5033ED21K00F
A1R512	322-3293-00			RES,FXD,FILM:11K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 11K0
A1R513	313-1470-00			RES,FXD,FILM:47 OHM,5%,0.2W	57668	TR20JE 47E
A1R518	313-1680-00			RES,FXD,FILM:68 OHM,0.2W,5%	57668	TR20JT68 68E
A1R519	313-1621-00			RES,FXD,FILM:620 OHM,5%,0.2W	57668	TR20JE 620E
A1R520	313-1393-00			RES,FXD,FILM:39K OHM,5%,0.2W	57668	TR20JE 39K
A1R521	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R527	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R529	313-1561-00			RES,FXD,FILM:560 OHM,5%,0.2W	57668	TR20JE 560E
A1R537	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R538	313-1621-00			RES,FXD,FILM:620 OHM,5%,0.2W	57668	TR20JE 620E
A1R542	313-1680-00			RES,FXD,FILM:68 OHM,0.2W,5%	57668	TR20JT68 68E
A1R543	313-1621-00			RES,FXD,FILM:620 OHM,5%,0.2W	57668	TR20JE 620E
A1R544	313-1393-00			RES,FXD,FILM:39K OHM,5%,0.2W	57668	TR20JE 39K
A1R545	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R550	313-1471-00			RES,FXD,FILM:470 OHM,5%,0.2W	57668	TR20JE 470E
A1R551	321-1682-07			RES,FXD,FILM:5.7K OHM,0.1%,0.125W,TC=T9	19701	5033RE5K701B
A1R552	321-0641-07			RES,FXD,FILM:1.8K OHM,0.1,0.125W,TC=T9	07716	CEAE 18000B
A1R553	322-3210-00			RES,FXD,FILM:1.5K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K50
A1R554	322-3213-00			RES,FXD,FILM:1.62K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K62
A1R555	321-0294-00			RES,FXD,FILM:11.3K OHM,1%,0.125W,TC=T0	19701	5043ED11K30F
A1R556	322-3282-00			RES,FXD,FILM:8.45K OHM,1%,0.2W,TC=T0	80009	322-3282-00
A1R557	321-0808-07			RES,FXD,FILM:300 OHM,0.1%,0.125W,TC=T9	24546	NE55E3000B
A1R558	321-0657-07			RES,FXD,FILM:60 OHM,0.1%,0.125W,TC=T9	57668	RB14BZE 60E
A1R560	313-1621-00			RES,FXD,FILM:620 OHM,5%,0.2W	57668	TR20JE 620E
A1R600	313-1270-00			RES,FXD,FILM:27 OHM 5%,0.2W	57668	TR20JT68 27E
A1R601	313-1750-00			RES,FXD,FILM:75 OHM,5%,0.2W	57668	TR20JE 75E
A1R602	313-1470-00			RES,FXD,FILM:47 OHM,5%,0.2W	57668	TR20JE 47E
A1R605	311-2227-00			RES,VAR,NONWW:TRMR,100 OHM,20%,0.5W LINEAR	TK1450	GF06UT 100
A1R606	313-1100-00			RES,FXD,FILM:10 OHM,5%,0.2W	57668	TR20JE10E0
A1R607	313-1100-00			RES,FXD,FILM:10 OHM,5%,0.2W	57668	TR20JE10E0
A1R614	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R615	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R617	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A1R618	311-2234-00			RES,VAR,NONWW:TRMR,5K OHM,20%,0.5W LINEAR	TK1450	GF06UT 5K
A1R619	315-0510-00			RES,FXD,FILM:51 OHM,5%,0.25W	19701	5043CX51R00J
A1R620	322-3258-00			RES,FXD,FILM:4.75K OHM,1%,0.2W,TC=T0	56845	ORDER BY DESCR
A1R622	322-3226-00			RES,FXD,FILM:2.21K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K21
A1R623	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R624	313-1100-00			RES,FXD,FILM:10 OHM,5%,0.2W	57668	TR20JE10E0
A1R637	322-3222-00			RES,FXD,FILM:2K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K00
A1R638	311-2234-00			RES,VAR,NONWW:TRMR,5K OHM,20%,0.5W LINEAR	TK1450	GF06UT 5K
A1R639	311-2230-00			RES,VAR,NONWW:TRMR,500 OHM,20%,0.50 LINEAR	TK1450	GF06UT 500
A1R642	313-1432-00			RES,FXD,FILM:4.3K OHM,5%,0.2W	57668	TR20JE 04K3
A1R643	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R644	322-3258-00			RES,FXD,FILM:4.75K OHM,1%,0.2W,TC=T0	56845	ORDER BY DESCR
A1R645	321-0265-00			RES,FXD,FILM:5.62K OHM,1%,0.125W,TC=T0	19701	5043ED5K620F
A1R646	321-0252-00			RES,FXD,FILM:4.12K OHM,1%,0.125W,TC=T0	07716	CEAD41200F
A1R649	322-3243-00			RES,FXD,FILM:3.32K OHM,1%,0.2W,TC=T0	80009	322-3243-00
A1R650	322-3318-00			RES,FXD,FILM:20K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 20K0
A1R651	322-3189-00			RES,FXD,FILM:909 OHM,1%,0.2W,TC=T0	57668	CRB 20 FXE 909E
A1R652	315-0274-00			RES,FXD,FILM:270K OHM,5%,0.25W	57668	NTR25J-E270K
A1R653	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R655	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R658	321-0278-00			RES,FXD,FILM:7.68K OHM,1%,0.125W,TC=T0	07716	CEAD76800F
A1R659	322-3197-00			RES,FXD,FILM:1.1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K10
A1R669	321-0995-00			RES,FXD,FILM:549K OHM,1%,0.125W,TC=T0	24546	NA55D5493F
A1R670	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R671	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R678	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R700	313-1221-00			RES,FXD,FILM:220 OHM,5%,0.2W	57668	TR20JE220E
A1R701	322-3223-00			RES,FXD,FILM:2.05K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K05
A1R702	321-0252-00			RES,FXD,FILM:4.12K OHM,1%,0.125W,TC=T0	07716	CEAD41200F
A1R707	322-3201-00			RES,FXD,FILM:1.21K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K21
A1R708	313-1242-00			RES,FXD,FILM:2.4K OHM,5%,0.2W	57668	TR20JE 02K4
A1R709	322-3258-00			RES,FXD,FILM:4.75K OHM,1%,0.2W,TC=T0	56845	ORDER BY DESCR
A1R710	315-0396-00			RES,FXD,FILM:39M OHM,5%,0.25W	01121	CB3965
A1R713	313-1822-00			RES,FXD,FILM:8.2K,OHM,5%,0.2W	57668	TR20JE 08K2
A1R723	321-0240-00			RES,FXD,FILM:3.09K OHM,1%,0.125W,TC=T0	07716	CEAD30900F
A1R724	321-0680-00			RES,FXD,FILM:35.3K OHM,0.5%,0.125W,TC=T2	19701	5033RC35K30D
A1R731	322-3306-00			RES,FXD,FILM:15K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 15K0
A1R732	322-3273-00			RES,FXD,FILM:6.81K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 6K81
A1R733	322-3218-00			RES,FXD,FILM:1.82K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K82
A1R734	313-1221-00			RES,FXD,FILM:220 OHM,5%,0.2W	57668	TR20JE220E
A1R735	313-1273-00			RES,FXD,FILM:27K OHM,5%,0.2W	57668	TR20JE 27K
A1R736	321-0217-00			RES,FXD,FILM:1.78K OHM,1%,0.125W,TC=T0	19701	5043D1K780F
A1R737	322-3263-00			RES,FXD,FILM:5.36K OHM,1%,0.2W,TC=T0	56845	ORDER BY DESCR
A1R738	322-3273-00			RES,FXD,FILM:6.81K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 6K81
A1R742	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A1R743	313-1331-00			RES,FXD,FILM:330 OHM,5%,0.2W	57668	TR20JE 330E
A1R744	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R745	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A1R746	301-0470-00			RES,FXD,FILM:47 OHM,5%,0.5W	19701	5053CX47R00J
A1R747	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R748	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R749	313-1333-00			RES,FXD,FILM:33K OHM,5%,0.2W	57668	TR20JE 33K
A1R750	313-1151-00			RES,FXD,FILM:150 OHM,5%,0.2W	57668	TR20JE150E
A1R753	313-1242-00			RES,FXD,FILM:2.4K OHM,5%,0.2W	57668	TR20JE 02K4
A1R754	313-1333-00			RES,FXD,FILM:33K OHM,5%,0.2W	57668	TR20JE 33K
A1R755	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R757	313-1151-00			RES,FXD,FILM:150 OHM,5%,0.2W	57668	TR20JE150E
A1R800	321-0147-00			RES,FXD,FILM:332 OHM,1%,0.125W,TC=T0	07716	CEAD332R0F

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A1R801	311-2230-00			RES,VAR,NONWW:TRMR,500 OHM,20%,0.50 LINEAR	TK1450	GF06UT 500
A1R802	313-1222-00			RES,FXD,FILM:2.2K OHM,5%,0.2W	57668	TR20JE 02K2
A1R803	313-1821-00			RES,FXD,FILM:820 OHM,5%,0.2W	57668	TR20JE 820E
A1R804	313-1151-00			RES,FXD,FILM:150 OHM,5%,0.2W	57668	TR20JE150E
A1R805	311-1242-00			RES,VAR,NONWW:TRMR,200K OHM,0.5W	32997	3386X-T07-204
A1R806	322-3414-00			RES,FXD,FILM:200K OHM,1%,0.2W,TC=T0	91637	CCF50G20002F
A1R809	313-1151-00			RES,FXD,FILM:150 OHM,5%,0.2W	57668	TR20JE150E
A1R811	301-0331-00			RES,FXD,FILM:330 OHM,5%,0.5W	19701	5053CX330R0J
A1R817	313-1221-00			RES,FXD,FILM:220 OHM,5%,0.2W	57668	TR20JE220E
A1R820	321-0337-00			RES,FXD,FILM:31.6K OHM,1%,0.125W,TC=T0	07716	CEAD31601F
A1R821	321-0330-00			RES,FXD,FILM:26.7K OHM,1%,0.125W,TC=T0	07716	CEAD26701F
A1R822	322-3139-00			RES,FXD,FILM:274 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 274E
A1R823	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A1R849	313-1333-00			RES,FXD,FILM:33K OHM,5%,0.2W	57668	TR20JE 33K
A1R850	311-2234-00			RES,VAR,NONWW:TRMR,5K OHM,20%,0.5W LINEAR	TK1450	GF06UT 5K
A1R852	313-1240-00			RES,FXD,FILM:24 OHM,5%,0.2W	57668	TR20JT6824E0
A1R853	313-1240-00			RES,FXD,FILM:24 OHM,5%,0.2W	57668	TR20JT6824E0
A1R855	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A1R856	322-3210-00			RES,FXD,FILM:1.5K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K50
A1R858	322-3239-00			RES,FXD,FILM:3.01K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 3K01
A1R860	311-2234-00			RES,VAR,NONWW:TRMR,5K OHM,20%,0.5W LINEAR	TK1450	GF06UT 5K
A1R900	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R901	322-3197-00			RES,FXD,FILM:1.1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K10
A1R903	322-3258-00			RES,FXD,FILM:4.75K OHM,1%,0.2W,TC=T0	56845	ORDER BY DESCR
A1R904	313-1124-00			RES,FXD,FILM:120K OHM,5%,0.2W	57668	TR20JE120K
A1R907	313-1471-00			RES,FXD,FILM:470 OHM,5%,0.2W	57668	TR20JE 470E
A1R910	315-0396-00			RES,FXD,FILM:39M OHM,5%,0.25W	01121	CB3965
A1R912	313-1822-00			RES,FXD,FILM:8.2K,OHM,5%,0.2W	57668	TR20JE 08K2
A1R924	322-3325-00			RES,FXD,FILM:23.7K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 23K7
A1R936	322-3225-00			RES,FXD,FILM:2.15K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K15
A1R937	322-3268-00			RES,FXD,FILM:6.04K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 6K04
A1R939	315-0332-00			RES,FXD,FILM:3.3K OHM,5%,0.25W	57668	NTR25J-E03K3
A1R940	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R941	313-1151-00			RES,FXD,FILM:150 OHM,5%,0.2W	57668	TR20JE150E
A1R942	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A1R943	313-1151-00			RES,FXD,FILM:150 OHM,5%,0.2W	57668	TR20JE150E
A1R944	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R945	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A1R946	313-1221-00			RES,FXD,FILM:220 OHM,5%,0.2W	57668	TR20JE220E
A1R947	322-3117-00			RES,FXD,FILM:162 OHM,1%,0.2W,TC=T0	57668	CRB 20 FXE 162E
A1R950	301-0470-00			RES,FXD,FILM:47 OHM,5%,0.5W	19701	5053CX47R00J
A1R951	308-0555-00			RES,FXD,WW:5 OHM,5%,3W	00213	1200S-5.0-5
A1R952	322-3085-00			RES,FXD,FILM:75 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 75E0
A1R956	322-3239-00			RES,FXD,FILM:3.01K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 3K01
A1R957	321-0291-00			RES,FXD,FILM:10.5K OHM,1%,0.125W,TC=T0	19701	5033ED10K50F
A1R972	313-1510-00			RES,FXD,FILM:51 OHM,5%,0.2W	80009	313-1510-00
A1R973	313-1513-00			RES,FXD,CMPSN:51K OHM,5%,0.2W	57668	TR20JE 51K
A1R981	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R982	321-0103-00			RES,FXD,FILM:115 OHM,1%,0.125W,TC=T0	01121	RNK1150F
A1R985	322-3243-00			RES,FXD,FILM:3.32K OHM,1%,0.2W,TC=T0	80009	322-3243-00
A1R986	322-3097-00			RES,FXD,FILM:100 OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100E
A1R995	313-1512-00			RES,FXD,FILM:5.1K OHM,5%,0.2W	57668	TR20JE 5K1
A1S615	260-1421-00			SWITCH,PUSH:1 BTN,2 POLE,INSTRUMENT ID	59821	ORDER BY DESCR
A1TP800	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL	22526	48283-036
A1U100	153-2235-03			MICROCKT,LINEAR:LOW NOISE VERT PREAMP	80009	153-2235-03
A1U110	156-1245-00			MICROCKT,LINEAR:7 XSTR,NPN,SI,HV/HIGH CUR	01295	ULN2003AN-P3
A1U120	156-1245-00			MICROCKT,LINEAR:7 XSTR,NPN,SI,HV/HIGH CUR	01295	ULN2003AN-P3
A1U130	156-1245-00			MICROCKT,LINEAR:7 XSTR,NPN,SI,HV/HIGH CUR	01295	ULN2003AN-P3

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A1U140	156-0651-00			MICROCKT,DGTL:8-BIT PRL-OUT SER SHF RGTR	80009	156-0651-00
A1U150	156-0651-00			MICROCKT,DGTL:8-BIT PRL-OUT SER SHF RGTR	80009	156-0651-00
A1U160	156-1200-01			MICROCKT,LINEAR:BIFET,QUAD OPNL AMPL,SCRN	80009	156-1200-01
A1U165	156-2854-00			MICROCKT,LINEAR:OPNL AMPL QUAD	80009	156-2854-00
A1U170	156-0513-03			MICROCKT,LINEAR:CMOS,8 CHAN ANALOG MUX	04713	MC14051BCL
A1U180	156-1191-01			MICROCKT,LINEAR:BIFET,DUAL OPNL AMPL,SCRN	80009	156-1191-01
A1U200	153-2235-03			MICROCKT,LINEAR:LOW NOISE VERT PREAMP	80009	153-2235-03
A1U300	155-0238-00			MICROCKT,LINEAR:TRIGGER PREAMP	80009	155-0238-00
A1U350	156-1191-01			MICROCKT,LINEAR:BIFET,DUAL OPNL AMPL,SCRN	80009	156-1191-01
A1U400	155-0236-00			MICROCKT,LINEAR:VERTICAL CHANNEL SWITCH	80009	155-0236-00
A1U450	156-0158-07			MICROCKT,LINEAR:DUAL OPNL AMPL,SCREENED	01295	MC1458JG4
A1U475	156-0048-00			MICROCKT,LINEAR:5 XSTR ARRAY	02735	CA3046
A1U485	156-0048-00			MICROCKT,LINEAR:5 XSTR ARRAY	02735	CA3046
A1U500	155-0239-02			MICROCKT,LINEAR:TRIGGER	80009	155-0239-02
A1U550	156-0048-00			MICROCKT,LINEAR:5 XSTR ARRAY	02735	CA3046
A1U600	155-0237-02			MICROCKT,LINEAR:VERTICAL OUTPUT,H843	80009	155-0237-02
A1U650	155-0244-01			MICROCKT,DGTL:SYSTEM LOGIC INTERFACE	31471	M 217
A1U700	155-0240-00			MICROCKT,LINEAR:SWEEP	80009	155-0240-00
A1U735	156-0048-00			MICROCKT,LINEAR:5 XSTR ARRAY	02735	CA3046
A1U800	155-0241-02			MICROCKT,DGTL:HORIZONTAL AMP SYS	80009	155-0241-02
A1U850	156-0515-00			MICROCKT,DGTL:CMOS,TRIPLE 2-CHAN MUX	02735	CD4053BF
A1U860	156-0515-00			MICROCKT,DGTL:CMOS,TRIPLE 2-CHAN MUX	02735	CD4053BF
A1U900	155-0240-00			MICROCKT,LINEAR:SWEEP	80009	155-0240-00
A1U910	156-1191-01			MICROCKT,LINEAR:BIFET,DUAL OPNL AMPL,SCRN	80009	156-1191-01
A1U950	155-0242-01			MICROCKT,LINEAR:Z AXIS AUTOFOCUS	80009	155-0242-01
A1U975	160-5062-00			MICROCKT,DGTL:STTL,DECA 20 INP AND/OR PLD	80009	160-5062-00
A1U980	156-1611-01			MICROCKT,DGTL:ASTTL,DUAL D-TYPE FF	80009	156-1611-01
A1U985	156-0341-00			MICROCKT,DGTL:DUAL 2-INP OR DRIVER	07263	75453BTC
A1VR112	152-0166-00			SEMICOND DVC,DI:ZEN,SI,6.2V,5%,400MW,DO-7	04713	SZ11738RL
A1VR125	152-0166-00			SEMICOND DVC,DI:ZEN,SI,6.2V,5%,400MW,DO-7	04713	SZ11738RL
A1VR152	152-0236-00			SEMICOND DVC,DI:ZEN,SI,12.5V,4%,0.4W,DO-7	04713	SZ13553RL
A1VR225	152-0166-00			SEMICOND DVC,DI:ZEN,SI,6.2V,5%,400MW,DO-7	04713	SZ11738RL
A1VR550	152-0195-00			SEMICOND DVC,DI:ZEN,SI,5.1V,5%,0.4W,DO-7	04713	SZ11755RL
A1W101	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A1W103	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A1W104	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A1W105	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A1W106	195-6500-02			LEAD,ELECTRICAL:22 AWG,1.75 L,9-N	TK1544	195-6500-02
A1W107	195-6500-02			LEAD,ELECTRICAL:22 AWG,1.75 L,9-N	TK1544	195-6500-02
A1W108	195-6500-02			LEAD,ELECTRICAL:22 AWG,1.75 L,9-N	TK1544	195-6500-02
A1W109	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A1W120	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A1W121	175-4594-01			CA ASSY,SP,ELEC:6,22 AWG,5.25 L	80009	175-4594-01
A1W122	175-4598-00			CA ASSY,SP,ELEC:8,26 AWG,7.0 L,RIBBON	80009	175-4598-00
A1W141	174-0385-00			CA ASSY,SP,ELEC:5,22 AWG,2.6 L,9-N	80009	174-0385-00
A1W151	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A1W160	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A1W500	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A1W610	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A1W850	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A1W918	195-3991-01			LEAD,ELECTRICAL:22 AWG,3.5 L,0-N	80009	195-3991-01
A1W919	195-3991-01			LEAD,ELECTRICAL:22 AWG,3.5 L,0-N	80009	195-3991-01
A1XU100	136-0863-00			SKT,PL-IN ELEK:MICROCIRCUIT,64 DIP,PCB MT	19613	264-5200-00-3303
A1XU119	136-0728-00			SKT,PL-IN ELEK:MICROCKT,14 CONTACT	09922	DILB14P-108
A1XU191	136-0263-07			SOCKET,PIN TERM:U/W 0.025 SQ PIN (QUANTITY OF 16)	22526	ORDER BY DESCR
A1XU200	136-0863-00			SKT,PL-IN ELEK:MICROCIRCUIT,64 DIP,PCB MT	19613	264-5200-00-3303
A1XU300	136-0764-00			SKT,PL-IN ELEK:48 LINE CONT IMPD HYBRID	00779	ORDER BY DESCR

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A1XU400	136-0863-00			SKT,PL-IN ELEK:MICROCIRCUIT,64 DIP,PCB MT	19613	264-5200-00-3303
A1XU500	136-0764-00			SKT,PL-IN ELEK:48 LINE CONT IMPD HYBRID	00779	ORDER BY DESCR
A1XU600	136-0764-00			SKT,PL-IN ELEK:48 LINE CONT IMPD HYBRID	00779	ORDER BY DESCR
A1XU650	136-0757-00			SKT,PL-IN ELEK:MICROCIRCUIT,40 DIP	09922	DILB40P-108
A1XU700	136-0764-00			SKT,PL-IN ELEK:48 LINE CONT IMPD HYBRID	00779	ORDER BY DESCR
A1XU900	136-0764-00			SKT,PL-IN ELEK:48 LINE CONT IMPD HYBRID	00779	ORDER BY DESCR
A1XU950	136-0764-00			SKT,PL-IN ELEK:48 LINE CONT IMPD HYBRID	00779	ORDER BY DESCR

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A2	672-1037-12			CIRCUIT BD ASSY:LV PWR SPLY MODULE	80009	672-1037-12

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A2A1	----- -----			CIRCUIT BD ASSY:REGULATOR (AVAILABLE AT THE 672-1037-XX LEVEL ONLY)		
A2A1C1016	285-1222-00			CAP,FXD,PLASTIC:0.068UF,20%,250V	55112	158/.068/M/250/H
A2A1C1018	285-1222-00			CAP,FXD,PLASTIC:0.068UF,20%,250V	55112	158/.068/M/250/H
A2A1C1208	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A2A1C1220	290-0939-00			CAP,FXD,ELCTLT:10UF,+100-10%,100V	56289	672D106H100CG2C
A2A1C1222	281-0783-00			CAP,FXD,CER DI:0.1 UF 20%,100V	04222	MA401C104MAA
A2A1C1226	281-0791-00			CAP,FXD,CER DI:270PF,10%,100V	04222	MA101C271KAA
A2A1C1240	290-0939-00			CAP,FXD,ELCTLT:10UF,+100-10%,100V	56289	672D106H100CG2C
A2A1C1245	281-0783-00			CAP,FXD,CER DI:0.1 UF 20%,100V	04222	MA401C104MAA
A2A1C1246	281-0791-00			CAP,FXD,CER DI:270PF,10%,100V	04222	MA101C271KAA
A2A1C1260	290-0942-00			CAP,FXD,ELCTLT:100UF,+100-10%,25V	55680	UPA1E101MAH
A2A1C1261	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A2A1C1270	281-0791-00			CAP,FXD,CER DI:270PF,10%,100V	04222	MA101C271KAA
A2A1C1272	281-0774-00			CAP,FXD,CER DI:0.022MFD,20%,100V	04222	MA201E223MAA
A2A1C1280	290-0942-00			CAP,FXD,ELCTLT:100UF,+100-10%,25V	55680	UPA1E101MAH
A2A1C1290	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A2A1C1291	290-0778-00			CAP,FXD,ELCTLT:1UF,20%,50V,NPLZD	54473	ECE-A50N1
A2A1C1292	290-0778-01			CAP,FXD,ELCTLT:1UF,+20%,50V	55680	UEB1H010MAAITD
A2A1C1300	290-0942-00			CAP,FXD,ELCTLT:100UF,+100-10%,25V	55680	UPA1E101MAH
A2A1C1330	290-0942-00			CAP,FXD,ELCTLT:100UF,+100-10%,25V	55680	UPA1E101MAH
A2A1C1331	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A2A1C1350	290-0942-00			CAP,FXD,ELCTLT:100UF,+100-10%,25V	55680	UPA1E101MAH
A2A1C1357	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A2A1C1374	281-0791-00			CAP,FXD,CER DI:270PF,10%,100V	04222	MA101C271KAA
A2A1C1400	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A2A1C1402	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A2A1C1974	290-0778-00			CAP,FXD,ELCTLT:1UF,20%,50V,NPLZD	54473	ECE-A50N1
A2A1CR1011	152-0750-00			SEMICOND DVC,DI:RECT,BRIDGE,SI,600V,3A	05828	RKBPC606-12
A2A1CR1220	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1221	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1241	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1242	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1243	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1244	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1260	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1261	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1262	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1CR1263	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1CR1264	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1CR1281	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1CR1282	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1CR1283	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1290	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1CR1294	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1CR1295	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1CR1300	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1CR1301	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1CR1302	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1CR1303	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1330	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1331	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1332	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1334	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1351	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A2A1CR1376	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A2A1E1001	119-0181-00			ARSR,ELEC SURGE:230,GAS FILLED	25088	B1-A230
A2A1E1002	119-0181-00			ARSR,ELEC SURGE:230,GAS FILLED	25088	B1-A230

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A2A1F1330	159-0295-00			FUSE,CARTRIDGE:5 X 20MM,125V,1AMP	TK0946	TSC-1
A2A1J121	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 6)	22526	48283-036
A2A1J122	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 7)	22526	48283-036
A2A1J201	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 4)	22526	48283-036
A2A1J202	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 4)	22526	48283-036
A2A1J203	131-2925-00			CONN,RCPT,ELEC:CKT BD,1 X 6,0.2 SPACING	27264	10-10-1064
A2A1J204	131-1048-00			TERM,QIK DISC.:CKT BD MT,0.11 X 0.02 BL	00779	61134-1
A2A1J205	131-1048-00			TERM,QIK DISC.:CKT BD MT,0.11 X 0.02 BL	00779	61134-1
A2A1J206	131-1048-00			TERM,QIK DISC.:CKT BD MT,0.11 X 0.02 BL	00779	61134-1
A2A1J207	131-1048-00			TERM,QIK DISC.:CKT BD MT,0.11 X 0.02 BL	00779	61134-1
A2A1J208	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL	22526	48283-036
A2A1L1011	108-0473-00			COIL,RF:FIXED,174UH	TK2042	ORDER BY DESCR
A2A1L1012	108-0473-00			COIL,RF:FIXED,174UH	TK2042	ORDER BY DESCR
A2A1L1402	108-0443-00			COIL,RF:FIXED,23.5UH	80009	108-0443-00
A2A1P208	131-3957-00			BUS,CONDUCTOR:SHUNT ASSEMBLY,BLACK	80009	131-3957-00
A2A1Q1220	151-0497-00			TRANSISTOR:NPN,SI,TO-220	80009	151-0497-00
A2A1Q1221	151-0347-00			TRANSISTOR:NPN,SI,TO-92	04713	SPS7951
A2A1Q1222	151-0347-00			TRANSISTOR:NPN,SI,TO-92	04713	SPS7951
A2A1Q1223	151-0347-00			TRANSISTOR:NPN,SI,TO-92	04713	SPS7951
A2A1Q1240	151-0464-00			TRANSISTOR:NPN,SI,TO-220	80009	151-0464-00
A2A1Q1241	151-0347-00			TRANSISTOR:NPN,SI,TO-92	04713	SPS7951
A2A1Q1243	151-0347-00			TRANSISTOR:NPN,SI,TO-92	04713	SPS7951
A2A1Q1245	151-0347-00			TRANSISTOR:NPN,SI,TO-92	04713	SPS7951
A2A1Q1280	151-0476-00			TRANSISTOR:NPN,SI,TO-220	80009	151-0476-00
A2A1Q1281	151-0347-00			TRANSISTOR:NPN,SI,TO-92	04713	SPS7951
A2A1Q1290	151-1059-00			TRANSISTOR:FET,N-CHAN,30MW,TO-92 CASE	04713	ORDER BY DESCR
A2A1Q1300	151-0482-00			TRANSISTOR:PNP,SI,TO-220	04713	SJE1977
A2A1Q1301	151-0342-00			TRANSISTOR:PNP,SI,TO-92	07263	S035928
A2A1Q1351	151-0429-00			TRANSISTOR:DARLINGTON,PNP,SI,TO-126	80009	151-0429-00
A2A1Q1354	151-0342-00			TRANSISTOR:PNP,SI,TO-92	07263	S035928
A2A1Q1370	151-0341-00			TRANSISTOR:NPN,SI,TO-106	04713	SPS6919
A2A1Q1376	151-0341-00			TRANSISTOR:NPN,SI,TO-106	04713	SPS6919
A2A1R1010	301-0150-00			RES,FXD,FILM:15 OHM,5%,0.5W	19701	5053CX15R00J
A2A1R1011	315-0560-00			RES,FXD,FILM:56 OHM,5%,0.25W	57668	NTR25J-E56E0
A2A1R1012	315-0560-00			RES,FXD,FILM:56 OHM,5%,0.25W	57668	NTR25J-E56E0
A2A1R1013	315-0683-00			RES,FXD,FILM:68K OHM,5%,0.25W	57668	NTR25J-E68K0
A2A1R1014	313-1363-00			RES,FXD,FILM:36K OHM,5%,0.2W	57668	TR20JE 36K
A2A1R1015	313-1363-00			RES,FXD,FILM:36K OHM,5%,0.2W	57668	TR20JE 36K
A2A1R1016	301-0680-00			RES,FXD,FILM:68 OHM,5%,0.5W	19701	5053CX68R00J
A2A1R1017	315-0474-00			RES,FXD,FILM:470K OHM,5%,0.25W	19701	5043CX470K0J92U
A2A1R1018	301-0300-00			RES,FXD,FILM:30 OHM,5%,0.5W	19701	5053CX30R00J
A2A1R1019	301-0150-00			RES,FXD,FILM:15 OHM,5%,0.5W	19701	5053CX15R00J
A2A1R1204	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A2A1R1208	313-1201-00			RES,FXD,FILM:200 OHM,5%,0.2W	57668	TR20JE200E
A2A1R1212	313-1393-00			RES,FXD,FILM:39K OHM,5%,0.2W	57668	TR20JE 39K
A2A1R1220	304-0822-00			RES,FXD,CMPSN:8.2K OHM,10%,1W	01121	GB8221
A2A1R1221	315-0100-02			RES,FXD,CMPSN:10 OHM,5%,0.25W	01121	CB1005
A2A1R1222	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A2A1R1223	313-1823-00			RES,FXD,FILM:82K OHM,5%,0.2W	57668	TR20JE 82K
A2A1R1226	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A2A1R1227	321-0634-00			RES,FXD,FILM:84.65K OHM,0.25%,0.125W,TC=T2	19701	5033RC84K65C
A2A1R1228	321-0293-03			RES,FXD,FILM:11.0K OHM,0.25%,0.125W,TC=T2	24546	NC55C1102C
A2A1R1229	313-1683-00			RES,FXD,FILM:68K OHM,5%,0.2W	57668	TR20JE 68K
A2A1R1240	303-0202-00			RES,FXD,CMPSN:2K OHM,5%,1W	01121	GB 2025

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A2A1R1241	307-0105-00			RES,FXD,CMPSN:3.9 OHM,5%,0.25W	01121	CB 39G5
A2A1R1242	313-1152-00			RES,FXD,FILM:1.5K OHM,5%,0.2W	57668	TR20JE01K5
A2A1R1243	313-1393-00			RES,FXD,FILM:39K OHM,5%,0.2W	57668	TR20JE 39K
A2A1R1244	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A2A1R1246	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A2A1R1247	321-0368-00			RES,FXD,FILM:66.5K OHM,1%,0.125W,TC=T0	07716	CEAD66501F
A2A1R1248	321-0319-00			RES,FXD,FILM:20.5K OHM,1%,0.125W,TC=T0	19701	5033ED20K50F
A2A1R1249	313-1473-00			RES,FXD,FILM:47K OHM,5%,0.2W	57668	TR20JE 47K
A2A1R1261	321-0289-00			RES,FXD,FILM:10.0K OHM,1%,0.125W,TC=T0	19701	5033ED10K0F
A2A1R1262	321-0318-00			RES,FXD,FILM:20.0K OHM,1%,0.125W,TC=T0	19701	5033ED20K00F
A2A1R1264	313-1473-00			RES,FXD,FILM:47K OHM,5%,0.2W	57668	TR20JE 47K
A2A1R1270	313-1432-00			RES,FXD,FILM:4.3K OHM,5%,0.2W	57668	TR20JE 04K3
A2A1R1273	313-1473-00			RES,FXD,FILM:47K OHM,5%,0.2W	57668	TR20JE 47K
A2A1R1274	313-1683-00			RES,FXD,FILM:68K OHM,5%,0.2W	57668	TR20JE 68K
A2A1R1280	303-0470-00			RES,FXD,CMPSN:47 OHM,5%,1W	01121	GB4705
A2A1R1281	308-0839-00			RES,FXD,WW:0.1 OHM,5%,1.0W	75042	BW-20-R1000J
A2A1R1282	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A2A1R1283	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A2A1R1284	321-0318-00			RES,FXD,FILM:20.0K OHM,1%,0.125W,TC=T0	19701	5033ED20K00F
A2A1R1285	321-0318-00			RES,FXD,FILM:20.0K OHM,1%,0.125W,TC=T0	19701	5033ED20K00F
A2A1R1286	313-1243-00			RES,FXD,FILM:24K OHM,5%,0.2W	80009	313-1243-00
A2A1R1287	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A2A1R1291	321-0334-00			RES,FXD,FILM:29.4K OHM,1%,0.125W,TC=T0	07716	CEAD29401F
A2A1R1292	311-2258-00			RES,VAR,NONWW:TRMR,1K OHM,20%,0.5W	TK1450	GF06VT 1 K OHM
A2A1R1293	321-0639-00			RES,FXD,FILM:9.6K OHM,1%,0.125W,TC=T0	19701	5043ED9K600F
A2A1R1294	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A2A1R1295	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A2A1R1296	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A2A1R1297	322-3289-00			RES,FXD,FILM:10K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 10K0
A2A1R1298	322-3275-00			RES,FXD,FILM:7.15K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 7K15
A2A1R1299	313-1224-00			RES,FXD,FILM:220K,5%,0.2W	57668	TR20JE 220K
A2A1R1300	303-0470-00			RES,FXD,CMPSN:47 OHM,5%,1W	01121	GB4705
A2A1R1301	308-0839-00			RES,FXD,WW:0.1 OHM,5%,1.0W	75042	BW-20-R1000J
A2A1R1302	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A2A1R1304	313-1243-00			RES,FXD,FILM:24K OHM,5%,0.2W	80009	313-1243-00
A2A1R1305	321-0289-06			RES,FXD,FILM:10.0K OHM,0.25%,0.125W,TC=T9	19701	5033RE10K00C
A2A1R1306	321-0318-03			RES,FXD,FILM:20.0K OHM,0.125%,0.125W,TC=T2	19701	5033RC20K00C
A2A1R1307	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A2A1R1309	313-1222-00			RES,FXD,FILM:2.2K OHM,5%,0.2W	57668	TR20JE 02K2
A2A1R1331	321-0685-00			RES,FXD,FILM:30K OHM,0.5%,0.125W,TC=T2	19701	5033RC30K00D
A2A1R1332	321-0318-03			RES,FXD,FILM:20.0K OHM,0.125%,0.125W,TC=T2	19701	5033RC20K00C
A2A1R1333	313-1751-00			RES,FXD,FILM:750 OHM,5%,0.2W	57668	TR20JE 750E
A2A1R1334	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A2A1R1351	313-1202-00			RES,FXD,FILM:2K OHM,5%,0.2W	57668	TR20JE02K0
A2A1R1352	301-0150-00			RES,FXD,FILM:15 OHM,5%,0.5W	19701	5053CX15R00J
A2A1R1353	301-0150-00			RES,FXD,FILM:15 OHM,5%,0.5W	19701	5053CX15R00J
A2A1R1354	313-1222-00			RES,FXD,FILM:2.2K OHM,5%,0.2W	57668	TR20JE 02K2
A2A1R1355	313-1682-00			RES,FXD,FILM:6.8K OHM,5%,0.2W	57668	TR20JE 06K8
A2A1R1356	313-1512-00			RES,FXD,FILM:5.1K OHM,5%,0.2W	57668	TR20JE 5K1
A2A1R1357	321-0318-03			RES,FXD,FILM:20.0K OHM,0.125%,0.125W,TC=T2	19701	5033RC20K00C
A2A1R1358	321-0689-00			RES,FXD,FILM:24.9K OHM,0.5%,0.125W,TC=T0	19701	5033RD24K90D
A2A1R1359	313-1682-00			RES,FXD,FILM:6.8K OHM,5%,0.2W	57668	TR20JE 06K8
A2A1R1370	321-0363-00			RES,FXD,FILM:59.0K OHM,1%,0.125W,TC=T0	07716	CEAD59001F
A2A1R1372	321-0299-00			RES,FXD,FILM:12.7K OHM,1%,0.125W,TC=T0	19701	5033ED12K70F
A2A1R1374	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A2A1R1376	321-0318-03			RES,FXD,FILM:20.0K OHM,0.125%,0.125W,TC=T2	19701	5033RC20K00C
A2A1R1378	313-1202-00			RES,FXD,FILM:2K OHM,5%,0.2W	57668	TR20JE02K0
A2A1R1400	315-0101-03			RES,FXD,CMPSN:100 OHM,5%,0.25W	01121	CB1015

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A2A1R1402	315-0101-03			RES,FXD,CMPSN:100 OHM,5%,0.25W	01121	CB1015
A2A1RT1010	307-0350-00			RES,THERMAL:7.5 OHM,10%,3.9%/DEG C	80009	307-0350-00
A2A1RT1016	307-0746-00			RES,THERMAL:5 OHM,10%,7A/DEG C	15454	SG200-S
A2A1S350	260-1849-00			SWITCH,PUSH:DPDT,4A,250VAC	31918	NE15/F2U103EE
A2A1T1229	120-1401-00			XFMR,TRIGGER:LINE,1:1 TURNS RATIO	54937	DMI 500-2044
A2A1U1260	156-1161-00			MICROCKT,LINEAR:VOLTAGE REGULATOR,POS,ADJ	12969	UC317T
A2A1U1270	156-0495-00			MICROCKT,LINEAR:OPNL AMPL	01295	LM324N
A2A1U1281	156-0158-07			MICROCKT,LINEAR:DUAL OPNL AMPL,SCREENED	01295	MC1458JG4
A2A1U1290	156-1173-00			MICROCKT,LINEAR:VOLTAGE REFERENCE	04713	MC1403UDS
A2A1U1300	156-0495-00			MICROCKT,LINEAR:OPNL AMPL	01295	LM324N
A2A1U1330	156-0872-00			MICROCKT,LINEAR:VOLTAGE REGULATOR	04713	MC7912CT
A2A1U1371	156-0495-00			MICROCKT,LINEAR:OPNL AMPL	01295	LM324N
A2A1VR1293	152-0055-00			SEMICOND DVC,DI:ZEN,SI,11V,5%,0.4W,DO-7	14433	Z5407
A2A1W251	175-4585-00			CA ASSY,SP,ELEC:20,28 AWG,13.0 L	80009	175-4585-00

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A3	----- -----			CIRCUIT BD ASSY:INVERTER (AVAILABLE AT THE 672-1037-XX LEVEL ONLY)		
A3C1020	285-1192-00			CAP,FXD,PPR DI:0.0022 UF,20%,250VAC	TK0515	PME271Y510
A3C1021	290-0971-00			CAP,FXD,ELCTLT:290UF +50-10%,200V	56289	39DX1314
A3C1022	290-0971-00			CAP,FXD,ELCTLT:290UF +50-10%,200V	56289	39DX1314
A3C1023	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A3C1025	290-0942-00			CAP,FXD,ELCTLT:100UF,+100-10%,25V	55680	UPA1E101MAH
A3C1032	281-0812-00			CAP,FXD,CER DI:1000PF,10%,100V	04222	MA101C102KAA
A3C1033	281-0772-00			CAP,FXD,CER DI:4700PF,10%,100V	04222	MA201C472KAA
A3C1034	290-0524-00			CAP,FXD,ELCTLT:4.7UF,20%,10V	05397	T368A475M010AZ
A3C1035	281-0772-00			CAP,FXD,CER DI:4700PF,10%,100V	04222	MA201C472KAA
A3C1040	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A3C1042	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A3C1048	281-0826-00			CAP,FXD,CER DI:2200PF,10%,100V	20932	401EM100AD222K
A3C1050	285-1254-00			CAP,FXD,PLASTIC:0.22UF,10%,400WVDC	56289	730P0167
A3C1051	285-1192-00			CAP,FXD,PPR DI:0.0022 UF,20%,250VAC	TK0515	PME271Y510
A3C1052	285-1196-00			CAP,FXD,PPR DI:0.01UF,20%,250V	TK0515	PME 265 MB 510
A3C1062	281-0850-00			CAP,FXD,CER DI:820PF,5%,50VDC	04222	SA101A821JAA
A3C1065	285-1190-00			CAP,FXD,MTLZD:0.056 UF,5%,250 V	05292	PMT3R ADVISE
A3C1066	290-0782-01			CAP,FXD,ELCTLT:4.7UF,20%,35VDC	55680	UVX1V4R7MAA1TD
A3C1067	281-0850-00			CAP,FXD,CER DI:820PF,5%,50VDC	04222	SA101A821JAA
A3C1071	281-0772-00			CAP,FXD,CER DI:4700PF,10%,100V	04222	MA201C472KAA
A3C1072	290-0806-00			CAP,FXD,ELCTLT:3.3UF,+75-10%,350VDC	55680	UHU2V3R3TPA
A3C1075	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A3C1101	290-0942-00			CAP,FXD,ELCTLT:100UF,+100-10%,25V	55680	UPA1E101MAH
A3C1102	290-0942-00			CAP,FXD,ELCTLT:100UF,+100-10%,25V	55680	UPA1E101MAH
A3C1110	290-0800-00			CAP,FXD,ELCTLT:250UF,+100-10%,20V	56289	672D257H020DM5C
A3C1111	290-0800-00			CAP,FXD,ELCTLT:250UF,+100-10%,20V	56289	672D257H020DM5C
A3C1112	290-0782-01			CAP,FXD,ELCTLT:4.7UF,20%,35VDC	55680	UVX1V4R7MAA1TD
A3C1113	290-0798-00			CAP,FXD,ELCTLT:180UF,+100-10%,40V	56289	672D187H040DM5C
A3C1114	290-0800-00			CAP,FXD,ELCTLT:250UF,+100-10%,20V	56289	672D257H020DM5C
A3C1115	290-0800-00			CAP,FXD,ELCTLT:250UF,+100-10%,20V	56289	672D257H020DM5C
A3C1116	290-0798-00			CAP,FXD,ELCTLT:180UF,+100-10%,40V	56289	672D187H040DM5C
A3C1120	290-0939-00			CAP,FXD,ELCTLT:10UF,+100-10%,100V	56289	672D106H100CG2C
A3C1130	290-0939-00			CAP,FXD,ELCTLT:10UF,+100-10%,100V	56289	672D106H100CG2C
A3C1132	290-0880-00			CAP,FXD,ELCTLT:10UF,+50-10%,160V	54473	ECE-A160V10U
A3CR1022	152-0333-00			SEMICOND DVC,DI:SW,SI,55V,200MA,DO-35	07263	FDH-6012
A3CR1023	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A3CR1028	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A3CR1030	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A3CR1034	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A3CR1035	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A3CR1040	152-0075-00			SEMICOND DVC,DI:SW,GE,22V,80MW,DO-7	80009	152-0075-00
A3CR1050	152-0661-01			SEMICOND DVC,DI:RECT,SI,600V,3A	04713	S.R.3523-1RL
A3CR1060	152-0040-00			SEMICOND DVC,DI:RECT,SI,600V,1A,DO-41	80009	152-0040-00
A3CR1062	152-0333-00			SEMICOND DVC,DI:SW,SI,55V,200MA,DO-35	07263	FDH-6012
A3CR1063	152-0333-00			SEMICOND DVC,DI:SW,SI,55V,200MA,DO-35	07263	FDH-6012
A3CR1064	152-0333-00			SEMICOND DVC,DI:SW,SI,55V,200MA,DO-35	07263	FDH-6012
A3CR1065	152-0333-00			SEMICOND DVC,DI:SW,SI,55V,200MA,DO-35	07263	FDH-6012
A3CR1070	152-0040-00			SEMICOND DVC,DI:RECT,SI,600V,1A,DO-41	80009	152-0040-00
A3CR1072	152-0066-00			SEMICOND DVC,DI:RECT,SI,400V,1A,DO-41	05828	GP10G-020
A3CR1101	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3CR1102	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3CR1103	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3CR1104	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3CR1105	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3CR1106	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3CR1110	152-0794-00			SEMICOND DVC,DI:RECT,SI,10A,30V,TO-220	81483	95-4269

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A3CR1113	152-0946-00			SEMICOND DVC,DI:RECT,SI,40V,3.0A	80009	152-0946-00
A3CR1114	152-0946-00			SEMICOND DVC,DI:RECT,SI,40V,3.0A	80009	152-0946-00
A3CR1115	152-0946-00			SEMICOND DVC,DI:RECT,SI,40V,3.0A	80009	152-0946-00
A3CR1116	152-0946-00			SEMICOND DVC,DI:RECT,SI,40V,3.0A	80009	152-0946-00
A3CR1121	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3CR1122	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3CR1123	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3CR1124	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3CR1131	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3CR1132	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A3F1101	159-0255-00			FUSE,CARTRIDGE:FAST BLOW,4A,125V	80009	159-0255-00
A3F1102	159-0059-00			FUSE,WIRE LEAD:5A,125V	71400	A5
A3J301	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 3)	22526	48283-036
A3J302	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 3)	22526	48283-036
A3J303	131-2926-00			CONN;RCPT,ELEC:CKT BD,1 X 2,0.2 SPACING	27264	10-10-1024
A3L1110	108-0554-00			COIL,RF:FIXED,5UH,+/-20%	TK1345	108-0554-00
A3L1113	108-1144-00			COIL,RF:FIXED,27 UH,20%	34479	RL1284
A3L1114	108-1144-00			COIL,RF:FIXED,27 UH,20%	34479	RL1284
A3L1115	108-1144-00			COIL,RF:FIXED,27 UH,20%	34479	RL1284
A3L1116	108-1144-00			COIL,RF:FIXED,27 UH,20%	34479	RL1284
A3Q1021	151-0301-00			TRANSISTOR:PNP,SI,TO-18	80009	151-0301-00
A3Q1022	151-0192-00			TRANSISTOR:NPN,SI,TO-92	04713	SPS8801
A3Q1029	151-0254-00			TRANSISTOR:DARLINGTON,NPN,SI,625MW,TO-92	03508	X38L3118
A3Q1030	151-0301-00			TRANSISTOR:PNP,SI,TO-18	80009	151-0301-00
A3Q1040	151-0302-00			TRANSISTOR:NPN,SI,TO-18	04713	ST899
A3Q1050	151-1152-00			TRANSISTOR:MOSFE,N-CHANNEL,SI,TO-220	04713	IRF820
A3Q1060	151-1152-00			TRANSISTOR:MOSFE,N-CHANNEL,SI,TO-220	04713	IRF820
A3Q1062	151-0302-00			TRANSISTOR:NPN,SI,TO-18	04713	ST899
A3Q1070	151-1152-00			TRANSISTOR:MOSFE,N-CHANNEL,SI,TO-220	04713	IRF820
A3Q1110	151-0188-00			TRANSISTOR:PNP,SI,TO-92	80009	151-0188-00
A3R1018	313-1394-00			RES,FXD,FILM:390K,5%,0.2W	57668	TR20JE 390K
A3R1019	313-1394-00			RES,FXD,FILM:390K,5%,0.2W	57668	TR20JE 390K
A3R1020	301-0274-00			RES,FXD,FILM:270K OHM,5%,0.5W	19701	5053CX270K0J
A3R1022	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A3R1023	313-1122-00			RES,FXD,FILM:1.2K OHM,5%,0.2W	57668	TR20JE01K2
A3R1024	313-1473-00			RES,FXD,FILM:47K OHM,5%,0.2W	57668	TR20JE 47K
A3R1025	313-1302-00			RES,FXD,FILM:3K OHM,5%,0.2W	57668	TR20JE 03K0
A3R1027	321-0431-00			RES,FXD,FILM:301K OHM,1%,0.125W,TC=T0	07716	CEAD30102F
A3R1028	321-0481-04			RES,FXD,FILM:1M OHM,0.1%,0.125W,TC=T2	91637	CMF55116D10003B
A3R1029	313-1152-00			RES,FXD,FILM:1.5K OHM,5%,0.2W	57668	TR20JE01K5
A3R1030	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A3R1031	313-1334-00			RES,FXD,FILM:330K OHM,5%,0.2W	80009	313-1334-00
A3R1032	321-0335-00			RES,FXD,FILM:30.1K OHM,1%,0.125W,TC=T0	57668	RB14FXE30K1
A3R1033	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A3R1034	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A3R1035	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A3R1036	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A3R1037	313-1272-00			RES,FXD,FILM:2.7K OHM,5%,0.2W	57668	TR20JE 02K7
A3R1040	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A3R1041	313-1471-00			RES,FXD,FILM:470 OHM,5%,0.2W	57668	TR20JE 470E
A3R1042	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A3R1044	321-0334-00			RES,FXD,FILM:29.4K OHM,1%,0.125W,TC=T0	07716	CEAD29401F
A3R1045	321-0289-00			RES,FXD,FILM:10.0K OHM,1%,0.125W,TC=T0	19701	5033ED10K0F
A3R1046	321-0422-00			RES,FXD,FILM:243K OHM,1%,0.125W,TC=T0	07716	CEAD24302F
A3R1050	308-0843-00			RES,FXD,WW:0.2 OHM,5%,1/0W	91637	RS1A-90-R2J
A3R1052	313-1470-00			RES,FXD,FILM:47 OHM,5%,0.2W	57668	TR20JE 47E

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A3R1060	313-1470-00			RES,FXD,FILM:47 OHM,5%,0.2W	57668	TR20JE 47E
A3R1061	313-1202-00			RES,FXD,FILM:2K OHM,5%,0.2W	57668	TR20JE02K0
A3R1062	313-1682-00			RES,FXD,FILM:6.8K OHM,5%,0.2W	57668	TR20JE 06K8
A3R1063	313-1202-00			RES,FXD,FILM:2K OHM,5%,0.2W	57668	TR20JE02K0
A3R1064	313-1202-00			RES,FXD,FILM:2K OHM,5%,0.2W	57668	TR20JE02K0
A3R1065	315-0154-00			RES,FXD,FILM:150K OHM,5%,0.25W	57668	NTR25J-E150K
A3R1066	313-1202-00			RES,FXD,FILM:2K OHM,5%,0.2W	57668	TR20JE02K0
A3R1067	313-1682-00			RES,FXD,FILM:6.8K OHM,5%,0.2W	57668	TR20JE 06K8
A3R1068	313-1202-00			RES,FXD,FILM:2K OHM,5%,0.2W	57668	TR20JE02K0
A3R1069	303-0363-00			RES,FXD,CMPSN:36K OHM,5%,1W	01121	GB3635
A3R1070	313-1470-00			RES,FXD,FILM:47 OHM,5%,0.2W	57668	TR20JE 47E
A3R1071	315-0431-00			RES,FXD,FILM:430 OHM,5%,0.25W	19701	5043CX430R0J
A3R1072	321-0318-03			RES,FXD,FILM:20.0K OHM,0.125%,0.125W,TC=T2	19701	5033RC20K00C
A3R1075	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A3R1110	321-0219-00			RES,FXD,FILM:1.87K OHM,1%,0.125W,TC=T0	07716	CEAD18700F
A3R1111	315-0510-00			RES,FXD,FILM:51 OHM,5%,0.25W	19701	5043CX51R00J
A3R1112	321-0271-00			RES,FXD,FILM:6.49K OHM,1%,0.125W,TC=T0	07716	CEAD64900F
A3R1113	321-0271-00			RES,FXD,FILM:6.49K OHM,1%,0.125W,TC=T0	07716	CEAD64900F
A3R1114	321-0297-00			RES,FXD,FILM:12.1K OHM,1%,0.125W,TC=T0	07716	CEAD12101F
A3R1115	301-0301-00			RES,FXD,FILM:300 OHM,5%,0.5W	19701	5053CX300R0J
A3R1129	313-1474-00			RES,FXD,FILM:470K OHM,5%,0.2W	80009	313-1474-00
A3R1130	313-1273-00			RES,FXD,FILM:27K OHM,5%,0.2W	57668	TR20JE 27K
A3RL1060	108-0329-00			COIL,RF:FIXED,2.4UH	TK2042	ORDER BY DESCR
A3RT1110	307-0124-00			RES,THERMAL:5K OHM,10%,NTC	15454	1DC502K-220-EC
A3S1020	260-0907-01			SWITCH,THRMSTC:NC,OPEN 97.8,CL 75.6,10A	93410	430-1537
A3T1020	120-1244-00			TRANSFORMER,RF:COMMON MODE,13MH,0.5A	20462	4096
A3T1060	120-1437-00			XFMR,PWR,STPDN:	02113	C1310
A3U1029	156-0885-05			CPLR,OPTOELECTR:LED,5KV,ISOLATION	09019	H11AX1139R
A3U1030	156-1627-00			MICROCKT,LINEAR:BIPOLAR,PWM PWR SPLY CONT	12969	UC494ACN
A3U1040	156-0885-05			CPLR,OPTOELECTR:LED,5KV,ISOLATION	09019	H11AX1139R
A3U1064	156-0366-00			MICROCKT,DGTL:DUAL D FLIP-FLOP	02735	CD4013BF
A3U1066	156-0328-00			MICROCKT,DGTL:DUAL MOS CLOCK DRIVER	04713	MMH0026CP1D
A3U1110	156-1161-00			MICROCKT,LINEAR:VOLTAGE REGULATOR,POS,ADJ	12969	UC317T
A3VR1020	152-0166-00			SEMICOND DVC,DI:ZEN,SI,6.2V,5%,400MW,DO-7	04713	SZ11738RL
A3VR1062	152-0168-00			SEMICOND DVC,DI:ZEN,SI,12V,5%,0.4W,DO-763B	14552	TD331689
A3W1021	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A3W1022	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A3W1050	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A3W1060	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A3W1102	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A3Y1050	120-1417-00			TRANSFORMER,RF:POWER HIGH FREQUENCY	54937	500-2311

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A4	670-9493-02			CIRCUIT BD ASSY:READOUT	80009	670-9493-02
A4C2830	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2835	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2851	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2855	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2860	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2885	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2901	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2911	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A4C2912	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2913	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2926	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2940	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2950	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2960	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2970	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2980	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4C2990	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A4R2805	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A4R2830	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A4R2841	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A4R2842	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A4R2843	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A4R2844	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A4R2850	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A4R2901	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A4R2902	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A4R2903	321-1296-03			RES,FXD,FILM:12.0K OHM,0.25%,0.125W,TC=T2	07716	CEAC12001C
A4R2905	321-0816-03			RES,FXD,FILM:5K OHM,0.25%,0.125W,TC=T2	19701	5033RC5K000C
A4R2910	321-0685-00			RES,FXD,FILM:30K OHM,0.5%,0.125W,TC=T2	19701	5033RC30K00D
A4R2911	321-0685-00			RES,FXD,FILM:30K OHM,0.5%,0.125W,TC=T2	19701	5033RC30K00D
A4R2912	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A4R2913	321-0198-00			RES,FXD,FILM:1.13K OHM,1%,0.125W,TC=T0	07716	CEAD11300F
A4R2914	322-3306-00			RES,FXD,FILM:15K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 15K0
A4R2915	313-1202-00			RES,FXD,FILM:2K OHM,5%,0.2W	57668	TR20JE02K0
A4R2916	322-3414-00			RES,FXD,FILM:200K OHM,1%,0.2W,TC=T0	91637	CCF50G20002F
A4R2917	322-3385-00			RES,FXD,FILM:100K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100K
A4R2918	311-2270-00			RES,VAR,NONWW:TRMR,10K OHM,20%,0.5W	TK1450	GF06VT 10 K OHM
A4R2919	321-0756-00			RES,FXD,FILM:50K OHM,1%,0.125W,TC=T0	24546	NA55D5002F
A4R2920	313-1334-00			RES,FXD,FILM:330K OHM,5%,0.2W	80009	313-1334-00
A4R2921	322-3297-00			RES,FXD,FILM:12.1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 12K1
A4R2922	321-0756-00			RES,FXD,FILM:50K OHM,1%,0.125W,TC=T0	24546	NA55D5002F
A4R2923	321-0385-00			RES,FXD,FILM:100K OHM,1%,0.125W,TC=T0	19701	5033ED100K0F
A4R2924	322-3414-00			RES,FXD,FILM:200K OHM,1%,0.2W,TC=T0	91637	CCF50G20002F
A4R2925	321-0235-02			RES,FXD,FILM:2.74K OHM,0.5%,0.125W,TC=T2	24546	NC55C2741D
A4R2926	322-3222-00			RES,FXD,FILM:2K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K00
A4R2927	322-3318-00			RES,FXD,FILM:20K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 20K0
A4R2928	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A4R2929	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A4R2930	313-1152-00			RES,FXD,FILM:1.5K OHM,5%,0.2W	57668	TR20JE01K5
A4R2931	311-2258-00			RES,VAR,NONWW:TRMR,1K OHM,20%,0.5W	TK1450	GF06VT 1 K OHM
A4R2932	322-3414-00			RES,FXD,FILM:200K OHM,1%,0.2W,TC=T0	91637	CCF50G20002F
A4R2933	322-3385-00			RES,FXD,FILM:100K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 100K
A4R2934	322-3297-00			RES,FXD,FILM:12.1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 12K1
A4R2940	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A4R2945	313-1471-00			RES,FXD,FILM:470 OHM,5%,0.2W	57668	TR20JE 470E
A4R2975	313-1472-00			RES,FXD,FILM:4.7K OHM,5%,0.2W	57668	TR20JE 04K7
A4R2985	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0

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A4U2800	156-0514-00			MICROCKT,DGTL:CMOS,DIFF 4-CHANNEL MUX	02735	CD4052BF-98
A4U2805	156-0514-00			MICROCKT,DGTL:CMOS,DIFF 4-CHANNEL MUX	02735	CD4052BF-98
A4U2810	156-0382-00			MICROCKT,DGTL:QUAD 2-INP NAND GATE	01295	SN74LS00(N OR J)
A4U2820	156-1191-01			MICROCKT,LINEAR:BIFET,DUAL OPNL AMPL,SCRN	80009	156-1191-01
A4U2830	156-1172-00			MICROCKT,DGTL:DUAL 4 BIT BIN CNTR	80009	156-1172-00
A4U2835	156-0479-00			MICROCKT,DGTL:QUAD 2-INP OR GATE	80009	156-0479-00
A4U2850	156-0388-00			MICROCKT,DGTL:DUAL D FLIP-FLOP	01295	SN74LS74 N OR J
A4U2855	156-0383-00			MICROCKT,DGTL:QUAD 2-INP NOR GATE	01295	SN74LS02 N OR J
A4U2860	156-0975-00			MICROCKT,DGTL:UNIV SHIFT/STORAGE REGISTER	34335	SN74LS299N
A4U2865	156-0796-00			MICROCKT,DGTL:8 STG SHF & STORE BUS RGTR	02735	CD4094BF
A4U2870	156-1172-00			MICROCKT,DGTL:DUAL 4 BIT BIN CNTR	80009	156-1172-00
A4U2880	156-0388-00			MICROCKT,DGTL:DUAL D FLIP-FLOP	01295	SN74LS74 N OR J
A4U2885	156-0386-00			MICROCKT,DGTL:TRIPLE 3-INP NAND GATE	01295	SN74LS10(N OR J)
A4U2890	156-0382-00			MICROCKT,DGTL:QUAD 2-INP NAND GATE	01295	SN74LS00(N OR J)
A4U2900	156-0386-00			MICROCKT,DGTL:TRIPLE 3-INP NAND GATE	01295	SN74LS10(N OR J)
A4U2905	156-1702-00			MICROCKT,DGTL:STTL,10 BIT REGISTER	34335	AM29821DCB
A4U2910	156-1555-00			MICROCKT,LINEAR:D/A CONVERTER	34335	AM6080PC
A4U2920	156-1594-00			IC,MEMORY:NMOS,SRAM;2K X 8,150NS;,DIP24.6	65786	CY6116-55PC
A4U2930	160-1631-02			MICROCKT,DGTL:4096 X 8 EPROM,PRGM	80009	160-1631-02
A4U2935	156-0956-00			MICROCKT,DGTL:OCTAL BFR W/3 STATE OUT	18324	N74LS244(N OR F)
A4U2940	156-1172-00			MICROCKT,DGTL:DUAL 4 BIT BIN CNTR	80009	156-1172-00
A4U2950	156-0388-00			MICROCKT,DGTL:DUAL D FLIP-FLOP	01295	SN74LS74 N OR J
A4U2960	156-0796-00			MICROCKT,DGTL:8 STG SHF & STORE BUS RGTR	02735	CD4094BF
A4U2965	156-0382-00			MICROCKT,DGTL:QUAD 2-INP NAND GATE	01295	SN74LS00(N OR J)
A4U2970	156-0480-00			MICROCKT,DGTL:TTL,QUAD 2-INP AND GATE	80009	156-0480-00
A4U2980	156-0382-00			MICROCKT,DGTL:QUAD 2-INP NAND GATE	01295	SN74LS00(N OR J)
A4U2985	156-0768-00			MICROCKT,DGTL:BIDIRECT UNIV SR	01295	SN74LS194AN
A4U2990	156-0381-00			MICROCKT,DGTL:QUAD 2-INP ECXL OR GATE	01295	SN74LS86 N OR J
A4U2995	156-0651-00			MICROCKT,DGTL:8-BIT PRL-OUT SER SHF RGTR	80009	156-0651-00
A4VR2805	152-0217-00			SEMICOND DVC,DI:ZEN,SI,8.2V,5%,0.4W,DO-7	04713	SZG20
A4VR2925	152-0662-00			SEMICOND DVC,DI:ZEN,SI,5V,1%,400MW,DO-7	04713	SZG195RL
A4W411	175-4581-01			CA ASSY,SP,ELEC:26,28 AWG,2.25 L,RIBBON	22526	ORDER BY DESCR
A4W2851	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A4W2913	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07

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A5	670-9052-02	121001	152999	CIRCUIT BD ASSY:DIGITAL CONTROL	80009	670-9052-02
A5BT2570	146-0049-00			BATTERY,STORAGE:3.5V,750MAH	81855	LTC-7P
A5C2010	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A5C2011	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A5C2101	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2110	281-0814-00			CAP,FXD,CER DI:100 PF,10%,100V	04222	MA101A101KAA
A5C2111	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2112	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2113	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A5C2160	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2220	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2221	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2230	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2240	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2320	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2321	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2322	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A5C2330	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2331	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A5C2332	285-1300-01			CAP,FXD,MTLZD:0.1UF,10%,63V	55112	185/0.1/K/63/ABA
A5C2333	285-1300-01			CAP,FXD,MTLZD:0.1UF,10%,63V	55112	185/0.1/K/63/ABA
A5C2340	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2350	290-0527-00			CAP,FXD,ELCTLT:15UF,20%,20V	05397	T368B156M020AS
A5C2351	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2360	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2420	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A5C2421	285-1300-01			CAP,FXD,MTLZD:0.1UF,10%,63V	55112	185/0.1/K/63/ABA
A5C2422	281-0791-00			CAP,FXD,CER DI:270PF,10%,100V	04222	MA101C271KAA
A5C2430	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2431	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A5C2432	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A5C2450	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2470	290-0527-00			CAP,FXD,ELCTLT:15UF,20%,20V	05397	T368B156M020AS
A5C2501	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2510	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2511	281-0791-00			CAP,FXD,CER DI:270PF,10%,100V	04222	MA101C271KAA
A5C2520	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2521	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2530	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2550	281-0819-00			CAP,FXD,CER DI:33 PF,5%,50V	04222	GC105A330J
A5C2551	281-0816-00			CAP,FXD,CER DI:82 PF,5%,100V	04222	MA106A820JAA
A5C2552	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2601	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2610	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2620	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2621	285-1300-01			CAP,FXD,MTLZD:0.1UF,10%,63V	55112	185/0.1/K/63/ABA
A5C2622	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A5C2630	285-1300-01			CAP,FXD,MTLZD:0.1UF,10%,63V	55112	185/0.1/K/63/ABA
A5C2631	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A5C2632	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2640	285-1300-01			CAP,FXD,MTLZD:0.1UF,10%,63V	55112	185/0.1/K/63/ABA
A5C2650	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2660	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2720	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2721	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A5C2730	285-1348-00			CAP,FXD,MTLZD:0.22UF,10%,63V	TK1573	ORDER BY DESCR
A5C2731	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2732	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB

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A5C2733	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2740	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A5C2800	281-0812-00			CAP,FXD,CER DI:1000PF,10%,100V	04222	MA101C102KAA
A5CR2070	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2071	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2170	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2230	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2231	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2232	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2233	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2370	152-0322-00			SEMICOND DVC,DI:SCHOTTKY,SI,15V,1.2PF,DO-35	50434	5082-2672
A5CR2371	152-0322-00			SEMICOND DVC,DI:SCHOTTKY,SI,15V,1.2PF,DO-35	50434	5082-2672
A5CR2420	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2610	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2620	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2621	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2622	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2630	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2631	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2640	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A5CR2770	152-0322-00			SEMICOND DVC,DI:SCHOTTKY,SI,15V,1.2PF,DO-35	50434	5082-2672
A5J251	131-3360-00			CONN,RCPT,ELEC:HEADER,STR,20 PIN	53387	3592-6002
A5J500	131-3364-00			CONN,RCPT,ELEC:HEADER,STRAIGHT,34 PIN	53387	3594-6002
A5J501	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 3)	22526	48283-036
A5J503	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 3)	22526	48283-036
A5J651	131-3360-00			CONN,RCPT,ELEC:HEADER,STR,20 PIN	53387	3592-6002
A5J652	131-3360-00			CONN,RCPT,ELEC:HEADER,STR,20 PIN	53387	3592-6002
A5Q2070	151-0342-00			TRANSISTOR:PNP,SI,TO-92	07263	S035928
A5Q2270	151-0342-00			TRANSISTOR:PNP,SI,TO-92	07263	S035928
A5Q2320	151-0341-00			TRANSISTOR:NPN,SI,TO-106	04713	SPS6919
A5R2001	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2002	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2004	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2005	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2006	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2007	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2010	311-2234-00			RES,VAR,NONWW:TRMR,5K OHM,20%,0.5W LINEAR	TK1450	GF06UT 5K
A5R2011	322-3431-00			RES,FXD,FILM:301K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 301K
A5R2012	322-3289-02			RES,FXD,FILM:10K OHM,0.5%,0.2W,TC=T2	57668	CRB 20 DYE 10K0
A5R2013	322-3289-02			RES,FXD,FILM:10K OHM,0.5%,0.2W,TC=T2	57668	CRB 20 DYE 10K0
A5R2101	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2102	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2103	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2104	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2110	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2170	313-1512-00			RES,FXD,FILM:5.1K OHM,5%,0.2W	57668	TR20JE 5K1
A5R2170	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A5R2171	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2172	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2201	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2202	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2203	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2204	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2205	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2206	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2220	313-1681-00			RES,FXD,FILM:680 OHM,5%,0.2W	57668	TR20JE 680E

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A5R2230	322-3482-02			RES,FXD,FILM:14.2K OHM,0.5%,0.2W,TC=T0	57668	CRB 20 DYE 14K2
A5R2231	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2232	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2241	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2242	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2244	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2250	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2251	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2301	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2302	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2303	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2304	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2305	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2306	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2320	313-1203-00			RES,FXD,FILM:20K OHM,5%,0.2W	57668	TR20JE20K
A5R2330	322-3360-02			RES,FXD,FILM:54.9K OHM,0.5%,0.2W,TC=T2	57668	CRB20 DYE 54K9
A5R2331	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A5R2332	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A5R2333	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A5R2334	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A5R2340	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2341	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2342	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2343	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2344	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2345	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2346	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2370	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2401	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2402	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2403	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2404	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2405	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2406	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2407	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2408	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2409	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2410	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2411	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2412	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2413	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2414	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2415	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2416	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2417	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2420	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2421	322-3300-02			RES,FXD,FILM:13K OHM,0.5%,0.2W,TC=T2	57668	CRB20 DYE 13K0
A5R2422	322-3482-02			RES,FXD,FILM:14.2K OHM,0.5%,0.2W,TC=T0	57668	CRB 20 DYE 14K2
A5R2430	322-3289-02			RES,FXD,FILM:10K OHM,0.5%,0.2W,TC=T2	57668	CRB 20 DYE 10K0
A5R2431	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2432	322-3325-00			RES,FXD,FILM:23.7K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 23K7
A5R2433	322-3289-02			RES,FXD,FILM:10K OHM,0.5%,0.2W,TC=T2	57668	CRB 20 DYE 10K0
A5R2434	322-3289-02			RES,FXD,FILM:10K OHM,0.5%,0.2W,TC=T2	57668	CRB 20 DYE 10K0
A5R2440	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2441	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2442	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2443	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2444	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0

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A5R2470	313-1681-00			RES,FXD,FILM:680 OHM,5%,0.2W	57668	TR20JE 680E
A5R2471	313-1681-00			RES,FXD,FILM:680 OHM,5%,0.2W	57668	TR20JE 680E
A5R2501	313-1331-00			RES,FXD,FILM:330 OHM,5%,0.2W	57668	TR20JE 330E
A5R2502	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2503	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2504	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2505	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2506	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A5R2510	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2511	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2512	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2513	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2520	322-3177-02			RES,FXD,FILM:681 OHM,0.5%,0.2W,TC=T2	57668	CRB 20 DYE 681E
A5R2521	322-3177-02			RES,FXD,FILM:681 OHM,0.5%,0.2W,TC=T2	57668	CRB 20 DYE 681E
A5R2522	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2523	313-1683-00			RES,FXD,FILM:68K OHM,5%,0.2W	57668	TR20JE 68K
A5R2524	313-1683-00			RES,FXD,FILM:68K OHM,5%,0.2W	57668	TR20JE 68K
A5R2530	315-0106-00			RES,FXD,FILM:10M OHM,5%,0.25W	01121	CB1065
A5R2531	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2532	313-1683-00			RES,FXD,FILM:68K OHM,5%,0.2W	57668	TR20JE 68K
A5R2533	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A5R2534	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A5R2535	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A5R2536	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2537	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2540	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2541	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2542	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2543	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2544	313-1681-00			RES,FXD,FILM:680 OHM,5%,0.2W	57668	TR20JE 680E
A5R2545	313-1331-00			RES,FXD,FILM:330 OHM,5%,0.2W	57668	TR20JE 330E
A5R2560	313-1222-00			RES,FXD,FILM:2.2K OHM,5%,0.2W	57668	TR20JE 02K2
A5R2601	313-1331-00			RES,FXD,FILM:330 OHM,5%,0.2W	57668	TR20JE 330E
A5R2602	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2603	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2604	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A5R2610	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2611	313-1104-00			RES,FXD,FILM:100K OHM,5%,0.2W	57668	TR20JE100K
A5R2612	313-1512-00			RES,FXD,FILM:5.1K OHM,5%,0.2W	57668	TR20JE 5K1
A5R2613	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2620	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2621	313-1222-00			RES,FXD,FILM:2.2K OHM,5%,0.2W	57668	TR20JE 02K2
A5R2622	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2623	313-1222-00			RES,FXD,FILM:2.2K OHM,5%,0.2W	57668	TR20JE 02K2
A5R2624	313-1512-00			RES,FXD,FILM:5.1K OHM,5%,0.2W	57668	TR20JE 5K1
A5R2630	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A5R2631	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A5R2632	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A5R2640	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2641	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2642	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2643	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2644	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2645	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2660	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2661	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2701	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2702	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A5R2703	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2704	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2705	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2706	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2707	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2708	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2709	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2710	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2711	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2712	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2720	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2721	313-1203-00			RES,FXD,FILM:20K OHM,5%,0.2W	57668	TR20JE20K
A5R2730	313-1203-00			RES,FXD,FILM:20K OHM,5%,0.2W	57668	TR20JE20K
A5R2731	315-0107-00			RES,FXD,FILM:100M OHM,5%,0.25W	01121	CB1075
A5R2732	315-0107-00			RES,FXD,FILM:100M OHM,5%,0.25W	01121	CB1075
A5R2733	322-3235-00			RES,FXD,FILM:2.74K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 2K74
A5R2734	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2735	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A5R2740	322-3193-00			RES,FXD,FILM:1K OHM,1%,0.2W,TC=T0	57668	CRB20 FXE 1K00
A5R2741	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A5R2742	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5R2770	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A5TP2070	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL	22526	48283-036
A5TP2420	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL	22526	48283-036
A5TP2421	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL	22526	48283-036
A5TP2701	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL	22526	48283-036
A5U2101	156-1589-00			MICROCKT,LINEAR:D/A CONV,12 BIT,HS,MONO	06665	DAC312FR
A5U2140	156-1342-01			MICROCKT,DGTL:NMOS,MPU,8-BIT W/CLK	04713	SC67127P
A5U2160	160-5370-04			MICROCKT,DGTL:65536 X 8 EPROM,PRGM	80009	160-5370-04
A5U2201	156-0865-00			MICROCKT,DGTL:OCTAL D FF W/CLR	80009	156-0865-00
A5U2210	156-0391-00			MICROCKT,DGTL:LSTTL,HEX D TYPE FF W/CLEAR	04713	74LS174(N OR J)
A5U2220	156-0956-00			MICROCKT,DGTL:OCTAL BFR W/3 STATE OUT	18324	N74LS244(N OR F)
A5U2240	156-2396-00			MICROCKT,LINEAR:BIPOLAR,MPU RESET GENERATOR	01295	TL7705 ACP
A5U2250	160-5061-00			MICROCKT,DGTL:PROGRAMMABLE LOGIC DEVICE	80009	160-5061-00
A5U2260	160-5371-04			MICROCKT,DGTL:65536 X 8 EPROM,PRGM	80009	160-5371-04
A5U2301	156-0865-00			MICROCKT,DGTL:OCTAL D FF W/CLR	80009	156-0865-00
A5U2310	156-0865-00			MICROCKT,DGTL:OCTAL D FF W/CLR	80009	156-0865-00
A5U2350	156-0956-00			MICROCKT,DGTL:OCTAL BFR W/3 STATE OUT	18324	N74LS244(N OR F)
A5U2401	156-0513-03			MICROCKT,LINEAR:CMOS,8 CHAN ANALOG MUX	04713	MC14051BCL
A5U2410	156-1486-00			MICROCKT,DGTL:CMOS,8 CHANNEL DATA SEL	02735	CD4512BFX
A5U2420	156-1200-01			MICROCKT,LINEAR:BIFET,QUAD OPNL AMPL,SCRN	80009	156-1200-01
A5U2430	156-1200-01			MICROCKT,LINEAR:BIFET,QUAD OPNL AMPL,SCRN	80009	156-1200-01
A5U2440	156-0388-00			MICROCKT,DGTL:DUAL D FLIP-FLOP	01295	SN74LS74 N OR J
A5U2450	156-1065-00			MICROCKT,DGTL:OCTAL D TYPE TRANS LATCHES	01295	SN74LS373N
A5U2460	156-2473-00			IC,MEMORY:CMOS,SRAM;8K X 8,200NS,10UA	TK0961	uPD4464C-20
A5U2501	156-0513-03			MICROCKT,LINEAR:CMOS,8 CHAN ANALOG MUX	04713	MC14051BCL
A5U2510	156-1126-01			MICROCKT,LINEAR:VOLTAGE COMPARATOR,SELECTED	01295	LM311JG4
A5U2520	156-1191-01			MICROCKT,LINEAR:BIFET,DUAL OPNL AMPL,SCRN	80009	156-1191-01
A5U2521	156-0513-03			MICROCKT,LINEAR:CMOS,8 CHAN ANALOG MUX	04713	MC14051BCL
A5U2530	156-0513-03			MICROCKT,LINEAR:CMOS,8 CHAN ANALOG MUX	04713	MC14051BCL
A5U2540	156-1722-00			MICROCKT,DGTL:FTTL,HEX INVERTER	04713	MC74F04ND
A5U2550	156-0469-00			MICROCKT,DGTL:3-LINE TO 8-LINE DECODER	01295	SN74LS138N
A5U2601	156-0513-03			MICROCKT,LINEAR:CMOS,8 CHAN ANALOG MUX	04713	MC14051BCL
A5U2620	156-1200-01			MICROCKT,LINEAR:BIFET,QUAD OPNL AMPL,SCRN	80009	156-1200-01
A5U2630	156-1200-01			MICROCKT,LINEAR:BIFET,QUAD OPNL AMPL,SCRN	80009	156-1200-01
A5U2640	156-0895-00			MICROCKT,DGTL:14-BIT BINARY COUNTER	04713	MC14020BCL
A5U2650	156-0804-00			MICROCKT,DGTL:QUADRUPLE S-R LATCH	04713	74LS279(N OR J)
A5U2660	156-1026-00			MICROCKT,DGTL:4 LINE TO 1 LINE DECODER	18324	74LS154N

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A5VR2420	152-0278-00			SEMICOND DVC,DI:ZEN,SI,3V,5%,0.4W,DO-7	80009	152-0278-00
A5W511	174-0002-00			CA ASSY,SP,ELEC:26,28 AWG,2.0 L	80009	174-0002-00
A5W512	174-0001-00			CA ASSY,SP,ELEC:34,28 AWG,2.0 L	80009	174-0001-00
A5W2070	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A5W2540	131-1817-01			BUS,CONDUCTOR:22 AWG,2.0 TO 2.125 SPACING	TK1492	ORDER BY DESCR
A5W2610	131-1817-01			BUS,CONDUCTOR:22 AWG,2.0 TO 2.125 SPACING	TK1492	ORDER BY DESCR
A5W2701	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07
A5Y2540	158-0248-01			XTAL UNIT,QTZ:10.000MHZ,0.01% SER RESONANT	14301	011-669-02923

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A5	671-0965-00	152001		CIRCUIT BD ASSY:CONTROL/READOUT/BUFFER	80009	671-0965-00
A5C2010	290-5009-00			CAP,FXD,ELCTLT:15UF,25V	56289	293D156X0025D2T
A5C2011	290-5009-00			CAP,FXD,ELCTLT:15UF,25V	56289	293D156X0025D2T
A5C2101	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2110	283-5188-00			CAP,FXD,CER DI:100PF,5%,100V	04222	12061A101J1T050R
A5C2111	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2113	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A5C2160	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2220	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2221	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2222	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2230	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2240	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2241	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2250	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2321	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2322	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2323	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2324	283-5003-00			CAP,FXD,CER DI:0.01UF,10%,50V	14674	12065C103KAT060R
A5C2325	283-5003-00			CAP,FXD,CER DI:0.01UF,10%,50V	14674	12065C103KAT060R
A5C2330	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2331	290-0943-02			CAP,FXD,ELCTLT:47UF,20%,25V	55680	UVX1E470MAA1TD
A5C2332	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2333	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2350	290-5009-00			CAP,FXD,ELCTLT:15UF,25V	56289	293D156X0025D2T
A5C2352	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2360	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2415	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2420	290-5009-00			CAP,FXD,ELCTLT:15UF,25V	56289	293D156X0025D2T
A5C2421	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2422	283-5197-00			CAP,FXD,CER DI:330PF,5%,100V	TK2282	W1206C331J3B05
A5C2425	283-5003-00			CAP,FXD,CER DI:0.01UF,10%,50V	14674	12065C103KAT060R
A5C2430	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2431	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2432	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2433	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2434	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2440	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2450	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2451	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2452	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2460	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2465	283-5188-00			CAP,FXD,CER DI:100PF,5%,100V	04222	12061A101J1T050R
A5C2501	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2510	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2511	283-5197-00			CAP,FXD,CER DI:330PF,5%,100V	TK2282	W1206C331J3B05
A5C2520	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2521	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2530	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2540	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2542	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2550	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2610	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2621	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2622	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2623	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2629	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2630	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04

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A5C2631	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2632	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2633	283-5003-00			CAP,FXD,CER DI:0.01UF,10%,50V	14674	12065C103KAT060R
A5C2634	283-5003-00			CAP,FXD,CER DI:0.01UF,10%,50V	14674	12065C103KAT060R
A5C2640	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2641	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2650	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2720	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2721	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2722	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2730	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2731	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2732	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2733	285-1301-01			CAP,FXD,MTLZD:0.47UF,10%,50V	55112	1850.47K50ABB
A5C2734	283-5114-00			CAP,FXD,CER DI:0.1UF,10%,50V,X7R,1206 PKG	TK2282	W1206X104K2B04
A5C2820	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2821	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2830	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2831	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2835	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2836	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2850	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2851	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2855	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2860	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2861	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2870	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2875	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2885	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2890	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2901	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2905	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2911	283-5003-00			CAP,FXD,CER DI:0.01UF,10%,50V	14674	12065C103KAT060R
A5C2913	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2926	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2940	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2950	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2960	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2965	290-5009-00			CAP,FXD,ELCTLT:15UF,25V	56289	293D156X0025D2T
A5C2970	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2980	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2981	290-5009-00			CAP,FXD,ELCTLT:15UF,25V	56289	293D156X0025D2T
A5C2990	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5C2995	283-5098-00			CAP,FXD,CER DI:0.1UF,50WVDC	TK2282	W1206Z104Z2B04
A5CR2230	152-5004-00			SEMICOND DVC,DI:SI,SW,SER PR,70V	04713	BAV99T1
A5CR2332	152-5004-00			SEMICOND DVC,DI:SI,SW,SER PR,70V	04713	BAV99T1
A5CR2420	152-5004-00			SEMICOND DVC,DI:SI,SW,SER PR,70V	04713	BAV99T1
A5CR2421	152-5004-00			SEMICOND DVC,DI:SI,SW,SER PR,70V	04713	BAV99T1
A5CR2422	152-5004-00			SEMICOND DVC,DI:SI,SW,SER PR,70V	04713	BAV99T1
A5CR2423	152-5004-00			SEMICOND DVC,DI:SI,SW,SER PR,70V	04713	BAV99T1
A5CR2610	152-5005-00			SEMICOND DVC,DI:DUAL,COMMON ANODE,70V,BAW56	04713	MBAW56TI
A5CR2620	152-5005-00			SEMICOND DVC,DI:DUAL,COMMON ANODE,70V,BAW56	04713	MBAW56TI
A5CR2621	152-5005-00			SEMICOND DVC,DI:DUAL,COMMON ANODE,70V,BAW56	04713	MBAW56TI
A5CR2640	152-5005-00			SEMICOND DVC,DI:DUAL,COMMON ANODE,70V,BAW56	04713	MBAW56TI
A5J251	131-3360-00			CONN,RCPT,ELEC:HEADER,STR,20 PIN	53387	3592-6002
A5J411	131-3362-00			CONN,RCPT,ELEC:HEADER,STR,26 PIN	53387	3593-6002
A5J501	131-4671-00			CONN,RCPT,ELEC:1 X 3,0.1 SPACING	80009	131-4671-00
A5J503	131-4671-00			CONN,RCPT,ELEC:1 X 3,0.1 SPACING	80009	131-4671-00

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A5J504	131-4671-00			CONN,RCPT,ELEC:1 X 3,0.1 SPACING	80009	131-4671-00
A5J511	131-3362-00			CONN,RCPT,ELEC:HEADER,STR,26 PIN	53387	3593-6002
A5J512	131-3364-00			CONN,RCPT,ELEC:HEADER,STRAIGHT,34 PIN	53387	3594-6002
A5J651	131-3360-00			CONN,RCPT,ELEC:HEADER,STR,20 PIN	53387	3592-6002
A5J652	131-3360-00			CONN,RCPT,ELEC:HEADER,STR,20 PIN	53387	3592-6002
A5J4241	131-3323-00			CONN,RCPT,ELEC:HEADER,STR,2 X 20,0.1 CTR	22526	66506-025
A5J4330	131-3152-00			CONN,RCPT,ELEC:HEADER,2 X 8 0.1 SPACING	22526	66506-043
A5P501	131-0993-00			BUS,CONDUCTOR:SHUNT ASSEMBLY,BLACK	22526	65474-005
A5P503	131-0993-00			BUS,CONDUCTOR:SHUNT ASSEMBLY,BLACK	22526	65474-005
A5P504	131-0993-00			BUS,CONDUCTOR:SHUNT ASSEMBLY,BLACK	22526	65474-005
A5Q2320	151-5001-00			TRANSISTOR:NPN,SI,SOT-23	80009	151-5001-00
A5Q2805	151-5001-00			TRANSISTOR:NPN,SI,SOT-23	80009	151-5001-00
A5R2001	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2002	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2004	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2005	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2006	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2007	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2010	311-5038-00			RES,VAR,NONWW:TRMR,20K OHM,25%,0.1W	32997	3314A-1-203E
A5R2011	321-5026-00			RES,FXD,FILM:4.75K,1%,0.125W	01121	BCK4751FT
A5R2012	321-5165-00			RES,FXD,FILM:10K OHM,0.1%,0.125W,TC=T9	80009	321-5165-00
A5R2013	321-5165-00			RES,FXD,FILM:10K OHM,0.1%,0.125W,TC=T9	80009	321-5165-00
A5R2014	321-5167-00			RES,FXD,FILM:221K OHM,1%,0.125W	80009	321-5167-00
A5R2015	321-5041-00			RES,FXD,FILM:82.5K,1%,0.125W	01121	BCK8252FT
A5R2016	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2101	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2102	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2103	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2104	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2201	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2202	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2203	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2204	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2205	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2210	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2211	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2212	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2213	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2214	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2215	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2220	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2230	321-5165-00			RES,FXD,FILM:10K OHM,0.1%,0.125W,TC=T9	80009	321-5165-00
A5R2231	321-5022-00			RES,FXD,FILM:2.21K,1%,0.125W	01121	BCK2211FT
A5R2232	321-5022-00			RES,FXD,FILM:2.21K,1%,0.125W	01121	BCK2211FT
A5R2241	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2242	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2244	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2251	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2301	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2302	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2303	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2304	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2305	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2320	321-5034-00			RES,FXD,FILM:22.1K,1%,0.125W	01121	BCK2212FT
A5R2321	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2322	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2323	321-5032-00			RES,FXD,FILM:15.0K,1%,0.125W	01121	BCK1502FT
A5R2329	321-5036-00			RES,FXD,FILM:33.2K,1%,0.125W	01121	BCK3322FT

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A5R2330	321-5027-00			RES,FXD,FILM:5.62K,1%,0.125W	01121	BCK5621FT
A5R2331	321-5023-00			RES,FXD,FILM:2.74K,1%,0.125W	01121	BCK2741FT
A5R2332	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2333	321-5023-00			RES,FXD,FILM:2.74K,1%,0.125W	01121	BCK2741FT
A5R2334	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2340	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2341	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2342	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2343	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2344	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2345	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2346	321-5022-00			RES,FXD,FILM:2.21K,1%,0.125W	01121	BCK2211FT
A5R2401	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2402	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2403	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2404	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2405	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2406	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2407	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2408	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2409	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2410	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2411	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2412	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2413	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2414	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2415	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2416	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2417	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2420	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2421	321-5165-00			RES,FXD,FILM:10K OHM,0.1%,0.125W,TC=T9	80009	321-5165-00
A5R2422	321-5165-00			RES,FXD,FILM:10K OHM,0.1%,0.125W,TC=T9	80009	321-5165-00
A5R2423	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2424	321-5031-00			RES,FXD,FILM:12.1K,1%,0.125W	01121	BCK1212FT
A5R2430	321-5165-00			RES,FXD,FILM:10K OHM,0.1%,0.125W,TC=T9	80009	321-5165-00
A5R2431	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2432	321-5036-00			RES,FXD,FILM:33.2K,1%,0.125W	01121	BCK3322FT
A5R2433	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2434	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2435	321-5041-00			RES,FXD,FILM:82.5K,1%,0.125W	01121	BCK8252FT
A5R2440	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2441	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2442	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2443	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2444	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2461	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2465	321-5016-00			RES,FXD,FILM:681 OHM,1%,0.125W	01121	BCK6810FT
A5R2501	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2502	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2503	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2504	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2505	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2511	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2512	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2513	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2520	321-5164-00			RES,FXD,FILM:681 OHM,0.1%,0.125W,TC=T9	80009	321-5164-00
A5R2521	321-5164-00			RES,FXD,FILM:681 OHM,0.1%,0.125W,TC=T9	80009	321-5164-00
A5R2522	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A5R2523	321-5040-00			RES,FXD,FILM:68.1K,1%,0.125W	01121	BCK6812FT
A5R2524	321-5040-00			RES,FXD,FILM:68.1K,1%,0.125W	01121	BCK6812FT
A5R2531	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2532	321-5040-00			RES,FXD,FILM:68.1K,1%,0.125W	01121	BCK6812FT
A5R2533	321-5023-00			RES,FXD,FILM:2.74K,1%,0.125W	01121	BCK2741FT
A5R2534	321-5023-00			RES,FXD,FILM:2.74K,1%,0.125W	01121	BCK2741FT
A5R2535	321-5023-00			RES,FXD,FILM:2.74K,1%,0.125W	01121	BCK2741FT
A5R2536	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2537	321-5022-00			RES,FXD,FILM:2.21K,1%,0.125W	01121	BCK2211FT
A5R2540	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2560	321-5022-00			RES,FXD,FILM:2.21K,1%,0.125W	01121	BCK2211FT
A5R2601	321-5012-00			RES,FXD,FILM:332 OHM,1%,0.125W	01121	BCK3320FT
A5R2602	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2603	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2611	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2612	321-5026-00			RES,FXD,FILM:4.75K,1%,0.125W	01121	BCK4751FT
A5R2613	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2620	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2621	321-5022-00			RES,FXD,FILM:2.21K,1%,0.125W	01121	BCK2211FT
A5R2622	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2623	321-5022-00			RES,FXD,FILM:2.21K,1%,0.125W	01121	BCK2211FT
A5R2624	321-5026-00			RES,FXD,FILM:4.75K,1%,0.125W	01121	BCK4751FT
A5R2625	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2626	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2630	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2631	321-5023-00			RES,FXD,FILM:2.74K,1%,0.125W	01121	BCK2741FT
A5R2632	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2640	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2643	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2644	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2645	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2646	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2647	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2648	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2649	321-5012-00			RES,FXD,FILM:332 OHM,1%,0.125W	01121	BCK3320FT
A5R2701	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2702	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2703	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2704	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2705	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2706	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2707	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2708	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2709	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2710	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2711	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2712	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2720	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2721	321-5034-00			RES,FXD,FILM:22.1K,1%,0.125W	01121	BCK2212FT
A5R2730	321-5034-00			RES,FXD,FILM:22.1K,1%,0.125W	01121	BCK2212FT
A5R2731	321-5199-00			RES,FXD,FILM:100M OHM,10%,0.0625 W	80009	321-5199-00
A5R2732	321-5199-00			RES,FXD,FILM:100M OHM,10%,0.0625 W	80009	321-5199-00
A5R2733	321-5023-00			RES,FXD,FILM:2.74K,1%,0.125W	01121	BCK2741FT
A5R2734	321-5022-00			RES,FXD,FILM:2.21K,1%,0.125W	01121	BCK2211FT
A5R2735	321-5022-00			RES,FXD,FILM:2.21K,1%,0.125W	01121	BCK2211FT
A5R2740	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2741	321-5006-00			RES,FXD,FILM:100 OHM,1%,0.125W	01121	BCK1000FT
A5R2830	321-5012-00	B050000	B050689	RES,FXD,FILM:332 OHM,1%,0.125W	01121	BCK3320FT

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A5R2830	321-5051-00	B050690		RES,FXD,FILM:0 OHM,1%,0.125W	80009	321-5051-00
A5R2865	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2866	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2885	321-5018-00	B050000	B050352	RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2885	321-5012-00	B050353		RES,FXD,FILM:332 OHM,1%,0.125W	01121	BCK3320FT
A5R2890	321-5018-00	B050000	B050352	RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2890	321-5012-00	B050353		RES,FXD,FILM:332 OHM,1%,0.125W	01121	BCK3320FT
A5R2902	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2903	321-5165-00			RES,FXD,FILM:10K OHM,0.1%,0.125W,TC=T9	80009	321-5165-00
A5R2904	321-5051-00	B050000	B050689	RES,FXD,FILM:0 OHM,1%,0.125W	80009	321-5051-00
A5R2904	321-0512-00	B050690		RES,FXD,FILM:2.1M OHM,1%,0.125W,TC=T0	80009	321-0512-00
A5R2905	321-5028-00			RES,FXD,FILM:6.81K,1%,0.125W	01121	BCK6811FT
A5R2906	321-5165-00			RES,FXD,FILM:10K OHM,0.1%,0.125W,TC=T9	80009	321-5165-00
A5R2907	321-5033-00			RES,FXD,FILM:18.2K,1%,0.125W	01121	BCK1822FT
A5R2908	321-5032-00			RES,FXD,FILM:15.0K,1%,0.125W	01121	BCK1502FT
A5R2909	321-5032-00			RES,FXD,FILM:15.0K,1%,0.125W	01121	BCK1502FT
A5R2910	321-5032-00			RES,FXD,FILM:15.0K,1%,0.125W	01121	BCK1502FT
A5R2911	321-5032-00			RES,FXD,FILM:15.0K,1%,0.125W	01121	BCK1502FT
A5R2912	321-5018-00			RES,FXD,FILM:1.00K,1%,0.125W	01121	BCK1001FT
A5R2913	321-5015-00			RES,FXD,FILM:562 OHM,1%,0.125W	01121	BCK5620FT
A5R2914	321-5032-00			RES,FXD,FILM:15.0K,1%,0.125W	01121	BCK1502FT
A5R2915	321-5015-00			RES,FXD,FILM:562 OHM,1%,0.125W	01121	BCK5620FT
A5R2916	321-5064-00			RES,FXD,FILM:200K,1%,0.125W,1206,8MM	80009	321-5064-00
A5R2917	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2918	311-5038-00			RES,VAR,NONWW:TRMR,20K OHM,25%,0.1W	32997	3314A-1-203E
A5R2919	321-5038-00			RES,FXD,FILM:47.5K,1%,0.125W	01121	BCK4752FT
A5R2920	321-5064-00			RES,FXD,FILM:200K,1%,0.125W,1206,8MM	80009	321-5064-00
A5R2921	321-5031-00			RES,FXD,FILM:12.1K,1%,0.125W	01121	BCK1212FT
A5R2922	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2923	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2924	321-5064-00			RES,FXD,FILM:200K,1%,0.125W,1206,8MM	80009	321-5064-00
A5R2925	321-5023-00			RES,FXD,FILM:2.74K,1%,0.125W	01121	BCK2741FT
A5R2926	321-5020-00			RES,FXD,FILM:1.50K,1%,0.125W	01121	BCK1501FT
A5R2927	321-5026-00			RES,FXD,FILM:4.75K,1%,0.125W	01121	BCK4751FT
A5R2928	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2929	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2930	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2931	311-5040-00			RES,VAR,NONWW:TRMR,10K OHM,25%,0.1W	32997	3314J-1-103E
A5R2932	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2933	321-5064-00			RES,FXD,FILM:200K,1%,0.125W,1206,8MM	80009	321-5064-00
A5R2934	321-5064-00			RES,FXD,FILM:200K,1%,0.125W,1206,8MM	80009	321-5064-00
A5R2935	321-5047-00			RES,FXD,FILM:100K,1%,0.125W	01121	BCK1003FT
A5R2960	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2961	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5R2995	321-5030-00			RES,FXD,FILM:10.0K,1%,0.125W	01121	BCK1002FT
A5U2101	156-5157-01			MICROCKT,INTFC:DAC,BIPOLAR,12 BIT	80009	156-5157-01
A5U2140	156-1342-01			MICROCKT,DGTL:NMOS,MPU,8-BIT W/CLK	04713	SC67127P
A5U2160	160-5876-00			MICROCKT,DGTL:8K X 8 EPROM,PRGM	80009	160-5876-00
A5U2201	156-5147-01			MICROCKT,DGTL:CMOS,OCTAL D TYPE FF W/RESET	80009	156-5147-01
A5U2210	156-5147-01			MICROCKT,DGTL:CMOS,OCTAL D TYPE FF W/RESET	80009	156-5147-01
A5U2220	156-5071-01			MICROCKT,DGTL:CMOS,OCTAL BUS TRANS	80009	156-5071-01
A5U2240	156-5489-01			MICROCKT,LINEAR:MPU RESET GEN FOR 5V SYS	80009	156-5489-01
A5U2250	160-5874-00			MICROCKT,DGTL:LOGIC DEVICE,PRGM	80009	160-5874-00
A5U2301	156-5147-01			MICROCKT,DGTL:CMOS,OCTAL D TYPE FF W/RESET	80009	156-5147-01
A5U2310	156-5147-01			MICROCKT,DGTL:CMOS,OCTAL D TYPE FF W/RESET	80009	156-5147-01
A5U2350	156-5071-01			MICROCKT,DGTL:CMOS,OCTAL BUS TRANS	80009	156-5071-01
A5U2360	160-5877-00			MICROCKT,DGTL:16K X 8 X 8 EPROM,PRGM	80009	160-5877-00
A5U2401	156-5050-01			MICROCKT,DGTL:HCMOS,ANALOG SW,8 CHAN	80009	156-5050-01

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A5U2405	156-5409-01			MICROCKT,DGTL:HCMOS,OCTAL D-TYPE TRANS	80009	156-5409-01
A5U2410	156-5459-01			MICROCKT,DGTL:CMOS,OCTAL BUS TRANSCEIVER,	80009	156-5459-01
A5U2415	156-5409-01			MICROCKT,DGTL:HCMOS,OCTAL D-TYPE TRANS	80009	156-5409-01
A5U2420	156-2051-01			MICROCKT,LINEAR:OPNL AMPL,QUAD,JET INPUT	80009	156-2051-01
A5U2425	156-5409-01			MICROCKT,DGTL:HCMOS,OCTAL D-TYPE TRANS	80009	156-5409-01
A5U2430	156-2051-01			MICROCKT,LINEAR:OPNL AMPL,QUAD,JET INPUT	80009	156-2051-01
A5U2440	156-5145-01			MICROCKT,DGTL:HCMOS,DUAL D-TYPE FF	80009	156-5145-01
A5U2450	156-5409-01			MICROCKT,DGTL:HCMOS,OCTAL D-TYPE TRANS	80009	156-5409-01
A5U2460	156-2991-00			IC,MEMORY:CMOS,NVRAM;8K X 8,200NS,SRAM	80009	156-2991-00
A5U2501	156-5050-01			MICROCKT,DGTL:HCMOS,ANALOG SW,8 CHAN	80009	156-5050-01
A5U2510	156-5000-01			MICROCKT,LINEAR:VOLTAGE COMPARATOR	80009	156-5000-01
A5U2520	156-5138-01			MICROCKT,LINEAR:OP AMP,BIFET,DUAL	80009	156-5138-01
A5U2521	156-5050-01			MICROCKT,DGTL:HCMOS,ANALOG SW,8 CHAN	80009	156-5050-01
A5U2530	156-5050-01			MICROCKT,DGTL:HCMOS,ANALOG SW,8 CHAN	80009	156-5050-01
A5U2540	156-5081-01			MICROCKT,DGTL:HCMOS,HEX INVERTER	80009	156-5081-01
A5U2550	156-5088-01			MICROCKT,DGTL:CMOS,3 TO 8 DECODER/	80009	156-5088-01
A5U2560	156-5145-01			MICROCKT,DGTL:HCMOS,DUAL D-TYPE FF	80009	156-5145-01
A5U2570	156-5145-01			MICROCKT,DGTL:HCMOS,DUAL D-TYPE FF	80009	156-5145-01
A5U2601	156-5050-01			MICROCKT,DGTL:HCMOS,ANALOG SW,8 CHAN	80009	156-5050-01
A5U2620	156-2051-01			MICROCKT,LINEAR:OPNL AMPL,QUAD,JET INPUT	80009	156-2051-01
A5U2630	156-2051-01			MICROCKT,LINEAR:OPNL AMPL,QUAD,JET INPUT	80009	156-2051-01
A5U2640	156-5567-01			MICROCKT,DGTL:CMOS,14 STAGES BIN CNTR	80009	156-5567-01
A5U2650	156-5088-01			MICROCKT,DGTL:CMOS,3 TO 8 DECODER/	80009	156-5088-01
A5U2660	156-5088-01			MICROCKT,DGTL:CMOS,3 TO 8 DECODER/	80009	156-5088-01
A5U2800	156-5120-01			MICROCKT,DGTL:CMOS,DUAL 4 CHAN ANALOG MUX	80009	156-5120-01
A5U2805	156-5120-01			MICROCKT,DGTL:CMOS,DUAL 4 CHAN ANALOG MUX	80009	156-5120-01
A5U2810	156-5098-01			MICROCKT,DGTL:HCMOS,QUAD 2-INPUT NAND GATE	80009	156-5098-01
A5U2820	156-2051-01			MICROCKT,LINEAR:OPNL AMPL,QUAD,JET INPUT	80009	156-2051-01
A5U2830	156-5306-01			MICROCKT,DGTL:CMOS,DUAL 4 BIT	80009	156-5306-01
A5U2835	156-5085-01			MICROCKT,DGTL:CMOS,QUAD 2-INPUT OR GATE	80009	156-5085-01
A5U2850	156-5145-01			MICROCKT,DGTL:HCMOS,DUAL D-TYPE FF	80009	156-5145-01
A5U2855	156-5106-01			MICROCKT,DGTL:CMOS,QUAD 2 INPUT N OR GATE	80009	156-5106-01
A5U2860	156-5569-01			MICROCKT,DGTL:CMOS,8-BIT UNIVERSIAL SHIFT	80009	156-5569-01
A5U2865	156-5021-01			MICROCKT,DGTL:CMOS,8 STATE SHIFT ANS STOR	80009	156-5021-01
A5U2870	156-5306-01			MICROCKT,DGTL:CMOS,DUAL 4 BIT	80009	156-5306-01
A5U2875	156-5145-01			MICROCKT,DGTL:HCMOS,DUAL D-TYPE FF	80009	156-5145-01
A5U2880	156-5145-01			MICROCKT,DGTL:HCMOS,DUAL D-TYPE FF	80009	156-5145-01
A5U2885	156-5130-01			MICROCKT,DGTL:CMOS,TRIPLE 3-INPUT N AND D	80009	156-5130-01
A5U2890	156-5098-01			MICROCKT,DGTL:HCMOS,QUAD 2-INPUT NAND GATE	80009	156-5098-01
A5U2900	156-5130-01			MICROCKT,DGTL:CMOS,TRIPLE 3-INPUT N AND D	80009	156-5130-01
A5U2905	156-5147-01			MICROCKT,DGTL:CMOS,OCTAL D TYPE FF W/RESET	80009	156-5147-01
A5U2910	156-1555-00			MICROCKT,LINEAR:D/A CONVERTER	34335	AM6080PC
A5U2920	156-5011-00			IC,MEMORY:CMOS,SRAM;8K X 8,150NS	62786	HM6264LFP-15
A5U2930	160-5875-00			MICROCKT,DGTL:8K X 8 EPROM,PRGM	80009	160-5875-00
A5U2935	156-5071-01			MICROCKT,DGTL:CMOS,OCTAL BUS TRANS	80009	156-5071-01
A5U2940	156-5306-01			MICROCKT,DGTL:CMOS,DUAL 4 BIT	80009	156-5306-01
A5U2950	156-5145-01			MICROCKT,DGTL:HCMOS,DUAL D-TYPE FF	80009	156-5145-01
A5U2960	156-5021-01			MICROCKT,DGTL:CMOS,8 STATE SHIFT ANS STOR	80009	156-5021-01
A5U2965	156-5098-01			MICROCKT,DGTL:HCMOS,QUAD 2-INPUT NAND GATE	80009	156-5098-01
A5U2970	156-5098-01			MICROCKT,DGTL:HCMOS,QUAD 2-INPUT NAND GATE	80009	156-5098-01
A5U2975	156-5098-01			MICROCKT,DGTL:HCMOS,QUAD 2-INPUT NAND GATE	80009	156-5098-01
A5U2980	156-5098-01			MICROCKT,DGTL:HCMOS,QUAD 2-INPUT NAND GATE	80009	156-5098-01
A5U2985	156-5568-01			MICROCKT,DGTL:HCMOS,4-BIT BIDIRECTIONAL	80009	156-5568-01
A5U2990	156-5198-01			MICROCKT,DGTL:CMOS,QUAD 2-INPUT X OR GATE	80009	156-5198-01
A5U2995	156-5135-01			MICROCKT,DGTL:CMOS,8 BIT SER/PAR SHIFT	80009	156-5135-01
A5W411	174-1366-00			CA ASSY,SP,ELEC:26,28 AWG,3.0 L	TK1899	ORDER BY DESCR
A5W511	174-1501-00			CA ASSY,SP,ELEC:26,28 AWG,2.0 L,RIBBON	80009	174-1501-00
A5W512	174-1502-00			CA ASSY,SP,ELEC:34,28 AWG,2.0 L,RIBBON	80009	174-1502-00

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A5XU236O	136-0755-00			SKT,PL-IN ELEK:MICROCIRCUIT,28 DIP	09922	DILB28P-108
A5Y2540	158-5005-00			OSC,XTAL CLOCK:10MHZ	80009	158-5005-00

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A6	614-0825-00			FRONT PNL ASSY:STANDARD,2445B/55B/65B & 67B (STANDARD)	80009	614-0825-00
A6	614-0826-00			FRONT PNL ASSY:TV OPTION,2445B/55B/65B/67B (OPTION 05)	80009	614-0826-00
A6P3001	131-3478-01			CONN,RCPT,ELEC:VERT,2 X 10,0.1 SPACING	80009	131-3478-01
A6R3007	311-2318-00			RES,VAR,NONWW:5K OHM,30%,0.5W	32997	ORDER BY DESCR
A6R3008	311-2316-00			RES,VAR,NONWW:2K OHM,20%,0.5W	32997	ORDER BY DESCR
A6R3009	311-2317-00			RES,VAR,NONWW:5K OHM,30%,0.25W	32997	ORDER BY DESCR
A6R3010	311-2318-00			RES,VAR,NONWW:5K OHM,30%,0.5W	32997	ORDER BY DESCR
A6R3011	311-2316-00			RES,VAR,NONWW:2K OHM,20%,0.5W	32997	ORDER BY DESCR
A6R3012	311-2317-00			RES,VAR,NONWW:5K OHM,30%,0.25W	32997	ORDER BY DESCR
A6R3013	311-2316-00			RES,VAR,NONWW:2K OHM,20%,0.5W	32997	ORDER BY DESCR
A6R3014	311-2318-00			RES,VAR,NONWW:5K OHM,30%,0.5W	32997	ORDER BY DESCR
A6R3015	311-2316-00			RES,VAR,NONWW:2K OHM,20%,0.5W	32997	ORDER BY DESCR
A6R3016	311-2316-00			RES,VAR,NONWW:2K OHM,20%,0.5W	32997	ORDER BY DESCR
A6R3017	311-2316-00			RES,VAR,NONWW:2K OHM,20%,0.5W	32997	ORDER BY DESCR
A6R3018	311-2318-00			RES,VAR,NONWW:5K OHM,30%,0.5W	32997	ORDER BY DESCR
A6R3019	311-2316-00			RES,VAR,NONWW:2K OHM,20%,0.5W	32997	ORDER BY DESCR

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A6A1	----- -----			CIRCUIT BD ASSY:FRONT PANEL (REPLACEABLE AT A6 LEVEL ONLY)		
A6A1C3001	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A6A1C3002	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A6A1C3019	281-0909-00			CAP,FXD,CER DI:0.022UF,20%,50V	54583	MA12X7R1H223M-T
A6A1CR3001	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3002	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3003	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3004	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3005	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3006	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3007	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3008	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3009	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3010	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3011	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3012	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3013	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3014	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3015	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3016	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3017	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3018	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3019	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3020	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3021	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3022	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3023	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3024	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3025	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3026	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3027	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3028	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3029	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3030	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3031	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3032	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3033	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3034	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3035	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3036	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3037	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3038	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3039	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3040	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3041	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3042	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3043	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1CR3044	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A6A1DS3001	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3002	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3003	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3004	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3005	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3006	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3007	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3008	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3009	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A6A1DS3010	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3011	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3012	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3013	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3014	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3015	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3016	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3017	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3018	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3019	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3020	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3021	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3022	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3023	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3024	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3025	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3026	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3027	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3028	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3029	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3030	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3031	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3032	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3033	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3034	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3035	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3036	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3037	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3038	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3039	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3040	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3041	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3042	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3043	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3044	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3045	150-1161-00			LT EMITTING DIO:YELLOW	50434	QLMP 1487
A6A1DS3046	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3047	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3048	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1DS3049	150-1160-00			LT EMITTING DIO:GREEN	50434	QLMP 1587
A6A1R3001	307-0486-00			RES NTWK,FXD,FI:100 OHM,20%,1.125W	11236	750-101-R100 OHM
A6A1R3002	307-0695-00			RES NTWK,FXD,FI:9,150 OHM,2%,0.2W EA	11236	750-101-R150 OHM
A6A1R3003	307-0486-00			RES NTWK,FXD,FI:100 OHM,20%,1.125W	11236	750-101-R100 OHM
A6A1R3004	313-1151-00			RES,FXD,FILM:150 OHM,5%,0.2W	57668	TR20JE150E
A6A1R3005	313-1151-00			RES,FXD,FILM:150 OHM,5%,0.2W	57668	TR20JE150E
A6A1R3006	313-1101-00			RES,FXD,FILM:100 OHM,5%,0.2W	57668	TR20JE100E
A6A1S3001	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3002	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3003	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3004	260-2283-00			SWITCH,ROTARY:VOLTS/DIV	80009	260-2283-00
A6A1S3005	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3006	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3007	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3008	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3009	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3010	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3011	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3012	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A6A1S3013	260-2283-00			SWITCH,ROTARY:VOLTS/DIV	80009	260-2283-00
A6A1S3014	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3015	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3016	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3017	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3018	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3019	260-2283-00			SWITCH,ROTARY:VOLTS/DIV	80009	260-2283-00
A6A1S3020	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3021	260-2164-01			SWITCH,SLIDE:SPDT,4A,20VAC	09353	1101 AV2 BE2
A6A1S3022	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3023	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3024	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3025	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3026	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3027	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3028	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3029	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3030	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3031	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3032	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3033	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3034	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1S3035	260-2280-00			SW,PUSH BUTTON:MINI MOM.SPST,NORM OPEN	80009	260-2280-00
A6A1U3001	156-2120-00			MICROCKT,DGTL:SER-IN PRL-OUT SHIFT RGTR	80009	156-2120-00
A6A1U3002	156-2120-00			MICROCKT,DGTL:SER-IN PRL-OUT SHIFT RGTR	80009	156-2120-00
A6A1U3003	156-2120-00			MICROCKT,DGTL:SER-IN PRL-OUT SHIFT RGTR	80009	156-2120-00
A6A1U3004	156-2120-00			MICROCKT,DGTL:SER-IN PRL-OUT SHIFT RGTR	80009	156-2120-00
A6A1U3005	156-2120-00			MICROCKT,DGTL:SER-IN PRL-OUT SHIFT RGTR	80009	156-2120-00
A6A1U3006	156-2120-00			MICROCKT,DGTL:SER-IN PRL-OUT SHIFT RGTR	80009	156-2120-00

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A8	670-7280-00			CIRCUIT BD ASSY:SCALE ILLUM	80009	670-7280-00
A8DS100	150-0057-01			LAMP,INCAND:5V,0.115A,WIRE LD,AGED & SEL	71744	7153 AS 15
A8DS101	150-0057-01			LAMP,INCAND:5V,0.115A,WIRE LD,AGED & SEL	71744	7153 AS 15
A8DS102	150-0057-01			LAMP,INCAND:5V,0.115A,WIRE LD,AGED & SEL	71744	7153 AS 15

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A9	670-7277-09			CIRCUIT BD ASSY:HIGH VOLTAGE	80009	670-7277-09
A9C91	283-0084-00			CAP,FXD,CER DI:270PF,5%,1000V	59660	838533X5F02715
A9C1812	285-1430-00			CAP,FXD,PLASTIC:0.047UF,10%,400V	80009	285-1430-00
A9C1813	285-1430-00			CAP,FXD,PLASTIC:0.047UF,10%,400V	80009	285-1430-00
A9C1814	285-1430-00			CAP,FXD,PLASTIC:0.047UF,10%,400V	80009	285-1430-00
A9C1815	285-1430-00			CAP,FXD,PLASTIC:0.047UF,10%,400V	80009	285-1430-00
A9C1870	281-0773-00			CAP,FXD,CER DI:0.01UF,10%,100V	04222	MA201C103KAA
A9C1885	285-1430-00			CAP,FXD,PLASTIC:0.047UF,10%,400V	80009	285-1430-00
A9C1886	285-1430-00			CAP,FXD,PLASTIC:0.047UF,10%,400V	80009	285-1430-00
A9C1888	285-1430-00			CAP,FXD,PLASTIC:0.047UF,10%,400V	80009	285-1430-00
A9C1889	285-1430-00			CAP,FXD,PLASTIC:0.047UF,10%,400V	80009	285-1430-00
A9C1890	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A9C1891	281-0775-00			CAP,FXD,CER DI:0.1UF,20%,50V	04222	MA205E104MAA
A9C1912	281-0798-00			CAP,FXD,CER DI:51PF,1%,100V	04222	MA101A510GAA
A9C1915	281-0783-00			CAP,FXD,CER DI:0.1 UF 20%,100V	04222	MA401C104MAA
A9C1932	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A9C1950	281-0766-00			CAP,FXD,CER DI:100PF,20%,200V	04222	MA106A101MAA
A9C1951	290-0269-01			CAP,FXD,ELCTLT:0.22UF,5%,35V,1KHZ,TANTULUM	56289	173D224X5035U
A9C1971	285-1430-00			CAP,FXD,PLASTIC:0.047UF,10%,400V	80009	285-1430-00
A9C1972	290-0747-00			CAP,FXD,ELCTLT:100UF,+50-20%,25WVDC	54473	ECE-B25V100L
A9C1973	281-0826-00			CAP,FXD,CER DI:2200PF,10%,100V	20932	401EM100AD222K
A9C1980	281-0771-00			CAP,FXD,CER DI:2200PF,20%,200V	04222	SA106E222MAA
A9C1990	285-1096-00			CAP,FXD,PLASTIC:1UF,10%,50V	14752	230B1A105K
A9C1991	281-0826-00			CAP,FXD,CER DI:2200PF,10%,100V	20932	401EM100AD222K
A9CR1894	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A9CR1895	152-0400-00			SEMICOND DVC,DI:RECT,SI,400V,1A	04713	SR1977KRL
A9CR1915	152-0061-00			SEMICOND DVC,DI:SW,SI,175V,0.1A,DO-35	07263	FDH2161
A9CR1930	152-0061-00			SEMICOND DVC,DI:SW,SI,175V,0.1A,DO-35	07263	FDH2161
A9CR1950	152-0061-00			SEMICOND DVC,DI:SW,SI,175V,0.1A,DO-35	07263	FDH2161
A9CR1953	152-0061-00			SEMICOND DVC,DI:SW,SI,175V,0.1A,DO-35	07263	FDH2161
A9CR1990	152-0141-02			SEMICOND DVC,DI:SW,SI,30V,150MA,30V,DO-35	03508	DA2527 (1N4152)
A9DS90	150-0030-00			LAMP,GLOW:60-90V MAX,0.6MA,A28-T,WIRE LEADS	58224	A2B-T
A9DS91	150-0030-00			LAMP,GLOW:60-90V MAX,0.6MA,A28-T,WIRE LEADS	58224	A2B-T
A9J901	131-0589-00			TERMINAL,PIN:0.46 L X 0.025 SQ,PH BRZ (QUANTITY OF 9)	22526	48283-029
A9J902	131-0589-00			TERMINAL,PIN:0.46 L X 0.025 SQ,PH BRZ (QUANTITY OF 2)	22526	48283-029
A9J903	131-0589-00			TERMINAL,PIN:0.46 L X 0.025 SQ,PH BRZ (QUANTITY OF 2)	22526	48283-029
A9J904	131-0589-00			TERMINAL,PIN:0.46 L X 0.025 SQ,PH BRZ	22526	48283-029
A9L1921	108-0262-00			COIL,RF:FIXED,505NH	80009	108-0262-00
A9L1974	108-0318-00			COIL,RF:FIXED,100UH	32159	81000M
A9P191	131-3461-00			HEADER,MICROCKT:14 PIN,0.5 L,GOLD PL (SUBPART OF A9W900)	80009	131-3461-00
A9P900	----- -----					
A9Q1851	151-0443-00			TRANSISTOR:PNP,SI,TO-92	04713	SPS7950
A9Q1852	151-0443-00			TRANSISTOR:PNP,SI,TO-92	04713	SPS7950
A9Q1890	151-0443-00			TRANSISTOR:PNP,SI,TO-92	04713	SPS7950
A9Q1891	151-0745-00			TRANSISTOR:PNP,SI,TO-220	61271	2SA1077G
A9Q1980	151-0444-00			TRANSISTOR:NPN,SI,TO-92	04713	SPS797
A9R1812	315-0100-02			RES,FXD,CMPSN:10 OHM,5%,0.25W	01121	CB1005
A9R1813	315-0100-02			RES,FXD,CMPSN:10 OHM,5%,0.25W	01121	CB1005
A9R1814	315-0103-00			RES,FXD,FILM:10K OHM,5%,0.25W	19701	5043CX10K00J
A9R1815	315-0103-00			RES,FXD,FILM:10K OHM,5%,0.25W	19701	5043CX10K00J
A9R1833	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A9R1834	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A9R1842	311-2234-00			RES,VAR,NONWW:TRMR,5K OHM,20%,0.5W LINEAR	TK1450	GF06UT 5K
A9R1848	311-2234-00			RES,VAR,NONWW:TRMR,5K OHM,20%,0.5W LINEAR	TK1450	GF06UT 5K
A9R1853	321-0447-00			RES,FXD,FILM:442K OHM,1%,0.125W,TC=T0	24546	NA55D4423F

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A9R1854	321-0435-00			RES,FXD,FILM:332K OHM,1%,0.125W,TC=T0	07716	CEAD33202F
A9R1855	321-0407-00			RES,FXD,FILM:169K OHM,1%,0.125W,TC=T0	07716	CEAD16902F
A9R1856	321-0367-00			RES,FXD,FILM:64.9K OHM,1%,0.125W,TC=T0	07716	CEAD64901F
A9R1857	321-0364-00			RES,FXD,FILM:60.4K OHM,1%,0.125W,TC=T0	19701	5043ED60K40F
A9R1858	313-1105-00			RES,FXD,FILM:1M OHM,5%,0.2W	57668	TR20JE1M
A9R1864	311-2236-00			RES,VAR,NONWW:TRMR,20K OHM,20%,0.5W LINEAR	TK1450	GF06UT 20K
A9R1870	311-2239-00			RES,VAR,NONWW:TRMR,100K OHM,20%,0.5W LINEAR	TK1450	GF06UT 100K
A9R1871	315-0154-00			RES,FXD,FILM:150K OHM,5%,0.25W	57668	NTR25J-E150K
A9R1872	315-0184-00			RES,FXD,FILM:180K OHM,5%,0.25W	19701	5043CX180K0J
A9R1873	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A9R1878	311-2239-00			RES,VAR,NONWW:TRMR,100K OHM,20%,0.5W LINEAR	TK1450	GF06UT 100K
A9R1880	315-0434-00			RES,FXD,FILM:430K OHM,5%,0.25W	57668	NTR25J-E430K
A9R1881	321-0385-00			RES,FXD,FILM:100K OHM,1%,0.125W,TC=T0	19701	5033ED100K0F
A9R1885	315-0103-00			RES,FXD,FILM:10K OHM,5%,0.25W	19701	5043CX10K00J
A9R1888	315-0100-02			RES,FXD,CMPSN:10 OHM,5%,0.25W	01121	CB1005
A9R1890	313-1473-00			RES,FXD,FILM:47K OHM,5%,0.2W	57668	TR20JE 47K
A9R1891	321-0481-04			RES,FXD,FILM:1M OHM,0.1%,0.125W,TC=T2	91637	CMF55116D10003B
A9R1892	321-0639-00			RES,FXD,FILM:9.6K OHM,1%,0.125W,TC=T0	19701	5043ED9K600F
A9R1892	321-0693-00			RES,FXD,FILM:68.1K OHM,0.5%,0.125W,TC=T0	19701	5033RD6812DB2980
A9R1893	321-0481-04			RES,FXD,FILM:1M OHM,0.1%,0.125W,TC=T2	91637	CMF55116D10003B
A9R1895	313-1302-00			RES,FXD,FILM:3K OHM,5%,0.2W	57668	TR20JE 03K0
A9R1896	315-0100-02			RES,FXD,CMPSN:10 OHM,5%,0.25W	01121	CB1005
A9R1897	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A9R1898	313-1102-00			RES,FXD,FILM:1K OHM,5%,0.2W	57668	TR20JE01K0
A9R1901	315-0101-03			RES,FXD,CMPSN:100 OHM,5%,0.25W	01121	CB1015
A9R1910	321-0271-00			RES,FXD,FILM:6.49K OHM,1%,0.125W,TC=T0	07716	CEAD64900F
A9R1911	321-0245-00			RES,FXD,FILM:3.48K OHM,1%,0.125W,TC=T0	19701	5033ED3K48F
A9R1913	315-0101-03			RES,FXD,CMPSN:100 OHM,5%,0.25W	01121	CB1015
A9R1920	315-0152-00			RES,FXD,FILM:1.5K OHM,5%,0.25W	57668	NTR25J-E01K5
A9R1922	315-0331-03			RES,FXD,CMPSN:330 OHM 5%,0.25W	01121	CB3315
A9R1941	313-1201-00			RES,FXD,FILM:200 OHM,5%,0.2W	57668	TR20JE200E
A9R1944	321-0306-00			RES,FXD,FILM:15.0K OHM,1%,0.125W,TC=T0	19701	5033ED15J00F
A9R1945	321-0963-07			RES,FXD,FILM:98.73K OHM,0.1%,0.125W,TC=T9	07716	CEA 98.73KOHM 1%
A9R1950	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A9R1951	313-1220-00			RES,FXD,FILM:22 OHM,5%,0.2W	57668	TR20JE22E
A9R1952	313-0202-00			RES,FXD,FILM:2K OHM,5%,0.166W	80009	313-0202-00
A9R1953	313-1393-00			RES,FXD,FILM:39K OHM,5%,0.2W	57668	TR20JE 39K
A9R1971	313-1202-00			RES,FXD,FILM:2K OHM,5%,0.2W	57668	TR20JE02K0
A9R1972	313-1224-00			RES,FXD,FILM:220K,5%,0.2W	57668	TR20JE 220K
A9R1973	313-1124-00			RES,FXD,FILM:120K OHM,5%,0.2W	57668	TR20JE120K
A9R1990	321-0693-00			RES,FXD,FILM:68.1K OHM,0.5%,0.125W,TC=T0	19701	5033RD6812DB2980
A9R1991	315-0107-00			RES,FXD,FILM:100M OHM,5%,0.25W	01121	CB1075
A9R1992	313-1394-00			RES,FXD,FILM:390K,5%,0.2W	57668	TR20JE 390K
A9R1994	321-0402-00			RES,FXD,FILM:150K OHM,1%,0.125W,TC=T0	19701	5033ED150K0F
A9T1970	120-1418-01			XFMR,PWR,SDN&SU:HIGH VOLTAGE	80009	120-1418-01
A9U1062	156-0411-02			MICROCKT,LINEAR:QUAD COMPARATOR,SCREENED	04713	LM339JDS
A9U1830	152-0805-00			SEMICOND DVC,DI:HV MULTR,4.67KV INPUT,+14KV	S4431	MSR8506
A9U1890	156-1191-01			MICROCKT,LINEAR:BIFET,DUAL OPNL AMPL,SCRN	80009	156-1191-01
A9U1956	156-0158-07			MICROCKT,LINEAR:DUAL OPNL AMPL,SCREENED	01295	MC1458JG4
A9VR1891	152-0282-00			SEMICOND DVC,DI:ZEN,SI,30V,2%,400MW,DO-35	14552	1N972B
A9W900	198-4603-01			WIRE SET,ELEC:W/CRT SOCKET	80009	198-4603-01
A9W1909	131-0566-00			BUS,CONDUCTOR:DUMMY RES,0.094 OD X 0.225 L	24546	OMA 07

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
A14	670-8000-00			CIRCUIT BD ASSY:DYNAMIC CENTERING	80009	670-8000-00
A14C3401	281-0775-01			CAP,FXD,CER DI:0.1UF,20%,50V	04222	SA105E104MAA
A14J141	131-0608-00			TERMINAL,PIN:0.365 L X 0.025 BRZ GLD PL (QUANTITY OF 5)	22526	48283-036
A14R3401	311-2234-00			RES,VAR,NONWW:TRMR,5K OHM,20%,0.5W LINEAR	TK1450	GF06UT 5K
A14R3402	313-1222-00			RES,FXD,FILM:2.2K OHM,5%,0.2W	57668	TR20JE 02K2
A14R3403	313-1750-00			RES,FXD,FILM:75 OHM,5%,0.2W	57668	TR20JE 75E
A14R3404	321-0284-00			RES,FXD,FILM:8.87K OHM,1%,0.125W,TC=T0	19701	5043ED8K870F
A14R3405	313-1750-00			RES,FXD,FILM:75 OHM,5%,0.2W	57668	TR20JE 75E
A14R3407	311-2234-00			RES,VAR,NONWW:TRMR,5K OHM,20%,0.5W LINEAR	TK1450	GF06UT 5K
A14R3408	321-0284-00			RES,FXD,FILM:8.87K OHM,1%,0.125W,TC=T0	19701	5043ED8K870F
A14R3409	313-1222-00			RES,FXD,FILM:2.2K OHM,5%,0.2W	57668	TR20JE 02K2
A14R3410	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A14R3411	313-1103-00			RES,FXD,FILM:10K OHM,5%,0.2W	57668	TR20JE10K0
A14U3401	156-0130-00			MICROCKT,LINEAR:MODULATOR/DEMODULATOR	80009	156-0130-00
A14U3402	156-0130-00			MICROCKT,LINEAR:MODULATOR/DEMODULATOR	80009	156-0130-00
A14VR3401	152-0227-00			SEMICOND DVC,DI:ZEN,SI,6.2V,5%,0.4W,DO-7	04713	SZ13903

Component No.	Tektronix Part No.	Serial/Assembly No. Effective	Dscont	Name & Description	Mfr. Code	Mfr. Part No.
C10	281-0697-00			CAP,FXD,CER DI:5000PF,+100-0%,100V	72982	2425-003W5W0502Z
F90	159-0021-00			FUSE,CARTRIDGE:3AG,2A,250V,FAST BLOW	71400	AGC-CW-2
L90	119-1478-01			COIL,TUBE DEFL:FXD,TRACE ROTATION	80009	119-1478-01
R134	311-2312-01			RES,VAR,NONWW:PNL,5K OHM,20%,0.5W	80009	311-2312-01
R351	311-2312-01			RES,VAR,NONWW:PNL,5K OHM,20%,0.5W	80009	311-2312-01
R352	311-2312-01			RES,VAR,NONWW:PNL,5K OHM,20%,0.5W	80009	311-2312-01
R975	311-2312-01			RES,VAR,NONWW:PNL,5K OHM,20%,0.5W	80009	311-2312-01
R976	311-2312-01			RES,VAR,NONWW:PNL,5K OHM,20%,0.5W	80009	311-2312-01
R977	311-2312-01			RES,VAR,NONWW:PNL,5K OHM,20%,0.5W	80009	311-2312-01
S90	260-1967-00			SWITCH,SLIDE:DPDT 5A/250V 10A/125V MKD	TK0935	4021.0512
V900	154-0850-01			CRT ASSEMBLY:FINISHED 2445	80009	154-0850-01